Regarding the change of names mentioned in the document, such as Hitachi Electric and Hitachi XX, to Renesas Technology Corp.

The semiconductor operations of Mitsubishi Electric and Hitachi were transferred to Renesas Technology Corporation on April 1st 2003. These operations include microcomputer, logic, analog and discrete devices, and memory chips other than DRAMs (flash memory, SRAMs etc.)

Accordingly, although Hitachi, Hitachi, Ltd., Hitachi Semiconductors, and other Hitachi brand names are mentioned in the document, these names have in fact all been changed to Renesas Technology Corp.

Thank you for your understanding. Except for our corporate trademark, logo and corporate statement, no changes whatsoever have been made to the contents of the document, and these changes do not constitute any alteration to the contents of the document itself.

Renesas Technology Home Page: www.renesas.com

Renesas Technology Corp.
Customer Support Dept.
April 1, 2003

Hitachi Single-Chip Microcomputer

$$
\mathrm{H} 8 / 3052 \mathrm{~F}^{\mathrm{F}} \mathrm{ZTAT}^{\mathrm{TM}}
$$

# HD64F3052TE, HD64F3052F, HD64F3052BTE, HD64F3052BF, HD64F3052BVTE, HD64F3052BVF 

Hardware Manual

## HITACHI

## Cautions

1. Hitachi neither warrants nor grants licenses of any rights of Hitachi's or any third party's patent, copyright, trademark, or other intellectual property rights for information contained in this document. Hitachi bears no responsibility for problems that may arise with third party's rights, including intellectual property rights, in connection with use of the information contained in this document.
2. Products and product specifications may be subject to change without notice. Confirm that you have received the latest product standards or specifications before final design, purchase or use.
3. Hitachi makes every attempt to ensure that its products are of high quality and reliability. However, contact Hitachi's sales office before using the product in an application that demands especially high quality and reliability or where its failure or malfunction may directly threaten human life or cause risk of bodily injury, such as aerospace, aeronautics, nuclear power, combustion control, transportation, traffic, safety equipment or medical equipment for life support.
4. Design your application so that the product is used within the ranges guaranteed by Hitachi particularly for maximum rating, operating supply voltage range, heat radiation characteristics, installation conditions and other characteristics. Hitachi bears no responsibility for failure or damage when used beyond the guaranteed ranges. Even within the guaranteed ranges, consider normally foreseeable failure rates or failure modes in semiconductor devices and employ systemic measures such as fail-safes, so that the equipment incorporating Hitachi product does not cause bodily injury, fire or other consequential damage due to operation of the Hitachi product.
5. This product is not designed to be radiation resistant.
6. No one is permitted to reproduce or duplicate, in any form, the whole or part of this document without written approval from Hitachi.
7. Contact Hitachi's sales office for any questions regarding this document or Hitachi semiconductor products.

## Preface

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ is a series of high-performance microcontrollers that integrate system supporting functions together with an $\mathrm{H} 8 / 300 \mathrm{H}$ CPU core.

The H8/300H CPU has a 32 -bit internal architecture with sixteen 16-bit general registers, and a concise, optimized instruction set designed for speed. It can address a 16-Mbyte linear address space.

The on-chip supporting functions include ROM, RAM, a 16-bit integrated timer unit (ITU), a programmable timing pattern controller (TPC), a watchdog timer (WDT), a serial communication interface (SCI), an A/D converter, a D/A converter, I/O ports, a direct memory access controller (DMAC), a refresh controller, and other facilities. Of the two SCI channels, one has been expanded to support the ISO/IEC7816-3 smart card interface. Functions have also been added to reduce power consumption in battery-powered applications: individual modules can be placed in standby, and the frequency of the system clock supplied to the chip can be divided down under software control.

The address space is divided into eight areas. The data bus width and access cycle length can be selected independently in each area, simplifying the connection of different types of memory. Seven operating modes (modes 1 to 7 ) are provided, offering a choice of data bus width and address space size.

With these features, the $\mathrm{H} 8 / 3052 \mathrm{~F}$ can be used to implement compact, high-performance systems easily.

The H8/3052F has an F-ZTAT ${ }^{\text {TM } *}$ version with on-chip flash memory that can be programmed on-board. These versions enable users to respond quickly and flexibly to changing application specifications.

This manual describes the H8/3052F hardware. For details of the instruction set, refer to the H8/300H Series Programming Manual.

Note: * F-ZTAT ${ }^{\text {TM }}$ (Flexible-Zero Turn Around Time) is a trademark of Hitachi, Ltd.

## List of Items Revised or Added for This Version

| Section | Page | Item | Description |
| :---: | :---: | :---: | :---: |
| All | - |  | Amendments due to introduction of the H8/3052F-ZTAT B mask version. |
| 1.1 Overview | 5 | Table 1.1 Feature | Product lineup amended |
| 1.2 Block Diagram | 6 | Figure1.1 Block Diagram | $\mathrm{V}_{\mathrm{cL}}$ amended to $\mathrm{V}_{\mathrm{cL}} / \mathrm{V}_{\mathrm{cc}}$ |
| 1.3.2 Pin Assignments in Each Mode | 8,12 | Table 1.2 Pin Assignments in Each Mode(FP-100B or TFP-100B) | Pin 1 and note 1 amended |
| 1.3.3 Pin Functions | 17 | Table1.3 Pin Functions | Note added |
| 18.10.1 Socket Adapters and Memory Map | 603 | Table 18.12 H8/3052F Socket Adapter Product Codes | Product codes added |
| 19.2 Oscillator Circuit | 611 | 19.2.1 Connecting a Crystal Resonator "Circuit Configuration" | Description added |
|  |  | Figure 19.2 Connecting of Crystal Resonator (Example) | Amended |
|  |  | Table 19.1(1) Damping Resistance Value | 25 MHz added |
|  |  | Table 19.1(2) External Capacitance Values | Added |
|  | 612 | Table 19.2 Crystal Resonator Parameters | 20 and 25 MHz added |
| 19.2.2 External Clock Input | 614 | Table 19.3 Clock Timing | Amended |
| 21.1 Absolute Maximum Ratings | 635 | Table 21.1 Absolute Maximum Ratings | Amended |
|  | 637 | Table 21.2 (1)DC Characteristics | Current dissipation amended |
|  | $\begin{aligned} & 639, \\ & 640 \end{aligned}$ | Table 21.2(2) DC Characteristics | Added |
|  | 641 | Table 21.3 Permissible Output Currents | Conditions amended |
| 21.2.2 AC <br> Characteristics | $\begin{aligned} & 643, \\ & 644 \end{aligned}$ | Table 21.4 Bus Timing | Amended |
|  | 645 | Table 21.5 Refresh Controller Bus Timing | Amended |
|  | 646 | Table 21.6 Control Signal Timing | Amended |


| Section | Page | Item | Description |
| :--- | :--- | :--- | :--- |
| 21.2 .2 AC <br> Characteristics | 647 | Table 21.7 Timing of On-Chip <br> Supporting Modules | Amended |
| 21.2.3 A/D <br> Conversion <br> Characteristics | 649 | Table 21.8 A/D Converter <br> Characteristics | Amended |
| 21.2 .4 D/A <br> Conversion <br> Characteristics | 650 | Table 21.9 D/A Converter <br> Characteristics | Amended |
| 21.2 .5 Flash <br> Memory <br> Characteristics | 651 | Table 21.10 Flash Memory <br> Characteristics | Conditions amended |
| Appendix F Product <br> Code Lineup | 816 | Table F.1 H8/3052F Product Code <br> Lineup | Product types added |
| Appendix H | 819 | Table H.1 Differences between | H8/3052F-ZTAT <br> Differences from <br> H8/3048F-ZTAT |
|  | H8/3052F-ZTAT and H8/3048F-ZTAT | amended |  |

## Contents

Section 1 Overview ..... 1
1.1 Overview ..... 1
1.2 Block Diagram ..... 6
1.3 Pin Description ..... 7
1.3.1 Pin Arrangement ..... 7
1.3.2 Pin Assignments in Each Mode ..... 8
1.3.3 Pin Functions ..... 13
Section 2 CPU ..... 19
2.1 Overview ..... 19
2.1.1 Features ..... 19
2.1.2 Differences from H8/300 CPU ..... 20
2.2 CPU Operating Modes ..... 21
2.3 Address Space ..... 22
2.4 Register Configuration ..... 23
2.4.1 Overview ..... 23
2.4.2 General Registers ..... 24
2.4.3 Control Registers ..... 25
2.4.4 Initial CPU Register Values ..... 26
2.5 Data Formats ..... 27
2.5.1 General Register Data Formats ..... 27
2.5.2 Memory Data Formats ..... 29
2.6 Instruction Set ..... 30
2.6.1 Instruction Set Overview ..... 30
2.6.2 Instructions and Addressing Modes ..... 31
2.6.3 Tables of Instructions Classified by Function ..... 32
2.6.4 Basic Instruction Formats ..... 42
2.6.5 Notes on Use of Bit Manipulation Instructions ..... 43
2.7 Addressing Modes and Effective Address Calculation ..... 44
2.7.1 Addressing Modes ..... 44
2.7.2 Effective Address Calculation ..... 47
2.8 Processing States ..... 51
2.8.1 Overview ..... 51
2.8.2 Program Execution State ..... 52
2.8.3 Exception-Handling State ..... 52
2.8.4 Exception-Handling Sequences ..... 54
2.8.5 Bus-Released State ..... 55
2.8.6 Reset State ..... 55
2.8.7 Power-Down State ..... 55
2.9 Basic Operational Timing ..... 56
2.9.1 Overview ..... 56
2.9.2 On-Chip Memory Access Timing ..... 56
2.9.3 On-Chip Supporting Module Access Timing ..... 57
2.9.4 Access to External Address Space ..... 58
Section 3 MCU Operating Modes ..... 59
3.1 Overview ..... 59
3.1.1 Operating Mode Selection ..... 59
3.1.2 Register Configuration ..... 60
3.2 Mode Control Register (MDCR) ..... 60
3.3 System Control Register (SYSCR) ..... 61
3.4 Operating Mode Descriptions ..... 63
3.4.1 Mode 1 ..... 63
3.4.2 Mode 2 ..... 63
3.4.3 Mode 3 ..... 63
3.4.4 Mode 4 ..... 63
3.4.5 Mode 5 ..... 63
3.4.6 Mode 6 ..... 64
3.4.7 Mode 7 ..... 64
3.5 Pin Functions in Each Operating Mode ..... 64
3.6 Memory Map in Each Operating Mode ..... 65
Section 4 Exception Handling ..... 69
4.1 Overview ..... 69
4.1.1 Exception Handling Types and Priority ..... 69
4.1.2 Exception Handling Operation ..... 69
4.1.3 Exception Sources and Vector Table ..... 70
4.2 Reset ..... 72
4.2.1 Overview ..... 72
4.2.2 Reset Sequence ..... 72
4.2.3 Interrupts after Reset ..... 75
4.3 Interrupts ..... 76
4.4 Trap Instruction. ..... 77
4.5 Stack Status after Exception Handling ..... 77
4.6 Notes on Use of the Stack ..... 78
Section 5 Interrupt Controller ..... 79
5.1 Overview ..... 79
5.1.1 Features ..... 79
5.1.2 Block Diagram ..... 80
5.1.3 Pin Configuration ..... 81
5.1.4 Register Configuration ..... 81
5.2 Register Descriptions ..... 82
5.2.1 System Control Register (SYSCR) ..... 82
5.2.2 Interrupt Priority Registers A and B (IPRA, IPRB) ..... 83
5.2.3 IRQ Status Register (ISR) ..... 90
5.2.4 IRQ Enable Register (IER) ..... 91
5.2.5 IRQ Sense Control Register (ISCR) ..... 92
5.3 Interrupt Sources ..... 93
5.3.1 External Interrupts ..... 93
5.3.2 Internal Interrupts ..... 94
5.3.3 Interrupt Exception Vector Table ..... 94
5.4 Interrupt Operation ..... 98
5.4.1 Interrupt Handling Process ..... 98
5.4.2 Interrupt Exception Handling Sequence ..... 103
5.4.3 Interrupt Response Time ..... 104
5.5 Usage Notes ..... 105
5.5.1 Contention between Interrupt Generation and Disabling ..... 105
5.5.2 Instructions that Inhibit Interrupts. ..... 106
5.5.3 Interrupts during EEPMOV Instruction Execution ..... 106
5.5.4 Notes on Use of External Interrupts ..... 106
Section 6 Bus Controller ..... 109
6.1 Overview ..... 109
6.1.1 Features ..... 109
6.1.2 Block Diagram ..... 110
6.1.3 Pin Configuration ..... 111
6.1.4 Register Configuration ..... 112
6.2 Register Descriptions ..... 112
6.2.1 Bus Width Control Register (ABWCR) ..... 112
6.2.2 Access State Control Register (ASTCR) ..... 113
6.2.3 Wait Control Register (WCR) ..... 114
6.2.4 Wait State Controller Enable Register (WCER) ..... 115
6.2.5 Bus Release Control Register (BRCR) ..... 116
6.2.6 Chip Select Control Register (CSCR) ..... 118
6.3 Operation ..... 119
6.3.1 Area Division ..... 119
6.3.2 Chip Select Signals ..... 121
6.3.3 Data Bus. ..... 122
6.3.4 Bus Control Signal Timing ..... 123
6.3.5 Wait Modes ..... 131
6.3.6 Interconnections with Memory (Example) ..... 137
6.3.7 Bus Arbiter Operation ..... 139
6.4 Usage Notes ..... 142
6.4.1 Connection to Dynamic RAM and Pseudo-Static RAM ..... 142
6.4.2 Register Write Timing ..... 142
6.4.3 BREQ Input Timing ..... 144
6.4.4 Transition To Software Standby Mode ..... 144
Section 7 Refresh Controller ..... 145
7.1 Overview ..... 145
7.1.1 Features ..... 145
7.1.2 Block Diagram ..... 147
7.1.3 Pin Configuration ..... 148
7.1.4 Register Configuration ..... 148
7.2 Register Descriptions ..... 149
7.2.1 Refresh Control Register (RFSHCR) ..... 149
7.2.2 Refresh Timer Control/Status Register (RTMCSR) ..... 152
7.2.3 Refresh Timer Counter (RTCNT) ..... 153
7.2.4 Refresh Time Constant Register (RTCOR) ..... 154
7.3 Operation ..... 155
7.3.1 Overview ..... 155
7.3.2 DRAM Refresh Control ..... 157
7.3.3 Pseudo-Static RAM Refresh Control ..... 172
7.3.4 Interval Timing ..... 176
7.4 Interrupt Source ..... 182
7.5 Usage Notes ..... 182
Section 8 DMA Controller ..... 185
8.1 Overview ..... 185
8.1.1 Features ..... 185
8.1.2 Block Diagram ..... 186
8.1.3 Functional Overview ..... 187
8.1.4 Pin Configuration ..... 189
8.1.5 Register Configuration ..... 189
8.2 Register Descriptions (Short Address Mode) ..... 191
8.2.1 Memory Address Registers (MAR) ..... 192
8.2.2 I/O Address Registers (IOAR) ..... 193
8.2.3 Execute Transfer Count Registers (ETCR) ..... 194
8.2.4 Data Transfer Control Registers (DTCR) ..... 195
8.3 Register Descriptions (Full Address Mode) ..... 198
8.3.1 Memory Address Registers (MAR) ..... 198
8.3.2 I/O Address Registers (IOAR) ..... 199
8.3.3 Execute Transfer Count Registers (ETCR) ..... 199
8.3.4 Data Transfer Control Registers (DTCR) ..... 201
8.4 Operation ..... 206
8.4.1 Overview ..... 206
8.4.2 I/O Mode ..... 208
8.4.3 Idle Mode ..... 210
8.4.4 Repeat Mode ..... 213
8.4.5 Normal Mode ..... 217
8.4.6 Block Transfer Mode ..... 220
8.4.7 DMAC Activation ..... 225
8.4.8 DMAC Bus Cycle ..... 227
8.4.9 DMAC Multiple-Channel Operation ..... 233
8.4.10 External Bus Requests, Refresh Controller, and DMAC ..... 234
8.4.11 NMI Interrupts and DMAC ..... 235
8.4.12 Aborting a DMA Transfer ..... 236
8.4.13 Exiting Full Address Mode ..... 237
8.4.14 DMAC States in Reset State, Standby Modes, and Sleep Mode ..... 238
8.5 Interrupts ..... 239
8.6 Usage Notes ..... 240
8.6.1 Note on Word Data Transfer ..... 240
8.6.2 DMAC Self-Access ..... 240
8.6.3 Longword Access to Memory Address Registers ..... 240
8.6.4 Note on Full Address Mode Setup ..... 240
8.6.5 Note on Activating DMAC by Internal Interrupts ..... 240
8.6.6 NMI Interrupts and Block Transfer Mode ..... 242
8.6.7 Memory and I/O Address Register Values ..... 242
8.6.8 Bus Cycle when Transfer is Aborted ..... 243
Section 9 I/O Ports ..... 245
9.1 Overview ..... 245
9.2 Port 1 ..... 249
9.2.1 Overview ..... 249
9.2.2 Register Configuration ..... 250
9.3 Port 2 ..... 252
9.3.1 Overview ..... 252
9.3.2 Register Configuration ..... 253
9.4 Port 3 ..... 256
9.4.1 Overview ..... 256
9.4.2 Register Configuration ..... 256
9.5 Port 4 ..... 258
9.5.1 Overview ..... 258
9.5.2 Register Configuration ..... 259
9.6 Port 5 ..... 262
9.6.1 Overview ..... 262
9.6.2 Register Configuration ..... 263
9.7 Port 6 ..... 266
9.7.1 Overview ..... 266
9.7.2 Register Configuration ..... 267
9.8 Port 7 ..... 270
9.8.1 Overview ..... 270
9.8.2 Register Configuration ..... 271
9.9 Port 8 ..... 272
9.9.1 Overview ..... 272
9.9.2 Register Configuration ..... 273
9.10 Port 9 ..... 277
9.10.1 Overview ..... 277
9.10.2 Register Configuration ..... 278
9.11 Port A ..... 282
9.11.1 Overview ..... 282
9.11.2 Register Configuration ..... 284
9.11.3 Pin Functions ..... 286
9.12 Port B ..... 294
9.12.1 Overview ..... 294
9.12.2 Register Configuration ..... 296
9.12.3 Pin Functions ..... 298
Section 10 16-Bit Integrated Timer Unit (ITU) ..... 305
10.1 Overview ..... 305
10.1.1 Features ..... 305
10.1.2 Block Diagrams ..... 308
10.1.3 Pin Configuration ..... 313
10.1.4 Register Configuration ..... 314
10.2 Register Descriptions ..... 317
10.2.1 Timer Start Register (TSTR) ..... 317
10.2.2 Timer Synchro Register (TSNC) ..... 318
10.2.3 Timer Mode Register (TMDR) ..... 320
10.2.4 Timer Function Control Register (TFCR) ..... 323
10.2.5 Timer Output Master Enable Register (TOER) ..... 325
10.2.6 Timer Output Control Register (TOCR) ..... 327
10.2.7 Timer Counters (TCNT) ..... 328
10.2.8 General Registers (GRA, GRB) ..... 329
10.2.9 Buffer Registers (BRA, BRB) ..... 330
10.2.10 Timer Control Registers (TCR) ..... 331
10.2.11 Timer I/O Control Register (TIOR) ..... 333
10.2.12 Timer Status Register (TSR) ..... 335
10.2.13 Timer Interrupt Enable Register (TIER) ..... 337
10.3 CPU Interface ..... 339
10.3.1 16-Bit Accessible Registers ..... 339
10.3.2 8-Bit Accessible Registers ..... 341
10.4 Operation ..... 343
10.4.1 Overview ..... 343
10.4.2 Basic Functions ..... 344
10.4.3 Synchronization ..... 353
10.4.4 PWM Mode. ..... 355
10.4.5 Reset-Synchronized PWM Mode. ..... 359
10.4.6 Complementary PWM Mode ..... 362
10.4.7 Phase Counting Mode ..... 371
10.4.8 Buffering ..... 373
10.4.9 ITU Output Timing ..... 380
10.5 Interrupts ..... 382
10.5.1 Setting of Status Flags ..... 382
10.5.2 Clearing of Status Flags ..... 384
10.5.3 Interrupt Sources and DMA Controller Activation. ..... 385
10.6 Usage Notes ..... 386
Section 11 Programmable Timing Pattern Controller. ..... 401
11.1 Overview ..... 401
11.1.1 Features ..... 401
11.1.2 Block Diagram ..... 402
11.1.3 Pin Configuration ..... 403
11.1.4 Register Configuration ..... 404
11.2 Register Descriptions ..... 405
11.2.1 Port A Data Direction Register (PADDR) ..... 405
11.2.2 Port A Data Register (PADR) ..... 405
11.2.3 Port B Data Direction Register (PBDDR) ..... 406
11.2.4 Port B Data Register (PBDR) ..... 406
11.2.5 Next Data Register A (NDRA) ..... 407
11.2.6 Next Data Register B (NDRB) ..... 409
11.2.7 Next Data Enable Register A (NDERA) ..... 411
11.2.8 Next Data Enable Register B (NDERB) ..... 412
11.2.9 TPC Output Control Register (TPCR) ..... 413
11.2.10 TPC Output Mode Register (TPMR) ..... 415
11.3 Operation ..... 417
11.3.1 Overview ..... 417
11.3.2 Output Timing ..... 418
11.3.3 Normal TPC Output ..... 419
11.3.4 Non-Overlapping TPC Output ..... 421
11.3.5 TPC Output Triggering by Input Capture ..... 423
11.4 Usage Notes ..... 424
11.4.1 Operation of TPC Output Pins ..... 424
11.4.2 Note on Non-Overlapping Output. ..... 424
Section 12 Watchdog Timer ..... 427
12.1 Overview ..... 427
12.1.1 Features ..... 427
12.1.2 Block Diagram ..... 428
12.1.3 Register Configuration ..... 428
12.2 Register Descriptions ..... 429
12.2.1 Timer Counter (TCNT) ..... 429
12.2.2 Timer Control/Status Register (TCSR) ..... 430
12.2.3 Reset Control/Status Register (RSTCSR) ..... 432
12.2.4 Notes on Register Access ..... 433
12.3 Operation ..... 434
12.3.1 Watchdog Timer Operation ..... 434
12.3.2 Interval Timer Operation ..... 435
12.3.3 Timing of Setting of Overflow Flag (OVF) ..... 435
12.3.4 Timing of Setting of Watchdog Timer Reset Bit (WRST) ..... 436
12.4 Interrupts ..... 437
12.5 Usage Notes ..... 437
Section 13 Serial Communication Interface ..... 439
13.1 Overview ..... 439
13.1.1 Features ..... 439
13.1.2 Block Diagram ..... 441
13.1.3 Pin Configuration ..... 442
13.1.4 Register Configuration ..... 442
13.2 Register Descriptions ..... 443
13.2.1 Receive Shift Register (RSR) ..... 443
13.2.2 Receive Data Register (RDR) ..... 443
13.2.3 Transmit Shift Register (TSR) ..... 444
13.2.4 Transmit Data Register (TDR) ..... 444
13.2.5 Serial Mode Register (SMR) ..... 445
13.2.6 Serial Control Register (SCR) ..... 448
13.2.7 Serial Status Register (SSR) ..... 452
13.2.8 Bit Rate Register (BRR) ..... 456
13.3 Operation ..... 466
13.3.1 Overview ..... 466
13.3.2 Operation in Asynchronous Mode ..... 468
13.3.3 Multiprocessor Communication ..... 477
13.3.4 Synchronous Operation. ..... 484
13.4 SCI Interrupts ..... 492
13.5 Usage Notes ..... 493
Section 14 Smart Card Interface ..... 499
14.1 Overview ..... 499
14.1.1 Features ..... 499
14.1.2 Block Diagram ..... 500
14.1.3 Pin Configuration ..... 501
14.1.4 Register Configuration. ..... 501
14.2 Register Descriptions ..... 502
14.2.1 Smart Card Mode Register (SCMR) ..... 502
14.2.2 Serial Status Register (SSR) ..... 503
14.2.3 Serial Mode Register (SMR). ..... 505
14.2.4 Serial Control Register (SCR) ..... 506
14.3 Operation ..... 507
14.3.1 Overview ..... 507
14.3.2 Pin Connections ..... 507
14.3.3 Data Format ..... 509
14.3.4 Register Settings ..... 510
14.3.5 Clock ..... 512
14.3.6 Transmitting and Receiving Data ..... 514
14.4 Usage Notes ..... 521
Section 15 A/D Converter ..... 525
15.1 Overview ..... 525
15.1.1 Features ..... 525
15.1.2 Block Diagram ..... 526
15.1.3 Pin Configuration. ..... 527
15.1.4 Register Configuration ..... 528
15.2 Register Descriptions ..... 529
15.2.1 A/D Data Registers A to D (ADDRA to ADDRD) ..... 529
15.2.2 A/D Control/Status Register (ADCSR) ..... 530
15.2.3 A/D Control Register (ADCR) ..... 532
15.3 CPU Interface. ..... 533
15.4 Operation ..... 534
15.4.1 Single Mode $(\mathrm{SCAN}=0)$ ..... 534
15.4.2 Scan Mode (SCAN = 1) ..... 536
15.4.3 Input Sampling and A/D Conversion Time ..... 538
15.4.4 External Trigger Input Timing ..... 539
15.5 Interrupts ..... 540
15.6 Usage Notes ..... 540
Section 16 D/A Converter ..... 547
16.1 Overview ..... 547
16.1.1 Features ..... 547
16.1.2 Block Diagram ..... 548
16.1.3 Pin Configuration ..... 549
16.1.4 Register Configuration ..... 549
16.2 Register Descriptions ..... 550
16.2.1 D/A Data Registers 0 and 1 (DADR0/1) ..... 550
16.2.2 D/A Control Register (DACR) ..... 550
16.2.3 D/A Standby Control Register (DASTCR) ..... 552
16.3 Operation ..... 553
16.4 D/A Output Control ..... 554
Section 17 RAM ..... 555
17.1 Overview ..... 555
17.1.1 Block Diagram ..... 555
17.1.2 Register Configuration ..... 556
17.2 System Control Register (SYSCR) ..... 556
17.3 Operation ..... 557
Section 18 ROM ..... 559
18.1 Features ..... 559
18.2 Overview ..... 560
18.2.1 Block Diagram ..... 560
18.2.2 Mode Transitions ..... 560
18.2.3 On-Board Programming Modes ..... 563
18.2.4 Flash Memory Emulation in RAM ..... 565
18.2.5 Differences between Boot Mode and User Program Mode ..... 566
18.2.6 Block Configuration ..... 567
18.3 Pin Configuration ..... 568
18.4 Register Configuration ..... 568
18.5 Register Descriptions ..... 569
18.5.1 Flash Memory Control Register 1 (FLMCR1) ..... 569
18.5.2 Flash Memory Control Register 2 (FLMCR2) ..... 571
18.5.3 Erase Block Register 1 (EBR1) ..... 574
18.5.4 Erase Block Register 2 (EBR2) ..... 575
18.5.5 RAM Control Register (RAMCR) ..... 576
18.6 On-Board Programming Modes ..... 577
18.6.1 Boot Mode ..... 578
18.6.2 User Program Mode ..... 583
18.7 Programming/Erasing Flash Memory ..... 585
18.7.1 Program Mode ..... 587
18.7.2 Program-Verify Mode. ..... 588
18.7.3 Notes on Program/Program-Verify Procedure ..... 588
18.7.4 Erase Mode ..... 592
18.7.5 Erase-Verify Mode ..... 592
18.8 Protection ..... 594
18.8.1 Hardware Protection ..... 594
18.8.2 Software Protection. ..... 596
18.8.3 Error Protection ..... 597
18.8.4 NMI Input Disable Conditions ..... 599
18.9 Flash Memory Emulation in RAM ..... 600
18.10 Flash Memory PROM Mode ..... 602
18.10.1 Socket Adapters and Memory Map ..... 602
18.10.2 Notes on Use of PROM Mode ..... 603
18.11 Notes on Flash Memory Programming/Erasing ..... 604
Section 19 Clock Pulse Generator ..... 609
19.1 Overview ..... 609
19.1.1 Block Diagram ..... 610
19.2 Oscillator Circuit ..... 611
19.2.1 Connecting a Crystal Resonator. ..... 611
19.2.2 External Clock Input. ..... 613
19.3 Duty Adjustment Circuit ..... 615
19.4 Prescalers ..... 615
19.5 Frequency Divider ..... 615
19.5.1 Register Configuration ..... 616
19.5.2 Division Control Register (DIVCR) ..... 616
19.5.3 Usage Notes ..... 617
Section 20 Power-Down State ..... 619
20.1 Overview ..... 619
20.2 Register Configuration ..... 621
20.2.1 System Control Register (SYSCR) ..... 621
20.2.2 Module Standby Control Register (MSTCR) ..... 623
20.3 Sleep Mode ..... 625
20.3.1 Transition to Sleep Mode ..... 625
20.3.2 Exit from Sleep Mode ..... 625
20.4 Software Standby Mode ..... 626
20.4.1 Transition to Software Standby Mode ..... 626
20.4.2 Exit from Software Standby Mode ..... 626
20.4.3 Selection of Waiting Time for Exit from Software Standby Mode ..... 627
20.4.4 Sample Application of Software Standby Mode ..... 629
20.4.5 Note ..... 629
20.5 Hardware Standby Mode ..... 630
20.5.1 Transition to Hardware Standby Mode ..... 630
20.5.2 Exit from Hardware Standby Mode ..... 630
20.5.3 Timing for Hardware Standby Mode ..... 630
20.6 Module Standby Function ..... 631
20.6.1 Module Standby Timing ..... 631
20.6.2 Read/Write in Module Standby ..... 631
20.6.3 Usage Notes ..... 632
20.7 System Clock Output Disabling Function ..... 633
Section 21 Electrical Characteristics (Preliminary) ..... 635
21.1 Absolute Maximum Ratings ..... 635
21.2 Electrical Characteristics ..... 636
21.2.1 DC Characteristics ..... 636
21.2.2 AC Characteristics ..... 643
21.2.3 A/D Conversion Characteristics. ..... 649
21.2.4 D/A Conversion Characteristics ..... 650
21.2.5 Flash Memory Characteristics ..... 651
21.3 Operational Timing ..... 652
21.3.1 Bus Timing ..... 652
21.3.2 Refresh Controller Bus Timing ..... 656
21.3.3 Control Signal Timing ..... 661
21.3.4 Clock Timing ..... 663
21.3.5 TPC and I/O Port Timing ..... 663
21.3.6 ITU Timing ..... 664
21.3.7 SCI Input/Output Timing ..... 665
21.3.8 DMAC Timing ..... 666
Appendix A Instruction Set. ..... 667
A. 1 Instruction List ..... 667
A. 2 Operation Code Map ..... 682
A. 3 Number of States Required for Execution ..... 685
Appendix B Internal I/O Register ..... 695
B. 1 Addresses ..... 695
B. 2 Function ..... 703
Appendix C I/O Port Block Diagrams ..... 784
C. 1 Port 1 Block Diagram ..... 784
C. 2 Port 2 Block Diagram ..... 785
C. 3 Port 3 Block Diagram ..... 786
C. 4 Port 4 Block Diagram ..... 787
C. 5 Port 5 Block Diagram ..... 788
C. 6 Port 6 Block Diagrams ..... 789
C. 7 Port 7 Block Diagrams ..... 793
C. 8 Port 8 Block Diagrams ..... 794
C. 9 Port 9 Block Diagrams ..... 797
C. 10 Port A Block Diagrams ..... 801
C. 11 Port B Block Diagrams ..... 805
Appendix D Pin States ..... 809
D. 1 Port States in Each Mode ..... 809
D. 2 Pin States at Reset ..... 812
Appendix E Timing of Transition to and Recovery from Hardware Standby Mode ..... 815
E. 1 Timing of Transition to Hardware Standby Mode ..... 815
E. 2 Timing of Recovery from Hardware Standby Mode. ..... 815
Appendix F Product Code Lineup ..... 816
Appendix G Package Dimensions ..... 817
Appendix H Differences from H8/3048F-ZTAT. ..... 819

## Section 1 Overview

### 1.1 Overview

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ is a series of microcontrollers (MCUs) that integrate system supporting functions together with an $\mathrm{H} 8 / 300 \mathrm{H}$ CPU core having an original Hitachi architecture.

The H8/300H CPU has a 32 -bit internal architecture with sixteen 16 -bit general registers, and a concise, optimized instruction set designed for speed. It can address a 16-Mbyte linear address space. Its instruction set is upward-compatible at the object-code level with the H8/300 CPU, enabling easy porting of software from the $\mathrm{H} 8 / 300$ Series.

The on-chip system supporting functions include ROM, RAM, a 16-bit integrated timer unit (ITU), a programmable timing pattern controller (TPC), a watchdog timer (WDT), a serial communication interface (SCI), an A/D converter, a D/A converter, I/O ports, a direct memory access controller (DMAC), a refresh controller, and other facilities.

The H8/3052F has 512 kbytes of ROM and 8 kbytes of RAM.
Seven MCU operating modes offer a choice of data bus width and address space size. The modes (modes 1 to 7) include one single-chip mode and six expanded modes.

The H8/3052F has an F-ZTAT ${ }^{\text {TM } *}$ version with on-chip flash memory that can be programmed on-board.

Table 1.1 summarizes the features of the $\mathrm{H} 8 / 3052 \mathrm{~F}$.
Note: * F-ZTAT (Flexible-Zero Turn Around Time) is a trademark of Hitachi, Ltd.

## Table 1.1 Features

## Feature Description

CPU
Upward-compatible with the H8/300 CPU at the object-code level

- General-register machine
- Sixteen 16-bit general registers
(also usable as + eight 16-bit registers or eight 32-bit registers)
- High-speed operation
— Maximum clock rate: 25 MHz
— Add/subtract: 80 ns
— Multiply/divide: 560 ns
- 16-Mbyte address space
- Instruction features
- 8/16/32-bit data transfer, arithmetic, and logic instructions
- Signed and unsigned multiply instructions ( 8 bits $\times 8$ bits, 16 bits $\times 16$ bits)
- Signed and unsigned divide instructions (16 bits $\div 8$ bits, 32 bits $\div 16$ bits)
- Bit accumulator function
- Bit manipulation instructions with register-indirect specification of bit positions

Memory - Flash memory: 512 kbytes

- RAM: 8 kbytes

Interrupt
controller

- Seven external interrupt pins: NMI, $\overline{\mathrm{IRQ}}_{0}$ to $\overline{\mathrm{IRQ}}_{5}$
- 30 internal interrupts
- Three selectable interrupt priority levels
Feature Description

Bus controller

- Address space can be partitioned into eight areas, with independent bus specifications in each area
- Chip select output available for areas 0 to 7
- 8-bit access or 16-bit access selectable for each area
- Two-state or three-state access selectable for each area
- Selection of four wait modes
- Bus arbitration function

Refresh controller

- DRAM refresh
- Directly connectable to 16-bit-wide DRAM
- CAS-before-RAS refresh
- Self-refresh mode selectable
- Pseudo-static RAM refresh
- Self-refresh mode selectable
- Usable as an interval timer

DMA controller (DMAC)

- Short address mode
- Maximum four channels available
- Selection of I/O mode, idle mode, or repeat mode
- Can be activated by compare match/input capture A interrupts from ITU channels 0 to 3, transmit-data-empty and receive-data-full interrupts from SCl channel 0 , or external requests
- Full address mode
- Maximum two channels available
- Selection of normal mode or block transfer mode
- Can be activated by compare match/input capture A interrupts from ITU channels 0 to 3, external requests, or auto-request

16-bit integrated timer unit (ITU)

- Five 16 -bit timer channels, capable of processing up to 12 pulse outputs or 10 pulse inputs
- 16 -bit timer counter (channels 0 to 4 )
- Two multiplexed output compare/input capture pins (channels 0 to 4 )
- Operation can be synchronized (channels 0 to 4 )
- PWM mode available (channels 0 to 4 )
- Phase counting mode available (channel 2 )
- Buffering available (channels 3 and 4)
- Reset-synchronized PWM mode available (channels 3 and 4)
- Complementary PWM mode available (channels 3 and 4 )
- DMAC can be activated by compare match/input capture A interrupts (channels 0 to 3 )

Programmable timing pattern controller (TPC)

- Maximum 16-bit pulse output, using ITU as time base
- Up to four 4-bit pulse output groups (or one 16-bit group, or two 8-bit groups)
- Non-overlap mode available
- Output data can be transferred by DMAC

Watchdog timer
(WDT),
1 channel

- Reset signal can be generated by overflow
- Usable as an interval timer

Serial communication interface (SCI), 2 channels

- Selection of asynchronous or synchronous mode
- Full duplex: can transmit and receive simultaneously
- On-chip baud-rate generator
- Smart card interface functions added (SCIO only)

A/D converter

- Resolution: 10 bits
- Eight channels, with selection of single or scan mode
- Variable analog conversion voltage range
- Sample-and-hold function
- A/D conversion can be externally triggered

| Feature | Description |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| D/A converter | - Resolution: 8 bits <br> - Two channels <br> - D/A outputs can be sustained in software standby mode |  |  |  |  |
| I/O ports | - 70 input/output pins <br> - 9 input-only pins |  |  |  |  |
| Operating modes | - Seven MCU operating modes |  |  |  |  |
|  | Mode 1 | 1 Mbyte | $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$ | 8 bits | 16 bits |
|  | Mode 2 | 1 Mbyte | $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$ | 16 bits | 16 bits |
|  | Mode 3 | 16 Mbytes | $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$ | 8 bits | 16 bits |
|  | Mode 4 | 16 Mbytes | $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$ | 16 bits | 16 bits |
|  | Mode 5 | 1 Mbyte | $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$ | 8 bits | 16 bits |
|  | Mode 6 | 16 Mbytes | $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$ | 8 bits | 16 bits |
|  | Mode 7 | 1 Mbyte | - | - | - |

- On-chip ROM is disabled in modes 1 to 4

Power-down state

- Sleep mode
- Software standby mode
- Hardware standby mode
- Module standby function
- Programmable system clock frequency division

Other features

- On-chip clock pulse generator

| Product lineup | Product Type |  | Product Code | Package(Hitachi Package Code) |
| :--- | :--- | :--- | :--- | :--- |
|  | H8/3052F-ZTAT | 5 FV version | HD64F3052F | 100-pin QFP (FP-100B) |
|  |  |  | HD64F3052TE | 100-pin TQFP (TFP-100B) |
|  | H8/3052F-ZTAT | 5 F version | HD64F3052BF | 100-pin QFP (FP-100B) |
|  | B mask version |  | HD64F3052BTE | 100-pin TQFP (TFP-100B) |
|  |  | 3V version | HD64F3052BVF | 100-pin QFP (FP-100B) |
|  |  |  | HD64F3052BVTE | 100-pin TQFP (TFP-100B) |

### 1.2 Block Diagram

Figure 1.1 shows an internal block diagram.


Figure 1.1 Block Diagram

## 1.3 Pin Description

### 1.3.1 Pin Arrangement

Figure 1.2 shows the pin arrangement of the $\mathrm{H} 8 / 3052 \mathrm{~F}$.


Note: * This pin functions as the $\mathrm{V}_{\text {CL }}$ pin during 5 V operation and as the $\mathrm{V}_{\mathrm{CC}}$ pin during 3 V operation. An external capacitor must be connected to the $\mathrm{V}_{\mathrm{CL}}$ pin.

Figure 1.2 Pin Arrangement (FP-100B or TFP-100B, Top View)

### 1.3.2 Pin Assignments in Each Mode

Table 1.2 lists the pin assignments in each mode.
Table 1.2 Pin Assignments in Each Mode (FP-100B or TFP-100B)
Pin Name

| Pin No. | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{V}_{\mathrm{cL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{CL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{cL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{CL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{cL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{cL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ | $\mathrm{V}_{\mathrm{cL}}\left(\mathrm{V}_{\mathrm{cc}}\right)^{* 1}$ |
| 2 | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ |
| 3 | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{9} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{\mathrm{g}} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{\mathrm{g}} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{\mathrm{g}} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{9} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{9} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{1} / \mathrm{TP}_{\mathrm{g}} / \\ & \mathrm{TIOCB}_{3} \end{aligned}$ |
| 4 | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4} \end{aligned}$ |
| 5 | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} / \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4} / \end{aligned}$ |
| 6 | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \text { TOCXA }_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4} \end{aligned}$ |
| 7 | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{5} / \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4} \end{aligned}$ |
| 8 | $\begin{aligned} & \mathrm{PB}_{6} / \mathrm{TP}_{14} / \\ & \mathrm{DREQ}_{0} / \\ & \mathrm{CS}_{7} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{6} / \mathrm{TP}_{14} /{ }^{\prime}{ }^{\mathrm{DREQ}_{0} /} \\ & \mathrm{CS}_{7} \end{aligned}$ | $\begin{aligned} & \frac{\mathrm{PB}_{6} / \mathrm{TP}_{14} /}{\mathrm{DREQ}_{0} /} \\ & \mathrm{CS}_{7} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{6} / \mathrm{TP}_{14} /{ }^{\mathrm{DREQ}_{0} /} \\ & \mathrm{CRE}_{7} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{6} / \mathrm{TP}_{14} / \\ & \mathrm{DREQ}_{0} / \\ & \mathrm{CS}_{7} \end{aligned}$ | $\begin{aligned} & {\frac{\mathrm{PB}_{6} / \mathrm{TP}_{14} /}{}}_{\mathrm{DREQ}_{0} /}^{\overline{\mathrm{CS}}_{7}} \end{aligned}$ | ${\frac{\mathrm{PB}_{6} / \mathrm{TP}^{14}}{}{ }^{\mathrm{DREQ}}}_{0}$ |
| 9 | $\frac{\mathrm{PB}_{7} / \mathrm{TP}_{15} /{ }^{15} /}{\mathrm{DREQ}_{1} /}$ | $\frac{\mathrm{PB}_{i} / \mathrm{TP}_{15} /}{\mathrm{DREQ}_{1} /} \text { ADTRG}$ | $\frac{\mathrm{PB}_{7} / \mathrm{TP}_{15} /{ }^{5}{ }^{\mathrm{DREQ}}{ }_{1}}{\mathrm{ADTRG}}$ | $\frac{\mathrm{PB}_{\nabla} / \mathrm{TP}_{15} /{ }^{\mathrm{DREQ}} /}{\text { ARTRG }}$ |  | $\frac{\mathrm{PB}_{7} / \mathrm{TP}_{15} /}{\mathrm{DREQ}_{1} /} \text { (ADTRG}$ | $\frac{\mathrm{PB}_{7} / \mathrm{TP}_{15} /}{\mathrm{DREQ}_{1} /}$ |
| 10 | FEW | FWE | FWE | FWE | FWE | FWE | FWE |
| 11 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\mathrm{ss}}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 12 | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{P} 9_{0} / \mathrm{TxD}{ }_{0}$ | $\mathrm{Pg}_{0} / \mathrm{TxD}{ }_{0}$ |
| 13 | $\mathrm{Pg}_{1} / \mathrm{TxD}_{1}$ | $\mathrm{Pg}_{1} / \mathrm{TxD}_{1}$ | $\mathrm{Pg}_{1} / \mathrm{TxD}{ }_{1}$ | $\mathrm{Pg}_{1} / \mathrm{TxD}_{1}$ | $\mathrm{Pg}_{1} / \mathrm{TxD}{ }_{1}$ | $\mathrm{P} 9_{1} / \mathrm{TxD}_{1}$ | $\mathrm{Pg}_{1} / \mathrm{TxD}{ }_{1}$ |
| 14 | $\mathrm{Pg}_{2} / \mathrm{RxD} 0$ | $\mathrm{Pg}_{2} / \mathrm{RxD} 0$ | $\mathrm{Pg}_{2} / \mathrm{RxD} \mathrm{D}_{0}$ | $\mathrm{Pg}_{2} / \mathrm{RxD} \mathrm{D}_{0}$ | $\mathrm{Pg}_{2} / \mathrm{RxD} \mathrm{D}_{0}$ | $\mathrm{Pg}_{2} / \mathrm{RxD} 0$ | $\mathrm{Pg}_{2} / \mathrm{RxD} 0$ |
| 15 | $\mathrm{Pg}_{3} / \mathrm{RxD}{ }_{1}$ | $\mathrm{Pg}_{3} / \mathrm{RxD} \mathrm{D}_{1}$ | $\mathrm{Pg}_{3} / \mathrm{RxD}{ }_{1}$ | $\mathrm{Pg}_{3} / \mathrm{RxD} \mathrm{i}_{1}$ | $\mathrm{P9}_{3} / \mathrm{RxD}{ }_{1}$ | $\mathrm{Pg}_{3} / \mathrm{RxD} \mathrm{l}_{1}$ | $\mathrm{Pg}_{3} / \mathrm{RxD}$ |
| 16 | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{IQQ}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{TRQ}_{4} \end{aligned}$ | $\frac{\mathrm{Pg}_{4} / \mathrm{SCK}_{0} /}{\mathrm{RQ}_{4}}$ | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{TQQ}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{IQQ}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{PQ}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{Pg}_{4} / \mathrm{SCK}_{0} / \\ & \mathrm{IQQ}_{4} \end{aligned}$ |
| 17 | $\begin{aligned} & \mathrm{Pg}_{5} / \mathrm{SCK}_{1} / \\ & \mathrm{IRQ}_{5} \end{aligned}$ | $\frac{\mathrm{Pg}_{5} / \mathrm{SCK}_{1} /}{\mathrm{IRQ}_{5}}$ | $\frac{\mathrm{Pg}_{5} / \mathrm{SCK}_{1} /}{\mathrm{IRQ}_{5}}$ | $\begin{aligned} & \mathrm{Pg}_{5} / \mathrm{SCK}_{1} / \\ & \mathrm{IRQ}_{5} \end{aligned}$ | $\frac{\mathrm{Pg}_{5} / \mathrm{SCK}_{1} /}{\mathrm{IRQ}_{5}}$ | $\frac{\mathrm{Pg}_{5} / \mathrm{SCK}_{1} /}{\mathrm{IRQ}_{5}}$ | $\frac{\mathrm{Pg}_{5} / \mathrm{SCK}_{1} /}{\mathrm{RQ}_{5}}$ |

Rev. 2.0, 03/01, page 8 of 822

| Pin No. | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 18 | $\mathrm{P} 4 \mathrm{f}_{0} \mathrm{D}^{* *}$ | $\mathrm{P} 4 \mathrm{f}^{\prime} \mathrm{D}^{* 3}$ | $\mathrm{P} 4 \mathrm{f}^{\prime} \mathrm{D}_{0}{ }^{* 2}$ | $\mathrm{P} 4 \mathrm{f}^{\prime} \mathrm{D}_{0}{ }^{* 3}$ | $\mathrm{P} 4 \mathrm{f}^{\prime} \mathrm{D}_{0}{ }^{* 2}$ | $\mathrm{P} 4 \mathrm{f}^{\prime} \mathrm{D}_{0}{ }^{* 2}$ | P4 |
| 19 | $\mathrm{P} 4, / \mathrm{D}_{1}{ }^{* 2}$ | $\mathrm{P} 4, / \mathrm{D}_{1}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{D}_{1}{ }^{* 2}$ | $\mathrm{P} 4 / \mathrm{/D}{ }^{*}{ }^{* 3}$ | $\mathrm{P} 4, \mathrm{D}_{1}{ }^{*}{ }^{2}$ | $\mathrm{P} 4, \mathrm{D}_{1}{ }^{*}{ }^{2}$ | P4 ${ }_{1}$ |
| 20 | $\mathrm{P} 4_{2} / \mathrm{D}_{2}{ }^{* 2}$ | $\mathrm{P} 4 / \mathrm{D}_{2}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{I}_{2}{ }^{* 2}$ | $\mathrm{P} 4 / 2 / \mathrm{D}^{* 3}$ | $\mathrm{P} 4 / \mathrm{ID}_{2}{ }^{2}$ | $\mathrm{P} 4 / \mathrm{D}_{2}{ }^{* 2}$ | $\mathrm{P} 4{ }_{2}$ |
| 21 | $\mathrm{P} 4 / 3 / \mathrm{D}_{3}{ }^{2}$ | $\mathrm{P} 4 / /_{3}{ }^{* 3}$ | $\mathrm{P} 4 / /_{3}{ }^{* 2}$ | $\mathrm{P} 4{ }_{3} / \mathrm{D}_{3}{ }^{3}$ | $\mathrm{P}_{3} / \mathrm{D}_{3}{ }^{2}$ | $\mathrm{P}_{3} / \mathrm{D}_{3}{ }^{*}{ }^{2}$ | P4 ${ }_{3}$ |
| 22 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 23 | $\mathrm{P} 44 / \mathrm{D}_{4}{ }^{* 2}$ | $\mathrm{P} 44 / \mathrm{D}_{4}{ }^{* 3}$ | $\mathrm{P} 44 / \mathrm{D}_{4}{ }^{* 2}$ | $\mathrm{P} 44 / \mathrm{D}_{4}{ }^{* 3}$ | $\mathrm{P}_{4} / \mathrm{D}_{4}{ }^{*}{ }^{2}$ | $\mathrm{P}_{4} / \mathrm{D}_{4}{ }^{* 2}$ | P 44 |
| 24 | P4/ $/ \mathrm{D}_{5}{ }^{* 2}$ | $\mathrm{P} 4 / \mathrm{D}_{5}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{D}_{5}{ }^{* 2}$ | $\mathrm{P} 4 /_{5} / \mathrm{D}_{5}{ }^{3}$ | $\mathrm{P} 4 \mathrm{~s}^{\prime} \mathrm{D}_{5}{ }^{* 2}$ | $\mathrm{P} 4 \mathrm{~s}^{\prime} \mathrm{D}_{5}{ }^{* 2}$ | P45 |
| 25 | $\mathrm{P} 4 / \mathrm{ID}_{6}{ }^{* 2}$ | $\mathrm{P} 4 / \mathrm{D}_{6}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{D}_{6}{ }^{* 2}$ | $\mathrm{P} 46 / \mathrm{D}_{6}{ }^{* 3}$ | $\mathrm{P} 4 \mathrm{f}_{6} \mathrm{D}_{6}{ }^{2}$ | $\mathrm{P} 46 / \mathrm{D}_{6}{ }^{2}$ | P4 ${ }_{6}$ |
| 26 | $\mathrm{P} 4 / \mathrm{/} \mathrm{D}_{7}{ }^{*}$ | $\mathrm{P} 4 / \mathrm{/} \mathrm{D}_{7}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{/D} \mathrm{D}^{* 2}$ | $\mathrm{P} 4 / \mathrm{D}_{7}{ }^{* 3}$ | $\mathrm{P} 4 / \mathrm{/} \mathrm{D}_{7}{ }^{2}$ | $\mathrm{P} 4 / \mathrm{D}_{7}{ }^{*}{ }^{2}$ | P 47 |
| 27 | $\mathrm{D}_{8}$ | $\mathrm{D}_{8}$ | $\mathrm{D}_{8}$ | $\mathrm{D}_{8}$ | $\mathrm{D}_{8}$ | $\mathrm{D}_{8}$ | P3 ${ }_{0}$ |
| 28 | $\mathrm{D}_{9}$ | $\mathrm{D}_{9}$ | $\mathrm{D}_{9}$ | $\mathrm{D}_{9}$ | $\mathrm{D}_{9}$ | $\mathrm{D}_{9}$ | P3 ${ }_{1}$ |
| 29 | $\mathrm{D}_{10}$ | $\mathrm{D}_{10}$ | $\mathrm{D}_{10}$ | $\mathrm{D}_{10}$ | $\mathrm{D}_{10}$ | $\mathrm{D}_{10}$ | $\mathrm{P}_{2}$ |
| 30 | $\mathrm{D}_{11}$ | $\mathrm{D}_{11}$ | $\mathrm{D}_{11}$ | $\mathrm{D}_{11}$ | $\mathrm{D}_{11}$ | $\mathrm{D}_{11}$ | $\mathrm{P}_{3}$ |
| 31 | $\mathrm{D}_{12}$ | $\mathrm{D}_{12}$ | $\mathrm{D}_{12}$ | $\mathrm{D}_{12}$ | $\mathrm{D}_{12}$ | $\mathrm{D}_{12}$ | $\mathrm{P}_{4}$ |
| 32 | $\mathrm{D}_{13}$ | $\mathrm{D}_{13}$ | $\mathrm{D}_{13}$ | $\mathrm{D}_{13}$ | $\mathrm{D}_{13}$ | $\mathrm{D}_{13}$ | $\mathrm{P}_{5}$ |
| 33 | $\mathrm{D}_{14}$ | $\mathrm{D}_{14}$ | $\mathrm{D}_{14}$ | $\mathrm{D}_{14}$ | $\mathrm{D}_{14}$ | $\mathrm{D}_{14}$ | $\mathrm{P}_{6}$ |
| 34 | $\mathrm{D}_{15}$ | $\mathrm{D}_{15}$ | $\mathrm{D}_{15}$ | $\mathrm{D}_{15}$ | $\mathrm{D}_{15}$ | $\mathrm{D}_{15}$ | $\mathrm{P}_{7}$ |
| 35 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\text {c }}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\text {c }}$ | $\mathrm{V}_{\text {c }}$ | $\mathrm{V}_{\text {c }}$ |
| 36 | $\mathrm{A}_{0}$ | $\mathrm{A}_{0}$ | $\mathrm{A}_{0}$ | $\mathrm{A}_{0}$ | $\mathrm{P} 1_{0} / \mathrm{A}_{0}$ | $\mathrm{P} 1 \mathrm{c}_{0} / \mathrm{A}_{0}$ | P1。 |
| 37 | $\mathrm{A}_{1}$ | $\mathrm{A}_{1}$ | $\mathrm{A}_{1}$ | $\mathrm{A}_{1}$ | $\mathrm{P} 1_{1} / \mathrm{A}_{1}$ | $\mathrm{P} 1 . / \mathrm{A}_{1}$ | $\mathrm{P} 1_{1}$ |
| 38 | $\mathrm{A}_{2}$ | $\mathrm{A}_{2}$ | $\mathrm{A}_{2}$ | $\mathrm{A}_{2}$ | $\mathrm{P} 1_{2} / \mathrm{A}_{2}$ | $\mathrm{P} 1_{2} / \mathrm{A}_{2}$ | $\mathrm{P} 1_{2}$ |
| 39 | $\mathrm{A}_{3}$ | $\mathrm{A}_{3}$ | $\mathrm{A}_{3}$ | $\mathrm{A}_{3}$ | $\mathrm{P}_{3} / \mathrm{A}_{3}$ | $\mathrm{P} 1_{3} / \mathrm{A}_{3}$ | $\mathrm{P} 1_{3}$ |
| 40 | $\mathrm{A}_{4}$ | $\mathrm{A}_{4}$ | $\mathrm{A}_{4}$ | $\mathrm{A}_{4}$ | $\mathrm{P1}_{4} / \mathrm{A}_{4}$ | $\mathrm{P} 144 / \mathrm{A}_{4}$ | $\mathrm{P} 14^{4}$ |
| 41 | $\mathrm{A}_{5}$ | $\mathrm{A}_{5}$ | $\mathrm{A}_{5}$ | $\mathrm{A}_{5}$ | $\mathrm{P} 1_{5} / \mathrm{A}_{5}$ | $\mathrm{P} 1 \mathrm{~s}_{5} / \mathrm{A}_{5}$ | $\mathrm{P} 1_{5}$ |
| 42 | $\mathrm{A}_{6}$ | $\mathrm{A}_{6}$ | $\mathrm{A}_{6}$ | $\mathrm{A}_{6}$ | $\mathrm{P} 1_{6} / \mathrm{A}_{6}$ | $\mathrm{P} 1 \mathrm{C}_{6} / \mathrm{A}_{6}$ | $\mathrm{P} 1_{6}$ |
| 43 | $\mathrm{A}_{7}$ | $\mathrm{A}_{7}$ | $\mathrm{A}_{7}$ | $\mathrm{A}_{7}$ | $\mathrm{P} 17^{7} / \mathrm{A}_{7}$ | $\mathrm{P}_{7} / \mathrm{A}_{7}$ | $\mathrm{Pl}_{7}$ |
| 44 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 45 | $\mathrm{A}_{8}$ | $\mathrm{A}_{8}$ | $\mathrm{A}_{8}$ | $\mathrm{A}_{8}$ | $\mathrm{P}_{2} / \mathrm{A}_{8}$ | $\mathrm{P} 2_{0} / \mathrm{A}_{8}$ | P2 ${ }_{0}$ |
| 46 | $\mathrm{A}_{9}$ | $\mathrm{A}_{9}$ | $\mathrm{A}_{9}$ | $\mathrm{A}_{9}$ | P2, $/ \mathrm{A}_{9}$ | $\mathrm{P} 2, / \mathrm{A}_{9}$ | P2 ${ }_{1}$ |

## Pin Name

| Pin No. | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 47 | $\mathrm{A}_{10}$ | $\mathrm{A}_{10}$ | $\mathrm{A}_{10}$ | $\mathrm{A}_{10}$ | $\mathrm{P}_{2} / \mathrm{A}_{10}$ | $\mathrm{P}_{2} / \mathrm{A}_{10}$ | $\mathrm{P}_{2}$ |
| 48 | $\mathrm{A}_{11}$ | $\mathrm{A}_{11}$ | $\mathrm{A}_{11}$ | $\mathrm{A}_{11}$ | $\mathrm{P}_{3} / \mathrm{A}_{11}$ | $\mathrm{P}_{3} / \mathrm{A}_{11}$ | $\mathrm{P}_{3}$ |
| 49 | $\mathrm{A}_{12}$ | $\mathrm{A}_{12}$ | $\mathrm{A}_{12}$ | $\mathrm{A}_{12}$ | $\mathrm{P}_{4} / \mathrm{A}_{12}$ | $\mathrm{P}_{4} / \mathrm{A}_{12}$ | P2 ${ }_{4}$ |
| 50 | $\mathrm{A}_{13}$ | $\mathrm{A}_{13}$ | $\mathrm{A}_{13}$ | $\mathrm{A}_{13}$ | $\mathrm{P}_{5} / \mathrm{A}_{13}$ | $\mathrm{P}_{5} / \mathrm{A}_{13}$ | $\mathrm{P} 2_{5}$ |
| 51 | $\mathrm{A}_{14}$ | $\mathrm{A}_{14}$ | $\mathrm{A}_{14}$ | $\mathrm{A}_{14}$ | $\mathrm{P}_{2} / \mathrm{A}_{14}$ | $\mathrm{P}_{6} / \mathrm{A}_{14}$ | $\mathrm{P}_{6}{ }_{6}$ |
| 52 | $\mathrm{A}_{15}$ | $\mathrm{A}_{15}$ | $\mathrm{A}_{15}$ | $\mathrm{A}_{15}$ | $\mathrm{P} 2_{7} / \mathrm{A}_{15}$ | $\mathrm{P} 2_{7} / \mathrm{A}_{15}$ | $\mathrm{P} 2_{7}$ |
| 53 | $\mathrm{A}_{16}$ | $\mathrm{A}_{16}$ | $\mathrm{A}_{16}$ | $\mathrm{A}_{16}$ | $\mathrm{P} 5 / \mathrm{A}_{16}$ | P5 $/ \mathrm{A}_{16}$ | P5 |
| 54 | $\mathrm{A}_{17}$ | $\mathrm{A}_{17}$ | $\mathrm{A}_{17}$ | $\mathrm{A}_{17}$ | P5 $\mathrm{l}_{1} / \mathrm{A}_{17}$ | P5, $/ \mathrm{A}_{17}$ | P5 ${ }_{1}$ |
| 55 | $\mathrm{A}_{18}$ | $\mathrm{A}_{18}$ | $\mathrm{A}_{18}$ | $\mathrm{A}_{18}$ | $\mathrm{P} 5 / \mathrm{A}_{18}$ | P5 $\mathrm{S}_{2}$ ( ${ }_{18}$ | $\mathrm{P5}_{2}$ |
| 56 | $\mathrm{A}_{19}$ | $\mathrm{A}_{19}$ | $\mathrm{A}_{19}$ | $\mathrm{A}_{19}$ | $\mathrm{P5}_{3} / \mathrm{A}_{19}$ | $\mathrm{P}_{3} / \mathrm{A}_{19}$ | $\mathrm{P5}_{3}$ |
| 57 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 58 | P6, $\overline{\text { WAIT }}$ | P60 $/ \overline{\text { WAIT }}$ | P6/ /WAIT | $\mathrm{P} 6_{0} \overline{\text { WAIT }}$ | P6, $\overline{\text { WAIT }}$ | P60 $\overline{\text { WAIT }}$ | P6 ${ }_{0}$ |
| 59 | P6, $\overline{\text { BREQ }}$ | P6, $\overline{\text { BREQ }}$ | P6,//RREQ | P6, $\overline{\text { BREQ }}$ | P6, $\overline{\text { BREQ }}$ | P6, $\overline{\text { BREQ }}$ | P61 |
| 60 | P6/ $/ \overline{\text { BACK }}$ | P6/ $/ \overline{\text { BACK }}$ | P6, ${ }_{2}$ BACK | $\mathrm{P} 6_{2} / \overline{\text { BACK }}$ | P6, $\overline{\text { BACK }}$ | P6, $\overline{\text { BACK }}$ | $\mathrm{P}_{2}$ |
| 61 | $\phi$ | $\phi$ | $\phi$ | $\phi$ | $\phi$ | $\phi$ | $\phi$ |
| 62 | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ | $\overline{\text { STBY }}$ |
| 63 | RES | RES | RES | $\overline{\text { RES }}$ | RES | RES | RES |
| 64 | NMI | NMI | NMI | NMI | NMI | NMI | NMI |
| 65 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 66 | EXTAL | EXTAL | EXTAL | EXTAL | EXTAL | EXTAL | EXTAL |
| 67 | XTAL | XTAL | XTAL | XTAL | XTAL | XTAL | XTAL |
| 68 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\text {cc }}$ |
| 69 | $\overline{\text { AS }}$ | $\overline{\text { AS }}$ | $\overline{\text { AS }}$ | $\overline{\text { AS }}$ | $\overline{\text { AS }}$ | $\overline{\text { AS }}$ | $\mathrm{P6}_{3}$ |
| 70 | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{RD}}$ | $\mathrm{Pb}_{4}$ |
| 71 | HWR | HWR | HWR | HWR | HWR | HWR | $\mathrm{P6}_{5}$ |
| 72 | $\overline{\text { LWR }}$ | $\overline{\text { LWR }}$ | LWR | $\overline{\text { LWR }}$ | $\overline{\text { LWR }}$ | $\overline{\text { LWR }}$ | P66 |
| 73 | MD. | MD。 | MD ${ }_{0}$ | MD ${ }_{0}$ | MD ${ }_{0}$ | MD ${ }_{0}$ | MD |
| 74 | MD ${ }_{1}$ | MD ${ }_{1}$ | MD ${ }_{1}$ | MD ${ }_{1}$ | MD ${ }_{1}$ | MD ${ }_{1}$ | MD ${ }_{1}$ |
| 75 | MD | MD2 | $\mathrm{MD}_{2}$ | $\mathrm{MD}_{2}$ | $\mathrm{MD}_{2}$ | $\mathrm{MD}_{2}$ | $\mathrm{MD}_{2}$ |

Rev. 2.0, 03/01, page 10 of 822

Pin Name

| Pin No. | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 76 | $\mathrm{AV}_{\text {cc }}$ | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{AV}_{\mathrm{cc}}$ |
| 77 | $V_{\text {REF }}$ | $V_{\text {REF }}$ | $V_{\text {REF }}$ | $V_{\text {REF }}$ | $V_{\text {REF }}$ | $V_{\text {REF }}$ | $V_{\text {REF }}$ |
| 78 | P7 $/$ AN ${ }_{0}$ | $\mathrm{P} 7 \mathrm{l}^{\prime} \mathrm{AN}{ }_{0}$ | $\mathrm{P} 7 \mathrm{l}^{\prime} \mathrm{AN}{ }_{0}$ | $\mathrm{P} 7_{0} / \mathrm{AN}_{0}$ | $\mathrm{P} 7 \mathrm{~J}^{\prime} \mathrm{AN}{ }_{0}$ | $\mathrm{P} 7 \mathrm{l}^{\prime} \mathrm{AN}{ }_{0}$ | P7 $/ \mathrm{AN}_{0}$ |
| 79 | P7//AN ${ }_{1}$ | P7, $/ \mathrm{AN}_{1}$ | P7/ $/$ AN ${ }_{1}$ | P7/ $/$ AN ${ }_{1}$ | P7//AN ${ }_{1}$ | P7 ${ }_{1} / \mathrm{AN}_{1}$ | P7/ $/$ AN ${ }_{1}$ |
| 80 | $\mathrm{P}_{2} / \mathrm{AN}_{2}$ | $\mathrm{P} 7_{2} / \mathrm{AN}_{2}$ | $\mathrm{P} 7_{2} / \mathrm{AN}_{2}$ | $\mathrm{P} 7{ }_{2} / \mathrm{AN}_{2}$ | $\mathrm{P} 7_{2} / \mathrm{AN}_{2}$ | $\mathrm{P} 7_{2} / \mathrm{AN}_{2}$ | $\mathrm{P}_{2} / \mathrm{AN}_{2}$ |
| 81 | $\mathrm{P}_{3} / \mathrm{AN}_{3}$ | $\mathrm{P} 7{ }_{3} / \mathrm{AN}_{3}$ | $\mathrm{P}_{3} / \mathrm{AN}_{3}$ | $\mathrm{P}_{3} / \mathrm{AN}_{3}$ | $\mathrm{P7}_{3} / \mathrm{AN}_{3}$ | $\mathrm{P} 7{ }_{3} / \mathrm{AN}_{3}$ | $\mathrm{P}_{3} / \mathrm{AN}_{3}$ |
| 82 | $\mathrm{P7}_{4} / \mathrm{AN}_{4}$ | $\mathrm{P} 7_{4} / \mathrm{AN}_{4}$ | $\mathrm{P} 7_{4} / \mathrm{AN}_{4}$ | $\mathrm{P} 7_{4} / \mathrm{AN}_{4}$ | $\mathrm{P} 7_{4} / \mathrm{AN}_{4}$ | $\mathrm{P} 7_{4} / \mathrm{AN}_{4}$ | $\mathrm{P7}_{4} / \mathrm{AN}_{4}$ |
| 83 | $\mathrm{P}_{5} / \mathrm{AN}_{5}$ | $\mathrm{P} 7{ }_{5} / \mathrm{AN}_{5}$ | $\mathrm{P} 7_{5} / \mathrm{AN}_{5}$ | $\mathrm{P} 7{ }_{5} / \mathrm{AN}_{5}$ | $\mathrm{P} 7_{5} / \mathrm{AN}_{5}$ | $\mathrm{P} 7 / \mathrm{AN}_{5}$ | $\mathrm{P}_{5} / \mathrm{AN}_{5}$ |
| 84 | $\begin{aligned} & \mathrm{P}_{6} / \mathrm{AN}_{6} / \\ & \mathrm{DA}_{0} \end{aligned}$ | $\begin{aligned} & {\mathrm{P} 7{ }_{6} / \mathrm{AN}_{6} /}^{\mathrm{DA}_{0}} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{6} / \mathrm{AN}_{\mathrm{s}} / \\ & \mathrm{DA}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{6} / \mathrm{AN}_{6} / \\ & \mathrm{DA}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{6} / \mathrm{AN}_{6} / \\ & \mathrm{DA}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{7} / \mathrm{AN}_{6} / \\ & \mathrm{DA}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{6} / \mathrm{AN}_{6} / \\ & \mathrm{DA}_{0} \end{aligned}$ |
| 85 | $\begin{aligned} & \mathrm{P}_{7} / \mathrm{AN}_{7} / \\ & \mathrm{DA}_{1} \end{aligned}$ | $\begin{aligned} & {\mathrm{P} 7_{/} / \mathrm{AN}_{7} /}^{\mathrm{DA}_{1}} \end{aligned}$ | $\begin{aligned} & {\mathrm{P} 7_{/} / \mathrm{AN}_{7} /}^{\mathrm{DA}_{1}} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{7} / \mathrm{AN}_{7} / \\ & \mathrm{DA}_{1} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{7} / / \mathrm{AN}_{7} / \\ & \mathrm{DA}_{1} \end{aligned}$ | $\begin{aligned} & {\mathrm{P} 7_{/} / \mathrm{AN}_{7} /}^{\mathrm{DA}_{1}} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{7} / \mathrm{AN}_{7} / \\ & \mathrm{DA}_{1} \end{aligned}$ |
| 86 | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{AV}_{\text {ss }}$ |
| 87 | $\begin{aligned} & \mathrm{P} 8 / \sqrt{\prime} \overline{\mathrm{RFSH}} /_{\mathrm{RQ}_{0}} \end{aligned}$ | $\begin{aligned} & {\mathrm{P} 8_{0} \overline{\mathrm{RFSH}} /}_{\overline{\mathrm{IQQ}}_{0}} \end{aligned}$ | $\begin{aligned} & \mathrm{P} 8 / 0^{\overline{\mathrm{RFSH}} /} \\ & \mathrm{IRQ}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{0} \overline{\mathrm{RFSH}} / \\ & \mathrm{IRQ}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P} 8 / \sigma^{\overline{\mathrm{RFSH}} /} \\ & \mathrm{IRQ}_{0} \end{aligned}$ | $\begin{aligned} & \mathrm{P} 8 / 0^{\overline{\mathrm{RFSH}} /} \\ & \mathrm{IRQ}_{0} \end{aligned}$ | $\mathrm{P} 8_{0} /{\overline{\mathrm{RQ}}{ }_{0}}$ |
| 88 | $\frac{\mathrm{P}_{8} / \overline{\mathrm{CS}_{3}} /}{\mathrm{IRQ}_{1}}$ | $\frac{\mathrm{PB}_{8} / \overline{\mathrm{CS}_{3}} /}{\mathrm{IRQ}_{1}}$ | $\begin{aligned} & \mathrm{PB}_{2} /{\overline{\mathrm{CS}_{3}} /}_{\mathrm{IRQ}_{1}} \\ & \hline \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{/} / \overline{\mathrm{CS}}_{3} / \\ & \mathrm{IRQ}_{1} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{8} / \overline{\mathrm{CS}_{3}} / \\ & \mathrm{IRQ}_{1} \end{aligned}$ | $\frac{\mathrm{PB}_{1} / \overline{\mathrm{CS}}_{3} /}{\mathrm{IRQ}_{1}}$ | $\mathrm{P} 8_{1} / \overline{\mathrm{IRQ}}_{1}$ |
| 89 | $\begin{aligned} & \mathrm{PB}_{2} / \overline{\mathrm{CS}_{2}} / \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PQ}_{2} / \overline{\overline{\mathrm{CS}}_{2} /} \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \overline{\mathrm{CS}_{2}} / \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \overline{\mathrm{CS}}_{2} / \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{2} / \overline{\mathrm{CS}}_{2} / \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PQ}_{2} / \overline{\overline{\mathrm{CS}}_{2} /} \\ & \mathrm{IRQ}_{2} \end{aligned}$ | $\mathrm{P} 8_{2} / \overline{\mathrm{RQ}}_{2}$ |
| 90 | $\begin{aligned} & \mathrm{PB}_{3} / \overline{\mathrm{CS}} / \mathrm{I} \\ & \overline{\mathrm{RQ}}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \overline{\mathrm{CS}_{1} /} \\ & \overline{\mathrm{TRQ}}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \overline{\mathrm{CS}_{1} / I} \\ & \mathrm{IRQ}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{P}_{3} \overline{\overline{\mathrm{CS}}_{1} /} \\ & {\overline{\mathrm{R} Q_{3}}}^{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} \overline{\overline{\mathrm{CS}}_{1} /} \overline{\mathrm{IRQ}}_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{PB}_{3} / \overline{\mathrm{CS}_{1} /} \\ & \overline{\mathrm{IRQ}}_{3} \end{aligned}$ | $\mathrm{P}_{3} / \overline{\mathrm{IRQ}}_{3}$ |
| 91 | $\mathrm{P}_{4} / \overline{\mathrm{CS}} 0$ | $\mathrm{P8}_{4} / \overline{\mathrm{CS}}$ | $\mathrm{P8}_{4} / \overline{\mathrm{CS}}$ | $\mathrm{P8}_{4} / \overline{\mathrm{CS}}$ | $\mathrm{P8}_{4} / \overline{\mathrm{CS}}$ | $\mathrm{P8}_{4} / \overline{\mathrm{CS}}$ | $\mathrm{P8}_{4}$ |
| 92 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 93 | $\begin{aligned} & \mathrm{PA}_{d} / \mathrm{TP}_{d}{ }^{\prime} \\ & \hline \mathrm{TEND}_{d} / \\ & \text { TCLKA } \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{d} / \mathrm{TP}_{d} / \\ & \mathrm{TEND}_{d} / \\ & \mathrm{TCLKA}^{2} \end{aligned}$ |  | $\begin{aligned} & \mathrm{PA}_{0} / \mathrm{TP}_{d} / \\ & \mathrm{TEND}_{o} / \\ & \mathrm{TCLKA}^{2} \end{aligned}$ |  | $\begin{aligned} & \mathrm{PA}_{0} / \mathrm{TP}_{d} / \\ & \mathrm{TEND}_{0} / \\ & \mathrm{TCLKA}^{2} \end{aligned}$ | $\begin{aligned} & \hline \mathrm{PA}_{d} / \mathrm{TP}_{d} / \\ & \mathrm{TEND}_{d} / \\ & \text { TCLKA } \end{aligned}$ |
| 94 | $\begin{aligned} & \hline \mathrm{PA}_{1} / \mathrm{TP}_{1} / \\ & \mathrm{TEND}_{1} / \\ & \mathrm{TCLKB} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{1} / \mathrm{TP} / P_{1} / \\ & \mathrm{TEND}_{1} / \\ & \mathrm{TCLLKB}^{2} \end{aligned}$ | $\begin{aligned} & \hline \mathrm{PA}_{1} / \mathrm{TP} / P_{1} / \\ & \mathrm{TEND}_{1} / \\ & \mathrm{TCLKB}^{2} \end{aligned}$ | $\begin{aligned} & \hline \mathrm{PA}_{1} / \mathrm{TP}_{1} / / \\ & \mathrm{TEND}_{1} / \\ & \mathrm{TCLKB}^{2} \end{aligned}$ | $\begin{aligned} & \hline \mathrm{PA}_{1} / \mathrm{TP}_{1} / \\ & \hline \mathrm{TEND}_{1} / \\ & \mathrm{TCLKB} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{1} / \mathrm{TP}_{1} / \\ & \mathrm{TEND}_{1} / \\ & \text { TCLKB } \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{1} / \mathrm{TP} / 1 \\ & \mathrm{TEND}_{1} / \\ & \mathrm{TCLKB}_{1} \end{aligned}$ |
| 95 | $\mathrm{PA}_{2} / \mathrm{TP}_{2} /$ TIOCA ${ }_{0}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2}$ TIOCA $_{0}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2}$ TIOCA ${ }_{0}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2} /$ TIOCA ${ }^{\prime}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2} /$ TIOCA ${ }_{0}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2}$ TIOCA ${ }_{0}$ TCLKC | $\mathrm{PA}_{2} / \mathrm{TP}_{2}$ TIOCA $_{0} /$ TCLKC |

## Pin Name

| Pin No. | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 96 | $\mathrm{PA}_{3} / \mathrm{TP}_{3} /$ TIOCB ${ }^{1}$ TCLKD | $\mathrm{PA}_{3} / \mathrm{TP}_{3} /$ TIOCB TCLKD | $\mathrm{PA}_{3} / \mathrm{TP}_{3} /$ TIOCB TCLKD | $\begin{aligned} & \mathrm{PA}_{3} / \mathrm{TP}_{3} / \\ & \text { TIOCB }_{j} \\ & \text { TCLKD } \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{3} / \mathrm{TP}_{3} / \\ & \text { TIOCB }_{d} \\ & \text { TCLKD } \end{aligned}$ | $\mathrm{PA}_{3} / \mathrm{TP}_{3}$ TIOCB $_{0} /$ TCLKD | $\mathrm{PA}_{3} / \mathrm{TP}_{3} /$ TIOCB $_{0} /$ TCLKD |
| 97 | $\begin{aligned} & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{4} / \\ & \overline{\mathrm{CS}}_{6} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{4} / \\ & \overline{\mathrm{CS}}_{6} \end{aligned}$ | $\mathrm{PA}_{4} / \mathrm{TP}_{4} /$ TIOCA/ $/$ $\overline{\mathrm{CS}}{ }_{6}$ | $\begin{aligned} & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{1} / \\ & \mathrm{CS}_{6} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{4} / \\ & \overline{\mathrm{CS}}_{6} \end{aligned}$ | $\mathrm{PA}_{4} / \mathrm{TP}_{4}$ TIOCA $/$ $\mathrm{A}_{23} / \mathrm{CS}_{6}$ | $\begin{aligned} & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{1} \end{aligned}$ |
| 98 | $\mathrm{PA}_{5} / \mathrm{TP}_{5} /$ $\mathrm{TIOCB}_{5}$ $\overline{\mathrm{CS}_{5}}$ | $\mathrm{PA}_{5} / \mathrm{TP}_{5} /$ $\overline{\mathrm{CS}}$ | $\mathrm{PA}_{5} / \mathrm{TP}_{5} /$ TIOCB/ $/$ $\overline{\mathrm{CS}}_{5}$ | $\begin{aligned} & \mathrm{PA}_{5} / \mathrm{TP}_{5} / \\ & \mathrm{TIOCBB}_{1} / \\ & \mathrm{CS}_{5} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{5} / \mathrm{TP}_{5} / \\ & \mathrm{TIOCBB}_{1} / \\ & \mathrm{CS}_{5} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{5} / \mathrm{TP}_{5} / \\ & \mathrm{TIOCB}_{8} / \\ & \mathrm{A}_{22} / \mathrm{CS}_{5} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{5} / \mathrm{TP}_{5} / \\ & \mathrm{TIOCB} \end{aligned}$ |
| 99 | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \text { TIOCA }_{2} \end{aligned}$ $\overline{\mathrm{CS}}_{4}$ | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \mathrm{TIOCA}_{2} \end{aligned}$ $\overline{\mathrm{CS}}_{4}$ | $\mathrm{PA}_{6} / \mathrm{TP}_{6} /$ $\mathrm{TIOCA}_{2}$ $\overline{\mathrm{CS}}_{4}$ | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \mathrm{TIOCA}_{2} / \\ & \mathrm{CS}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \mathrm{TIOCA}_{2} / \\ & \mathrm{CS}_{4} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \mathrm{TIOCA}_{2} / \\ & \mathrm{A}_{21} / \mathrm{CS}_{4}^{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}{ }_{\mathrm{s}} / \\ & \mathrm{TIOCA} \end{aligned}$ |
| 100 | $\begin{aligned} & \mathrm{PA}_{/} / \mathrm{TP}_{7} / \\ & \mathrm{TIOCB}_{2} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{/} / \mathrm{TP}_{7} / \\ & \mathrm{TIOCB}_{2} \end{aligned}$ | $\mathrm{A}_{20}$ | $\mathrm{A}_{20}$ | $\begin{aligned} & \mathrm{PA}_{/} / \mathrm{TP}_{/} / \\ & \mathrm{TIOCB}_{2} \end{aligned}$ | $\mathrm{A}_{20}$ | $\begin{aligned} & \mathrm{PA}_{/} / \mathrm{TP}_{7} / \\ & \mathrm{TIOCB}_{2} \end{aligned}$ |

Notes: 1. This pin functions as the $\mathrm{V}_{\mathrm{c}}$ pin during 5 V operation and as the $\mathrm{V}_{\mathrm{cc}}$ pin during 3 V operation. An external capacitor must be connected when this pin functions as the $\mathrm{V}_{\mathrm{c}}$ pin.
2. In modes $1,3,5$, and 6 the $P 4_{0}$ to $P 4_{7}$ functions of pins $P 4_{0} / D_{0}$ to $P 4_{7} / D_{7}$ are selected after a reset, but they can be changed by software.
3. In modes 2 and 4 the $D_{0}$ to $D_{7}$ functions of pins $\mathrm{P}_{0} / D_{0}$ to $P 4_{7} / D_{7}$ are selected after a reset, but they can be changed by software.

### 1.3.3 Pin Functions

Table 1.3 summarizes the pin functions.

## Table 1.3 Pin Functions

| Type | Symbol | Pin No. | 1/0 | Name and Function |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power | $\mathrm{V}_{\text {cc }}$ | 35, 68 | Input | Power: For connection to the power supply. Connect all $\mathrm{V}_{\mathrm{cc}}$ pins to the system power supply. |  |  |  |
|  | $\mathrm{V}_{\text {ss }}$ | $\begin{aligned} & 11,22,44, \\ & 57,65,92 \end{aligned}$ | Input | Ground: For connection to ground (0 V). Connect all $\mathrm{V}_{\text {ss }}$ pins to the $0-\mathrm{V}$ system power supply. |  |  |  |
|  | $\mathrm{V}_{\mathrm{CL}}$ | 1* | Input | Connect an external capacitor between this pin and GND ( 0 V ). |  |  |  |
| Clock | XTAL | 67 | Input | For connection to a crystal resonator. For examples of crystal resonator and external clock input, see section 19, Clock Pulse Generator. |  |  |  |
|  | EXTAL | 66 | Input | For connection to a crystal resonator or input of an external clock signal. For examples of crystal resonator and external clock input, see section 19, Clock Pulse Generator. |  |  |  |
|  | $\phi$ | 61 | Output | System clock: Supplies the system clock to external devices. |  |  |  |
| Operating mode control | $\mathrm{MD}_{2} \text { to } \mathrm{MD}_{0}$ | 75 to 73 | Input | Mode 2 to mode 0: For setting the operating mode, as follows. Inputs at these pins must not be changed during operation. |  |  |  |
|  |  |  |  | $\mathrm{MD}_{2}$ | MD ${ }_{1}$ | MD ${ }_{0}$ | Operating Mode |
|  |  |  |  | 0 | 0 | 0 | - |
|  |  |  |  | 0 | 0 | 1 | Mode 1 |
|  |  |  |  | 0 | 1 | 0 | Mode 2 |
|  |  |  |  | 0 | 1 | 1 | Mode 3 |
|  |  |  |  | 1 | 0 | 0 | Mode 4 |
|  |  |  |  | 1 | 0 | 1 | Mode 5 |
|  |  |  |  | 1 | 1 | 0 | Mode 6 |
|  |  |  |  | 1 | 1 | 1 | Mode 7 |


| Type | Symbol | Pin No. | 1/0 | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| System control | RES | 63 | Input | Reset input: When driven low, this pin resets the chip |
|  | FWE | 10 | Input | Flash write enable: Allows program mode setting. |
|  | $\overline{\text { STBY }}$ | 62 | Input | Standby: When driven low, this pin forces a transition to hardware standby mode |
|  | $\overline{\text { BREQ }}$ | 59 | Input | Bus request: Used by an external bus master to request the bus right |
|  | $\overline{\text { BACK }}$ | 60 | Output | Bus request acknowledge: Indicates that the bus has been granted to an external bus master |
| Interrupts | NMI | 64 | Input | Nonmaskable interrupt: Requests a nonmaskable interrupt |
|  | $\begin{aligned} & \overline{\mathrm{IRQ}}_{5} \text { to } \\ & \overline{\mathrm{IRQ}}_{0} \end{aligned}$ | $\begin{aligned} & 17,16, \\ & 90 \text { to } 87 \end{aligned}$ | Input | Interrupt request 5 to 0: Maskable interrupt request pins |
| Address bus | $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$ | $\begin{aligned} & 97 \text { to } 100, \\ & 56 \text { to } 45, \\ & 43 \text { to } 36 \end{aligned}$ | Output | Address bus: Outputs address signals |
| Data bus | $\mathrm{D}_{15}$ to $\mathrm{D}_{0}$ | $\begin{aligned} & 34 \text { to } 23, \\ & 21 \text { to } 18 \end{aligned}$ | Input/ output | Data bus: Bidirectional data bus |
| Bus control | $\overline{\mathrm{CS}}_{7}$ to $\overline{\mathrm{CS}}_{0}$ | $\begin{aligned} & 8,97 \text { to } 99, \\ & 88 \text { to } 91 \end{aligned}$ | Output | Chip select: Select signals for areas 7 to 0 |
|  | $\overline{\text { AS }}$ | 69 | Output | Address strobe: Goes low to indicate valid address output on the address bus |
|  | $\overline{\mathrm{RD}}$ | 70 | Output | Read: Goes low to indicate reading from the external address space |
|  | $\overline{\text { HWR }}$ | 71 | Output | High write: Goes low to indicate writing to the external address space; indicates valid data on the upper data bus ( $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ ). |
|  | $\overline{\text { LWR }}$ | 72 | Output | Low write: Goes low to indicate writing to the external address space; indicates valid data on the lower data bus ( $D_{7}$ to $D_{0}$ ). |
|  | $\overline{\text { WAIT }}$ | 58 | Input | Wait: Requests insertion of wait states in bus cycles during access to the external address space |

Rev. 2.0, 03/01, page 14 of 822

| Type | Symbol | Pin No. | 1/0 | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| Refresh controller | $\overline{\text { RFSH }}$ | 87 | Output | Refresh: Indicates a refresh cycle |
|  | $\overline{\mathrm{CS}}_{3}$ | 88 | Output | Row address strobe $\overline{\text { RAS: Row address }}$ strobe signal for DRAM connected to area 3 |
|  | $\overline{\overline{R D}}$ | 70 | Output | Column address strobe $\overline{\mathrm{CAS}}$ : Column address strobe signal for DRAM connected to area 3 ; used with $2 \overline{W E}$ DRAM. |
|  |  |  |  | Write enable $\overline{W E}$ : Write enable signal for DRAM connected to area 3; used with 2 $\overline{\mathrm{CAS}}$ DRAM. |
|  | $\overline{\text { HWR }}$ | 71 | Output | Upper write $\overline{\mathrm{UW}}$ : Write enable signal for DRAM connected to area 3 ; used with $2 \overline{\mathrm{WE}}$ DRAM. |
|  |  |  |  | Upper column address strobe $\overline{\text { UCAS: }}$ Column address strobe signal for DRAM connected to area 3; used with 2 $\overline{\mathrm{CAS}}$ DRAM. |
|  | $\overline{\overline{L W R}}$ | 72 | Output | Lower write $\overline{\text { LW: }}$ Write enable signal for DRAM connected to area 3 ; used with $2 \overline{W E}$ DRAM. |
|  |  |  |  | Lower column address strobe $\overline{\text { LCAS: }}$ Column address strobe signal for DRAM connected to area 3; used with 2 $\overline{\mathrm{CAS}}$ DRAM. |
| DMA controller (DMAC) | $\begin{aligned} & \overline{\operatorname{DREQ}}_{1}, \\ & \overline{\mathrm{DREQ}}_{0} \end{aligned}$ | 9, 8 | Input | DMA request 1 and 0: DMAC activation requests |
|  | $\begin{aligned} & \overline{\mathrm{TEND}}_{1}, \\ & \mathrm{TEND}_{0} \end{aligned}$ | 94, 93 | Output | Transfer end 1 and 0 : These signals indicate that the DMAC has ended a data transfer |
| 16-bit integrated timer unit (ITU) | TCLKD to TCLKA | 96 to 93 | Input | Clock input D to A: External clock inputs |
|  | $\begin{aligned} & \text { TIOCA }_{4} \text { to } \\ & \text { TIOCA }_{0} \end{aligned}$ | $\begin{aligned} & 4,2,99 \\ & 97,95 \end{aligned}$ | Input/ output | Input capture/output compare A4 to A0: GRA4 to GRA0 output compare or input capture, or PWM output |
|  | $\begin{aligned} & \text { TIOCB }_{4} \text { to } \\ & \text { TIOCB }_{0} \end{aligned}$ | $\begin{aligned} & 5,3,100 \\ & 98,96 \end{aligned}$ | Input/ output | Input capture/output compare B4 to B0: GRB4 to GRB0 output compare or input capture, or PWM output |
|  | $\mathrm{TOCXA}_{4}$ | 6 | Output | Output compare XA4: PWM output |
|  | $\mathrm{TOCXB}_{4}$ | 7 | Output | Output compare XB4: PWM output |


| Type | Symbol | Pin No. | 1/0 | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| Programmable timing pattern controller (TPC) | $\mathrm{TP}_{15}$ to $\mathrm{TP}_{0}$ | $\begin{aligned} & 9 \text { to } 2, \\ & 100 \text { to } 93 \end{aligned}$ | Output | TPC output 15 to 0: Pulse output |
| Serial communication interface (SCl) | TxD ${ }_{1}, T x D_{0}$ | 13, 12 | Output | Transmit data (channels 0 and 1): SCI data output |
|  | RxD ${ }_{1}, R \times D_{0}$ | 15, 14 | Input | Receive data (channels 0 and 1): SCl data input |
|  | $\mathrm{SCK}_{1}, \mathrm{SCK}_{0}$ | 17, 16 | Input/ output | Serial clock (channels 0 and 1): SCI clock input/output |
| A/D converter | $\mathrm{AN}_{7}$ to $\mathrm{AN}_{0}$ | 85 to 78 | Input | Analog 7 to 0: Analog input pins |
|  | $\overline{\text { ADTRG }}$ | 9 | Input | A/D trigger: External trigger input for starting A/D conversion |
| D/A converter | $\mathrm{DA}_{1}, \mathrm{DA}_{0}$ | 85, 84 | Output | Analog output: Analog output from the D/A converter |
| A/D and D/A converters | $\mathrm{AV}_{\text {cc }}$ | 76 | Input | Power supply pin for the A/D and D/A converters. Connect to the system power supply ( +5 V ) when not using the A/D and D/A converters. |
|  | $\mathrm{AV}_{\text {ss }}$ | 86 | Input | Ground pin for the A/D and D/A converters. Connect to system ground (0 V). |
|  | $\mathrm{V}_{\text {REF }}$ | 77 | Input | Reference voltage input pin for the $A / D$ and D/A converters. Connect to the system power supply ( +5 V ) when not using the $A / D$ and D/A converters. |
| I/O ports | $\mathrm{P} 1_{7}$ to $\mathrm{P} 1_{0}$ | 43 to 36 | Input/ output | Port 1: Eight input/output pins. <br> The direction of each pin can be selected in the port 1 data direction register (P1DDR). |
|  | $\mathrm{P} 2_{7}$ to $\mathrm{P} 2_{0}$ | 52 to 45 | Input/ output | Port 2: Eight input/output pins. The direction of each pin can be selected in the port 2 data direction register (P2DDR). |
|  | $\mathrm{P}_{7}$ to P 30 | 34 to 27 | Input/ output | Port 3: Eight input/output pins. <br> The direction of each pin can be selected in the port 3 data direction register (P3DDR). |
|  | $\mathrm{P} 4_{7}$ to P 40 | $\begin{aligned} & 26 \text { to } 23, \\ & 21 \text { to } 18 \end{aligned}$ | Input/ output | Port 4: Eight input/output pins. <br> The direction of each pin can be selected in the port 4 data direction register (P4DDR). |

Rev. 2.0, 03/01, page 16 of 822

| Type | Symbol | Pin No. | 1/0 | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| I/O ports | $\mathrm{P5}_{3}$ to $\mathrm{P5}{ }_{0}$ | 56 to 53 | Input/ output | Port 5: Four input/output pins. The direction of each pin can be selected in the port 5 data direction register (P5DDR). |
|  | P6, to P60 | $\begin{aligned} & 72 \text { to } 69, \\ & 60 \text { to } 58 \end{aligned}$ | Input/ output | Port 6: Seven input/output pins. The direction of each pin can be selected in the port 6 data direction register (P6DDR). |
|  | $\mathrm{P7}_{7}$ to $\mathrm{P} 7_{0}$ | 85 to 78 | Input | Port 7: Eight input pins |
|  | P84 to P80 | 91 to 87 | Input/ output | Port 8: Five input/output pins. The direction of each pin can be selected in the port 8 data direction register (P8DDR). |
|  | $\mathrm{P9}_{5}$ to $\mathrm{P9}$ 。 | 17 to 12 | Input/ output | Port 9: Six input/output pins. The direction of each pin can be selected in the port 9 data direction register (P9DDR). |
|  | PA, to $\mathrm{PA}_{0}$ | 100 to 93 | Input/ output | Port A: Eight input/output pins. The direction of each pin can be selected in the port A data direction register (PADDR). |
|  | $\mathrm{PB}_{7}$ to $\mathrm{PB}_{0}$ | 9 to 2 | Input/ output | Port B: Eight input/output pins. The direction of each pin can be selected in the port $B$ data direction register (PBDDR). |
| Notes: | pin functions ected to the | as the $V_{c\llcorner } p$ stem pow | during 5 upply) | operation and as the $\mathrm{V}_{\mathrm{cc}}$ pin (should be ing 3 V operation. |

## Section 2 CPU

### 2.1 Overview

The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU is a high-speed central processing unit with an internal 32-bit architecture that is upward-compatible with the $\mathrm{H} 8 / 300 \mathrm{CPU}$. The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU has sixteen 16-bit general registers, can address a $16-\mathrm{Mbyte}$ linear address space, and is ideal for realtime control.

### 2.1.1 Features

The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU has the following features.

- Upward compatibility with $\mathrm{H} 8 / 300 \mathrm{CPU}$

Can execute H8/300 Series object programs

- General-register architecture

Sixteen 16-bit general registers (also usable as sixteen 8-bit registers or eight 32-bit registers)

- Sixty-two basic instructions
- 8/16/32-bit data transfer and arithmetic and logic instructions
- Multiply and divide instructions
- Powerful bit-manipulation instructions
- Eight addressing modes
— Register direct [Rn]
— Register indirect [@ERn]
— Register indirect with displacement [@(d:16, ERn) or @(d:24, ERn)]
— Register indirect with post-increment or pre-decrement [@ERn+ or @-ERn]
— Absolute address [@aa:8, @aa:16, or @aa:24]
— Immediate [\#xx:8, \#xx:16, or \#xx:32]
— Program-counter relative [@(d:8, PC) or @(d:16, PC)]
— Memory indirect [@ @aa:8]
- 16-Mbyte linear address space
- High-speed operation
- All frequently-used instructions execute in two to four states
— Maximum clock frequency: $\quad 25 \mathrm{MHz}$
- 8/16/32-bit register-register add/subtract: 80 ns
- $8 \times 8$-bit register-register multiply: 560 ns
- $16 \div 8$-bit register-register divide: 560 ns
- $16 \times 16$-bit register-register multiply: $\quad 0.88 \mu \mathrm{~s}$
$-32 \div$ 16-bit register-register divide: $\quad 0.88 \mu$ s
- Two CPU operating modes
- Normal mode (not available in the H8/3052F)
- Advanced mode
- Low-power mode

Transition to power-down state by SLEEP instruction

### 2.1.2 Differences from H8/300 CPU

In comparison to the $\mathrm{H} 8 / 300 \mathrm{CPU}$, the $\mathrm{H} 8 / 300 \mathrm{H}$ has the following enhancements.

- More general registers

Eight 16-bit registers have been added.

- Expanded address space
- Advanced mode supports a maximum 16-Mbyte address space.
- Normal mode supports the same 64-kbyte address space as the $\mathrm{H} 8 / 300 \mathrm{CPU}$.
(Normal mode is not available in the $\mathrm{H} 8 / 3052 \mathrm{~F}$.)
- Enhanced addressing

The addressing modes have been enhanced to make effective use of the 16-Mbyte address space.

- Enhanced instructions
- Data transfer, arithmetic, and logic instructions can operate on 32-bit data.
- Signed multiply/divide instructions and other instructions have been added.


### 2.2 CPU Operating Modes

The H8/300H CPU has two operating modes: normal and advanced. Normal mode supports a maximum 64-kbyte address space. Advanced mode supports up to 16 Mbytes.

The H8/3052F can be used only in advanced mode. (Information from this point on will apply to advanced mode unless otherwise stated.)


Figure 2.1 CPU Operating Modes

### 2.3 Address Space

The maximum address space of the $\mathrm{H} 8 / 300 \mathrm{H}$ CPU is 16 Mbytes. The $\mathrm{H} 8 / 3052 \mathrm{~F}$ has various operating modes (MCU modes), some providing a 1-Mbyte address space, the others supporting the full 16 Mbytes.

Figure 2.2 shows the address ranges of the $\mathrm{H} 8 / 3052 \mathrm{~F}$. For further details see section 3.6, Memory Map in Each Operating Mode.

The 1-Mbyte operating modes use 20-bit addressing. The upper 4 bits of effective addresses are ignored.

| H'00000 | H'000000 |  |
| :---: | :---: | :---: |
| H'FFFFF |  |  |
|  | H'FFFFFF |  |
|  |  | b. 16-Mbyte modes |

Figure 2.2 Memory Map

### 2.4 Register Configuration

### 2.4.1 Overview

The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU has the internal registers shown in figure 2.3. There are two types of registers: general registers and control registers.

General Registers (ERn)

| 15 |  |  |  | 07 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ER0 | E0 |  | ROH | ROL |  |
| ER1 | E1 |  | R1H | R1L |  |
| ER2 | E2 |  | R2H | R2L |  |
| ER3 | E3 |  | R3H | R3L |  |
| ER4 | E4 |  | R4H | R4L |  |
| ER5 | E5 |  | R5H | R5L |  |
| ER6 | E6 |  | R6H | R6L |  |
| ER7 | E7 | (SP) | R7H | R7L |  |

## Control Registers (CR)



Legend
SP: Stack pointer
PC: Program counter
CCR: Condition code register
I: Interrupt mask bit
UI: User bit or interrupt mask bit
H: Half-carry flag
U: User bit
$\mathrm{N}: \quad$ Negative flag
Z: Zero flag
V: Overflow flag
C: Carry flag

Figure 2.3 CPU Internal Registers

### 2.4.2 General Registers

The H8/300H CPU has eight 32-bit general registers. These general registers are all functionally alike and can be used without distinction between data registers and address registers. When a general register is used as a data register, it can be accessed as a 32-bit, 16 -bit, or 8 -bit register. When the general registers are used as 32 -bit registers or as address registers, they are designated by the letters ER (ER0 to ER7).

The ER registers divide into 16 -bit general registers designated by the letters E (E0 to E7) and R (R0 to R7). These registers are functionally equivalent, providing a maximum sixteen 16-bit registers. The E registers (E0 to E7) are also referred to as extended registers.

The R registers divide into 8 -bit general registers designated by the letters RH ( R 0 H to R 7 H ) and RL (R0L to R7L). These registers are functionally equivalent, providing a maximum sixteen 8 -bit registers.

Figure 2.4 illustrates the usage of the general registers. The usage of each register can be selected independently.


Figure 2.4 Usage of General Registers

General register ER7 has the function of stack pointer (SP) in addition to its general-register function, and is used implicitly in exception handling and subroutine calls. Figure 2.5 shows the stack.


Figure 2.5 Stack

### 2.4.3 Control Registers

The control registers are the 24-bit program counter (PC) and the 8-bit condition code register (CCR).

Program Counter (PC): This 24-bit counter indicates the address of the next instruction the CPU will execute. The length of all CPU instructions is 2 bytes (one word) or a multiple of 2 bytes, so the least significant PC bit is ignored. When an instruction is fetched, the least significant PC bit is regarded as 0 .

Condition Code Register (CCR): This 8-bit register contains internal CPU status information, including the interrupt mask bit (I) and half-carry (H), negative (N), zero (Z), overflow (V), and carry (C) flags.

- Bit 7—Interrupt Mask Bit (I)

Masks interrupts other than NMI when set to 1 . NMI is accepted regardless of the I bit setting. The I bit is set to 1 at the start of an exception-handling sequence.

- Bit 6—User Bit or Interrupt Mask Bit (UI)

Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions. This bit can also be used as an interrupt mask bit. For details see section 5, Interrupt Controller.

- Bit 5—Half-Carry Flag (H)

When the ADD.B, ADDX.B, SUB.B, SUBX.B, CMP.B, or NEG.B instruction is executed, this flag is set to 1 if there is a carry or borrow at bit 3 , and cleared to 0 otherwise. When the ADD.W, SUB.W, CMP.W, or NEG.W instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 11 , and cleared to 0 otherwise. When the ADD.L, SUB.L, CMP.L, or

NEG.L instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 27, and cleared to 0 otherwise.

- Bit 4—User Bit (U)

Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions.

- Bit 3—Negative Flag (N)

Indicates the most significant bit (sign bit) of data.

- Bit 2—Zero Flag (Z)

Set to 1 to indicate zero data, and cleared to 0 to indicate non-zero data.

- Bit 1—Overflow Flag (V)

Set to 1 when an arithmetic overflow occurs, and cleared to 0 at other times.

- Bit 0-Carry Flag (C)

Set to 1 when a carry occurs, and cleared to 0 otherwise. Used by:

- Add instructions, to indicate a carry
- Subtract instructions, to indicate a borrow
- Shift and rotate instructions, to store the value shifted out of the end bit

The carry flag is also used as a bit accumulator by bit manipulation instructions.
Some instructions leave flag bits unchanged. Operations can be performed on CCR by the LDC, STC, ANDC, ORC, and XORC instructions. The N, Z, V, and C flags are used by conditional branch (Bcc) instructions.

For the action of each instruction on the flag bits, see appendix A.1, Instruction List. For the I and UI bits, see section 5, Interrupt Controller.

### 2.4.4 Initial CPU Register Values

In reset exception handling, PC is initialized to a value loaded from the vector table, and the I bit in CCR is set to 1 . The other CCR bits and the general registers are not initialized. The initial value of the stack pointer (ER7) is undefined. The stack pointer must therefore be initialized by an MOV.L instruction executed immediately after a reset.

### 2.5 Data Formats

The H8/300H CPU can process 1-bit, 4-bit (BCD), 8-bit (byte), 16-bit (word), and 32-bit (longword) data. Bit-manipulation instructions operate on 1-bit data by accessing bit $\mathrm{n}(\mathrm{n}=0,1,2$, $\ldots, 7)$ of byte operand data. The DAA and DAS decimal-adjust instructions treat byte data as two digits of 4-bit BCD data.

### 2.5.1 General Register Data Formats

Figure 2.6 shows the data formats in general registers.


Figure 2.6 General Register Data Formats (1)

Data Type | General |
| :--- |
| Register | Data Format

Word data $\quad \mathrm{Rn}$
Word data
Longword data $\quad$ ERn

Figure 2.6 General Register Data Formats (2)

### 2.5.2 Memory Data Formats

Figure 2.7 shows the data formats on memory. The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU can access word data and longword data on memory, but word or longword data must begin at an even address. If an attempt is made to access word or longword data at an odd address, no address error occurs but the least significant bit of the address is regarded as 0 , so the access starts at the preceding address. This also applies to instruction fetches.


Figure 2.7 Memory Data Formats
When ER7 (SP) is used as an address register to access the stack, the operand size should be word size or longword size.

### 2.6 Instruction Set

### 2.6.1 Instruction Set Overview

The H8/300H CPU has 62 types of instructions, which are classified in table 2.1.
Table 2.1 Instruction Classification

| Function | Instruction | Types |
| :--- | :--- | :--- |
| Data transfer | MOV, PUSH*1, POP*1, MOVTPE*2, MOVFPE*2 | 3 |
| Arithmetic operations | ADD, SUB, ADDX, SUBX, INC, DEC, ADDS, SUBS, DAA, | 18 |
|  | DAS, MULXU, MULXS, DIVXU, DIVXS, CMP, NEG, EXTS, |  |
|  | EXTU | 4 |
| Logic operations | AND, OR, XOR, NOT | 4 |
| Shift operations | SHAL, SHAR, SHLL, SHLR, ROTL, ROTR, ROTXL, ROTXR | 8 |
| Bit manipulation | BSET, BCLR, BNOT, BTST, BAND, BIAND, BOR, BIOR, | 14 |
|  | BXOR, BIXOR, BLD, BILD, BST, BIST | 14 |
| Branch | Bcc**, JMP, BSR, JSR, RTS | 5 |
| System control | TRAPA, RTE, SLEEP, LDC, STC, ANDC, ORC, XORC, NOP | 9 |
| Block data transfer | EEPMOV | 1 |
|  |  | Total 62 types |

Notes: 1. POP.W Rn is identical to MOV.W @SP+, Rn. PUSH.W Rn is identical to MOV.W Rn, @-SP. POP.L ERn is identical to MOV.L @SP+, Rn. PUSH.L ERn is identical to MOV.L Rn, @-SP.
2. Not available in the $\mathrm{H} 8 / 3052 \mathrm{~F}$.
3. Bcc is a generic branching instruction.

### 2.6.2 Instructions and Addressing Modes

Table 2.2 indicates the instructions available in the $\mathrm{H} 8 / 300 \mathrm{H}$ CPU
Table 2.2 Instructions and Addressing Modes

| Function | Instruction | Addressing Modes |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | X | ¢ |  |  |  |  | $\begin{aligned} & \text { © } \\ & \text { © } \\ & \text { © } \end{aligned}$ |  |  |  |  |  | 1 |
| Data transfer | MOV | BWL | BWL | BWL | BWL | BWL | BWL | B | BWL | BWL | - | - | - | - |
|  | POP, PUSH | - | - | - | - | - | - | - | - | - | - | - | - | WL |
|  | MOVFPE,* MOVTPE* | - | - | - | - | - | - | - | B | - | - | - | - | - |
| Arithmeticoperations | ADD, CMP | BWL | BWL | - | - | - | - | - | - | - | - | - | - | - |
|  | SUB | WL | BWL | - | - | - | - | - | - | - | - | - | - | - |
|  | ADDX, SUBX | B | B | - | - | - | - | - | - | - | - | - | - | - |
|  | ADDS, SUBS | - | L | - | - | - | - | - | - | - | - | - | - | - |
|  | INC, DEC | - | BWL | - | - | - | - | - | - | - | - | - | - | - |
|  | DAA, DAS | - | B | - | - | - | - | - | - | - | - | - | - | - |
|  | MULXU, MULXS, DIVXU, DIVXS | - | BW | - | - | - | - | - | - | - | - | - | - | - |
|  | NEG | - | BWL | - | - | - | - | - | - | - | - | - | - | - |
|  | EXTU, EXTS | - | WL | - | - | - | - | - | - | - | - | - | - | - |
| $\begin{aligned} & \text { Logic } \\ & \text { operations } \end{aligned}$ | AND, OR, XOR | BWL | BWL | - | - | - | - | - | - | - | - | - | - | - |
|  | NOT | - | BWL | - | - | - | - | - | - | - | - | - | - | - |
| Shift instructions |  | - | BWL | - | - | - | - | - | - | - | - | - | - | - |
| Bit manipulation |  | - | B | B | - | - | - | B | - | - | - | - | - | - |
| Branch | Bcc, BSR | - | - | - | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | - | - |
|  | JMP, JSR | - | - | $\bigcirc$ | - | - | - | - | - | $\bigcirc$ | - | - | $\bigcirc$ | - |
|  | RTS | - | - | - | - | - | - | - | - | - | - | - | - | $\bigcirc$ |
| System control | TRAPA | - | - | - | - | - | - | - | - | - | - | - | - | $\bigcirc$ |
|  | RTE | - | - | - | - | - | - | - | - | - | - | - | - | $\bigcirc$ |
|  | SLEEP | - | - | - | - | - | - | - | - | - | - | - | - | $\bigcirc$ |
|  | LDC | B | B | w | w | w | w | - | w | w | - | - | - | - |
|  | STC | - | B | w | w | w | w | - | w | w | - | - | - | - |
|  | ANDC, ORC, XORC | B | - | - | - | - | - | - | - | - | - | - | - | - |
|  | NOP | - | - | - | - | - | - | - | - | - | - | - | - | $\bigcirc$ |
| Block data transer |  | - | - | - | - | - | - | - | - | - | - | - | - | BW |
| Legend |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| B: Byte |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| W: Word |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| L: Longword |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| Note: * Not available in the H8/3052F. |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

### 2.6.3 Tables of Instructions Classified by Function

Tables 2.3 to 2.10 summarize the instructions in each functional category. The operation notation used in these tables is defined next.

## Operation Notation

| Rd | General register (destination)* |
| :---: | :---: |
| Rs | General register (source)* |
| Rn | General register* |
| ERn | General register (32-bit register or address register) |
| (EAd) | Destination operand |
| (EAs) | Source operand |
| CCR | Condition code register |
| N | $N$ (negative) flag of CCR |
| Z | Z (zero) flag of CCR |
| V | $V$ (overflow) flag of CCR |
| C | C (carry) flag of CCR |
| PC | Program counter |
| SP | Stack pointer |
| \#IMM | Immediate data |
| disp | Displacement |
| + | Addition |
| - | Subtraction |
| $\times$ | Multiplication |
| $\div$ | Division |
| $\wedge$ | AND logical |
| $\checkmark$ | OR logical |
| $\oplus$ | Exclusive OR logical |
| $\rightarrow$ | Move |
| ᄀ | NOT (logical complement) |
| :3/:8/:16/:24 | 3-, 8-, 16-, or 24-bit length |
| Note: * General registers include 8-bit registers (R0H to R7H, ROL to R7L), 16-bit registers (R0 to R7, E0 to E7), and 32-bit data or address registers (ER0 to ER7). |  |

Table 2.3 Data Transfer Instructions

| Instruction | Size* | Function |
| :---: | :---: | :---: |
| MOV | B/W/L | (EAs) $\rightarrow \mathrm{Rd}$, Rs $\rightarrow$ (EAd) |
|  |  | Moves data between two general registers or between a general register and memory, or moves immediate data to a general register. |
| MOVFPE | B | (EAs) $\rightarrow$ Rd |
|  |  | Cannot be used in the H8/3052F. |
| MOVTPE | B | Rs $\rightarrow$ (EAs) |
|  |  | Cannot be used in the H8/3052F. |
| POP | W/L | @SP $+\rightarrow$ Rn |
|  |  | Pops a general register from the stack. POP.W Rn is identical to MOV.W @SP+, Rn. Similarly, POP.L ERn is identical to MOV.L @SP+, ERn. |
| PUSH | W/L | Rn $\rightarrow$ @-SP |
|  |  | Pushes a general register onto the stack. PUSH.W Rn is identical to MOV.W Rn, @-SP. Similarly, PUSH.L ERn is identical to MOV.L ERn, @-SP. |

Note: * Size refers to the operand size.
B: Byte
W: Word
L: Longword

Table 2.4 Arithmetic Operation Instructions

| Instruction | Size* | Function |
| :---: | :---: | :---: |
| ADD, SUB | B/W/L | $\mathrm{Rd} \pm \mathrm{Rs} \rightarrow \mathrm{Rd}, \mathrm{Rd} \pm$ \#IMM $\rightarrow \mathrm{Rd}$ |
|  |  | Performs addition or subtraction on data in two general registers, or on immediate data and data in a general register. (Immediate byte data cannot be subtracted from data in a general register. Use the SUBX or ADD instruction.) |
| ADDX, SUBX | B | $R d \pm R s \pm C \rightarrow R d, R d \pm \# I M M \pm C \rightarrow R d$ |
|  |  | Performs addition or subtraction with carry or borrow on data in two general registers, or on immediate data and data in a general register. |
| INC, DEC | B/W/L | $\mathrm{Rd} \pm 1 \rightarrow \mathrm{Rd}, \mathrm{Rd} \pm 2 \rightarrow \mathrm{Rd}$ |
|  |  | Increments or decrements a general register by 1 or 2. (Byte operands can be incremented or decremented by 1 only.) |
| ADDS, SUBS | L | $\mathrm{Rd} \pm 1 \rightarrow \mathrm{Rd}, \mathrm{Rd} \pm 2 \rightarrow \mathrm{Rd}, \mathrm{Rd} \pm 4 \rightarrow \mathrm{Rd}$ |
|  |  | Adds or subtracts the value 1, 2 , or 4 to or from data in a 32-bit register. |
| DAA, DAS | B | Rd decimal adjust $\rightarrow$ Rd |
|  |  | Decimal-adjusts an addition or subtraction result in a general register by referring to CCR to produce 4 -bit BCD data. |
| MULXU | B/W | $\mathrm{Rd} \times \mathrm{Rs} \rightarrow \mathrm{Rd}$ |
|  |  | Performs unsigned multiplication on data in two general registers: either 8 bits $\times 8$ bits $\rightarrow 16$ bits or 16 bits $\times 16$ bits $\rightarrow 32$ bits. |
| MULXS | B/W | $\mathrm{Rd} \times \mathrm{Rs} \rightarrow \mathrm{Rd}$ |
|  |  | Performs signed multiplication on data in two general registers: either 8 bits $\times 8$ bits $\rightarrow 16$ bits or 16 bits $\times 16$ bits $\rightarrow 32$ bits. |


| Instruction | Size* | Function |
| :---: | :---: | :---: |
| DIVXU | B/W | $\mathrm{Rd} \div \mathrm{Rs} \rightarrow \mathrm{Rd}$ |
|  |  | Performs unsigned division on data in two general registers: either 16 bits $\div 8$ bits $\rightarrow 8$-bit quotient and 8 -bit remainder or 32 bits $\div 16$ bits $\rightarrow 16$-bit quotient and 16 -bit remainder. |
| DIVXS | B/W | $\mathrm{Rd} \div \mathrm{Rs} \rightarrow \mathrm{Rd}$ |
|  |  | Performs signed division on data in two general registers: either 16 bits $\div 8$ bits $\rightarrow 8$-bit quotient and 8 -bit remainder, or 32 bits $\div 16$ bits $\rightarrow 16$-bit quotient and 16-bit remainder. |
| CMP | B/W/L | Rd-Rs, Rd-\#IMM |
|  |  | Compares data in a general register with data in another general register or with immediate data, and sets CCR according to the result. |
| NEG | B/W/L | $0-\mathrm{Rd} \rightarrow \mathrm{Rd}$ |
|  |  | Takes the two's complement (arithmetic complement) of data in a general register. |
| EXTS | W/L | Rd (sign extension) $\rightarrow$ Rd |
|  |  | Extends byte data in the lower 8 bits of a 16-bit register to word data, or extends word data in the lower 16 bits of a 32-bit register to longword data, by extending the sign bit. |
| EXTU | W/L | Rd (zero extension) $\rightarrow$ Rd |
|  |  | Extends byte data in the lower 8 bits of a 16-bit register to word data, or extends word data in the lower 16 bits of a 32-bit register to longword data, by padding with zeros. |

Note: * Size refers to the operand size.
B: Byte
W: Word
L: Longword

Table 2.5 Logic Operation Instructions

| Instruction | Size* | Function |
| :---: | :---: | :---: |
| AND | B/W/L | $\mathrm{Rd} \wedge \mathrm{Rs} \rightarrow \mathrm{Rd}, \mathrm{Rd} \wedge$ \#IMM $\rightarrow \mathrm{Rd}$ |
|  |  | Performs a logical AND operation on a general register and another general register or immediate data. |
| OR | B/W/L | Rdv Rs $\rightarrow$ Rd, Rdv \#IMM $\rightarrow$ Rd |
|  |  | Performs a logical OR operation on a general register and another general register or immediate data. |
| XOR | B/W/L | $\mathrm{Rd} \oplus \mathrm{Rs} \rightarrow \mathrm{Rd}, \mathrm{Rd} \oplus$ \#IMM $\rightarrow \mathrm{Rd}$ |
|  |  | Performs a logical exclusive OR operation on a general register and another general register or immediate data. |
| NOT | B/W/L | $\neg \mathrm{Rd} \rightarrow \mathrm{Rd}$ |

Takes the one's complement of general register contents.
Note: * Size refers to the operand size.
B: Byte
W: Word
L: Longword

Table 2.6 Shift Instructions

| Instruction | Size* | Function |
| :--- | :--- | :--- |
| SHAL, | B/W/L | Rd (shift) $\rightarrow$ Rd <br> SHAR |
|  |  | Performs an arithmetic shift on general register contents. |
| SHLL, | B/W/L | Rd (shift) $\rightarrow$ Rd <br> SHLR |
| ROTforms a logical shift on general register contents. |  |  |
| ROTR | B/W/L | Rd (rotate) $\rightarrow$ Rd <br> Rotates general register contents. |
| ROTXL, | B/W/L | Rd (rotate) $\rightarrow$ Rd <br> ROTXR |
|  |  | Rotates general register contents through the carry bit. |

Note: * Size refers to the operand size.
B: Byte
W: Word
L: Longword

Table 2.7 Bit Manipulation Instructions
Instruction Size* Function

| BSET | B | $1 \rightarrow$ (<bit-No.> of <EAd>) |
| :---: | :---: | :---: |
|  |  | Sets a specified bit in a general register or memory operand to 1. The bit number is specified by 3-bit immediate data or the lower 3 bits of a general register. |
| BCLR | B | $0 \rightarrow$ (<bit-No.> of <EAd>) |
|  |  | Clears a specified bit in a general register or memory operand to 0 . The bit number is specified by 3 -bit immediate data or the lower 3 bits of a general register. |
| BNOT | B | $\neg$ (<bit-No.> of <EAd>) $\rightarrow$ (<bit-No.> of <EAd>) |
|  |  | Inverts a specified bit in a general register or memory operand. The bit number is specified by 3 -bit immediate data or the lower 3 bits of a general register. |
| BTST | B | $\neg$ (<bit-No.> of <EAd>) $\rightarrow$ Z |
|  |  | Tests a specified bit in a general register or memory operand and sets or clears the $Z$ flag accordingly. The bit number is specified by 3 -bit immediate data or the lower 3 bits of a general register. |
| BAND | B | $\mathrm{C} \wedge$ (<bit-No.> of <EAd>) $\rightarrow$ C |
|  |  | ANDs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag. |
| BIAND | B | $C \wedge[\neg(<$ bit-No. $>$ of $<E A d>$ ) $] \rightarrow$ C |
|  |  | ANDs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag. The bit number is specified by 3 -bit immediate data. |


| Instruction | Size* | Function |
| :---: | :---: | :---: |
| BOR | B | C $\vee$ (<bit-No.> of <EAd $>$ ) $\rightarrow$ C |
|  |  | ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag. |
| BIOR | B | $C \vee[\neg$ (<bit-No. $>$ of $<E A d>$ ) $] \rightarrow$ C |
|  |  | ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag. |
|  |  | The bit number is specified by 3-bit immediate data. |
| BXOR | B | C $\oplus$ (<bit-No.> of <EAd>) $\rightarrow$ C |
|  |  | Exclusive-ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag. |
| BIXOR | B | $\mathrm{C} \oplus[\neg$ (<bit-No.> of <EAd>) $] \rightarrow \mathrm{C}$ |
|  |  | Exclusive-ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag. |
|  |  | The bit number is specified by 3-bit immediate data. |
| BLD | B | (<bit-No.> of <EAd>) $\rightarrow$ C |
|  |  | Transfers a specified bit in a general register or memory operand to the carry flag. |
| BILD | B | $\neg$ (<bit-No.> of <EAd>) $\rightarrow$ C |
|  |  | Transfers the inverse of a specified bit in a general register or memory operand to the carry flag. |
|  |  | The bit number is specified by 3-bit immediate data. |
| BST | B | C $\rightarrow$ (<bit-No.> of <EAd>) |
|  |  | Transfers the carry flag value to a specified bit in a general register or memory operand. |
| BIST | B | $\mathrm{C} \rightarrow \neg$ (<bit-No.> of <EAd>) |
|  |  | Transfers the inverse of the carry flag value to a specified bit in a general register or memory operand. |
|  |  | The bit number is specified by 3-bit immediate data. |

Note: * Size refers to the operand size.
B: Byte

Table 2.8 Branching Instructions
Instruction Size Function
Bcc - Branches to a specified address if a specified condition is true. The branching conditions are listed below.

| Mnemonic | Description | Condition |
| :--- | :--- | :--- |
| BRA (BT) | Always (true) | Always |
| BRN (BF) | Never (false) | Never |
| BHI | High | $\mathrm{C} \vee \mathrm{Z}=0$ |
| BLS | Low or same | $\mathrm{C} \vee \mathrm{Z}=1$ |
| Bcc (BHS) | Carry clear (high or same) | $\mathrm{C}=0$ |
| BCS (BLO) | Carry set (low) | $\mathrm{C}=1$ |
| BNE | Not equal | $\mathrm{Z}=0$ |
| BEQ | Equal | $\mathrm{Z}=1$ |
| BVC | Overflow clear | $\mathrm{V}=0$ |
| BVS | Overflow set | $\mathrm{V}=1$ |
| BPL | Plus | $\mathrm{N}=0$ |
| BMI | Minus | $\mathrm{N}=1$ |
| BGE | Greater or equal | $\mathrm{N} \oplus \mathrm{V}=0$ |
| BLT | Less than | $\mathrm{N} \oplus \mathrm{V}=1$ |
| BGT | Greater than | $\mathrm{Z} \vee(\mathrm{N} \oplus \mathrm{V})=0$ |
| BLE | Less or equal | $\mathrm{Z} \vee(\mathrm{N} \oplus \mathrm{V})=1$ |


| JMP | - | Branches unconditionally to a specified address |
| :--- | :--- | :--- |
| BSR | - | Branches to a subroutine at a specified address |
| JSR | - | Branches to a subroutine at a specified address |
| RTS | - | Returns from a subroutine |

Table 2.9 System Control Instructions

| Instruction | Size* | Function |
| :--- | :--- | :--- |
| TRAPA | - | Starts trap-instruction exception handling |
| RTE | - | Returns from an exception-handling routine |
| SLEEP | - | Causes a transition to the power-down state |
| LDC | B/W | (EAs) $\rightarrow$ CCR <br> Moves the source operand contents to the condition code register. <br> The condition code register size is one byte, but in transfer from <br> memory, data is read by word access. |
| STC | B/W | CCR $\rightarrow$ (EAd) <br> Transfers the CCR contents to a destination location. The condition <br> code register size is one byte, but in transfer to memory, data is <br> written by word access. |
| ANDC | CCR $\wedge$ \#IMM $\rightarrow$ CCR <br> Logically AND the condition code register with immediate data. |  |
| ORC | CCR $\vee$ \#IMM $\rightarrow$ CCR <br> Logically ORs the condition code register with immediate data. |  |
| XORC | B | CCR $\oplus$ \#IMM $\rightarrow$ CCR <br> Logically exclusive-ORs the condition code register with immediate <br> data. |
| NOP | PC $+2 \rightarrow$ PC |  |
| Only increments the program counter. |  |  |

Note: * Size refers to the operand size.
B: Byte
W: Word

Table 2.10 Block Transfer Instruction
Instruction Size Function
EEPMOV.B - if R4L $\neq 0$ then
repeat @ER5+ $\rightarrow$ @ER6+, R4L-1 $\rightarrow$ R4L
until $\quad R 4 L=0$
else next;
EEPMOV.W - if $\mathrm{R} 4 \neq 0$ then
repeat @ER5+ $\rightarrow$ @ER6+, R4-1 $\rightarrow$ R4
until $\quad$ 4 $4=0$
else next;
Transfers a data block according to parameters set in general registers R4L or R4, ER5, and ER6.
R4L or R4: Size of block (bytes)
ER5: Starting source address
ER6: Starting destination address
Execution of the next instruction begins as soon as the transfer is completed.

### 2.6.4 Basic Instruction Formats

The H8/300H instructions consist of 2-byte (1-word) units. An instruction consists of an operation field (OP field), a register field ( r field), an effective address extension (EA field), and a condition field (cc).

Operation Field: Indicates the function of the instruction, the addressing mode, and the operation to be carried out on the operand. The operation field always includes the first 4 bits of the instruction. Some instructions have two operation fields.

Register Field: Specifies a general register. Address registers are specified by 3 bits, data registers by 3 bits or 4 bits. Some instructions have two register fields. Some have no register field.

Effective Address Extension: Eight, 16, or 32 bits specifying immediate data, an absolute address, or a displacement. A 24-bit address or displacement is treated as 32-bit data in which the first 8 bits are $0\left(\mathrm{H}^{\prime} 00\right)$.

Condition Field: Specifies the branching condition of Bcc instructions.
Figure 2.8 shows examples of instruction formats.

Operation field only
$\square$ op NOP, RTS, etc.

Operation field and register fields

| op | rn | rm |
| :---: | :---: | :---: |

Operation field, register fields, and effective address extension

| op | rn | rm |
| :---: | :---: | :---: |
| EA (disp) |  |  |

Operation field, effective address extension, and condition field

| op | cc | EA (disp) |
| :--- | :--- | :--- | BRA d:8

Figure 2.8 Instruction Formats

### 2.6.5 Notes on Use of Bit Manipulation Instructions

The BSET, BCLR, BNOT, BST, and BIST instructions read a byte of data, modify a bit in the byte, then write the byte back. Care is required when these instructions are used to access registers with write-only bits, or to access ports.

The BCLR instruction can be used to clear flags in the on-chip registers. In an interrupt-handling routine, for example, if it is known that the flag is set to 1 , it is not necessary to read the flag ahead of time.

| Step | Description |  |
| :--- | :--- | :--- |
| 1 | Read | Read data (byte unit) at the specified address |
| 2 | Bit manipulation | Modify the specified bit in the read data |
| 3 | Write | Write the modified data (byte unit) to the specified address |

In the following example, a BCLR instruction is executed on the data direction register (DDR) of port 4.
$\mathrm{P}_{7}$ and $\mathrm{P} 4_{6}$ are set as input pins, and are inputting low-level and high-level signals, respectively.
$\mathrm{P} 4_{5}$ to $\mathrm{P} 4_{0}$ are set as output pins, and are in the low-level output state.
In this example, the BCLR instruction is used to make $\mathrm{P} 4_{0}$ an input port.

## Before Execution of BCLR Instruction

|  | $\mathbf{P 4}_{\mathbf{7}}$ | $\mathbf{P 4}_{6}$ | $\mathbf{P 4}_{\mathbf{5}}$ | $\mathbf{P 4}_{\mathbf{4}}$ | $\mathbf{P 4}_{\mathbf{3}}$ | $\mathbf{P 4}_{\mathbf{2}}$ | $\mathbf{P 4}_{\mathbf{1}}$ | $\mathbf{P 4}_{\mathbf{0}}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Input/output | Input | Input | Output | Output | Output | Output | Output | Output |
| DDR | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| DR | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

## Execution of BCLR Instruction

BCLR \#0, @P4DDR ; Execute BCLR instruction on DDR

## After Execution of BCLR Instruction

|  | $\mathbf{P 4}_{\mathbf{7}}$ | $\mathbf{P 4}_{6}$ | $\mathbf{P 4}_{\mathbf{5}}$ | $\mathbf{P 4}_{\mathbf{4}}$ | $\mathbf{P 4}_{\mathbf{3}}$ | $\mathbf{P 4}_{\mathbf{2}}$ | $\mathbf{P 4}_{\mathbf{1}}$ | $\mathbf{P 4}_{\mathbf{0}}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Input/output | Output | Output | Output | Output | Output | Output | Output | Input |
| DDR | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |
| DR | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

## Explanation of BCLR Instruction

To execute the BCLR instruction, the CPU begins by reading P4DDR. Since P4DDR is a writeonly register, it is read as $H^{\prime} F F$, even though its true value is $H^{\prime} 3 F$.

Next the CPU clears bit 0 of the read data, changing the value to H'FE.
Finally, the CPU writes this value (H'FE) back to DDR to complete the BCLR instruction.
As a result, $\mathrm{P} 4_{0} \mathrm{DDR}$ is cleared to 0 , making $\mathrm{P} 44_{0}$ an input pin. In addition, $\mathrm{P} 4{ }_{7} \mathrm{DDR}$ and $\mathrm{P} 44_{6} \mathrm{DDR}$ are set to 1 , making $\mathrm{P} 4_{7}$ and $\mathrm{P} 4_{6}$ output pins.

The BCLR instruction can be used to clear flags in the internal I/O registers to 0 . In an interrupthandling routine, for example, if it is known that the flag is set to 1 , it is not necessary to read the flag ahead of time.

### 2.7 Addressing Modes and Effective Address Calculation

### 2.7.1 Addressing Modes

The H8/300H CPU supports the eight addressing modes listed in table 2.11. Each instruction uses a subset of these addressing modes. Arithmetic and logic instructions can use the register direct and immediate modes. Data transfer instructions can use all addressing modes except programcounter relative and memory indirect. Bit manipulation instructions use register direct, register indirect, or absolute (@aa:8) addressing mode to specify an operand, and register direct (BSET, BCLR, BNOT, and BTST instructions) or immediate (3-bit) addressing mode to specify a bit number in the operand.

Table 2.11 Addressing Modes

| No. | Addressing Mode | Symbol |
| :--- | :--- | :--- |
| 1 | Register direct | Rn |
| 2 | Register indirect | @ERn |
| 3 | Register indirect with displacement | @(d:16, ERn)/@(d:24, ERn) |
| 4 | Register indirect with post-increment | @ERn+ |
|  | Register indirect with pre-decrement | @-ERn |
| 5 | Absolute address | @aa:8/@aa:16/@aa:24 |
| 6 | Immediate | \#xx:8/\#xx:16/\#xx:32 |
| 7 | Program-counter relative | @(d:8, PC)/@(d:16, PC) |
| 8 | Memory indirect | @@aa:8 |

Register Direct-Rn: The register field of the instruction code specifies an 8-, 16-, or 32-bit register containing the operand. R 0 H to R 7 H and R 0 L to R 7 L can be specified as 8 -bit registers. R0 to R7 and E0 to E7 can be specified as 16-bit registers. ER0 to ER7 can be specified as 32-bit registers.

Register Indirect—@ERn: The register field of the instruction code specifies an address register (ERn), the lower 24 bits of which contain the address of the operand.

Register Indirect with Displacement—@(d:16, ERn) or @(d:24, ERn): A 16-bit or 24-bit displacement contained in the instruction code is added to the contents of an address register (ERn) specified by the register field of the instruction, and the lower 24 bits of the sum specify the address of a memory operand. A 16-bit displacement is sign-extended when added.

## Register Indirect with Post-Increment or Pre-Decrement—@ERn+ or @-ERn:

- Register indirect with post-increment—@ERn+

The register field of the instruction code specifies an address register (ERn) the lower 24 bits of which contain the address of a memory operand. After the operand is accessed, 1,2 , or 4 is added to the address register contents ( 32 bits) and the sum is stored in the address register. The value added is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the register value should be even.

- Register indirect with pre-decrement—@-ERn

The value 1,2 , or 4 is subtracted from an address register (ERn) specified by the register field in the instruction code, and the lower 24 bits of the result become the address of a memory operand. The result is also stored in the address register. The value subtracted is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the resulting register value should be even.


#### Abstract

Absolute Address—@aa:8, @aa:16, or @aa:24: The instruction code contains the absolute address of a memory operand. The absolute address may be 8 bits long ( $@$ aa: 8 ), 16 bits long (@aa:16), or 24 bits long (@aa:24). For an 8 -bit absolute address, the upper 16 bits are all assumed to be 1 (H'FFFF). For a 16-bit absolute address the upper 8 bits are a sign extension. A 24 -bit absolute address can access the entire address space. Table 2.12 indicates the accessible address ranges.


Table 2.12 Absolute Address Access Ranges

| Absolute Address | 1-Mbyte Modes | 16-Mbyte Modes |
| :--- | :--- | :--- |
| 8 bits (@aa:8) | H'FFF00 to H'FFFFF | H'FFFF00 to H'FFFFFF |
|  | $(1048320$ to 1048575) | $(16776960$ to 16777215) |
| 16 bits (@aa:16) | H'00000 to H'07FFF,H'F8000 to H'FFFFF <br>  <br>  <br> (0 to 32767, 1015808 to 1048575) | H'000000 to H'007FFF, $^{\text {H'FF8000 to H'FFFFFF }}$(0 to 32767, 16744448 to 16777215) <br> 24 bits (@aa:24)H'00000 to H'FFFFF <br>  <br> (0 to 1048575) |

Immediate-\#xx:8, \#xx:16, or \#xx:32: The instruction code contains 8-bit (\#xx:8), 16-bit (\#xx:16), or 32-bit (\#xx:32) immediate data as an operand.

The instruction codes of the ADDS, SUBS, INC, and DEC instructions contain immediate data implicitly. The instruction codes of some bit manipulation instructions contain 3-bit immediate data specifying a bit number. The TRAPA instruction code contains 2-bit immediate data specifying a vector address.

Program-Counter Relative-@(d:8, PC) or @(d:16, PC): This mode is used in the Bcc and BSR instructions. An 8-bit or 16-bit displacement contained in the instruction code is signextended to 24 bits and added to the 24 -bit PC contents to generate a 24 -bit branch address. The PC value to which the displacement is added is the address of the first byte of the next instruction, so the possible branching range is -126 to +128 bytes ( -63 to +64 words) or -32766 to +32768 bytes ( -16383 to +16384 words) from the branch instruction. The resulting value should be an even number.

Memory Indirect—@ @aa:8: This mode can be used by the JMP and JSR instructions. The instruction code contains an 8-bit absolute address specifying a memory operand. This memory operand contains a branch address. The memory operand is accessed by longword access. The first byte of the memory operand is ignored, generating a 24 -bit branch address. See figure 2.9. The upper bits of the 8 -bit absolute address are assumed to be 0 ( $\mathrm{H}^{\prime} 0000$ ), so the address range is 0 to 255 ( $\mathrm{H}^{\prime} 000000$ to $\mathrm{H}^{\prime} 0000 \mathrm{FF}$ ). Note that the first part of this range is also the exception vector area. For further details see section 5, Interrupt Controller.


## Figure 2.9 Memory-Indirect Branch Address Specification

When a word-size or longword-size memory operand is specified, or when a branch address is specified, if the specified memory address is odd, the least significant bit is regarded as 0 . The accessed data or instruction code therefore begins at the preceding address. See section 2.5.2, Memory Data Formats.

### 2.7.2 Effective Address Calculation

Table 2.13 explains how an effective address is calculated in each addressing mode. In the 1-Mbyte operating modes the upper 4 bits of the calculated address are ignored in order to generate a 20 -bit effective address.

Table 2.13 Effective Address Calculation


3 Register indirect with displacement @(d:16, ERn)/@(d:24, ERn)


4 Register indirect with post-increment or pre-decrement
Register indirect with post-increment @ERn+


Register indirect with pre-decrement @-ERn


1 for a byte operand, 2 for a word operand, 4 for a longword operand

Absolute address
@aa:8

| op | abs |
| :--- | :--- |

@aa:16

@aa:24


6 Immediate
Operand is immediate
\#xx:8, \#xx:16, or \#xx:32 data

| op | IMM |
| :---: | :---: |

7 Program-counter relative
@(d:8, PC) or @(d:16, PC)


Addressing Mode and
No. Instruction Format
Memory indirect @@aa:8

- Normal mode

| op | abs |
| :--- | :--- |



Advanced mode

| op | abs |
| :---: | :---: |



Legend:
r, rm, rn: Register field
op: Operation field
disp: Displacement
IMM: Immediate data
abs: Absolute address

### 2.8 Processing States

### 2.8.1 Overview

The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU has five processing states: the program execution state, exception-handling state, power-down state, reset state, and bus-released state. The power-down state includes sleep mode, software standby mode, and hardware standby mode. Figure 2.10 classifies the processing states. Figure 2.12 indicates the state transitions.


Figure 2.10 Processing States

### 2.8.2 Program Execution State

In this state the CPU executes program instructions in normal sequence.

### 2.8.3 Exception-Handling State

The exception-handling state is a transient state that occurs when the CPU alters the normal program flow due to a reset, interrupt, or trap instruction. The CPU fetches a starting address from the exception vector table and branches to that address. In interrupt and trap exception handling the CPU references the stack pointer (ER7) and saves the program counter and condition code register.

Types of Exception Handling and Their Priority: Exception handling is performed for resets, interrupts, and trap instructions. Table 2.14 indicates the types of exception handling and their priority. Trap instruction exceptions are accepted at all times in the program execution state.

Table 2.14 Exception Handling Types and Priority

| Priority | Type of <br> Exception | Detection Timing | Start of Exception Handling |
| :--- | :--- | :--- | :--- |
| High | Reset | Synchronized with clock | Exception handling starts immediately <br> when RES changes from low to high |
|  | Interrupt | End of instruction <br> execution or end of <br> exception handling* | When an interrupt is requested, <br> exception handling starts at the end of <br> the current instruction or current <br> exception-handling sequence |
| Lrap instruction | When TRAPA instruction <br> is executed | Exception handling starts when a trap <br> (TRAPA) instruction is executed |  |

Note: * Interrupts are not detected at the end of the ANDC, ORC, XORC, and LDC instructions, or immediately after reset exception handling.

Figure 2.11 classifies the exception sources. For further details about exception sources, vector numbers, and vector addresses, see section 4, Exception Handling, and section 5, Interrupt Controller.


Figure 2.11 Classification of Exception Sources


Notes: 1. From any state except hardware standby mode, a transition to the reset state occurs whenever RES goes low.
2. From any state, a transition to hardware standby mode occurs when $\overline{\text { STBY }}$ goes low.

Figure 2.12 State Transitions

### 2.8.4 Exception-Handling Sequences

Reset Exception Handling: Reset exception handling has the highest priority. The reset state is entered when the $\overline{\mathrm{RES}}$ signal goes low. Reset exception handling starts after that, when $\overline{\mathrm{RES}}$ changes from low to high. When reset exception handling starts the CPU fetches a start address from the exception vector table and starts program execution from that address. All interrupts, including NMI, are disabled during the reset exception-handling sequence and immediately after it ends.

Interrupt Exception Handling and Trap Instruction Exception Handling: When these exception-handling sequences begin, the CPU references the stack pointer (ER7) and pushes the program counter and condition code register on the stack. Next, if the UE bit in the system control register (SYSCR) is set to 1 , the CPU sets the I bit in the condition code register to 1 . If the UE bit is cleared to 0 , the CPU sets both the I bit and the UI bit in the condition code register to 1 . Then the CPU fetches a start address from the exception vector table and execution branches to that address.

Figure 2.13 shows the stack after the exception-handling sequence.


Figure 2.13 Stack Structure after Exception Handling

### 2.8.5 Bus-Released State

In this state the bus is released to a bus master other than the CPU, in response to a bus request. The bus masters other than the CPU are the DMA controller, the refresh controller, and an external bus master. While the bus is released, the CPU halts except for internal operations. Interrupt requests are not accepted. For details see section 6.3.7, Bus Arbiter Operation.

### 2.8.6 Reset State

When the $\overline{\operatorname{RES}}$ input goes low all current processing stops and the CPU enters the reset state. The I bit in the condition code register is set to 1 by a reset. All interrupts are masked in the reset state. Reset exception handling starts when the $\overline{\mathrm{RES}}$ signal changes from low to high.

The reset state can also be entered by a watchdog timer overflow. For details see section 12, Watchdog Timer.

### 2.8.7 Power-Down State

In the power-down state the CPU stops operating to conserve power. There are three modes: sleep mode, software standby mode, and hardware standby mode.

Sleep Mode: A transition to sleep mode is made if the SLEEP instruction is executed while the SSBY bit is cleared to 0 in the system control register (SYSCR). CPU operations stop immediately after execution of the SLEEP instruction, but the contents of CPU registers are retained.

Software Standby Mode: A transition to software standby mode is made if the SLEEP instruction is executed while the SSBY bit is set to 1 in SYSCR. The CPU and clock halt and all on-chip supporting modules stop operating. The on-chip supporting modules are reset, but as long as a specified voltage is supplied the contents of CPU registers and on-chip RAM are retained. The I/O ports also remain in their existing states.

Hardware Standby Mode: A transition to hardware standby mode is made when the $\overline{\text { STBY }}$ input goes low. As in software standby mode, the CPU and all clocks halt and the on-chip supporting modules are reset, but as long as a specified voltage is supplied, on-chip RAM contents are retained.

For further information see section 20, Power-Down State.

### 2.9 Basic Operational Timing

### 2.9.1 Overview

The $\mathrm{H} 8 / 300 \mathrm{H}$ CPU operates according to the system clock $(\phi)$. The interval from one rise of the system clock to the next rise is referred to as a "state." A memory cycle or bus cycle consists of two or three states. The CPU uses different methods to access on-chip memory, the on-chip supporting modules, and the external address space. Access to the external address space can be controlled by the bus controller.

### 2.9.2 On-Chip Memory Access Timing

On-chip memory is accessed in two states. The data bus is 16 bits wide, permitting both byte and word access. Figure 2.14 shows the on-chip memory access cycle. Figure 2.15 indicates the pin states.


Figure 2.14 On-Chip Memory Access Cycle


## Figure 2.15 Pin States during On-Chip Memory Access

### 2.9.3 On-Chip Supporting Module Access Timing

The on-chip supporting modules are accessed in three states. The data bus is 8 or 16 bits wide, depending on the register being accessed. Figure 2.16 shows the on-chip supporting module access timing. Figure 2.17 indicates the pin states.


Figure 2.16 Access Cycle for On-Chip Supporting Modules
$\phi$

Address bus
$\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{HWR}}, \overline{\mathrm{LWR}}$
$D_{15}$ to $D_{0}$


Figure 2.17 Pin States during Access to On-Chip Supporting Modules

### 2.9.4 Access to External Address Space

The external address space is divided into eight areas (areas 0 to 7). Bus-controller settings determine whether each area is accessed via an 8-bit or 16-bit bus, and whether it is accessed in two or three states. For details see section 6, Bus Controller.

## Section 3 MCU Operating Modes

### 3.1 Overview

### 3.1.1 Operating Mode Selection

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ has seven operating modes (modes 1 to 7 ) that are selected by the mode pins ( $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ ) as indicated in table 3.1. The input at these pins determines the size of the address space and the initial bus mode.

Table 3.1 Operating Mode Selection

| Operating Mode* ${ }^{3}$ | Mode Pins |  |  | Description |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MD ${ }_{2}$ | MD ${ }_{1}$ | MD ${ }_{0}$ | Address Space | Initial Bus Mode* | $\begin{aligned} & \text { On-Chip } \\ & \text { ROM } \end{aligned}$ | On-Chip RAM |
| - | 0 | 0 | 0 | - | - | - | - |
| Mode 1 | 0 | 0 | 1 | Expanded mode | 8 bits | Disabled | Enabled*2 |
| Mode 2 | 0 | 1 | 0 | Expanded mode | 16 bits | Disabled | Enabled*2 |
| Mode 3 | 0 | 1 | 1 | Expanded mode | 8 bits | Disabled | Enabled*2 |
| Mode 4 | 1 | 0 | 0 | Expanded mode | 16 bits | Disabled | Enabled*2 |
| Mode 5 | 1 | 0 | 1 | Expanded mode | 8 bits | Enabled | Enabled*2 |
| Mode 6 | 1 | 1 | 0 | Expanded mode | 8 bits | Enabled | Enabled*2 |
| Mode 7 | 1 | 1 | 1 | Single-chip advanced mode | - | Enabled | Enabled |

Notes: 1. In modes 1 to 6 , an 8 -bit or 16 -bit data bus can be selected on a per-area basis by settings made in the area bus width control register (ABWCR). For details see section 6, Bus Controller.
2. If the RAME bit in SYSCR is cleared to 0 , these addresses become external addresses.
3. These are the operating modes when the FWE pin is at 0 . For the operating modes when the FWE pin is at 1 , see section 18 , ROM.

For the address space size there are two choices: 1 Mbyte or 16 Mbytes . The external data bus is either 8 or 16 bits wide depending on ABWCR settings. If 8 -bit access is selected for all areas, the external data bus is 8 bits wide. For details see section 6, Bus Controller.

Modes 1 to 4 are externally expanded modes that enable access to external memory and peripheral devices and disable access to the on-chip ROM. Modes 1 and 2 support a maximum address space of 1 Mbyte. Modes 3 and 4 support a maximum address space of 16 Mbytes.

Modes 5 and 6 are externally expanded modes that enable access to external memory and peripheral devices and also enable access to the on-chip ROM. Mode 5 supports a maximum address space of 1 Mbyte. Mode 6 supports a maximum address space of 16 Mbytes.

Mode 7 is a single-chip mode that operates using the on-chip ROM, RAM, and Internal I/O registers, and makes all I/O ports available. Mode 7 supports a 1-Mbyte address space.

The H8/3052F can be used only in modes 1 to 7 . The inputs at the mode pins must select one of these seven modes. The inputs at the mode pins must not be changed during operation.

### 3.1.2 Register Configuration

The H8/3052F has a mode control register (MDCR) that indicates the inputs at the mode pins $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$, and a system control register (SYSCR). Table 3.2 summarizes these registers.

Table 3.2 Registers

| Address* | Name | Abbreviation | R/w | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFF1 | Mode control register | MDCR | R | Undetermined |
| H'FFF2 | System control register | SYSCR | R/W | H'OB |

Note: * The lower 16 bits of the address are indicated.

### 3.2 Mode Control Register (MDCR)

MDCR is an 8-bit read-only register that indicates the current operating mode of the $\mathrm{H} 8 / 3052 \mathrm{~F}$.


Note: * Determined by pins $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$.

Bits 7 and 6—Reserved: Read-only bits, always read as 1.
Bits 5 to 3—Reserved: Read-only bits, always read as 0.

Bits 2 to 0—Mode Select 2 to 0 (MDS2 to MDS0): These bits indicate the logic levels at pins $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ (the current operating mode). MDS2 to MDS0 correspond to $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$. MDS 2 to MDS0 are read-only bits. The mode pin $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$ levels are latched into these bits when MDCR is read.

### 3.3 System Control Register (SYSCR)

SYSCR is an 8-bit register that controls the operation of the H8/3052F.


Bit 7—Software Standby (SSBY): Enables transition to software standby mode. (For further information about software standby mode see section 20, Power-Down State.)

When software standby mode is exited by an external interrupt, this bit remains set to 1 . To clear this bit, write 0 .

## Bit 7: SSBY Description

| 0 | SLEEP instruction causes transition to sleep mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | SLEEP instruction causes transition to software standby mode |  |

Bits 6 to 4—Standby Timer Select (STS2 to STS0): These bits select the length of time the CPU and on-chip supporting modules wait for the internal clock oscillator to settle when software standby mode is exited by an external interrupt. When using a crystal oscillator, set these bits so that the waiting time will be at least 7 ms at the system clock rate. For further information about waiting time selection, see section 20.4.3, Selection of Waiting Time for Exit from Software Standby Mode.

| Bit 6: STS2 | Bit 5: STS1 | Bit 4: STS0 | Description |  |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Waiting time $=8,192$ states | (Initial value) |
|  |  | 1 | Waiting time $=16,384$ states |  |
| 1 | 0 | Waiting time $=32,768$ states |  |  |
| 1 | 0 | 0 | Waiting time $=65,536$ states |  |
|  |  | 1 | Waiting time $=131,072$ states |  |

Bit 3-User Bit Enable (UE): Selects whether to use the UI bit in the condition code register as a user bit or an interrupt mask bit.

| Bit 3: UE | Description |  |
| :--- | :--- | :--- |
| 0 | Ul bit in CCR is used as an interrupt mask bit |  |
| 1 | UI bit in CCR is used as a user bit | (Initial value) |

Bit 2—NMI Edge Select (NMIEG): Selects the valid edge of the NMI input.
Bit 2: NMIEG Description

| 0 | An interrupt is requested at the falling edge of NMI | (Initial value) |
| :--- | :--- | :--- |
| 1 | An interrupt is requested at the rising edge of NMI |  |

Bit 1—Reserved: Read-only bit, always read as 1 .
Bit 0—RAM Enable (RAME): Enables or disables the on-chip RAM. The RAME bit is initialized by the rising edge of the $\overline{\mathrm{RES}}$ signal. It is not initialized in software standby mode.
Bit 0: RAME Description

| 0 | On-chip RAM is disabled |  |
| :--- | :--- | :--- |
| 1 | On-chip RAM is enabled | (Initial value) |

### 3.4 Operating Mode Descriptions

### 3.4.1 Mode 1

Ports 1,2 , and 5 function as address pins $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$, permitting access to a maximum 1-Mbyte address space. The initial bus mode after a reset is 8 bits, with 8 -bit access to all areas. If at least one area is designated for 16 -bit access in ABWCR, the bus mode switches to 16 bits.

### 3.4.2 Mode 2

Ports 1,2 , and 5 function as address pins $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$, permitting access to a maximum 1-Mbyte address space. The initial bus mode after a reset is 16 bits, with 16 -bit access to all areas. If all areas are designated for 8 -bit access in ABWCR, the bus mode switches to 8 bits.

### 3.4.3 Mode 3

Ports 1, 2, and 5 and part of port A function as address pins $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$, permitting access to a maximum 16-Mbyte address space. The initial bus mode after a reset is 8 bits, with 8 -bit access to all areas. If at least one area is designated for 16 -bit access in ABWCR, the bus mode switches to 16 bits. $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ are valid when 0 is written in bits 7 to 5 of the bus release control register (BRCR). (In this mode $\mathrm{A}_{20}$ is always used for address output.)

### 3.4.4 Mode 4

Ports 1, 2, and 5 and part of port A function as address pins $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$, permitting access to a maximum 16-Mbyte address space. The initial bus mode after a reset is 16 bits, with 16 -bit access to all areas. If all areas are designated for 8-bit access in ABWCR, the bus mode switches to 8 bits. $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ are valid when 0 is written in bits 7 to 5 of BRCR. (In this mode $\mathrm{A}_{20}$ is always used for address output.)

### 3.4.5 Mode 5

Ports 1, 2, and 5 can function as address pins $\mathrm{A}_{19}$ to $\mathrm{A}_{0}$, permitting access to a maximum 1-Mbyte address space, but following a reset they are input ports. To use ports 1,2 , and 5 as an address bus, the corresponding bits in their data direction registers (P1DDR, P2DDR, and P5DDR) must be set to 1 . The initial bus mode after a reset is 8 bits, with 8 -bit access to all areas. If at least one area is designated for 16 -bit access in ABWCR, the bus mode switches to 16 bits.

Ports 1, 2, and 5 and part of port A function as address pins $\mathrm{A}_{23}$ to $\mathrm{A}_{0}$, permitting access to a maximum 16-Mbyte address space, but following a reset they are input ports. To use ports 1, 2, and 5 as an address bus, the corresponding bits in their data direction registers (P1DDR, P2DDR, and P5DDR) must be set to 1 . For $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ output, clear bits 7 to 5 of BRCR to 0 . (In this mode $\mathrm{A}_{20}$ is always used for address output.)

The initial bus mode after a reset is 8 bits, with 8 -bit access to all areas. If at least one area is designated for 16 -bit access in ABWCR, the bus mode switches to 16 bits.

### 3.4.7 Mode 7

This mode operates using the on-chip ROM, RAM, and registers. All I/O ports are available.
Mode 7 supports a 1-Mbyte address space.

### 3.5 Pin Functions in Each Operating Mode

The pin functions of ports 1 to 5 and port A vary depending on the operating mode. Table 3.3 indicates their functions in each operating mode.

## Table 3.3 Pin Functions in Each Mode

| Port | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Port 1 | $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ | $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ | $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ | $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ | P 17 to $\mathrm{P} 1_{0}{ }^{\text {*2 }}$ | P 17 to $\mathrm{P} 1_{0}{ }^{* 2}$ | $\mathrm{P} 1_{7}$ to $\mathrm{P} 1_{0}$ |
| Port 2 | $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ | $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ | $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ | $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ | P 2 , to $\mathrm{P} 2_{0}{ }^{* 2}$ | $\mathrm{P} 2_{7}$ to $\mathrm{P}_{0}{ }^{* 2}$ | P 2 7 to $\mathrm{P} 2_{0}$ |
| Port 3 | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ | P 3 , to $\mathrm{P} 3_{0}$ |
| Port 4 | P 47 to $\mathrm{P} 4_{0}{ }^{* 1}$ | $\mathrm{D}_{7}$ to $\mathrm{D}_{0}{ }^{* 1}$ | P 47 to $\mathrm{P} 4_{0}{ }^{* 1}$ | $\mathrm{D}_{7}$ to $\mathrm{D}_{0}{ }^{* 1}$ | P 47 to $\mathrm{P} 4{ }_{0}{ }^{* 1}$ | P 47 to $\mathrm{P} 4_{0}{ }^{* 1}$ | P 47 to P 40 |
| Port 5 | $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ | $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ | $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ | $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ | $\mathrm{P5}_{3}$ to $\mathrm{P} 50{ }^{* 2}$ | $\mathrm{P5}_{3}$ to $\mathrm{P5}{ }_{0}^{* 2}$ | $\mathrm{P}_{3}$ to P 50 |
| Port A | $\mathrm{PA}_{7}$ to $\mathrm{PA}_{4}$ | $\mathrm{PA}_{7}$ to $\mathrm{PA}_{4}$ | $\begin{aligned} & \mathrm{PA}_{7} \text { to } \mathrm{PA}_{5}^{* 3}, \\ & \mathrm{~A}_{20} \end{aligned}$ | $\begin{aligned} & \mathrm{PA}_{7} \text { to } \mathrm{PA}_{5}^{* 3}, \\ & \mathrm{~A}_{20} \end{aligned}$ | $\mathrm{PA}_{7}$ to $\mathrm{PA}_{4}$ | $\begin{aligned} & \mathrm{PA}_{7} \text { to } \mathrm{PA}_{5}, \\ & \mathrm{~A}_{20}{ }^{* 3} \end{aligned}$ | $\mathrm{PA}_{7}$ to $\mathrm{PA}_{4}$ |

Notes: 1. Initial state. The bus mode can be switched by settings in ABWCR. These pins function as $\mathrm{P} 4_{7}$ to $\mathrm{P} 4_{0}$ in 8 -bit bus mode, and as $\mathrm{D}_{7}$ to $\mathrm{D}_{0}$ in 16 -bit bus mode.
2. Initial state. These pins become address output pins when the corresponding bits in the data direction registers (P1DDR, P2DDR, P5DDR) are set to 1.
3. Initial state. $\mathrm{A}_{20}$ is always an address output pin. $\mathrm{PA}_{7}$ to $\mathrm{PA}_{5}$ are switched over to $\mathrm{A}_{23}$ to $A_{21}$ output by writing 0 in bits 7 to 5 of BRCR.

### 3.6 Memory Map in Each Operating Mode

Figure 3.1 shows a memory map of the $\mathrm{H} 8 / 3052 \mathrm{~F}$. The address space is divided into eight areas.
The initial bus mode differs between modes 1 and 2 , and also between modes 3 and 4 .
The address locations of the on-chip RAM and on-chip registers differ between the 1-Mbyte modes (modes $1,2,5$, and 7 ) and 16 -Mbyte modes (modes 3,4 , and 6 ). The address range specifiable by the CPU in the 8 - and 16 -bit absolute addressing modes ( $@ a a: 8$ and @aa:16) also differs.

Modes 1 and 2 (1-Mbyte expanded modes with on-chip ROM disabled)


Modes 3 and 4
(16-Mbyte expanded modes with on-chip ROM disabled)


Note: * External addresses can be accessed by disabling on-chip RAM.

Figure 3.1 H8/3052F Memory Map in Each Operating Mode

Rev. 2.0, 03/01, page 66 of 822
(16-Mbyte expanded mode with on-chip ROM enabled)


Mode 7 (single-chip advanced mode)


Note: * External addresses can be accessed by disabling on-chip RAM.

Figure 3.1 H8/3052F Memory Map in Each Operating Mode (cont)

## Section 4 Exception Handling

### 4.1 Overview

### 4.1.1 Exception Handling Types and Priority

As table 4.1 indicates, exception handling may be caused by a reset, trap instruction, or interrupt. Exception handling is prioritized as shown in table 4.1. If two or more exceptions occur simultaneously, they are accepted and processed in priority order. Trap instruction exceptions are accepted at all times in the program execution state.

Table 4.1 Exception Types and Priority

| Priority | Exception Type | Start of Exception Handling |
| :--- | :--- | :--- |
| High | Reset | $\frac{\text { Starts immediately after a low-to-high transition at the }}{\text { RES pin }}$ |

### 4.1.2 Exception Handling Operation

Exceptions originate from various sources. Trap instructions and interrupts are handled as follows.

1. The program counter (PC) and condition code register (CCR) are pushed onto the stack.
2. The CCR interrupt mask bit is set to 1 .
3. A vector address corresponding to the exception source is generated, and program execution starts from the address indicated in that address.

For a reset exception, steps 2 and 3 above are carried out.

### 4.1.3 Exception Sources and Vector Table

The exception sources are classified as shown in figure 4.1. Different vectors are assigned to different exception sources. Table 4.2 lists the exception sources and their vector addresses.


Figure 4.1 Exception Sources

Table 4.2 Exception Vector Table

| Exception Source | Vector Number | Vector Address*1 |
| :---: | :---: | :---: |
| Reset | 0 | H'0000 to H'0003 |
| Reserved for system use | 1 | H'0004 to H'0007 |
|  | 2 | $\mathrm{H}^{\prime} 0008$ to H'000B |
|  | 3 | H'000C to H'000F |
|  | 4 | H'0010 to H'0013 |
|  | 5 | H'0014 to H'0017 |
|  | 6 | H'0018 to H'001B |
| External interrupt (NMI) | 7 | H'001C to H'001F |
| Trap instruction (4 sources) | 8 | H'0020 to H'0023 |
|  | 9 | H'0024 to H'0027 |
|  | 10 | $\mathrm{H}^{\prime} 0028$ to H'002B |
|  | 11 | H'002C to H'002F |
| External interrupt $\mathrm{IRQ}_{0}$ | 12 | $\mathrm{H}^{\prime} 0030$ to H'0033 |
| External interrupt $\mathrm{IRQ}_{1}$ | 13 | H'0034 to H'0037 |
| External interrupt $\mathrm{IRQ}_{2}$ | 14 | H'0038 to H'003B |
| External interrupt $\mathrm{IRQ}_{3}$ | 15 | H'003C to H'003F |
| External interrupt $\mathrm{IRQ}_{4}$ | 16 | H'0040 to H'0043 |
| External interrupt $\mathrm{IRQ}_{5}$ | 17 | H'0044 to H'0047 |
| Reserved for system use | 18 | $\mathrm{H}^{\prime} 0048$ to H'004B |
|  | 19 | $\mathrm{H}^{\prime} 004 \mathrm{C}$ to H'004F |
| Internal interrupts*2 | $\begin{aligned} & 20 \\ & \text { to } \\ & 60 \end{aligned}$ | H'0050 to H'0053 to H'OOFO to H'OOF3 |

Notes: 1. Lower 16 bits of the address.
2. For the internal interrupt vectors, see section 5.3.3, Interrupt Vector Table.

### 4.2 Reset

### 4.2.1 Overview

A reset is the highest-priority exception. When the $\overline{\mathrm{RES}}$ pin goes low, all processing halts and the chip enters the reset state. A reset initializes the internal state of the CPU and the registers of the on-chip supporting modules. Reset exception handling begins when the $\overline{\mathrm{RES}}$ pin changes from low to high.

The chip can also be reset by overflow of the watchdog timer. For details see section 12, Watchdog Timer.

### 4.2.2 Reset Sequence

The chip enters the reset state when the $\overline{\operatorname{RES}}$ pin goes low.
To ensure that the chip is reset, hold the $\overline{\operatorname{RES}}$ pin low for at least 20 ms at power-up. To reset the chip during operation, hold the $\overline{\text { RES }}$ pin low for at least 20 system clock ( $\phi$ ) cycles. See appendix D.2, Pin States at Reset, for the states of the pins in the reset state.

When the $\overline{\text { RES }}$ pin goes high after being held low for the necessary time, the chip starts reset exception handling as follows.

- The internal state of the CPU and the registers of the on-chip supporting modules are initialized, and the I bit is set to 1 in CCR.
- The contents of the reset vector address ( $\mathrm{H}^{\prime} 0000$ to $\mathrm{H}^{\prime} 0003$ ) are read, and program execution starts from the address indicated in the vector address.

Figure 4.2 shows the reset sequence in modes 1 and 3 . Figure 4.3 shows the reset sequence in modes 2 and 4 . Figure 4.4 shows the reset sequence in modes 5 to 7 .


Note: After a reset, the wait-state controller inserts three wait states in every bus cycle.

(1), (3) Address of reset vector: (1) = H'000000, (3) = H'000002
(2), (4) Start address (contents of reset vector)
(5) Start address
(6) First instruction of program

Note: After a reset, the wait-state controller inserts three wait states in every bus cycle.
Figure 4.3 Reset Sequence (Modes 2 and 4)


Figure 4.4 Reset Sequence (Modes 5 to 7)

### 4.2.3 Interrupts after Reset

If an interrupt is accepted after a reset but before the stack pointer (SP) is initialized, PC and CCR will not be saved correctly, leading to a program crash. To prevent this, all interrupt requests, including NMI, are disabled immediately after a reset. The first instruction of the program is always executed immediately after the reset state ends. This instruction should initialize the stack pointer (example: MOV.L \#xx:32, SP).

### 4.3 Interrupts

Interrupt exception handling can be requested by seven external sources (NMI, $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ ) and 30 internal sources in the on-chip supporting modules. Figure 4.5 classifies the interrupt sources and indicates the number of interrupts of each type.

The on-chip supporting modules that can request interrupts are the watchdog timer (WDT), refresh controller, 16 -bit integrated timer unit (ITU), DMA controller (DMAC), serial communication interface (SCI), and A/D converter. Each interrupt source has a separate vector address.

NMI is the highest-priority interrupt and is always accepted. Interrupts are controlled by the interrupt controller. The interrupt controller can assign interrupts other than NMI to two priority levels, and arbitrate between simultaneous interrupts. Interrupt priorities are assigned in interrupt priority registers A and B (IPRA and IPRB) in the interrupt controller.

For details on interrupts see section 5, Interrupt Controller.
Interrupts $\left\{\begin{array}{l}\text { External interrupts }\left\{\begin{array}{l}\mathrm{NMI}(1) \\ \mathrm{IRQ} \text { to } \mathrm{IRQ}_{5}(6)\end{array}\right. \\ \text { Internal interrupts }\left\{\begin{array}{l}\mathrm{WDT}^{* 1}(1) \\ \operatorname{Refresh} \text { controller *2 (1) } \\ \operatorname{ITU}(15) \\ \mathrm{DMAC}(4) \\ \mathrm{SCI}(8) \\ \text { A/D converter (1) }\end{array}\right.\end{array}\right.$

Notes: Numbers in parentheses are the number of interrupt sources.

1. When the watchdog timer is used as an interval timer, it generates an interrupt request at every counter overflow.
2. When the refresh controller is used as an interval timer, it generates an interrupt request at compare match.

Figure 4.5 Interrupt Sources and Number of Interrupts

### 4.4 Trap Instruction

Trap instruction exception handling starts when a TRAPA instruction is executed. If the UE bit is set to 1 in the system control register (SYSCR), the exception handling sequence sets the I bit to 1 in CCR. If the UE bit is 0 , the I and UI bits are both set to 1 . The TRAPA instruction fetches a start address from a vector table entry corresponding to a vector number from 0 to 3 , which is specified in the instruction code.

### 4.5 Stack Status after Exception Handling

Figure 4.6 shows the stack after completion of trap instruction exception handling and interrupt exception handling.


Figure 4.6 Stack after Completion of Exception Handling

### 4.6 Notes on Use of the Stack

When accessing word data or longword data, the $\mathrm{H} 8 / 3052 \mathrm{~F}$ regards the lowest address bit as 0 . The stack should always be accessed by word access or longword access, and the value of the stack pointer (SP, ER7) should always be kept even. Use the following instructions to save registers:

PUSH.W Rn (or MOV.W Rn, @-SP)
PUSH.L ERn (or MOV.L ERn, @-SP)
Use the following instructions to restore registers:
POP.W Rn (or MOV.W @SP+, Rn)
POP.L ERn (or MOV.L @SP+, ERn)
Setting SP to an odd value may lead to a malfunction. Figure 4.7 shows an example of what happens when the SP value is odd.


Legend
CCR: Condition code register
PC: Program counter
R1L: General register R1L
SP: Stack pointer
Note: The diagram illustrates modes 3 and 4.

Figure 4.7 Operation when SP Value is Odd

### 5.1 Overview

### 5.1.1 Features

The interrupt controller has the following features:

- Interrupt priority registers (IPRs) for setting interrupt priorities

Interrupts other than NMI can be assigned to two priority levels on a source-by-source or module-by-module basis in interrupt priority registers A and B (IPRA and IPRB).

- Three-level masking by the I and UI bits in the CPU condition code register (CCR)
- Independent vector addresses

All interrupts are independently vectored; the interrupt service routine does not have to identify the interrupt source.

- Seven external interrupt pins

NMI has the highest priority and is always accepted; either the rising or falling edge can be selected. For each of $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$, falling edge or level sensing can be selected independently.

### 5.1.2 Block Diagram

Figure 5.1 shows a block diagram of the interrupt controller.


Legend
ISCR: IRQ sense control register
IER: IRQ enable register
ISR: IRQ status register
IPRA: Interrupt priority register A
IPRB: Interrupt priority register B
SYSCR: System control register

Figure 5.1 Interrupt Controller Block Diagram

### 5.1.3 Pin Configuration

Table 5.1 lists the interrupt pins.
Table 5.1 Interrupt Pins

| Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Nonmaskable interrupt | NMI | Input | Nonmaskable external interrupt, rising <br> edge or falling edge selectable |
| External interrupt request <br> 5 to 0 | $\overline{\mathrm{RQ}}_{5}$ to $\overline{\mathrm{IRQ}}_{0}$ | Input | Maskable external interrupts, falling <br> edge or level sensing selectable |

### 5.1.4 Register Configuration

Table 5.2 lists the registers of the interrupt controller.
Table 5.2 Interrupt Controller Registers

| Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: |
| H'FFF2 | System control register | SYSCR | R/W | H'0B |
| H'FFF4 | IRQ sense control register | ISCR | R/W | H'00 |
| H'FFF5 | IRQ enable register | IER | R/W | H'00 |
| H'FFF6 | IRQ status register | ISR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'00 |
| H'FFF8 | Interrupt priority register A | IPRA | R/W | H'00 |
| H'FFF9 | Interrupt priority register B | IPRB | R/W | H'00 |

Notes: 1. Lower 16 bits of the address.
2. Only 0 can be written, to clear flags.

### 5.2 Register Descriptions

### 5.2.1 System Control Register (SYSCR)

SYSCR is an 8-bit readable/writable register that controls software standby mode, selects the action of the UI bit in CCR, selects the NMI edge, and enables or disables the on-chip RAM.

Only bits 3 and 2 are described here. For the other bits, see section 3.3, System Control Register (SYSCR).

SYSCR is initialized to $\mathrm{H}^{\prime} 0 \mathrm{~B}$ by a reset and in hardware standby mode. It is not initialized in software standby mode.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SSBY | STS2 | STS1 | STS0 | UE | NMIEG | - | RAME |
| Initial value | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | - | R/W |
|  |  |  |  |  |  |  |  | M enable |
|  |  |  |  |  |  |  | rved |  |
|  |  |  | dby ti ct 2 to |  |  | MI edge elects the | ct <br> Il inp | edge |
|  | Softw | stand |  |  | bit | able |  |  |
|  |  |  |  |  | cts w | ther to us | he U |  |
|  |  |  |  |  | as a | er bit or | rrupt | sk bit |

Bit 3-User Bit Enable (UE): Selects whether to use the UI bit in CCR as a user bit or an interrupt mask bit.

| Bit 3: UE | Description |  |
| :--- | :--- | ---: |
| 0 | UI bit in CCR is used as interrupt mask bit |  |
| 1 | UI bit in CCR is used as user bit | (Initial value) |

Bit 2-NMI Edge Select (NMIEG): Selects the NMI input edge.

| Bit 2: NMIEG | Description |  |
| :--- | :--- | :--- |
| 0 | Interrupt is requested at falling edge of NMI input | (Initial value) |
| 1 | Interrupt is requested at rising edge of NMI input |  |

### 5.2.2 Interrupt Priority Registers A and B (IPRA, IPRB)

IPRA and IPRB are 8-bit readable/writable registers that control interrupt priority.

Interrupt Priority Register A (IPRA): IPRA is an 8-bit readable/writable register in which interrupt priority levels can be set.

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IPRA7 | IPRA6 | IPRA5 | IPRA4 | IPRA3 | IPRA2 | IPRA1 | IPRA0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  |  |  |  | Priority level AO Selects the priority level of ITU channel 2 interrupt requests |

Priority level A1 Selects the priority level of ITU channel 1 interrupt requests

Priority level A2
Selects the priority level of ITU channel 0 interrupt requests

Priority level A3
Selects the priority level of WDT and refresh controller interrupt requests

Priority level A4
Selects the priority level of $\mathrm{IRQ}_{4}$ and $\mathrm{IRQ}_{5}$ interrupt requests

Priority level A5
Selects the priority level of $I \mathrm{RQ}_{2}$ and $I \mathrm{RQ}_{3}$ interrupt requests

## Priority level A6

Selects the priority level of $\mathrm{IRQ}_{1}$ interrupt requests

## Priority level A7

Selects the priority level of $\mathrm{IRQ}_{0}$ interrupt requests

IPRA is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode.

Bit 7—Priority Level A7 (IPRA7): Selects the priority level of $\mathrm{IRQ}_{0}$ interrupt requests.

| Bit 7: IPRA7 | Description |  |
| :--- | :--- | :--- |
| 0 | $\mathrm{IRQ}_{0}$ interrupt requests have priority level 0 (Non-priority) | (Initial value) |
| 1 | $\mathrm{IRQ}_{0}$ interrupt requests have priority level 1 (Priority) |  |

Bit 6—Priority Level A6 (IPRA6): Selects the priority level of $\mathrm{IRQ}_{1}$ interrupt requests.

| Bit 6: IPRA6 | Description |  |
| :--- | :--- | :--- |
| 0 | $\mathrm{IRQ}_{1}$ interrupt requests have priority level 0 (Non-priority) | (Initial value) |
| 1 | $\mathrm{IRQ}_{1}$ interrupt requests have priority level 1 (Priority) |  |

Bit 5—Priority Level A5 (IPRA5): Selects the priority level of $I R Q_{2}$ and $I R Q_{3}$ interrupt requests.

| Bit 5: IPRA5 | Description |
| :--- | :--- |
| 0 | $\mathrm{IRQ}_{2}$ and $I \mathrm{RQ}_{3}$ interrupt requests have priority level 0 (Non-priority) |
| 1 | $\mathrm{IRQ}_{2}$ and $I R Q_{3}$ interrupt requests have priority level 1 (Priority) |

Bit 4—Priority Level A4 (IPRA4): Selects the priority level of $I R Q_{4}$ and $I R Q_{5}$ interrupt requests.
Bit 4: IPRA4 Description
$0 \quad \mathrm{IRQ}_{4}$ and $\mathrm{IRQ}_{5}$ interrupt requests have priority level 0 (Non-priority)
(Initial value)
$1 \quad \mathrm{IRQ}_{4}$ and $\mathrm{IRQ}_{5}$ interrupt requests have priority level 1 (Priority)

Bit 3—Priority Level A3 (IPRA3): Selects the priority level of WDT and refresh controller interrupt requests.

| Bit 3: IPRA3 | Description |
| :--- | :--- |
| 0 | WDT and refresh controller interrupt requests have priority level 0 <br> (Non-priority) |
| 1 | WDT and refresh controller interrupt requests have priority level 1 (Priority) |

Bit 2—Priority Level A2 (IPRA2): Selects the priority level of ITU channel 0 interrupt requests.
Bit 2: IPRA2 Description

ITU channel 0 interrupt requests have priority level 0 (Non-priority)
(Initial value)
ITU channel 0 interrupt requests have priority level 1 (Priority)

Bit 1—Priority Level A1 (IPRA1): Selects the priority level of ITU channel 1 interrupt requests.

| Bit 1: IPRA1 | Description |
| :--- | :--- |
| 0 | ITU channel 1 interrupt requests have priority level 0 (Non-priority) |
| 1 | ITU channel 1 interrupt requests have priority level 1 (Priority) |

Bit 0—Priority Level A0 (IPRA0): Selects the priority level of ITU channel 2 interrupt requests.

| Bit 0: IPRA0 | Description |
| :--- | :--- |
| 0 | ITU channel 2 interrupt requests have priority level 0 (Non-priority) |
| 1 | ITU channel 2 interrupt requests have priority level 1 (Priority) |

Interrupt Priority Register B (IPRB): IPRB is an 8-bit readable/writable register in which interrupt priority levels can be set.


IPRB is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode.

Bit 7—Priority Level B7 (IPRB7): Selects the priority level of ITU channel 3 interrupt requests.

## Bit 7: IPRB7 Description

| 0 | ITU channel 3 interrupt requests have priority level 0 (low priority) (Initial value) |
| :--- | :--- |
| 1 | ITU channel 3 interrupt requests have priority level 1 (high priority) |

Bit 6—Priority Level B6 (IPRB6): Selects the priority level of ITU channel 4 interrupt requests.
Bit 6: IPRB6 Description

| 0 | ITU channel 4 interrupt requests have priority level 0 (low priority) (Initial value) |
| :--- | :--- |
| 1 | ITU channel 4 interrupt requests have priority level 1 (high priority) |

Bit 5—Priority Level B5 (IPRB5): Selects the priority level of DMAC interrupt requests (channels 0 and 1).

| Bit 5: IPRB5 | Description |
| :--- | :--- |
| 0 | DMAC interrupt requests (channels 0 and 1) have priority level 0 (low priority) |
|  |  |
| 1 | DMAC interrupt requests (channels 0 and 1) have priority level 1 (high priority) |

Bit 4—Reserved: This bit can be written and read, but it does not affect interrupt priority.

Bit 3-Priority Level B3 (IPRB3): Selects the priority level of SCI channel 0 interrupt requests.
Bit 3: IPRB3 Description

| 0 | SClO interrupt requests have priority level 0 (low priority) | (Initial value) |
| :--- | :--- | :--- |
| 1 | SCl0 interrupt requests have priority level 1 (high priority) |  |

Bit 2—Priority Level B2 (IPRB2): Selects the priority level of SCI channel 1 interrupt requests.

| Bit 2: IPRB2 | Description |  |
| :--- | :--- | :--- |
| 0 | SCl1 interrupt requests have priority level 0 (low priority) | (Initial value) |
| 1 | SCI1 interrupt requests have priority level 1 (high priority) |  |

Bit 1—Priority Level B1 (IPRB1): Selects the priority level of A/D converter interrupt requests.

| Bit 1: IPRB1 | Description |
| :--- | :--- |
| 0 | A/D converter interrupt requests have priority level 0 (low priority) (Initial value) |
| 1 | A/D converter interrupt requests have priority level 1 (high priority) |

Bit 0—Reserved: This bit can be written and read, but it does not affect interrupt priority.

### 5.2.3 IRQ Status Register (ISR)

ISR is an 8-bit readable/writable register that indicates the status of $\operatorname{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ interrupt requests.


Note: * Only 0 can be written, to clear flags.

ISR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode.
Bits 7 and 6-Reserved: Read-only bits, always read as 0 .
Bits 5 to $0-I_{2}$ to IRQ $_{0}$ Flags (IRQ5F to IRQ0F): These bits indicate the status of $I R Q_{5}$ to $I R Q_{0}$ interrupt requests.

## Bits 5 to 0:

IRQ5F to IRQ0F Description
[Clearing conditions]
0 is written in $I R Q n F$ after reading the $I R Q n F$ flag when $I R Q n F=1$.
$\operatorname{IRQnSC}=0, \overline{\mathrm{IRQn}}$ input is high, and interrupt exception handling is carried out.
IRQnSC = 1 and IRQn interrupt exception handling is carried out.
1
[Setting conditions]
IRQnSC $=0$ and $\overline{\mathrm{IRQn}}$ input is low.
$\mathrm{IRQnSC}=1$ and a falling edge occurs in $\overline{\mathrm{IRQn}}$ input
Note: $\mathrm{n}=5$ to 0

Rev. 2.0, 03/01, page 90 of 822

### 5.2.4 IRQ Enable Register (IER)

IER is an 8-bit readable/writable register that enables or disables $\operatorname{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ interrupt requests.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | IRQ5E | IRQ4E | IRQ3E | IRQ2E | IRQ1E | IRQ0E |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Reserved bits |  |  | $I R Q_{5}$ to $I R Q_{0}$ enable |  |  |  |  |  |

IER is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode.

Bits 7 and 6-Reserved: These bits can be written and read, but they do not enable or disable interrupts.

Bits 5 to 0—IRQ to IRQ $_{0}$ Enable (IRQ5E to IRQ0E): These bits enable or disable $\mathrm{IRQ}_{5}$ to IRQ 0 interrupts.

Bits 5 to 0 :
IRQ5E to IRQ0E Description

| 0 | $\mathrm{IRQ}_{5}$ to $\mathrm{IRQ}_{0}$ interrupts are disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | $\mathrm{IRQ}_{5}$ to $\mathrm{IRQ}_{0}$ interrupts are enabled |  |

### 5.2.5 IRQ Sense Control Register (ISCR)

ISCR is an 8-bit readable/writable register that selects level sensing or falling-edge sensing of the inputs at pins $\overline{\mathrm{IRQ}}_{5}$ to $\overline{\mathrm{IRQ}}_{0}$.


ISCR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode.
Bits 7 and 6-Reserved: These bits can be written and read, but they do not select level or falling-edge sensing.

Bits 5 to 0-IRQ ${ }_{5}$ to IRQ $_{0}$ Sense Control (IRQ5SC to IRQ0SC): These bits select whether interrupts $\operatorname{IRQ}_{5}$ to $\mathrm{IRQ}_{0}$ are requested by level sensing of pins $\overline{\mathrm{IRQ}}_{5}$ to $\overline{\mathrm{IRQ}}_{0}$, or by falling-edge sensing.

Bits 5 to 0 :

## Description

IRQ5SC to IRQ0SC

| 0 | Interrupts are requested when $\overline{\mathrm{TQ}}_{5}$ to $\overline{\mathrm{RQ}}_{0}$ inputs are low |
| :--- | :--- |
| 1 | Interrupts are requested by falling-edge input at $\overline{\mathrm{TRQ}}_{5}$ to $\overline{\mathrm{IQ}}_{0}$ |

### 5.3 Interrupt Sources

The interrupt sources include external interrupts $\left(N M I, \operatorname{IRQ}_{0}\right.$ to $\left.\mathrm{IRQ}_{5}\right)$ and 30 internal interrupts.

### 5.3.1 External Interrupts

There are seven external interrupts: NMI, and $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$. Of these, $\mathrm{NMI}, \mathrm{IRQ}_{0}, \mathrm{IRQ}_{1}$, and $\mathrm{IRQ}_{2}$ can be used to exit software standby mode.

NMI: NMI is the highest-priority interrupt and is always accepted, regardless of the states of the I and UI bits in CCR. The NMIEG bit in SYSCR selects whether an interrupt is requested by the rising or falling edge of the input at the NMI pin. NMI interrupt exception handling has vector number 7.
$\mathbf{I R Q}_{\mathbf{0}}$ to $\mathrm{IRQ}_{5}$ Interrupts: These interrupts are requested by input signals at pins $\overline{\operatorname{IRQ}}_{0}$ to $\overline{\mathrm{IRQ}}_{5}$. The $I R Q_{0}$ to $I R Q_{5}$ interrupts have the following features.

- ISCR settings can select whether an interrupt is requested by the low level of the input at pins $\overline{\mathrm{IRQ}}_{0}$ to $\overline{\mathrm{IRQ}}_{5}$, or by the falling edge.
- IER settings can enable or disable the $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ interrupts. Interrupt priority levels can be assigned by four bits in IPRA (IPRA7 to IPRA4).
- The status of $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ interrupt requests is indicated in ISR. The ISR flags can be cleared to 0 by software.

Figure 5.2 shows a block diagram of interrupts $I R Q_{0}$ to $I R Q_{5}$.


Figure 5.2 Block Diagram of Interrupts $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$

Figure 5.3 shows the timing of the setting of the interrupt flags (IRQnF).


## Figure 5.3 Timing of Setting of IRQnF

Interrupts $\mathrm{IRQ}_{0}$ to $\mathrm{IRQ}_{5}$ have vector numbers 12 to 17 . These interrupts are detected regardless of whether the corresponding pin is set for input or output. When using a pin for external interrupt input, clear its DDR bit to 0 and do not use the pin for chip select output, refresh output, or SCI input or output.

### 5.3.2 Internal Interrupts

Thirty internal interrupts are requested from the on-chip supporting modules.

- Each on-chip supporting module has status flags for indicating interrupt status, and enable bits for enabling or disabling interrupts.
- Interrupt priority levels can be assigned in IPRA and IPRB.
- ITU and SCI interrupt requests can activate the DMAC, in which case no interrupt request is sent to the interrupt controller, and the I and UI bits are disregarded.


### 5.3.3 Interrupt Exception Vector Table

Table 5.3 lists the interrupt sources, their vector addresses, and their default priority order. In the default priority order, smaller vector numbers have higher priority. The priority of interrupts other than NMI can be changed in IPRA and IPRB. The priority order after a reset is the default order shown in table 5.3.

Table 5.3 Interrupt Sources, Vector Addresses, and Priority

| Interrupt Source | Origin | Vector Number | Vector Address* | IPR | Priority |
| :---: | :---: | :---: | :---: | :---: | :---: |
| NMI | External pins | 7 | H'001C to H'001F | - | High |
| $\mathrm{IRQ}_{0}$ |  | 12 | H'0030 to H'0033 | IPRA7 |  |
| $\mathrm{IRQ}_{1}$ |  | 13 | H'0034 to H0037 | IPRA6 |  |
| $\mathrm{IRQ}_{2}$ |  | 14 | H'0038 to H'003B | IPRA5 |  |
| $\mathrm{IRQ}_{3}$ |  | 15 | $\mathrm{H}^{\prime} 003 \mathrm{C}$ to H'003F |  |  |
| $\mathrm{IRQ}_{4}$ |  | 16 | H'0040 to H'0043 | IPRA4 |  |
| $\mathrm{IRQ}_{5}$ |  | 17 | $\mathrm{H}^{\prime} 0044$ to H'0047 |  |  |
| Reserved | - | 18 | H'0048 to H'004B |  |  |
|  |  | 19 | $\mathrm{H}^{\prime} 004 \mathrm{C}$ to H'004F |  |  |
| WOVI (interval timer) | Watchdog timer | 20 | $\mathrm{H}^{\prime} 0050$ to H'0053 | IPRA3 |  |
| CMI (compare match) | Refresh controller | 21 | H'0054 to H'0057 |  |  |
| Reserved | - | 22 | H'0058 to H'005B |  |  |
|  |  | 23 | H'005C to H'005F |  |  |
| IMIAO (compare match/ input capture A0) | ITU channel 0 | 24 | H'0060 to H'0063 | IPRA2 |  |
| IMIBO (compare match/ input capture B0) |  | 25 | H'0064 to H'0067 |  |  |
| OVIO (overflow 0) |  | 26 | H'0068 to H'006B |  |  |
| Reserved | - | 27 | $\mathrm{H}^{\prime} 006 \mathrm{C}$ to H'006F |  |  |
| IMIA1 (compare match/ input capture A1) | ITU channel 1 | 28 | $\mathrm{H}^{\prime} 0070$ to H'0073 | IPRA1 |  |
| IMIB1 (compare match/ input capture B1) |  | 29 | $\mathrm{H}^{\prime} 0074$ to H'0077 |  |  |
| OVI1 (overflow 1) |  | 30 | $\mathrm{H}^{\prime} 0078$ to H'007B |  |  |
| Reserved | - | 31 | $\mathrm{H}^{\prime} 007 \mathrm{C}$ to H'007F |  |  |


| Interrupt Source | Origin | Vector Number | Vector Address* | IPR | Priority |
| :---: | :---: | :---: | :---: | :---: | :---: |
| IMIA2 (compare match/ input capture A2) | ITU channel 2 | 32 | H'0080 to H'0083 | IPRA0 | High |
| IMIB2 (compare match/ input capture B2) |  | 33 | H'0084 to H'0087 |  |  |
| OVI2 (overflow 2) |  | 34 | $\mathrm{H}^{\prime} 0088$ to H'008B |  |  |
| Reserved | - | 35 | H'008C to H'008F |  |  |
| IMIA3 (compare match/ input capture A3) | ITU channel 3 | 36 | $\mathrm{H}^{\prime} 0090$ to H'0093 | IPRB7 |  |
| IMIB3 (compare match/ input capture B3) |  | 37 | H'0094 to H'0097 |  |  |
| OVI3 (overflow 3) |  | 38 | $\mathrm{H}^{\prime} 0098$ to H'009B |  |  |
| Reserved | - | 39 | H'009C to H'009F |  |  |
| IMIA4 (compare match/ input capture A4) | ITU channel 4 | 40 | H'00A0 to H'00A3 | IPRB6 |  |
| IMIB4 (compare match/ input capture B4) |  | 41 | H'00A4 to H'00A7 |  |  |
| OVI4 (overflow 4) |  | 42 | H'00A8 to H'00AB |  |  |
| Reserved | - | 43 | H'00AC to H'00AF |  |  |
| DENDOA | DMAC | 44 | H'00B0 to H'00B3 | IPRB5 |  |
| DEND0B |  | 45 | H'00B4 to H'00B7 |  |  |
| DEND1A |  | 46 | H'00B8 to H'00BB |  |  |
| DEND1B |  | 47 | H'00BC to H'OOBF |  |  |
| Reserved | - | 48 | H'00C0 to H'00C3 | - |  |
|  |  | 49 | H'00C4 to H'00C7 |  |  |
|  |  | 50 | $\mathrm{H}^{\prime} 00 \mathrm{C} 8$ to $\mathrm{H}^{\prime} 00 \mathrm{CB}$ |  |  |
|  |  | 51 | H'00CC to H'00CF |  |  |
|  |  |  |  |  |  |

Rev. 2.0, 03/01, page 96 of 822

| Interrupt Source | Origin | Vector Number | Vector Address* | IPR | Priority |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ERIO (receive error 0) | SCI channel 0 | 52 | H'00D0 to H'00D3 | IPRB3 | High |
| RXIO (receive data full 0) |  | 53 | H'00D4 to H'00D7 |  |  |
| TXIO (transmit data empty 0) |  | 54 | H'00D8 to H'00DB |  |  |
| TEIO (transmit end 0) |  | 55 | H'00DC to H'00DF |  |  |
| ERI1 (receive error 1) | SCI channel 1 | 56 | H'00E0 to H'00E3 | IPRB2 |  |
| RXI1 (receive data full 1) |  | 57 | H'00E4 to H'00E7 |  |  |
| TXI1 (transmit data empty 1) |  | 58 | H'00E8 to H'00EB |  |  |
| TEI1 <br> (transmit end 1) |  | 59 | H'00EC to H'00EF |  |  |
| ADI (A/D end) | A/D | 60 | H'00F0 to H'00F3 | IPRB1 | Low |

Note: * Lower 16 bits of the address.

### 5.4 Interrupt Operation

### 5.4.1 Interrupt Handling Process

The H8/3052F handles interrupts differently depending on the setting of the UE bit. When UE $=1$, interrupts are controlled by the I bit. When UE $=0$, interrupts are controlled by the I and UI bits. Table 5.4 indicates how interrupts are handled for all setting combinations of the UE, I, and UI bits.

NMI interrupts are always accepted except in the reset and hardware standby states. IRQ interrupts and interrupts from the on-chip supporting modules have their own enable bits. Interrupt requests are ignored when the enable bits are cleared to 0 .

Table 5.4 UE, I, and UI Bit Settings and Interrupt Handling

| SYSCR | CCR |  | Description |
| :---: | :---: | :---: | :---: |
| UE | 1 | UI |  |
| 1 | 0 | - | All interrupts are accepted. Interrupts with priority level 1 have higher priority. |
|  | 1 | - | No interrupts are accepted except NMI. |
| 0 | 0 | - | All interrupts are accepted. Interrupts with priority level 1 have higher priority. |
|  | 1 | 0 | NMI and interrupts with priority level 1 are accepted. |
|  |  | 1 | No interrupts are accepted except NMI. |

$\mathbf{U E}=1$ : Interrupts $I R Q^{0}$ to $I R Q_{5}$ and interrupts from the on-chip supporting modules can all be masked by the I bit in the CPU's CCR. Interrupts are masked when the I bit is set to 1 , and unmasked when the I bit is cleared to 0 . Interrupts with priority level 1 have higher priority. Figure 5.4 is a flowchart showing how interrupts are accepted when $\mathrm{UE}=1$.


Figure 5.4 Process Up to Interrupt Acceptance when UE = 1

1. If an interrupt condition occurs and the corresponding interrupt enable bit is set to 1 , an interrupt request is sent to the interrupt controller.
2. When the interrupt controller receives one or more interrupt requests, it selects the highestpriority request, following the IPR interrupt priority settings, and holds other requests pending. If two or more interrupts with the same IPR setting are requested simultaneously, the interrupt controller follows the priority order shown in table 5.3.
3. The interrupt controller checks the $I$ bit. If the $I$ bit is cleared to 0 , the selected interrupt request is accepted. If the I bit is set to 1 , only NMI is accepted; other interrupt requests are held pending.
4. When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
5. In interrupt exception handling, PC and CCR are saved to the stack area. The PC value that is saved indicates the address of the first instruction that will be executed after the return from the interrupt service routine.
6. Next the I bit is set to 1 in CCR, masking all interrupts except NMI.
7. The vector address of the accepted interrupt is generated, and the interrupt service routine starts executing from the address indicated by the contents of the vector address.
$\mathbf{U E}=\mathbf{0}$ : The I and UI bits in the CPU's CCR and the IPR bits enable three-level masking of IRQ $_{0}$ to $\mathrm{IRQ}_{5}$ interrupts and interrupts from the on-chip supporting modules.

- Interrupt requests with priority level 0 are enabled when the I bit is cleared to 0 , and disabled when the I bit is set to 1 .
- Interrupt requests with priority level 1 are enabled when the I bit or UI bit is cleared to 0 , and disabled when the I bit and UI bit are both set to 1 .
For example, if the interrupt enable bits of all interrupt requests are set to 1 , and IPRA and IPRB are set to $\mathrm{H}^{\prime} 20$ and $\mathrm{H}^{\prime} 00$, respectively (giving $\mathrm{IRQ}_{2}$ and $\mathrm{IRQ}_{3}$ interrupt requests priority over other interrupts), interrupts are enabled and disabled as follows:
a. If $\mathrm{I}=0$, all interrupts are enabled (priority order: $\mathrm{NMI}>\mathrm{IRQ}_{2}>\mathrm{IRQ}_{3}>\mathrm{IRQ}_{0} \ldots$ ).
b. If $\mathrm{I}=1$ and $\mathrm{UI}=0$, only $\mathrm{NMI}, \mathrm{IRQ}_{2}$, and $\mathrm{IRQ}_{3}$ are enabled.
c. If $\mathrm{I}=1$ and $\mathrm{UI}=1$, all interrupts are disabled except NMI.

Figure 5.5 shows the transitions among the above states.


Figure 5.5 Interrupt Enable/Disable State Transitions (Example)
Figure 5.6 is a flowchart showing how interrupts are accepted when UE $=0$.

1. If an interrupt condition occurs and the corresponding interrupt enable bit is set to 1 , an interrupt request is sent to the interrupt controller.
2. When the interrupt controller receives one or more interrupt requests, it selects the highestpriority request, following the IPR interrupt priority settings, and holds other requests pending. If two or more interrupts with the same IPR setting are requested simultaneously, the interrupt controller follows the priority order shown in table 5.3.
3. The interrupt controller checks the I bit. If the I bit is cleared to 0 , the interrupt request is accepted regardless of its IPR setting. The value of the UI bit is immaterial. If the I bit is set to 1 and the UI bit is cleared to 0 , only interrupt requests with priority level 1 are accepted; interrupt requests with priority level 0 are held pending. If the I bit and UI bit are both set to 1 , the interrupt request is held pending.
4. When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
5. In interrupt exception handling, PC and CCR are saved to the stack area. The PC value that is saved indicates the address of the first instruction that will be executed after the return from the interrupt service routine.
6. The I and UI bits are set to 1 in CCR, masking all interrupts except NMI.
7. The vector address of the accepted interrupt is generated, and the interrupt service routine starts executing from the address indicated by the contents of the vector address.


Figure 5.6 Process Up to Interrupt Acceptance when UE $=0$

### 5.4.2 Interrupt Exception Handling Sequence

Figure 5.7 shows the interrupt sequence in mode 2 when the program area and stack area are in 16-bit, two-state access space in external memory.


Figure 5.7 Interrupt Sequence (Mode 2, Two-State Access, Stack in External Memory)

### 5.4.3 Interrupt Response Time

Table 5.5 indicates the interrupt response time from the occurrence of an interrupt request until the first instruction of the interrupt service routine is executed.

Table 5.5 Interrupt Response Time

| No. | Item | On-Chip Memory | External Memory |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | 8-Bit Bus |  | 16-Bit Bus |  |
|  |  |  | 2 States | 3 States | 2 States | 3 States |
| 1 | Interrupt priority decision | $2^{* 1}$ | $2^{* 1}$ | $2^{* 1}$ | $2^{* 1}$ | $2^{* 1}$ |
| 2 | Maximum number of states until end of current instruction | 1 to 23 | 1 to 27 | 1 to $31^{* 4}$ | 1 to 23 | 1 to $25 * 4$ |
| 3 | Saving PC and CCR to stack | 4 | 8 | $12^{* 4}$ | 4 | $6^{* 4}$ |
| 4 | Vector fetch | 4 | 8 | $12^{* 4}$ | 4 | $6^{* 4}$ |
| 5 | Instruction prefetch*2 | 4 | 8 | $12^{* 4}$ | 4 | $6^{* 4}$ |
| 6 | Internal processing*3 | 4 | 4 | 4 | 4 | 4 |
| Total |  | 19 to 41 | 31 to 57 | 43 to 73 | 19 to 41 | 25 to 49 |

Notes: 1. 1 state for internal interrupts.
2. Prefetch after the interrupt is accepted and prefetch of the first instruction in the interrupt service routine.
3. Internal processing after the interrupt is accepted and internal processing after prefetch.
4. The number of states increases if wait states are inserted in external memory access.

### 5.5 Usage Notes

### 5.5.1 Contention between Interrupt Generation and Disabling

When an instruction clears an interrupt enable bit to 0 to disable the interrupt, the interrupt is not actually disabled until after execution of the instruction is completed. Thus, if an interrupt occurs while a BCLR, MOV, or other instruction is being executed to clear its interrupt enable bit to 0 , at the instant when execution of the instruction ends the interrupt is still enabled, so its interrupt exception handling is carried out. If a higher-priority interrupt is also requested, however, interrupt exception handling for the higher-priority interrupt is carried out, and the lower-priority interrupt is ignored. This also applies when an interrupt source flag is cleared to 0 .

Figure 5.8 shows an example in which an IMIEA bit is cleared to 0 in TIER of the ITU.


Figure 5.8 Contention between Interrupt and Interrupt-Disabling Instruction
This type of contention will not occur if the interrupt is masked when the interrupt enable bit or flag is cleared to 0 .

### 5.5.2 Instructions that Inhibit Interrupts

The LDC, ANDC, ORC, and XORC instructions inhibit interrupts. When an interrupt occurs, after determining the interrupt priority, the interrupt controller requests a CPU interrupt. If the CPU is currently executing one of these interrupt-inhibiting instructions, however, when the instruction is completed the CPU always continues by executing the next instruction.

### 5.5.3 Interrupts during EEPMOV Instruction Execution

The EEPMOV.B and EEPMOV.W instructions differ in their reaction to interrupt requests.
When the EEPMOV.B instruction is executing a transfer, no interrupts are accepted until the transfer is completed, not even NMI.

When the EEPMOV.W instruction is executing a transfer, interrupt requests other than NMI are not accepted until the transfer is completed. If NMI is requested, NMI exception handling starts at a transfer cycle boundary. The PC value saved on the stack is the address of the next instruction. Programs should be coded as follows to allow for NMI interrupts during EEPMOV.W execution:

```
L1: EEPMOV.W
    MOV.W R4,R4
    BNE L1
```


### 5.5.4 Notes on Use of External Interrupts

The specifications provide for the IRQnF flag to be cleared by first reading the flag while it is set to 1 , then writing 0 to it. However, there are cases in which the IRQnF flag is erroneously cleared, preventing execution of interrupt exception handling, simply by writing 0 to the flag, without first reading 1 from it. This occurs when the following conditions are fulfilled.

## Setting Conditions

1. When using multiple external interrupts (IRQa, IRQb)
2. When different clearing methods are used for the IRQaF flag and IRQbF flag, with the IRQaF flag cleared by writing 0 to it, and the IRQbF flag cleared by hardware.
3. IRQaF flag clears and bit operation command is being used for the IRQ status register (ISR) or the ISR is being read in bytes; IRQaF flag's bits clear and other bit values read in bits are written in bytes.

## Occurrence Conditions

1. If an ISR register read is executed to clear the IRQaF flag while $\operatorname{IRQaF}=1$, and then the IRQbF flag is cleared by the initiation of interrupt exception handling.
2. If there is contention between IRQaF flag clearing and IRQbF generation (IRQaF flag setting) (when IRQbF $=0$ at the time of the ISR read to clear the IRQaF flag, but IRQbF is set to 1 before the write to ISR).

If above setting conditions 1 to 3 and occurrence conditions 1 and 2 are all fulfilled, IRQbF will be cleared erroneously when the ISR write in occurrence condition 2 is executed, and so interrupt exception handling will not be carried out.

However, the IRQbF flag will not be cleared erroneously if 0 is written to it at least once between occurrence conditions 1 and 2.


Figure 5.9 IRQnF Flag when Interrupt Processing Is Not Conducted
Either of the following methods can be used to prevent this problem.

- Solution 1

When IRQaF flag clears, do not use the bit computation command, read the ISR in bytes.
When IRQaF only is 0 write all other bits as 1 in bytes.
For example, if $\mathrm{a}=0$
MOV.B @ISR,ROL
MOV.B \#HFE,ROL

MOV.B ROL,@ISR

- Solution 2

During IRQb interrupt processing, carry out IRQbF flag clear dummy processing.
For example, if $b=1$
IRQB MOV.B \#HFD,ROL
MOV.B ROL, @ISR

## Section 6 Bus Controller

### 6.1 Overview

The H8/3052F has an on-chip bus controller that divides the address space into eight areas and can assign different bus specifications to each. This enables different types of memory to be connected easily.

A bus arbitration function of the bus controller controls the operation of the DMA controller (DMAC) and refresh controller. The bus controller can also release the bus to an external device.

### 6.1.1 Features

Features of the bus controller are listed below.

- Independent settings for address areas 0 to 7
- 128-kbyte areas in 1-Mbyte modes; 2-Mbyte areas in 16-Mbyte modes.
- Chip select signals ( $\overline{\mathrm{CS}}_{0}$ to $\left.\overline{\mathrm{CS}}_{7}\right)$ can be output for areas 0 to 7 .
- Areas can be designated for 8-bit or 16-bit access.
- Areas can be designated for two-state or three-state access.
- Four wait modes
- Programmable wait mode, pin auto-wait mode, and pin wait modes 0 and 1 can be selected.
- Zero to three wait states can be inserted automatically.
- Bus arbitration function
- A built-in bus arbiter grants the bus right to the CPU, DMAC, refresh controller, or an external bus master.


### 6.1.2 Block Diagram

Figure 6.1 shows a block diagram of the bus controller.


Figure 6.1 Block Diagram of Bus Controller

### 6.1.3 Pin Configuration

Table 6.1 summarizes the bus controller's input/output pins.
Table 6.1 Bus Controller Pins

| Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Chip select 0 to 7 | $\overline{\mathrm{CS}}_{0}$ to $\overline{\mathrm{CS}}_{7}$ | Output | Strobe signals selecting areas 0 to 7 |
| Address strobe | $\overline{\mathrm{AS}}$ | Output | Strobe signal indicating valid address output <br> on the address bus |
| Read | $\overline{\mathrm{RD}}$ | Output | Strobe signal indicating reading from the <br> external address space |
| High write | $\overline{\mathrm{HWR}}$ | Output | Strobe signal indicating writing to the <br> external address space, with valid data on <br> the upper data bus $\left(\mathrm{D}_{15}\right.$ to $\left.\mathrm{D}_{8}\right)$ |
| Low write | $\overline{\mathrm{LWR}}$ | Output | Strobe signal indicating writing to the <br> external address space, with valid data on <br> the lower data bus ( $\mathrm{D}_{7}$ to $\left.\mathrm{D}_{0}\right)$ |
| Wait | $\overline{\mathrm{WAIT}}$ | Input | Wait request signal for access to external <br> three-state-access areas |
| Bus request | $\overline{\mathrm{BREQ}}$ | Output | Acknowledge signal indicating the bus is <br> released to an external device |
| Bus acknowledge | $\overline{\mathrm{BACK}}$ |  | external device |

### 6.1.4 Register Configuration

Table 6.2 summarizes the bus controller's registers.
Table 6.2 Bus Controller Registers

|  |  |  | Initial Value |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Address* | Name | Abbreviation | R/w | Modes | 1, 3, 5, $\mathbf{6}$ |
| Modes |  |  |  |  |  |
| $\mathbf{2 , 4 , 7}$ |  |  |  |  |  |$]$|  |  | ABWCR | R/W |
| :--- | :--- | :--- | :--- |
| H'FFEC | Bus width control register | H'FF | H'00 |
| H'FFED | Access state control register | ASTCR | R/W |
| H'FFEE | Wait control register | WCR | R/W |
| H'FFEF | Wait state controller enable <br> register | WCER | R/W |
|  | H'F3 | H'FF | H'FF |
| H'FFF3 | Bus release control register | BRCR | R/W |
| H'FF5F | Chip select control register | CSCR | R/W |

Note: * Lower 16 bits of the address.

### 6.2 Register Descriptions

### 6.2.1 Bus Width Control Register (ABWCR)

ABWCR is an 8-bit readable/writable register that selects 8-bit or 16-bit access for each area.


When ABWCR contains H'FF (selecting 8-bit access for all areas), the chip operates in 8-bit bus mode: the upper data bus $\left(\mathrm{D}_{15}\right.$ to $\left.\mathrm{D}_{8}\right)$ is valid, and port 4 is an input/output port. When at least one bit is cleared to 0 in ABWCR, the chip operates in 16-bit bus mode with a 16-bit data bus $\left(\mathrm{D}_{15}\right.$ to $\mathrm{D}_{0}$ ). In modes $1,3,5,6$, and 7 ABWCR is initialized to H'FF by a reset and in hardware standby mode. In modes 2 and 4 ABWCR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. ABWCR is not initialized in software standby mode.

Bits 7 to 0—Area 7 to 0 Bus Width Control (ABW7 to ABW0): These bits select 8-bit access or 16-bit access to the corresponding address areas.

## Bits 7 to 0 :

ABW7 to ABW0 Description

| 0 | Areas 7 to 0 are 16 -bit access areas |
| :--- | :--- |
| 1 | Areas 7 to 0 are 8 -bit access areas |

ABWCR specifies the bus width of external memory areas. The bus width of on-chip memory and registers is fixed and does not depend on ABWCR settings. These settings are therefore meaningless in single-chip mode (mode 7).

### 6.2.2 Access State Control Register (ASTCR)

ASTCR is an 8-bit readable/writable register that selects whether each area is accessed in two states or three states.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AST7 | AST6 | AST5 | AST4 | AST3 | AST2 | AST1 | ASTO |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

ASTCR is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 0—Area 7 to 0 Access State Control (AST7 to AST0): These bits select whether the corresponding area is accessed in two or three states.

## Bits 7 to 0 :

AST7 to AST0 Description

| 0 | Areas 7 to 0 are accessed in two states |  |
| :--- | :--- | :--- |
| 1 | Areas 7 to 0 are accessed in three states | (Initial value) |

ASTCR specifies the number of states in which external areas are accessed. On-chip memory and registers are accessed in a fixed number of states that does not depend on ASTCR settings. These settings are therefore meaningless in single-chip mode (mode 7).

### 6.2.3 Wait Control Register (WCR)

WCR is an 8-bit readable/writable register that selects the wait mode for the wait-state controller (WSC) and specifies the number of wait states.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | WMS1 | WMSO | WC1 | WCO |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 1 | 1 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
| Reserved bits |  |  |  |  |  | Wait count 1/0 <br> These bits select the number of wait states inserted |  |  |
| Wait mode select $1 / 0$ <br> These bits select the wait mode |  |  |  |  |  |  |  |  |

WCR is initialized to H'F3 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 4—Reserved: Read-only bits, always read as 1 .
Bits 3 and 2-Wait Mode Select 1 and 0 (WMS1/0): These bits select the wait mode.

| Bit 3: WMS1 | Bit 2: WMSO | Description | (Initial value) |
| :--- | :--- | :--- | :--- |
| 0 | 0 | Programmable wait mode |  |
|  | 1 | No wait states inserted by wait-state controller |  |
| 1 | 0 | Pin wait mode 1 |  |
|  | 1 | Pin auto-wait mode |  |

Bits 1 and 0—Wait Count 1 and 0 (WC1/0): These bits select the number of wait states inserted in access to external three-state-access areas.

| Bit 1: WC1 | Bit 0: WC0 | Description |  |
| :--- | :--- | :--- | :--- |
| 0 | 0 | No wait states inserted by wait-state controller |  |
|  | 1 | 1 state inserted |  |
| 1 | 0 | 2 states inserted |  |
| 1 | 3 states inserted | (Initial value) |  |

### 6.2.4 Wait State Controller Enable Register (WCER)

WCER is an 8-bit readable/writable register that enables or disables wait-state control of external three-state-access areas by the wait-state controller.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | WCE7 | WCE6 | WCE5 | WCE4 | WCE3 | WCE2 | WCE1 | WCEO |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Wait-state controller enable 7 to 0 |  |  |  |  |  |  |  |  |

WCER is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 0-Wait-State Controller Enable 7 to 0 (WCE7 to WCE0): These bits enable or disable wait-state control of external three-state-access areas.

| Bits 7 to 0: <br> WCE7 to WCE0 | Description |  |
| :--- | :--- | :--- |
| 0 | Wait-state control disabled (pin wait mode 0) |  |
| 1 | Wait-state control enabled | (Initial value) |

Since WCER enables or disables wait-state control of external three-state-access areas, these settings are meaningless in single-chip mode (mode 7).

### 6.2.5 Bus Release Control Register (BRCR)

BRCR is an 8-bit readable/writable register that enables address output on bus lines $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ and enables or disables release of the bus to an external device.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | A23E | A22E | A21E | - | - | - | - | BRLE |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |
| Modes 1, 2, 5, 7 <br> Modes 3,4,6 | - | - | - | - | - | - | - | R/W |
|  | R/W | R/W | R/W | - | - | - | - | R/W |
|  | Address 23 to 21 enable These bits enable $\mathrm{PA}_{6}$ to $\mathrm{PA}_{4}$ to be used for $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ address output |  |  | Reserved bits |  |  | Bus release enable Enables or disables release of the bus to an external device |  |
|  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |

BRCR is initialized to H'FE by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7—Address 23 Enable (A23E): Enables $\mathrm{PA}_{4}$ to be used as the $\mathrm{A}_{23}$ address output pin. Writing 0 in this bit enables $\mathrm{A}_{23}$ address output from $\mathrm{PA}_{4}$. In modes other than 3,4 , and 6 this bit cannot be modified and $\mathrm{PA}_{4}$ has its ordinary input/output functions.

Bit 7: A23E Description

| 0 | $\mathrm{PA}_{4}$ is the $\mathrm{A}_{23}$ address output pin |  |
| :--- | :--- | :--- |
| 1 | $\mathrm{PA}_{4}$ is the $\mathrm{PA}_{4} / T \mathrm{TP}_{4} /$ TIOCA ${ }_{1}$ input/output pin | (Initial value) |

Bit 6—Address 22 Enable (A22E): Enables $\mathrm{PA}_{5}$ to be used as the $\mathrm{A}_{22}$ address output pin. Writing 0 in this bit enables $\mathrm{A}_{22}$ address output from $\mathrm{PA}_{5}$. In modes other than 3,4 , and 6 this bit cannot be modified and $\mathrm{PA}_{5}$ has its ordinary input/output functions.

Bit 6: A22E Description

| 0 | $\mathrm{PA}_{5}$ is the $\mathrm{A}_{22}$ address output pin |  |
| :--- | :--- | :--- |
| 1 | $\mathrm{PA}_{5}$ is the $\mathrm{PA}_{5} / \mathrm{TP}_{5} /$ TIOCB ${ }_{1}$ input/output pin | (Initial value) |

Bit 5—Address 21 Enable (A21E): Enables $\mathrm{PA}_{6}$ to be used as the $\mathrm{A}_{21}$ address output pin. Writing 0 in this bit enables $\mathrm{A}_{21}$ address output from $\mathrm{PA}_{6}$. In modes other than 3, 4, and 6 this bit cannot be modified and $\mathrm{PA}_{6}$ has its ordinary input/output functions.

## Bit 5: A21E Description

| 0 | $\mathrm{PA}_{6}$ is the $\mathrm{A}_{21}$ address output pin |  |
| :--- | :--- | :--- |
| 1 | $\mathrm{PA}_{6}$ is the $\mathrm{PA}_{6} / T \mathrm{P}_{6} / T I O C A_{2}$ input/output pin | (Initial value) |

Bits 4 to 1—Reserved: Read-only bits, always read as 1.
Bit 0—Bus Release Enable (BRLE): Enables or disables release of the bus to an external device.

| Bit 0: BRLE | Description |
| :--- | :--- |
| 0 | The bus cannot be released to an external device; $\overline{\mathrm{BREQ}}$ and $\overline{\mathrm{BACK}}$ can be <br> (Initial value) |
| 1 | The bus can be released to an external device |

### 6.2.6 Chip Select Control Register (CSCR)

CSCR is an 8-bit readable/writable register that enables or disables output of chip select signals $\left(\overline{\mathrm{CS}}_{7}\right.$ to $\left.\overline{\mathrm{CS}}_{4}\right)$.

If a chip select signal $\left(\overline{\mathrm{CS}}_{7}\right.$ to $\left.\overline{\mathrm{CS}}_{4}\right)$ output is selected in this register, the corresponding pin functions as a chip select signal $\left(\overline{\mathrm{CS}}_{7}\right.$ to $\left.\overline{\mathrm{CS}}_{4}\right)$ output, this function taking priority over other functions. CSCR cannot be modified in single-chip mode.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CS7E | CS6E | CS5E | CS4E | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | - | - | - | - |
|  | Chip select 7 to 4 enable These bits enable or disable chip select signal output |  |  |  | Reserved bits |  |  |  |

CSCR is initialized to $\mathrm{H}^{\prime} 0 \mathrm{~F}$ by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 4—Chip Select 7 to 4 Enable (CS7E to CS4E): These bits enable or disable output of the corresponding chip select signal.

Bit n: CSnE
Description

| 0 | Output of chip select signal $\mathrm{CS}_{\mathrm{n}}$ is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | Output of chip select signal $\mathrm{CS}_{\mathrm{n}}$ is enabled |  |

Note: $\mathrm{n}=7$ to 4

Bits 3 to 0—Reserved: Read-only bits, always read as 1.

### 6.3 Operation

### 6.3.1 Area Division

The external address space is divided into areas 0 to 7 . Each area has a size of 128 kbytes in the 1 -Mbyte modes, or 2 Mbytes in the $16-\mathrm{Mbyte}$ modes. Figure 6.2 shows a general view of the memory map.


Figure 6.2 Access Area Map for Modes 1 to 6

Chip select signals $\left(\overline{\mathrm{CS}}_{7}\right.$ to $\left.\overline{\mathrm{CS}}_{0}\right)$ can be output for areas 7 to 0 . The bus specifications for each area can be selected in ABWCR, ASTCR, WCER, and WCR as shown in table 6.3.

Table 6.3 Bus Specifications

| ABWCR | ASTCR | WCER | WCR |  | Bus Specifications |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ABWn | ASTn | WCEn | WMS1 | WMSO | Bus Width | Access States | Wait Mode |
| 0 | 0 | - | - | - | 16 | 2 | Disabled |
|  | 1 | 0 | - | - | 16 | 3 | Pin wait mode 0 |
|  |  | 1 | 0 | 0 | 16 | 3 | Programmable wait mode |
|  |  |  |  | 1 | 16 | 3 | Disabled |
|  |  |  | 1 | 0 | 16 | 3 | Pin wait mode 1 |
|  |  |  |  | 1 | 16 | 3 | Pin auto-wait mode |
| 1 | 0 | - | - | - | 8 | 2 | Disabled |
|  | 1 | 0 | - | - | 8 | 3 | Pin wait mode 0 |
|  |  | 1 | 0 | 0 | 8 | 3 | Programmable wait mode |
|  |  |  |  | 1 | 8 | 3 | Disabled |
|  |  |  | 1 | 0 | 8 | 3 | Pin wait mode 1 |
|  |  |  |  | 1 | 8 | 3 | Pin auto-wait mode |

Note: $\mathrm{n}=7$ to 0

### 6.3.2 Chip Select Signals

For each of areas 7 to 0 , the $\mathrm{H} 8 / 3052 \mathrm{~F}$ can output a chip select signal $\left(\overline{\mathrm{CS}}_{7}\right.$ to $\left.\overline{\mathrm{CS}}_{0}\right)$ that goes low to indicate when the area is selected. Figure 6.3 shows the output timing of a $\overline{\mathrm{CS}}_{\mathrm{n}}$ signal ( $\mathrm{n}=7$ to 0 ).

Output of $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{0}$ : Output of $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{0}$ is enabled or disabled in the data direction register (DDR) of the corresponding port.

In the expanded modes with on-chip ROM disabled, a reset leaves pin $\overline{\mathrm{CS}}_{0}$ in the output state and pins $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{1}$ in the input state. To output chip select signals $\overline{\mathrm{CS}}_{3}$, to $\overline{\mathrm{CS}}_{1}$ the corresponding DDR bits must be set to 1 . In the expanded modes with on-chip ROM enabled, a reset leaves pins $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{0}$ in the input state. To output chip select signals $\overline{\mathrm{CS}}_{3}$, to $\overline{\mathrm{CS}}_{0}$ the corresponding DDR bits must be set to 1 . For details see section 9, I/O Ports.

Output of $\overline{\mathrm{CS}}_{7}$ to $\overline{\mathrm{CS}}_{4}$ : Output of $\overline{\mathrm{CS}}_{7}$ to $\overline{\mathrm{CS}}_{4}$ is enabled or disabled in the chip select control register (CSCR). A reset leaves pins $\overline{\mathrm{CS}}_{7}$ to $\overline{\mathrm{CS}}_{4}$ in the input state. To output chip select signals $\overline{\mathrm{CS}}_{7}$ to $\overline{\mathrm{CS}}_{4}$, the corresponding CSCR bits must be set to 1 . For details see section 9, I/O Ports.


Figure $6.3 \overline{\mathrm{CS}}_{\mathrm{n}}$ Output Timing ( $\mathrm{n}=7$ to 0)
When the on-chip ROM, on-chip RAM, and on-chip registers are accessed, $\overline{\mathrm{CS}}_{7}$ and $\overline{\mathrm{CS}}_{0}$ remain high. The $\overline{\mathrm{CS}}_{\mathrm{n}}$ signals are decoded from the address signals. They can be used as chip select signals for SRAM and other devices.

### 6.3.3 Data Bus

The $\mathrm{H} 8 / 3052$ F allows either 8 -bit access or 16 -bit access to be designated for each of areas 0 to 7 . An 8-bit-access area uses the upper data bus ( $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ ). A 16-bit-access area uses both the upper data bus ( $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ ) and lower data bus ( $\mathrm{D}_{7}$ to $\mathrm{D}_{0}$ ).

In read access the $\overline{\mathrm{RD}}$ signal applies without distinction to both the upper and lower data bus. In write access the $\overline{\mathrm{HWR}}$ signal applies to the upper data bus, and the $\overline{\mathrm{LWR}}$ signal applies to the lower data bus.

Table 6.4 indicates how the two parts of the data bus are used under different access conditions.
Table 6.4 Access Conditions and Data Bus Usage

| Area | Access Size | Read/W rite | Address | Valid Strobe | Upper Data Bus ( $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ ) | Lower Data Bus ( $D_{7}$ to $D_{0}$ ) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8-bit-access area | - | Read | - | $\overline{\mathrm{RD}}$ | Valid | Invalid |
|  |  | Write | - | $\overline{\text { HWR }}$ |  | Undetermined data |
| 16-bit-access area | Byte | Read | Even | $\overline{\mathrm{RD}}$ | Valid | Invalid |
|  |  |  | Odd |  | Invalid | Valid |
|  |  | Write | Even | HWR | Valid | Undetermined data |
|  |  |  | Odd | $\overline{\text { LWR }}$ | Undetermined data | Valid |
|  | Word | Read | - | $\overline{\mathrm{RD}}$ | Valid | Valid |
|  |  | Write | - | HWR, <br> LWR | Valid | Valid |

Note: Undetermined data means that unpredictable data is output. Invalid means that the bus is in the input state and the input is ignored.

### 6.3.4 <br> Bus Control Signal Timing

8-Bit, Three-State-Access Areas: Figure 6.4 shows the timing of bus control signals for an 8 -bit, three-state-access area. The upper address bus $\left(D_{15}\right.$ to $\left.D_{8}\right)$ is used to access these areas. The $\overline{\mathrm{LWR}}$ pin is always high. Wait states can be inserted.


Note: $\mathrm{n}=7$ to 0

Figure 6.4 Bus Control Signal Timing for 8-Bit, Three-State-Access Area

8-Bit, Two-State-Access Areas: Figure 6.5 shows the timing of bus control signals for an 8-bit, two-state-access area. The upper address bus $\left(\mathrm{D}_{15}\right.$ to $\left.\mathrm{D}_{8}\right)$ is used to access these areas. The $\overline{\mathrm{LWR}}$ pin is always high. Wait states cannot be inserted.


Note: $\mathrm{n}=7$ to 0
Figure 6.5 Bus Control Signal Timing for 8-Bit, Two-State-Access Area

16-Bit, Three-State-Access Areas: Figures 6.6 to 6.8 show the timing of bus control signals for a 16-bit, three-state-access area. In these areas, the upper address bus $\left(D_{15}\right.$ to $\left.D_{8}\right)$ is used to access even addresses and the lower address bus ( $D_{7}$ to $D_{0}$ ) is used to access odd addresses. Wait states can be inserted.


Figure 6.6 Bus Control Signal Timing for 16-Bit, Three-State-Access Area (1) (Byte Access to Even Address)


Note: $\mathrm{n}=7$ to 0

Figure 6.7 Bus Control Signal Timing for 16-Bit, Three-State-Access Area (2) (Byte Access to Odd Address)


Note: $\mathrm{n}=7$ to 0

Figure 6.8 Bus Control Signal Timing for 16-Bit, Three-State-Access Area (3) (Word Access)

16-Bit, Two-State-Access Areas: Figures 6.9 to 6.11 show the timing of bus control signals for a 16-bit, two-state-access area. In these areas, the upper address bus ( $D_{15}$ to $\left.D_{8}\right)$ is used to access even addresses and the lower address bus ( $D_{7}$ to $D_{0}$ ) is used to access odd addresses. Wait states cannot be inserted.


Figure 6.9 Bus Control Signal Timing for 16-Bit, Two-State-Access Area (1) (Byte Access to Even Address)


Note: $\mathrm{n}=7$ to 0

Figure 6.10 Bus Control Signal Timing for 16-Bit, Two-State-Access Area (2) (Byte Access to Odd Address)


Note: $\mathrm{n}=7$ to 0

Figure 6.11 Bus Control Signal Timing for 16-Bit, Two-State-Access Area (3) (Word Access)
6.3.5 Wait Modes

Four wait modes can be selected as shown in table 6.5.

Table 6.5 Wait Mode Selection

| ASTCR | WCER | WCR |  | WSC Control | Wait Mode |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ASTn Bit | WCEn Bit | WMS1 Bit | WMSO Bit |  |  |
| 0 | - | - | - | Disabled | No wait states |
| 1 | 0 | - | - | Disabled | Pin wait mode 0 |
|  | 1 | 0 | 0 | Enabled | Programmable wait mode |
|  |  |  | 1 | Enabled | No wait states |
|  |  | 1 | 0 | Enabled | Pin wait mode 1 |
|  |  |  | 1 | Enabled | Pin auto-wait mode |

Note: $\mathrm{n}=7$ to 0

Wait Mode in Areas Where Wait-State Controller is Disabled: External three-state access areas in which the wait-state controller is disabled ( $\mathrm{ASTn}=1, \mathrm{WCEn}=0$ ) operate in pin wait mode 0 . The other wait modes are unavailable. The settings of bits WMS1 and WMS0 are ignored in these areas.

## - Pin Wait Mode 0

Wait states can only be inserted by $\overline{\text { WAIT }}$ pin control. During access to an external three-stateaccess area, if the WAIT pin is low at the fall of the system clock $(\phi)$ in the $\mathrm{T}_{2}$ state, a wait state ( $\mathrm{T}_{\mathrm{w}}$ ) is inserted. If the $\overline{\text { WAIT }}$ pin remains low, wait states continue to be inserted until the $\overline{\text { WAIT }}$ signal goes high. Figure 6.12 shows the timing.


Figure 6.12 Pin Wait Mode 0

Wait Modes in Areas Where Wait-State Controller is Enabled: External three-state access areas in which the wait-state controller is enabled ( $\mathrm{ASTn}=1, \mathrm{WCEn}=1$ ) can operate in pin wait mode 1, pin auto-wait mode, or programmable wait mode, as selected by bits WMS1 and WMS0. Bits WMS1 and WMS0 apply to all areas, so all areas in which the wait-state controller is enabled operate in the same wait mode.

## - Pin Wait Mode 1

In all accesses to external three-state-access areas, the number of wait states ( $\mathrm{T}_{\mathrm{w}}$ ) selected by bits WC1 and WC0 are inserted. If the WAIT pin is low at the fall of the system clock ( $\phi$ ) in the last of these wait states, an additional wait state is inserted. If the $\overline{\text { WAIT }}$ pin remains low, wait states continue to be inserted until the WAIT signal goes high.
Pin wait mode 1 is useful for inserting four or more wait states, or for inserting different numbers of wait states for different external devices.
If the wait count is 0 , this mode operates in the same way as pin wait mode 0 .
Figure 6.13 shows the timing when the wait count is $1(\mathrm{WC1}=0, \mathrm{WC} 0=1)$ and one additional wait state is inserted by $\overline{\mathrm{WAIT}}$ input.


Figure 6.13 Pin Wait Mode 1

- Pin Auto-Wait Mode

If the $\overline{\mathrm{WAIT}} \mathrm{pin}$ is low, the number of wait states $\left(\mathrm{T}_{\mathrm{w}}\right)$ selected by bits WC 1 and WC 0 are inserted.
In pin auto-wait mode, if the $\overline{\text { WAIT }}$ pin is low at the fall of the system clock $(\phi)$ in the $\mathrm{T}_{2}$ state, the number of wait states $\left(\mathrm{T}_{\mathrm{w}}\right)$ selected by bits WC 1 and WC 0 are inserted. No additional wait states are inserted even if the $\overline{\mathrm{WAIT}}$ pin remains low. Pin auto-wait mode can be used for an easy interface to low-speed memory, simply by routing the chip select signal to the $\overline{\text { WAIT }}$ pin.
Figure 6.14 shows the timing when the wait count is 1 .


Figure 6.14 Pin Auto-Wait Mode

- Programmable Wait Mode

The number of wait states $\left(\mathrm{T}_{\mathrm{w}}\right)$ selected by bits WC 1 and WC 0 are inserted in all accesses to external three-state-access areas. Figure 6.15 shows the timing when the wait count is 1 (WC1 $=0, \mathrm{WC} 0=1$ ).


Figure 6.15 Programmable Wait Mode

Example of Wait State Control Settings: A reset initializes ASTCR and WCER to H'FF and WCR to H'F3, selecting programmable wait mode and three wait states for all areas. Software can select other wait modes for individual areas by modifying the ASTCR, WCER, and WCR settings. Figure 6.16 shows an example of wait mode settings.


Note: Wait states cannot be inserted in areas designated for two-state access by ASTCR.

Figure 6.16 Wait Mode Settings (Example)

### 6.3.6 Interconnections with Memory (Example)

For each area, the bus controller can select two- or three-state access and an 8- or 16-bit data bus width. In three-state-access areas, wait states can be inserted in a variety of modes, simplifying the connection of both high-speed and low-speed devices.

Figure 6.18 shows an example of interconnections between the $\mathrm{H} 8 / 3052$ F and memory. Figure 6.17 shows a memory map for this example.

A 256 -kword $\times 16$-bit EPROM is connected to area 0 . This device is accessed in three states via a 16-bit bus.

Two 32 -kword $\times 8$-bit SRAM devices (SRAM1 and SRAM2) are connected to area 1. These devices are accessed in two states via a 16-bit bus.

One 32 -kword $\times 8$-bit SRAM (SRAM3) is connected to area 2 . This device is accessed via an 8-bit bus, using three-state access with an additional wait state inserted in pin auto-wait mode.


Figure 6.17 Memory Map (Example)


Figure 6.18 Interconnections with Memory (Example)

### 6.3.7 Bus Arbiter Operation

The bus controller has a built-in bus arbiter that arbitrates between different bus masters. There are four bus masters: the CPU, DMA controller (DMAC), refresh controller, and an external bus master. When a bus master has the bus right it can carry out read, write, or refresh access. Each bus master uses a bus request signal to request the bus right. At fixed times the bus arbiter determines priority and uses a bus acknowledge signal to grant the bus to a bus master, which can then operate using the bus.

The bus arbiter checks whether the bus request signal from a bus master is active or inactive, and returns an acknowledge signal to the bus master if the bus request signal is active. When two or more bus masters request the bus, the highest-priority bus master receives an acknowledge signal. The bus master that receives an acknowledge signal can continue to use the bus until the acknowledge signal is deactivated.

The bus master priority order is:
(High) External bus master $>$ refresh controller $>\mathrm{DMAC}>\mathrm{CPU} \quad$ (Low)
The bus arbiter samples the bus request signals and determines priority at all times, but it does not always grant the bus immediately, even when it receives a bus request from a bus master with higher priority than the current bus master. Each bus master has certain times at which it can release the bus to a higher-priority bus master.

CPU: The CPU is the lowest-priority bus master. If the DMAC, refresh controller, or an external bus master requests the bus while the CPU has the bus right, the bus arbiter transfers the bus right to the bus master that requested it. The bus right is transferred at the following times:

- The bus right is transferred at the boundary of a bus cycle. If word data is accessed by two consecutive byte accesses, however, the bus right is not transferred between the two byte accesses.
- If another bus master requests the bus while the CPU is performing internal operations, such as executing a multiply or divide instruction, the bus right is transferred immediately. The CPU continues its internal operations.
- If another bus master requests the bus while the CPU is in sleep mode, the bus right is transferred immediately.

DMAC: When the DMAC receives an activation request, it requests the bus right from the bus arbiter. If the DMAC is bus master and the refresh controller or an external bus master requests the bus, the bus arbiter transfers the bus right from the DMAC to the bus master that requested the bus. The bus right is transferred at the following times.

The bus right is transferred when the DMAC finishes transferring 1 byte or 1 word. A DMAC transfer cycle consists of a read cycle and a write cycle. The bus right is not transferred between the read cycle and the write cycle.

There is a priority order among the DMAC channels. For details see section 8.4.9, MultipleChannel Operation.

Refresh Controller: When a refresh cycle is requested, the refresh controller requests the bus right from the bus arbiter. When the refresh cycle is completed, the refresh controller releases the bus. For details see section 7, Refresh Controller.

External Bus Master: When the BRLE bit is set to 1 in BRCR, the bus can be released to an external bus master. The external bus master has highest priority, and requests the bus right from the bus arbiter by driving the $\overline{\mathrm{BREQ}}$ signal low. Once the external bus master gets the bus, it keeps the bus right until the $\overline{\mathrm{BREQ}}$ signal goes high. While the bus is released to an external bus master, the H8/3052F holds the address bus and data bus control signals ( $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{HWR}}$, and $\overline{\mathrm{LWR}})$ in the high-impedance state, holds the chip select signals high ( $\overline{\mathrm{CS}}_{\mathrm{n}}: \mathrm{n}=7$ to 0$)$, and holds the $\overline{\mathrm{BACK}}$ pin in the low output state.

The bus arbiter samples the $\overline{\mathrm{BREQ}}$ pin at the rise of the system clock $(\phi)$. If $\overline{\mathrm{BREQ}}$ is low, the bus is released to the external bus master at the appropriate opportunity. The $\overline{\mathrm{BREQ}}$ signal should be held low until the $\overline{\mathrm{BACK}}$ signal goes low.

When the $\overline{\mathrm{BREQ}}$ pin is high in two consecutive samples, the $\overline{\mathrm{BACK}}$ signal is driven high to end the bus-release cycle.

Figure 6.19 shows the timing when the bus right is requested by an external bus master during a read cycle in a two-state-access area. There is a minimum interval of two states from when the $\overline{\mathrm{BREQ}}$ signal goes low until the bus is released.


Figure 6.19 External-Bus-Released State (Two-State-Access Area during Read Cycle)

### 6.4.1 Connection to Dynamic RAM and Pseudo-Static RAM

A different bus control signal timing applies when dynamic RAM or pseudo-static RAM is connected to area 3. For details see section 7, Refresh Controller.

### 6.4.2 Register Write Timing

ABWCR, ASTCR, and WCER Write Timing: Data written to ABWCR, ASTCR, or WCER takes effect starting from the next bus cycle. Figure 6.20 shows the timing when an instruction fetched from area 0 changes area 0 from three-state access to two-state access.


Figure 6.20 ASTCR Write Timing

DDR Write Timing: Data written to a data direction register (DDR) to change a $\overline{\mathrm{CS}}_{\mathrm{n}}$ pin from $\overline{\mathrm{CS}}_{\mathrm{n}}$ output to generic input, or vice versa, takes effect starting from the $\mathrm{T}_{3}$ state of the DDR write cycle. Figure 6.21 shows the timing when the $\overline{\mathrm{CS}}_{1}$ pin is changed from generic input to $\overline{\mathrm{CS}}_{1}$ output.


Figure 6.21 DDR Write Timing
BRCR Write Timing: Data written to switch between $A_{23}, A_{22}$, or $A_{21}$ output and generic input or output takes effect starting from the $\mathrm{T}_{3}$ state of the BRCR write cycle. Figure 6.22 shows the timing when a pin is changed from generic input to $\mathrm{A}_{23}, \mathrm{~A}_{22}$, or $\mathrm{A}_{21}$ output.


Figure 6.22 BRCR Write Timing

### 6.4.3 $\overline{\text { BREQ }}$ Input Timing

After driving the $\overline{\mathrm{BREQ}}$ pin low, hold it low until $\overline{\mathrm{BACK}}$ goes low. If $\overline{\mathrm{BREQ}}$ returns to the high level before $\overline{\mathrm{BACK}}$ goes low, the bus arbiter may operate incorrectly.

To terminate the external-bus-released state, hold the $\overline{\mathrm{BREQ}}$ signal high for at least three states. If $\overline{\mathrm{BREQ}}$ is high for too short an interval, the bus arbiter may operate incorrectly.

### 6.4.4 Transition To Software Standby Mode

If contention occurs between a transition to software standby mode and a bus request from an external bus master, the bus may be released for one state just before the transition to software standby mode (see figure 6.23 ). When using software standby mode, clear the BRLE bit to 0 in BRCR before executing the SLEEP instruction.


Figure 6.23 Contention between Bus-Released State and Software Standby Mode

## Section 7 Refresh Controller

### 7.1 Overview

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ has an on-chip refresh controller that enables direct connection of 16 -bit-wide DRAM or pseudo-static RAM (PSRAM).

DRAM or pseudo-static RAM can be directly connected to area 3 of the external address space.
A maximum 128 kbytes can be connected in modes 1 and 2 (1-Mbyte modes). A maximum 2 Mbytes can be connected in modes 3, 4, and 6 (16-Mbyte modes).

Systems that do not need to refresh DRAM or pseudo-static RAM can use the refresh controller as an 8 -bit interval timer.

When the refresh controller is not used, it can be independently halted to conserve power. For details see section 20.6, Module Standby Function.

Note: The refresh function cannot be used in modes 5 and 7.

### 7.1.1 Features

The refresh controller can be used for one of three functions: DRAM refresh control, pseudo-static RAM refresh control, or 8-bit interval timing. Features of the refresh controller are listed below.

Features as a DRAM Refresh Controller

- Enables direct connection of 16 -bit-wide DRAM
- Selection of $2 \overline{\mathrm{CAS}}$ or $2 \overline{\mathrm{WE}}$ mode
- Selection of 8-bit or 9-bit column address multiplexing for DRAM address input Examples:
- 1-Mbit DRAM: 8-bit row address $\times 8$-bit column address
- 4-Mbit DRAM: 9-bit row address $\times 9$-bit column address
— 4-Mbit DRAM: 10 -bit row address $\times 8$-bit column address
- $\overline{\mathrm{CAS}}$-before- $\overline{\mathrm{RAS}}$ refresh control
- Software-selectable refresh interval
- Software-selectable self-refresh mode
- Wait states can be inserted

Features as a Pseudo-Static RAM Refresh Controller

- $\overline{\mathrm{RFSH}}$ signal output for refresh control
- Software-selectable refresh interval
- Software-selectable self-refresh mode
- Wait states can be inserted

Features as an Interval Timer

- Refresh timer counter (RTCNT) can be used as an 8-bit up-counter
- Selection of seven counter clock sources: $\phi / 2, \phi / 8, \phi / 32, \phi / 128, \phi / 512, \phi / 2048, \phi / 4096$
- Interrupts can be generated by compare match between RTCNT and the refresh time constant register (RTCOR)


### 7.1.2 Block Diagram

Figure 7.1 shows a block diagram of the refresh controller.


Figure 7.1 Block Diagram of Refresh Controller

### 7.1.3 Pin Configuration

Table 7.1 summarizes the refresh controller's input/output pins.

## Table 7.1 Refresh Controller Pins

Signal

| Pin | Name | Abbr. | I/O | Function |
| :---: | :---: | :---: | :---: | :---: |
| $\overline{\mathrm{RFSH}}$ | Refresh | $\overline{\text { RFSH }}$ | Output | Goes low during refresh cycles; used to refresh DRAM and PSRAM |
| $\overline{\text { HWR }}$ | Upper write/upper column address strobe | $\overline{\text { UW }} / \overline{\text { UCAS }}$ | Output | Connects to the $\overline{\mathrm{UW}}$ pin of $2 \overline{\mathrm{WE}}$ DRAM or UCAS pin of 2 $\overline{C A S}$ DRAM |
| $\overline{\overline{L W R}}$ | Lower write/lower column address strobe | $\overline{\text { LW/LCAS }}$ | Output | Connects to the $\overline{\text { LW }}$ pin of $2 \overline{\mathrm{WE}}$ DRAM or LCAS pin of 2CAS DRAM |
| $\overline{\mathrm{RD}}$ | Column address strobe/ write enable | $\overline{\mathrm{CAS}} / \overline{\mathrm{WE}}$ | Output | Connects to the $\overline{\mathrm{CAS}}$ pin of $2 \overline{\mathrm{WE}}$ DRAM or $\overline{W E}$ pin of $2 \overline{\mathrm{CAS}}$ DRAM |
| $\overline{\mathrm{CS}_{3}}$ | Row address strobe | $\overline{\text { RAS }}$ | Output | Connects to the $\overline{\text { RAS }}$ pin of DRAM |

### 7.1.4 Register Configuration

Table 7.2 summarizes the refresh controller's registers.
Table 7.2 Refresh Controller Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFAC | Refresh control register | RFSHCR | R/W | H'02 |
| H'FFAD | Refresh timer control/status register | RTMCSR | R/W | H'07 |
| H'FFAE | Refresh timer counter | RTCNT | R/W | H'00 |
| H'FFAF | Refresh time constant register | RTCOR | R/W | H'FF |

Note: * Lower 16 bits of the address.

### 7.2 Register Descriptions

### 7.2.1 Refresh Control Register (RFSHCR)

RFSHCR is an 8-bit readable/writable register that selects the operating mode of the refresh controller.


RFSHCR is initialized to $\mathrm{H}^{\prime} 02$ by a reset and in hardware standby mode.

Bit 7—Self-Refresh Mode (SRFMD): Specifies DRAM or pseudo-static RAM self-refresh during software standby mode. When PSRAME $=1$ and DRAME $=0$, after the SRFMD bit is set to 1 , pseudo-static RAM can be self-refreshed when the H8/3052F enters software standby mode. When PSRAME $=0$ and DRAME $=1$, after the SRFMD bit is set to 1 , DRAM can be selfrefreshed when the H8/3052F enters software standby mode. In either case, the normal access state resumes on exit from software standby mode.

| Bit 7: SRFMD | Description |
| :--- | :--- |
| 0 | DRAM or PSRAM self-refresh is disabled in software standby mode |
|  |  |
| 1 | DRAM or PSRAM self-refresh is enabled in software standby mode |

Bit 6—PSRAM Enable (PSRAME) and Bit 5—DRAM Enable (DRAME): These bits enable or disable connection of pseudo-static RAM and DRAM to area 3 of the external address space.

When DRAM or pseudo-static RAM is connected, the bus cycle and refresh cycle of area 3 consist of three states, regardless of the setting in the access state control register (ASTCR). If AST3 $=0$ in ASTCR, wait states cannot be inserted.

When the PSRAME or DRAME bit is set to 1 , bits $0,2,3$, and 4 in RFSHCR and registers RTMCSR, RTCNT, and RTCOR are write-disabled, except that the CMF flag in RTMCSR can be cleared by writing 0 .

| Bit 6: PSRAME | Bit 5: DRAME | Description |
| :--- | :--- | :--- |
| 0 | 0 | Can be used as an interval timer <br>  <br>  1 |
| 1 | 0 | DRAM and PSRAM cannot be directly connected) |

Bit 4—Strobe Mode Select (CAS/ $\overline{\mathrm{WE}})$ : Selects $2 \overline{\mathrm{CAS}}$ or $2 \overline{\mathrm{WE}}$ mode. The setting of this bit is valid when PSRAME $=0$ and $\operatorname{DRAME}=1$. This bit is write-disabled when the PSRAME or DRAME bit is set to 1 .
Bit 4: CAS/WE Description

| 0 | $2 \overline{\mathrm{WE}}$ mode | (Initial value) |
| :--- | :--- | :---: |
| 1 | $2 \overline{\mathrm{CAS}}$ mode |  |

Rev. 2.0, 03/01, page 150 of 822

Bit 3—Address Multiplex Mode Select (M9/M8): Selects 8-bit or 9-bit column addressing. The setting of this bit is valid when PSRAME $=0$ and DRAME $=1$. This bit is write-disabled when the PSRAME or DRAME bit is set to 1 .

| Bit 3: $\mathbf{M 9} / \overline{\mathbf{M 8}}$ | Description |  |
| :--- | :--- | :--- |
| 0 | 8-bit column address mode | (Initial value) |
| 1 | 9 -bit column address mode |  |

Bit 2-Refresh Pin Enable (RFSHE): Enables or disables refresh signal output from the $\overline{\mathrm{RFSH}}$ pin. This bit is write-disabled when the PSRAME or DRAME bit is set to 1 .

## Bit 2: RFSHE Description

Refresh signal output at the $\overline{\text { RFSH }}$ pin is disabled (the $\overline{\text { RFSH }}$ pin can be used as a generic input/output port) (Initial value)
$1 \quad$ Refresh signal output at the $\overline{\mathrm{RFSH}}$ pin is enabled

Bit 1—Reserved: Read-only bit, always read as 1 .
Bit 0—Refresh Cycle Enable (RCYCE): Enables or disables insertion of refresh cycles.
The setting of this bit is valid when PSRAME $=1$ or DRAME $=1$. When PSRAME $=0$ and DRAME $=0$, refresh cycles are not inserted regardless of the setting of this bit.

| Bit 0: RCYCE | Description |  |
| :--- | :--- | :--- |
| 0 | Refresh cycles are disabled | (Initial value) |
| 1 | Refresh cycles are enabled for area 3 |  |

### 7.2.2 Refresh Timer Control/Status Register (RTMCSR)

RTMCSR is an 8-bit readable/writable register that selects the clock source for RTCNT. It also enables or disables interrupt requests when the refresh controller is used as an interval timer.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CMF | CMIE | CKS2 | CKS1 | CKS0 | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |
| Read/Write | $\underline{R /(W) *}$ | R/W | R/W | R/W | R/W | - | - | - |
|  |  | Compa <br> Enables | Cloc <br> The inter for in <br> match <br> disabl | select <br> bits sel clock ut to RT errupt the CM | 0 <br> an rce T <br> able terrupt |  | CMF |  |

## Compare match flag

Status flag indicating that RTCNT has matched RTCOR

Note: * Only 0 can be written, to clear the flag.

Bits 7 and 6 are initialized by a reset and in standby mode. Bits 5 to 3 are initialized by a reset and in hardware standby mode, but retain their previous values on transition to software standby mode.

Bit 7—Compare Match Flag (CMF): This status flag indicates that the RTCNT and RTCOR values have matched.

| Bit 7: CMF | Description |
| :--- | :--- |
| 0 | [Clearing condition] |
|  | Cleared by reading CMF when CMF $=1$, then writing 0 in CMF |
| 1 | [Setting condition] |
|  | When RTCNT $=$ RTCOR |

Bit 6—Compare Match Interrupt Enable (CMIE): Enables or disables the CMI interrupt requested when the CMF flag is set to 1 in RTMCSR. The CMIE bit is always cleared to 0 when PSRAME $=1$ or DRAME $=1$.

| Bit 6: CMIE | Description |  |
| :--- | :--- | :--- |
| 0 | The CMI interrupt requested by CMF is disabled | (Initial value) |
| 1 | The CMI interrupt requested by CMF is enabled |  |

Bits 5 to 3—Clock Select 2 to 0 (CKS2 to CKS0): These bits select an internal clock source for input to RTCNT. When used for refresh control, the refresh controller outputs a refresh request at periodic intervals determined by compare match between RTCNT and RTCOR. When used as an interval timer, the refresh controller generates CMI interrupts at periodic intervals determined by compare match. These bits are write-disabled when the PSRAME bit or DRAME bit is set to 1 .

| Bit 5: CKS2 | Bit 4: CKS1 | Bit 3: CKS0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Clock input is disabled |
|  |  | 1 | $\phi / 2$ clock source |
|  | 1 | 0 | $\phi / 8$ clock source |
| 1 | 1 | $\phi / 32$ clock source |  |
|  | 0 | 0 | $\phi / 128$ clock source |
|  | 1 | 0 | $\phi / 512$ clock source |

Bits 2 to 0—Reserved: Read-only bits, always read as 1.

### 7.2.3 Refresh Timer Counter (RTCNT)

RTCNT is an 8-bit readable/writable up-counter.


RTCNT is an up-counter that is incremented by an internal clock selected by bits CKS2 to CKS0 in RTMCSR. When RTCNT matches RTCOR (compare match), the CMF flag is set to 1 and RTCNT is cleared to $\mathrm{H}^{\prime} 00$.

RTCNT is write-disabled when the PSRAME bit or DRAME bit is set to 1 . RTCNT is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.

### 7.2.4 Refresh Time Constant Register (RTCOR)

RTCOR is an 8-bit readable/writable register that determines the interval at which RTCNT is compare matched.


RTCOR and RTCNT are constantly compared. When their values match, the CMF flag is set to 1 in RTMCSR, and RTCNT is simultaneously cleared to $\mathrm{H}^{\prime} 00$.

RTCOR is write-disabled when the PSRAME bit or DRAME bit is set to 1. RTCOR is initialized to H'FF by a reset and in hardware standby mode. In software standby mode it retains its previous value.

## $7.3 \quad$ Operation

### 7.3.1 Overview

One of three functions can be selected for the $\mathrm{H} 8 / 3052 \mathrm{~F}$ refresh controller: interfacing to DRAM connected to area 3 , interfacing to pseudo-static RAM connected to area 3 , or interval timing. Table 7.3 summarizes the register settings when these three functions are used.

## Table 7.3 Refresh Controller Settings

| Register Settings |  | Usage |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  | DRAM Interface | PSRAM Interface | Interval Timer |
| RFSHCR | SRFMD | Selects self-refresh mode | Selects self-refresh mode | Cleared to 0 |
|  | PSRAME | Cleared to 0 | Set to 1 | Cleared to 0 |
|  | DRAME | Set to 1 | Cleared to 0 | Cleared to 0 |
|  | CAS/ $\overline{\mathrm{WE}}$ | Selects 2 $\overline{\mathrm{CAS}}$ or 2 $\bar{W} E$ mode | - | - |
|  | M9/M8 | Selects column addressing mode | - | - |
|  | RFSHE | Selects $\overline{\mathrm{RFSH}}$ signal output | Selects $\overline{\text { RFSH }}$ signal output | Cleared to 0 |
|  | RCYCE | Selects insertion of refresh cycles | Selects insertion of refresh cycles | - |
| RTCOR |  | Refresh interval setting | Refresh interval setting | Interrupt interval setting |
| RTMCSR | $\begin{aligned} & \text { CKS2 to } \\ & \text { CKS0 } \end{aligned}$ |  |  |  |
|  | CMF | Set to 1 when <br> RTCNT = RTCOR | $\begin{aligned} & \text { Set to } 1 \text { when } \\ & \text { RTCNT = RTCOR } \end{aligned}$ | $\begin{aligned} & \text { Set to } 1 \text { when } \\ & \text { RTCNT = RTCOR } \end{aligned}$ |
|  | CMIE | Cleared to 0 | Cleared to 0 | Enables or disables interrupt requests |
| P8DDR | P8,DDR | Set to 1 ( $\overline{\mathrm{CS}}_{3}$ output) | Set to 1 ( $\overline{\mathrm{CS}}_{3}$ output) | Set to 0 or 1 |
| ABWCR | ABW3 | Cleared to 0 | - | - |

DRAM Interface: To set up area 3 for connection to 16-bit-wide DRAM, initialize RTCOR, RTMCSR, and RFSHCR in that order, clearing bit PSRAME to 0 and setting bit DRAME to 1 . Set bit $\mathrm{P} 8_{1} \mathrm{DDR}$ to 1 in the port 8 data direction register (P8DDR) to enable $\overline{\mathrm{CS}}_{3}$ output. In ABWCR, make area 3 a 16-bit-access area.

Pseudo-Static RAM Interface: To set up area 3 for connection to pseudo-static RAM, initialize RTCOR, RTMCSR, and RFSHCR in that order, setting bit PSRAME to 1 and clearing bit DRAME to 0 . Set bit $\mathrm{P}_{1}$ DDR to 1 in P8DDR to enable $\overline{\mathrm{CS}}_{3}$ output.

Interval Timer: When PSRAME $=0$ and $\operatorname{DRAME}=0$, the refresh controller operates as an interval timer. After setting RTCOR, select an input clock in RTMCSR and set the CMIE bit to 1. CMI interrupts will be requested at compare match intervals determined by RTCOR and bits CKS2 to CKS0 in RTMCSR.

When setting RTCOR, RTMCSR, and RFSHCR, make sure that PSRAME $=0$ and DRAME $=0$. Writing is disabled when either of these bits is set to 1 .

### 7.3.2 DRAM Refresh Control

Refresh Request Interval and Refresh Cycle Execution: The refresh request interval is determined by the settings of RTCOR and bits CKS2 to CKS0 in RTMCSR. Figure 7.2 illustrates the refresh request interval.


Figure 7.2 Refresh Request Interval (RCYCE = 1)
Refresh requests are generated at regular intervals as shown in figure 7.2, but the refresh cycle is not actually executed until the refresh controller gets the bus right.

Table 7.4 summarizes the relationship among area 3 settings, DRAM read/write cycles, and refresh cycles.

Table 7.4 Area 3 Settings, DRAM Access Cycles, and Refresh Cycles

| Area 3 Settings | Read/Write Cycle by CPU <br> or DMAC | Refresh Cycle |
| :--- | :--- | :--- |
| 2-state-access area | $\bullet 3$ states | $\bullet 3$ states |
| (AST3 $=0)$ | $\bullet$ Wait states cannot be inserted | $\bullet$ Wait states cannot be inserted |
| 3-state-access area | $\bullet 3$ states | $\bullet 3$ states |
| $($ AST3 $=1)$ | $\bullet$ Wait states can be inserted | $\bullet$ Wait states can be inserted |

To insert refresh cycles, set the RCYCE bit to 1 in RFSHCR. Figure 7.3 shows the state transitions for execution of refresh cycles.

When the first refresh request occurs after exit from the reset state or standby mode, the refresh controller does not execute a refresh cycle, but goes into the refresh request pending state. Note this point when using a DRAM that requires a refresh cycle for initialization.

When a refresh request occurs in the refresh request pending state, the refresh controller acquires the bus right, then executes a refresh cycle. If another refresh request occurs during execution of the refresh cycle, it is ignored.


Note: * A refresh request is ignored if it occurs while the refresh controller is requesting the bus right or executing a refresh cycle.

Figure 7.3 State Transitions for Refresh Cycle Execution

Address Multiplexing: Address multiplexing depends on the setting of the M9/M8 bit in RFSHCR, as described in table 7.5. Figure 7.4 shows the address output timing. Address output is multiplexed only in area 3.

## Table 7.5 Address Multiplexing




Figure 7.4 Multiplexed Address Output (Example without Wait States)
$2 \overline{\mathrm{CAS}}$ and $2 \overline{\mathrm{WE}}$ Modes: The CAS/ $\overline{\mathrm{WE}}$ bit in RFSHCR can select two control modes for 16-bitwide DRAM: one using $\overline{\text { UCAS }}$ and $\overline{\mathrm{LCAS}}$; the other using $\overline{\mathrm{UW}}$ and $\overline{\mathrm{LW}}$. These DRAM pins correspond to $\mathrm{H} 8 / 3052 \mathrm{~F}$ pins as shown in table 7.6.

Table 7.6 DRAM Pins and H8/3052F Pins

| H8/3052F Pin | DRAM Pin |  |
| :---: | :---: | :---: |
|  | $\overline{C A S / \overline{W E}}=0$ (2 $\overline{\text { WE }}$ Mode) | CAS/ $\overline{\mathrm{WE}}=1$ (2 $\overline{\mathrm{CAS}}$ Mode) |
| $\overline{\text { HWR }}$ | UW | $\overline{\text { UCAS }}$ |
| $\overline{\overline{\text { LWR }}}$ | $\overline{\text { LW }}$ | LCAS |
| $\overline{\overline{R D}}$ | $\overline{\text { CAS }}$ | $\overline{\text { WE }}$ |
| $\overline{\overline{C S}_{3}}$ | $\overline{\mathrm{RAS}}$ | $\overline{\text { RAS }}$ |

Figure 7.5 (1) shows the interface timing for $2 \overline{\mathrm{WE}}$ DRAM. Figure 7.5 (2) shows the interface timing for $2 \overline{\mathrm{CAS}}$ DRAM.


Figure 7.5 DRAM Control Signal Output Timing (1) ( $2 \overline{\mathrm{WE}}$ Mode)


Note: * 16-bit access

Figure 7.5 DRAM Control Signal Output Timing (2) (2 $\overline{\mathrm{CAS}}$ Mode)
Refresh Cycle Priority Order: When there are simultaneous bus requests, the priority order is:
(High) External bus master > refresh controller > DMA controller > CPU
For details see section 6.3.7, Bus Arbiter Operation.
Wait State Insertion: When bit AST3 is set to 1 in ASTCR, bus controller settings can cause wait states to be inserted into bus cycles and refresh cycles. For details see section 6.3.5, Wait Modes.

Self-Refresh Mode: Some DRAM devices have a self-refresh function. After the SRFMD bit is set to 1 in RFSHCR, when a transition to software standby mode occurs, the $\overline{\mathrm{CAS}}$ and $\overline{\text { RAS }}$ outputs go low in that order so that the DRAM self-refresh function can be used. On exit from software standby mode, the $\overline{\mathrm{CAS}}$ and $\overline{\mathrm{RAS}}$ outputs both go high.

Table 7.7 shows the pin states in software standby mode. Figure 7.6 shows the signal output timing.

Table 7.7 Pin States in Software Standby Mode (PSRAME = 0, DRAME $=1$ )
Software Standby Mode

|  | SRFMD $=\mathbf{0}$ |  |  | SRFMD $=\mathbf{1}$ (self-refresh mode) |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Signal | CAS $/ \overline{\mathrm{WE}}=\mathbf{0}$ | CAS $/ \overline{\mathrm{WE}}=\mathbf{1}$ |  | CAS $/ \overline{\mathrm{WE}}=\mathbf{0}$ | CAS $/ \overline{\mathbf{W E}}=\mathbf{1}$ |
| $\overline{\mathrm{HWR}}$ | High-impedance | High-impedance | High | Low |  |
| $\overline{\overline{\mathrm{LWR}}}$ | High-impedance | High-impedance |  | High | Low |
| $\overline{\mathrm{RD}}$ | High-impedance | High-impedance | Low | High |  |
| $\overline{\overline{\mathrm{CS}}_{3}}$ | High | High | Low | Low |  |
| $\overline{\mathrm{RFSH}}$ | High | High | Low | Low |  |



Figure 7.6 Signal Output Timing in Self-Refresh Mode (PSRAME =0, DRAME =1)

Operation in Power-Down State: The refresh controller operates in sleep mode. It does not operate in hardware standby mode. In software standby mode RTCNT is initialized, but RFSHCR, RTMCSR bits 5 to 3 , and RTCOR retain their settings prior to the transition to software standby mode.

Example 1: Connection to 2 $\overline{\mathbf{W E}}$ 1-Mbit DRAM (1-Mbyte Mode): Figure 7.7 shows typical interconnections to a $2 \overline{\mathrm{WE}} 1-\mathrm{Mbit}$ DRAM, and the corresponding address map. Figure 7.8 shows a setup procedure to be followed by a program for this example. After power-up the DRAM must be refreshed to initialize its internal state. Initialization takes a certain length of time, which can be measured by using an interrupt from another timer module, or by counting the number of times RTMCSR bit 7 (CMF) is set. Note that no refresh cycle is executed for the first refresh request after exit from the reset state or standby mode (the first time the CMF flag is set; see figure 7.3). When using this example, check the DRAM device characteristics carefully and use a procedure that fits them.


Figure 7.7 Interconnections and Address Map for 2WE 1-Mbit DRAM (Example)


DRAM can be accessed

Figure 7.8 Setup Procedure for $2 \overline{\mathrm{WE}}$ 1-Mbit DRAM (1-Mbyte Mode)

Example 2: Connection to 2WE 4-Mbit DRAM (16-Mbyte Mode): Figure 7.9 shows typical interconnections to a single $2 \overline{\mathrm{WE}} 4$-Mbit DRAM, and the corresponding address map. Figure 7.10 shows a setup procedure to be followed by a program for this example.

The DRAM in this example has 10-bit row addresses and 8-bit column addresses. Its address area is H'600000 to H'67FFFF.


Figure 7.9 Interconnections and Address Map for 2 $\overline{\mathbf{W E}} 4$-Mbit DRAM (Example)


Figure 7.10 Setup Procedure for 2WE 4-Mbit DRAM with 10-Bit Row Address and 8-Bit Column Address (16-Mbyte Mode)

Example 3: Connection to 2 $\overline{\text { CAS }}$ 4-Mbit DRAM (16-Mbyte Mode): Figure 7.11 shows typical interconnections to a single $2 \overline{\mathrm{CAS}} 4-\mathrm{Mbit}$ DRAM, and the corresponding address map.

Figure 7.12 shows a setup procedure to be followed by a program for this example.
The DRAM in this example has 9-bit row addresses and 9-bit column addresses. Its address area is H'600000 to H'67FFFF.


Figure 7.11 Interconnections and Address Map for 2 $\overline{\mathbf{C A S}}$ 4-Mbit DRAM (Example)


Figure 7.12 Setup Procedure for 2 $\overline{\text { CAS }}$ 4-Mbit DRAM with 9-Bit Row Address and 9-Bit Column Address (16-Mbyte Mode)

Example 4: Connection to Multiple 4-Mbit DRAM Chips (16-Mbyte Mode): Figure 7.13 shows an example of interconnections to two $2 \overline{\mathrm{CAS}} 4-\mathrm{Mbit}$ DRAM chips, and the corresponding address map. Up to four DRAM chips can be connected to area 3 by decoding upper address bits $\mathrm{A}_{19}$ and $\mathrm{A}_{20}$.

Figure 7.14 shows a setup procedure to be followed by a program for this example. The DRAM in this example has 9 -bit row addresses and 9 -bit column addresses. Both chips must be refreshed simultaneously, so the $\overline{\mathrm{RFSH}}$ pin must be used.


Figure 7.13 Interconnections and Address Map for Multiple 2 $\overline{\mathrm{CAS}} 4$-Mbit DRAM Chips (Example)


Figure 7.14 Setup Procedure for Multiple 2 $\overline{\text { CAS }}$ 4-Mbit DRAM Chips with 9-Bit Row Address and 9-Bit Column Address (16-Mbyte Mode)

### 7.3.3 Pseudo-Static RAM Refresh Control

Refresh Request Interval and Refresh Cycle Execution: The refresh request interval is determined as in a DRAM interface, by the settings of RTCOR and bits CKS2 to CKS0 in RTMCSR. The numbers of states required for pseudo-static RAM read/write cycles and refresh cycles are the same as for DRAM (see table 7.4). The state transitions are as shown in figure 7.3.

Pseudo-Static RAM Control Signals: Figure 7.15 shows the control signals for pseudo-static RAM read, write, and refresh cycles.


Figure 7.15 Pseudo-Static RAM Control Signal Output Timing

Refresh Cycle Priority Order: When there are simultaneous bus requests, the priority order is:
(High) External bus master > refresh controller > DMA controller > CPU
For details see section 6.3.7, Bus Arbiter Operation.
Wait State Insertion: When bit AST3 is set to 1 in ASTCR, the wait state controller (WSC) can insert wait states into bus cycles and refresh cycles. For details see section 6.3.5, Wait Modes.

Self-Refresh Mode: Some pseudo-static RAM devices have a self-refresh function. After the SRFMD bit is set to 1 in RFSHCR, when a transition to software standby mode occurs, the $\mathrm{H} 8 / 3052 \mathrm{~F}, \overline{\mathrm{CS}}_{3}$ output goes high and its $\overline{\mathrm{RFSH}}$ output goes low so that the pseudo-static RAM self-refresh function can be used. On exit from software standby mode, the $\overline{\mathrm{RFSH}}$ output goes high.

Table 7.8 shows the pin states in software standby mode. Figure 7.16 shows the signal output timing.

Table 7.8 Pin States in Software Standby Mode (PSRAME = 1, DRAME $=0$ )
Software Standby Mode

| Signal | SRFMD $=\mathbf{0}$ | SRFMD = 1 (Self-Refresh Mode) |
| :--- | :--- | :--- |
| $\overline{\mathrm{CS}}_{3}$ | High | High |
| $\overline{\mathrm{RD}}$ | High-impedance | High-impedance |
| $\overline{\mathrm{HWR}}$ | High-impedance | High-impedance |
| $\overline{\overline{\mathrm{LWR}}}$ | High-impedance | High-impedance |
| $\overline{\mathrm{RFSH}}$ | High | Low |



Figure 7.16 Signal Output Timing in Self-Refresh Mode (PSRAME = 1, DRAME = 0)

Operation in Power-Down State: The refresh controller operates in sleep mode. It does not operate in hardware standby mode. In software standby mode RTCNT is initialized, but RFSHCR, RTMCSR bits 5 to 3, and RTCOR retain their settings prior to the transition to software standby mode.

Example: Pseudo-static RAM may have separate $\overline{\mathrm{OE}}$ and $\overline{\mathrm{RFSH}}$ pins, or these may be combined into a single $\overline{\mathrm{OE}} / \overline{\mathrm{RFSH}}$ pin. Figure 7.17 shows an example of a circuit for generating an $\overline{\mathrm{OE}} / \overline{\mathrm{RFSH}}$ signal. Check the device characteristics carefully, and design a circuit that fits them. Figure 7.18 shows a setup procedure to be followed by a program.


Figure 7.17 Interconnection to Pseudo-Static RAM with $\overline{\mathrm{OE}} / \overline{\mathrm{RFSH}}$ Signal (Example)
Rev. 2.0, 03/01, page 174 of 822


Figure 7.18 Setup Procedure for Pseudo-Static RAM

### 7.3.4 Interval Timer

To use the refresh controller as an interval timer, clear the PSRAME and DRAME both to 0 . After setting RTCOR, select a clock source with bits CKS2 to CKS0 in RTMCSR, and set the CMIE bit to 1 .

Timing of Setting of Compare Match Flag and Clearing by Compare Match: The CMF flag in RTCSR is set to 1 by a compare match signal output when the RTCOR and RTCNT values match. The compare match signal is generated in the last state in which the values match (when RTCNT is updated from the matching value to a new value). Accordingly, when RTCNT and RTCOR match, the compare match signal is not generated until the next counter clock pulse. Figure 7.19 shows the timing.


Figure 7.19 Timing of Setting of CMF Flag
Operation in Power-Down State: The interval timer function operates in sleep mode. It does not operate in hardware standby mode. In software standby mode RTCNT and RTMCSR bits 7 and 6 are initialized, but RTMCSR bits 5 to 3 and RTCOR retain their settings prior to the transition to software standby mode.

Contention between RTCNT Write and Counter Clear: If a counter clear signal occurs in the $\mathrm{T}_{3}$ state of an RTCNT write cycle, clearing of the counter takes priority and the write is not performed. See figure 7.20.


Figure 7.20 Contention between RTCNT Write and Clear

Contention between RTCNT Write and Increment: If an increment pulse occurs in the $\mathrm{T}_{3}$ state of an RTCNT write cycle, writing takes priority and RTCNT is not incremented. See figure 7.21.


Figure 7.21 Contention between RTCNT Write and Increment

Contention between RTCOR Write and Compare Match: If a compare match occurs in the $\mathrm{T}_{3}$ state of an RTCOR write cycle, writing takes priority and the compare match signal is inhibited. See figure 7.22.


Figure 7.22 Contention between RTCOR Write and Compare Match
RTCNT Operation at Internal Clock Source Switchover: Switching internal clock sources may cause RTCNT to increment, depending on the switchover timing. Table 7.9 shows the relation between the time of the switchover (by writing to bits CKS2 to CKSO) and the operation of RTCNT.

The RTCNT input clock is generated from the internal clock source by detecting the falling edge of the internal clock. If a switchover is made from a high clock source to a low clock source, as in case No. 3 in table 7.9, the switchover will be regarded as a falling edge, an RTCNT clock pulse will be generated, and RTCNT will be incremented.

Table 7.9 Internal Clock Switchover and RTCNT Operation

## CKS2 to CKS0

No. Write Timing
1 Low $\rightarrow$ low switchover*1
RTCNT Operation


CKS bits rewritten
2 Low $\rightarrow$ high switchover*2


CKS bits rewritten

CKS2 to CKS0
No. Write Timing
RTCNT Operation
3 High $\rightarrow$ low switchover*3


CKS bits rewritten
$4 \quad$ High $\rightarrow$ high switchover


CKS bits rewritten
Notes: 1. Including switchovers from a low clock source to the halted state, and from the halted state to a low clock source.
2. Including switchover from the halted state to a high clock source.
3. Including switchover from a high clock source to the halted state.
4. The switchover is regarded as a falling edge, causing RTCNT to increment.

### 7.4 Interrupt Source

Compare match interrupts (CMI) can be generated when the refresh controller is used as an interval timer. Compare match interrupt requests are masked/unmasked with the CMIE bit of RTMCSR.

### 7.5 Usage Notes

When using the DRAM or pseudo-static RAM refresh function, note the following points:
With the refresh controller, if directly connected DRAM or PSRAM is disconnected*, the $\mathrm{P} 8_{0} / \overline{\mathrm{RFSH}} / \overline{\mathrm{IRQ}}_{0}$ pin and the $\mathrm{P}_{1} / \overline{\mathrm{CS}}_{3} / \overline{\mathrm{IRQ}}_{1}$ pin may both become low-level outputs simultaneously.

Note: * When the DRAM enable bit (DRAME) or PSRAM enable bit (PSRAME) in the refresh control register (RFSHCR) is cleared to 0 after being set to 1 .


Figure 7.23 Operation when DRAM/PSRAM Connection is Switched
Refresh cycles are not executed while the bus is released, during software standby mode, and when a bus cycle is greatly prolonged by insertion of wait states. When these conditions occur, other means of refreshing are required.

If refresh requests occur while the bus is released, the first request is held and one refresh cycle is executed after the bus-released state ends. Figure 7.24 shows the bus cycles in this case.


Figure 7.24 Refresh Cycles when Bus is Released
If a bus cycle is prolonged by insertion of wait states, the first refresh request is held, as in the busreleased state.

If there is contention with a bus request from an external bus master when making a transition to software standby mode, a one-state bus-released state may occur immediately before the transition to software standby mode (see figure 7.25).

When using software standby mode, clear the BRLE bit to 0 in BRCR before executing the SLEEP instruction.

When making a transition to self-refresh mode, the strobe waveform output may not be guaranteed due to the same kind of contention. This, too, can be prevented by clearing the BRLE bit to 0 in BRCR.


Figure 7.25 Contention between Bus-Released State and Software Standby Mode

## Section 8 DMA Controller

### 8.1 Overview

The H8/3052F has an on-chip DMA controller (DMAC) that can transfer data on up to four channels.

When the DMA controller is not used, it can be independently halted to conserve power. For details see section 20.6, Module Standby Function.

### 8.1.1 Features

DMAC features are listed below.

- Selection of short address mode or full address mode

Short address mode

- 8 -bit source address and 24 -bit destination address, or vice versa
- Maximum four channels available
- Selection of I/O mode, idle mode, or repeat mode

Full address mode

- 24-bit source and destination addresses
- Maximum two channels available
- Selection of normal mode or block transfer mode
- Directly addressable 16-Mbyte address space
- Selection of byte or word transfer
- Activation by internal interrupts, external requests, or auto-request (depending on transfer mode)
- 16-bit integrated timer unit (ITU) compare match/input capture interrupts (four)
- Serial communication interface (SCI channel 0) transmit-data-empty/receive-data-full interrupts
- External requests
- Auto-request


### 8.1.2 Block Diagram

Figure 8.1 shows a DMAC block diagram.


Figure 8.1 Block Diagram of DMAC

### 8.1.3 Functional Overview

Table 8.1 gives an overview of the DMAC functions.
Table 8.1 DMAC Functional Overview


## Address

Reg. Length

| Transfer | Mode | Activation | Source | Destination |
| :---: | :---: | :---: | :---: | :---: |
| Full address mode | Normal mode <br> - Auto-request <br> - Retains the transfer request internally <br> - Executes a specified number ( 1 to 65,536 ) of transfers continuously <br> - Selection of burst mode or cycle-steal mode <br> - External request <br> - Transfers one byte or one word per request <br> - Executes 1 to 65,536 transfers | - Auto-request <br> - External request | 24 | 24 |
|  | Block transfer <br> - Transfers one block of a specified size per request <br> - Executes 1 to 65,536 transfers <br> - Allows either the source or destination to be a fixed block area <br> - Block size can be 1 to 255 bytes or words | - Compare match/ input capture A interrupts from ITU channels 0 to 3 <br> - External request | 24 | 24 |

### 8.1.4 Pin Configuration

Table 8.2 lists the DMAC pins.
Table 8.2 DMAC Pins

| Channel | Name | Abbre- <br> viation | Input/ <br> Output | Function |
| :--- | :--- | :--- | :--- | :--- |
| 0 | DMA request 0 | $\overline{\mathrm{DREQ}}_{0}$ | Input | External request for DMAC channel 0 |
|  | Transfer end 0 | $\overline{\mathrm{TEND}}_{0}$ | Output | Transfer end on DMAC channel 0 |
| 1 | DMA request 1 | $\overline{\mathrm{DREQ}}_{1}$ | Input | External request for DMAC channel 1 |
|  | Transfer end 1 | $\overline{\mathrm{TEND}}_{1}$ | Output | Transfer end on DMAC channel 1 |

Note: External requests cannot be made to channel A in short address mode.

### 8.1.5 Register Configuration

Table 8.3 lists the DMAC registers.

Table 8.3 DMAC Registers

| Channel | Address* | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | H'FF20 | Memory address register OAR | MAROAR | R/W | Undetermined |
|  | H'FF21 | Memory address register OAE | MAROAE | R/W | Undetermined |
|  | H'FF22 | Memory address register OAH | MAROAH | R/W | Undetermined |
|  | H'FF23 | Memory address register OAL | MAROAL | R/W | Undetermined |
|  | H'FF26 | I/O address register 0A | IOAROA | R/W | Undetermined |
|  | H'FF24 | Execute transfer count register 0AH | ETCROAH | R/W | Undetermined |
|  | H'FF25 | Execute transfer count register 0AL | ETCROAL | R/W | Undetermined |
|  | H'FF27 | Data transfer control register 0A | DTCR0A | R/W | H'00 |
|  | H'FF28 | Memory address register OBR | MAROBR | R/W | Undetermined |
|  | H'FF29 | Memory address register OBE | MAR0BE | R/W | Undetermined |
|  | H'FF2A | Memory address register OBH | MAROBH | R/W | Undetermined |
|  | H'FF2B | Memory address register OBL | MAROBL | R/W | Undetermined |
|  | H'FF2E | I/O address register OB | IOAROB | R/W | Undetermined |
|  | H'FF2C | Execute transfer count register OBH | ETCROBH | R/W | Undetermined |
|  | H'FF2D | Execute transfer count register OBL | ETCROBL | R/W | Undetermined |
|  | H'FF2F | Data transfer control register 0B | DTCR0B | R/W | H'00 |
| 1 | H'FF30 | Memory address register 1AR | MAR1AR | R/W | Undetermined |
|  | H'FF31 | Memory address register 1AE | MAR1AE | R/W | Undetermined |
|  | H'FF32 | Memory address register 1AH | MAR1AH | R/W | Undetermined |
|  | H'FF33 | Memory address register 1AL | MAR1AL | R/W | Undetermined |
|  | H'FF36 | I/O address register 1A | IOAR1A | R/W | Undetermined |
|  | H'FF34 | Execute transfer count register 1AH | ETCR1AH | R/W | Undetermined |
|  | H'FF35 | Execute transfer count register 1AL | ETCR1AL | R/W | Undetermined |
|  | H'FF37 | Data transfer control register 1A | DTCR1A | R/W | H'00 |
|  | H'FF38 | Memory address register 1BR | MAR1BR | R/W | Undetermined |
|  | H'FF39 | Memory address register 1BE | MAR1BE | R/W | Undetermined |
|  | H'FF3A | Memory address register 1BH | MAR1BH | R/W | Undetermined |
|  | H'FF3B | Memory address register 1BL | MAR1BL | R/W | Undetermined |
|  | H'FF3E | I/O address register 1B | IOAR1B | R/W | Undetermined |
|  | H'FF3C | Execute transfer count register 1BH | ETCR1BH | R/W | Undetermined |
|  | H'FF3D | Execute transfer count register 1BL | ETCR1BL | R/W | Undetermined |
|  | H'FF3F | Data transfer control register 1B | DTCR1B | R/W | H'00 |

Note: * The lower 16 bits of the address are indicated.

### 8.2 Register Descriptions (Short Address Mode)

In short address mode, transfers can be carried out independently on channels A and B. Short address mode is selected by bits DTS2A and DTS1A in data transfer control register A (DTCRA) as indicated in table 8.4.

Table 8.4 Selection of Short and Full Address Modes

| Channel | Bit 2: <br> DTS2A | Bit 1: <br> DTS1A | Description |
| :--- | :--- | :--- | :--- | | 0 | 1 | 1 | DMAC channel 0 operates as one channel in full address <br> mode |
| :--- | :--- | :--- | :--- |
| Other than above | DMAC channels 0A and 0B operate as two independent <br> channels in short address mode |  |  |
| 1 | 1 | 1 | DMAC channel 1 operates as one channel in full address <br> mode |
|  | Other than above | DMAC channels 1A and 1B operate as two independent <br> channels in short address mode |  |

### 8.2.1 Memory Address Registers (MAR)

A memory address register (MAR) is a 32-bit readable/writable register that specifies a source or destination address. The transfer direction is determined automatically from the activation source.

An MAR consists of four 8-bit registers designated MARR, MARE, MARH, and MARL. All bits of MARR are reserved: they cannot be modified and always return an undetermined value when read.


Bit
Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Read/Write


Source or destination address

An MAR functions as a source or destination address register depending on how the DMAC is activated: as a destination address register if activation is by a receive-data-full interrupt from the serial communication interface (SCI) (channel 0), and as a source address register otherwise.

The MAR value is incremented or decremented each time one byte or word is transferred, automatically updating the source or destination memory address. For details, see section 8.2.4, Data Transfer Control Registers (DTCR).

The MARs are not initialized by a reset or in standby mode.

### 8.2.2 I/O Address Registers (IOAR)

An I/O address register (IOAR) is an 8 -bit readable/writable register that specifies a source or destination address. The IOAR value is the lower 8 bits of the address. The upper 16 address bits are all 1 (H'FFFF).


An IOAR functions as a source or destination address register depending on how the DMAC is activated: as a source address register if activation is by a receive-data-full interrupt from the SCI (channel 0), and as a destination address register otherwise.

The IOAR value is held fixed. It is not incremented or decremented when a transfer is executed.
The IOARs are not initialized by a reset or in standby mode.

### 8.2.3 Execute Transfer Count Registers (ETCR)

An execute transfer count register (ETCR) is a 16-bit readable/writable register that specifies the number of transfers to be executed. These registers function in one way in I/O mode and idle mode, and another way in repeat mode.

- I/O mode and idle mode

Bit

Initial value
Read/Write

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Undetermined
R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Transfer counter

In I/O mode and idle mode, ETCR functions as a 16-bit counter. The count is decremented by 1 each time one transfer is executed. The transfer ends when the count reaches $\mathrm{H}^{\prime} 0000$.

- Repeat mode


In repeat mode, ETCRH functions as an 8-bit transfer counter and ETCRL holds the initial transfer count. ETCRH is decremented by 1 each time one transfer is executed. When ETCRH reaches H'00, the value in ETCRL is reloaded into ETCRH and the same operation is repeated. The ETCRs are not initialized by a reset or in standby mode.

### 8.2.4 Data Transfer Control Registers (DTCR)

A data transfer control register (DTCR) is an 8-bit readable/writable register that controls the operation of one DMAC channel.


The DTCRs are initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.
Bit 7-Data Transfer Enable (DTE): Enables or disables data transfer on a channel. When the DTE bit is set to 1 , the channel waits for a transfer to be requested, and executes the transfer when activated as specified by bits DTS2 to DTS0. When DTE is 0 , the channel is disabled and does not accept transfer requests. DTE is set to 1 by reading the register when DTE is 0 , then writing 1 .

| Bit 7: DTE | Description |
| :--- | :--- |
| 0 | Data transfer is disabled. In I/O mode or idle mode, DTE is cleared to 0 when <br> the specified number of transfers have been completed. <br> (Initial value) |
| 1 | Data transfer is enabled |

If DTIE is set to 1, a CPU interrupt is requested when DTE is cleared to 0 .

Bit 6-Data Transfer Size (DTSZ): Selects the data size of each transfer.
Bit 6: DTSZ Description

| 0 | Byte-size transfer | (Initial value) |
| :--- | :--- | :--- |
| 1 | Word-size transfer |  |

Bit 5—Data Transfer Increment/Decrement (DTID): Selects whether to increment or decrement the memory address register (MAR) after a data transfer in I/O mode or repeat mode.

| Bit 5: DTID | Description |
| :---: | :---: |
| 0 | MAR is incremented after each data transfer <br> - If DTSZ $=0$, MAR is incremented by 1 after each transfer <br> - If DTSZ $=1$, MAR is incremented by 2 after each transfer |
| 1 | MAR is decremented after each data transfer <br> - If DTSZ $=0$, MAR is decremented by 1 after each transfer <br> - If $D T S Z=1, M A R$ is decremented by 2 after each transfer |

MAR is not incremented or decremented in idle mode.

Bit 4-Repeat Enable (RPE): Selects whether to transfer data in I/O mode, idle mode, or repeat mode.

| Bit 4: RPE | Bit 3: DTIE | Description |  |
| :--- | :--- | :--- | :--- |
| 0 | 0 | I/O mode | (Initial value) |
| 1 |  |  |  |
| 1 | 0 | Repeat mode |  |
| 1 | Idle mode |  |  |

Operations in these modes are described in sections 8.4.2, I/O Mode, 8.4.3, Idle Mode, and 8.4.4, Repeat Mode.

Bit 3-Data Transfer Interrupt Enable (DTIE): Enables or disables the CPU interrupt (DEND) requested when the DTE bit is cleared to 0 .

Bit 3: DTIE Description

| 0 | The DEND interrupt requested by DTE is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | The DEND interrupt requested by DTE is enabled |  |

Bits 2 to 0—Data Transfer Select (DTS2, DTS1, DTS0): These bits select the data transfer activation source. Some of the selectable sources differ between channels A and B.

Note: Refer to 8.3.4, Data Transfer Control Registers (DTCR).

| Bit 2: DTS2 | Bit 1: DTS1 | Bit 0: DTS0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Compare match/input capture A interrupt from ITU <br> channel 0 |
|  |  | 1 | Compare match/input capture A interrupt from ITU <br> channel 1 |
|  | 1 | 0 | Compare match/input capture A interrupt from ITU <br> channel 2 |
|  | 1 | 0 | Compare match/input capture A interrupt from ITU <br> channel 3 |
| 1 | 0 | 0 | Transmit-data-empty interrupt from SCI channel 0 |
|  | 1 | 1 | Receive-data-full interrupt from SCl channel 0 |

The same internal interrupt can be selected as an activation source for two or more channels at once. In that case the channels are activated in a priority order, highest-priority channel first. For the priority order, see section 8.4.9, Multiple-Channel Operation.

When a channel is enabled ( $\mathrm{DTE}=1$ ), its selected DMAC activation source cannot generate a CPU interrupt.

### 8.3 Register Descriptions (Full Address Mode)

In full address mode the A and B channels operate together. Full address mode is selected as indicated in table 8.4.

### 8.3.1 Memory Address Registers (MAR)

A memory address register (MAR) is a 32 -bit readable/writable register. MARA functions as the source address register of the transfer, and MARB as the destination address register.

An MAR consists of four 8 -bit registers designated MARR, MARE, MARH, and MARL. All bits of MARR are reserved: they cannot be modified and always return an undetermined value when read.

| Bit | 31 | 30 | 29 | 28 | 27 | 26 | 25 | 24 | 23 | 22 | 21 | 20 | 19 | 18 | 17 | 16 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  | Undetermined |  |  |  |  |  |  |  |  |
| Read/Write |  | - | - | - |  | - | - |  |  | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | MARR |  |  |  |  |  |  |  | MARE |  |  |  |  |  |  |  |



The MAR value is incremented or decremented each time one byte or word is transferred, automatically updating the source or destination memory address. For details, see section 8.3.4, Data Transfer Control Registers (DTCR).

The MARs are not initialized by a reset or in standby mode.

### 8.3.2 I/O Address Registers (IOAR)

The I/O address registers (IOARs) are not used in full address mode.

### 8.3.3 Execute Transfer Count Registers (ETCR)

An execute transfer count register (ETCR) is a 16-bit readable/writable register that specifies the number of transfers to be executed. The functions of these registers differ between normal mode and block transfer mode.

- Normal mode

ETCRA


ETCRB: Is not used in normal mode.

In normal mode ETCRA functions as a 16-bit transfer counter. The count is decremented by 1 each time one transfer is executed. The transfer ends when the count reaches H'0000. ETCRB is not used.

- Block transfer mode


## ETCRA

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

## ETCRB

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Undetermined
Read/Write

In block transfer mode, ETCRAH functions as an 8-bit block size counter. ETCRAL holds the initial block size. ETCRAH is decremented by 1 each time one byte or word is transferred. When the count reaches H'00, ETCRAH is reloaded from ETCRAL. Blocks consisting of an arbitrary number of bytes or words can be transferred repeatedly by setting the same initial block size value in ETCRAH and ETCRAL.

In block transfer mode ETCRB functions as a 16-bit block transfer counter. ETCRB is decremented by 1 each time one block is transferred. The transfer ends when the count reaches H'0000.

The ETCRs are not initialized by a reset or in standby mode.

### 8.3.4 Data Transfer Control Registers (DTCR)

The data transfer control registers (DTCRs) are 8-bit readable/writable registers that control the operation of the DMAC channels. A channel operates in full address mode when bits DTS2A and DTS1A are both set to 1 in DTCRA. DTCRA and DTCRB have different functions in full address mode.

## DTCRA

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DTE | DTSZ | SAID | SAIDE | DTIE | DTS2A | DTS1A | DTS0A |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | sfer en <br> disab fer | e |  |  |  |  |  | transfer <br> ct 0 A <br> cts block |
|  |  | ransfer <br> byte or <br> ize |  | inter <br> Enab CPU of the | or d errup ansfe | les the the end |  | mode |
|  |  |  | Source address increment/decrement Source address increment/ decrement enable These bits select whether the source address register (MARA) is incremented, decremented, or held fixed during the data transfer |  |  | Data transfer select <br> 2A and 1A <br> These bits must both be set to 1 |  |  |

DTCRA is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.
Bit 7—Data Transfer Enable (DTE): Together with the DTME bit in DTCRB, this bit enables or disables data transfer on the channel. When the DTME and DTE bits are both set to 1 , the channel is enabled. If auto-request is specified, data transfer begins immediately. Otherwise, the channel waits for transfers to be requested. When the specified number of transfers have been completed, the DTE bit is automatically cleared to 0 . When DTE is 0 , the channel is disabled and does not accept transfer requests. DTE is set to 1 by reading the register when DTE is 0 , then writing 1 .

| Bit 7: DTE | Description |
| :--- | :--- |
| 0 | Data transfer is disabled (DTE is cleared to 0 when the specified number of <br> transfers have been completed) |
| 1 | Data transfer is enabled |

If DTIE is set to 1, a CPU interrupt is requested when DTE is cleared to 0 .

Bit 6-Data Transfer Size (DTSZ): Selects the data size of each transfer.

| Bit 6: DTSZ | Description |  |
| :--- | :--- | :--- |
| 0 | Byte-size transfer | (Initial value) |
| 1 | Word-size transfer |  |

Bit 5—Source Address Increment/Decrement (SAID) and Bit 4—Source Address
Increment/Decrement Enable (SAIDE): These bits select whether the source address register (MARA) is incremented, decremented, or held fixed during the data transfer.

| Bit 5: SAID | Bit 4: SAIDE | Description |
| :---: | :---: | :---: |
| 0 | 0 | MARA is held fixed (Initial value) |
|  | 1 | MARA is incremented after each data transfer <br> - If DTSZ $=0$, MARA is incremented by 1 after each transfer <br> - If DTSZ $=1$, MARA is incremented by 2 after each transfer |
| 1 | 0 | MARA is held fixed |
|  | 1 | MARA is decremented after each data transfer <br> - If DTSZ $=0$, MARA is decremented by 1 after each transfer <br> - If DTSZ $=1$, MARA is decremented by 2 after each transfer |

Bit 3-Data Transfer Interrupt Enable (DTIE): Enables or disables the CPU interrupt (DEND) requested when the DTE bit is cleared to 0 .

Bit 3: DTIE Description

| 0 | The DEND interrupt requested by DTE is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | The DEND interrupt requested by DTE is enabled |  |

Bits 2 and 1—Data Transfer Select 2A and 1A (DTS2A, DTS1A): A channel operates in full address mode when DTS2A and DTS1A are both set to 1 .

Bit 0—Data Transfer Select 0A (DTS0A): Selects normal mode or block transfer mode.
Bit 0: DTS0A
Description

| 0 | Normal mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | Block transfer mode |  |

Rev. 2.0, 03/01, page 202 of 822

Operations in these modes are described in sections 8.4.5, Normal Mode, and 8.4.6, Block Transfer Mode.

## DTCRB

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DTME | - | DAID | DAIDE | TMS | DTS2B | DTS1B | DTSOB |
| Initial value Read/Write | 0 | $\begin{gathered} 0 \\ \text { R/W } \end{gathered}$ | 0 | 0 | 0 | 0 | 0 | 0 |
|  | R/W |  | R/W | R/W | R/W | R/W | R/W | R/W |
| Data transfer master enable Enables or disables data transfer, together with the DTE bit, and is cleared to 0 by an interrupt |  |  |  | Transfer mode select Selects whether the block area is the source or destination in block transfer mode |  |  |  |  |
|  |  | Destination address increment/decrement Destination address increment/decrement enable These bits select whether the destination address register (MARB) is incremented, decremented, or held fixed during the data transfer |  |  |  | Data transfer select 2B to 0B <br> These bits select the data transfer activation source |  |  |

DTCRB is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.
Bit 7—Data Transfer Master Enable (DTME): Together with the DTE bit in DTCRA, this bit enables or disables data transfer. When the DTME and DTE bits are both set to 1 , the channel is enabled. When an NMI interrupt occurs DTME is cleared to 0 , suspending the transfer so that the CPU can use the bus. The suspended transfer resumes when DTME is set to 1 again. For further information on operation in block transfer mode, see section 8.6.6, NMI Interrupts and Block Transfer Mode.

DTME is set to 1 by reading the register while DTME $=0$, then writing 1 .
Bit 7: DTME Description

Bit 6-Reserved: Although reserved, this bit can be written and read.

Bit 5—Destination Address Increment/Decrement (DAID) and Bit 4—Destination Address Increment/Decrement Enable (DAIDE): These bits select whether the destination address register (MARB) is incremented, decremented, or held fixed during the data transfer.

| Bit 5: DAID | Bit 4: DAIDE | Description |
| :---: | :---: | :---: |
| 0 | 0 | MARB is held fixed (Initial value) |
|  | 1 | MARB is incremented after each data transfer <br> - If DTSZ $=0$, MARB is incremented by 1 after each data transfer <br> - If DTSZ $=1$, MARB is incremented by 2 after each data transfer |
| 1 | 0 | MARB is held fixed |
|  | 1 | MARB is decremented after each data transfer <br> - If DTSZ $=0$, MARB is decremented by 1 after each data transfer <br> - If DTSZ $=1$, MARB is decremented by 2 after each data transfer |

Bit 3—Transfer Mode Select (TMS): Selects whether the source or destination is the block area in block transfer mode.

| Bit 3: TMS | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | Destination is the block area in block transfer mode |  |
| 1 | Source is the block area in block transfer mode |  |

Bits 2 to 0-Data Transfer Select 2B to 0B (DTS2B, DTS1B, DTS0B): These bits select the data transfer activation source. The selectable activation sources differ between normal mode and block transfer mode.

- Normal mode

| Bit 2: <br> DTS2B Bit 1: <br> DTS1B Bit 0: <br> DTSOB Description |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Auto-request (burst mode) | (Initial value) |
|  | 1 | 0 | Cannot be used |  |
| 1 | 0 | 0 | Auto-request (cycle-steal mode) |  |
| 1 | 1 Cannot be used  <br>   0 <br> 1 Cannot be used  <br>  1 Cannot be used |  | Falling edge of $\overline{\text { RREQ }}$ |  |

- Block transfer mode

| Bit 2: DTS2B | Bit 1: DTS1B | Bit 0: DTSOB | Description |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | Compare match/input capture A interrupt from ITU channel 0 <br> (Initial value) |
|  |  | 1 | Compare match/input capture A interrupt from ITU channel 1 |
|  | 1 | 0 | Compare match/input capture A interrupt from ITU channel 2 |
|  |  | 1 | Compare match/input capture A interrupt from ITU channel 3 |
| 1 | 0 | 0 | Cannot be used |
|  |  | 1 | Cannot be used |
|  | 1 | 0 | Falling edge of $\overline{\text { DREQ }}$ |
|  |  | 1 | Cannot be used |

The same internal interrupt can be selected to activate two or more channels. The channels are activated in a priority order, highest priority first. For the priority order, see section 8.4.9, DMAC Multiple-Channel Operation.

### 8.4 Operation

### 8.4.1 Overview

Table 8.5 summarizes the DMAC modes.
Table 8.5 DMAC Modes

| Transfer Mode |  | Activation | Notes |
| :---: | :---: | :---: | :---: |
| Short address mode | I/O mode <br> Idle mode <br> Repeat mode | Compare match/input capture A interrupt from ITU channels 0 to 3 <br> Transmit-data-empty and receive-data-full interrupts from SCl channel 0 | - Up to four channels can operate independently <br> - Only the B channels support external requests |
| Full address mode | Normal mode <br> Block transfer mode | Auto-request <br> External request <br> Compare match/input capture A interrupt from ITU channels 0 to 3 <br> External request | - $A$ and $B$ channels are paired; up to two channels are available <br> - Burst mode or cycle-steal mode can be selected for auto-requests |

A summary of operations in these modes follows.
I/O Mode: One byte or word is transferred per request. A designated number of these transfers are executed. A CPU interrupt can be requested at completion of the designated number of transfers. One 24 -bit address and one 8 -bit address are specified. The transfer direction is determined automatically from the activation source.

Idle Mode: One byte or word is transferred per request. A designated number of these transfers are executed. A CPU interrupt can be requested at completion of the designated number of transfers. One 24 -bit address and one 8 -bit address are specified. The addresses are held fixed. The transfer direction is determined automatically from the activation source.

Repeat Mode: One byte or word is transferred per request. A designated number of these transfers are executed. When the designated number of transfers are completed, the initial address and counter value are restored and operation continues. No CPU interrupt is requested. One 24-bit address and one 8 -bit address are specified. The transfer direction is determined automatically from the activation source.

## Normal Mode

- Auto-request

The DMAC is activated by register setup alone, and continues executing transfers until the designated number of transfers have been completed. A CPU interrupt can be requested at completion of the transfers. Both addresses are 24-bit addresses.

- Cycle-steal mode

The bus is released to another bus master after each byte or word is transferred.

- Burst mode

Unless requested by a higher-priority bus master, the bus is not released until the designated number of transfers have been completed.

- External request

One byte or word is transferred per request. A designated number of these transfers are executed. A CPU interrupt can be requested at completion of the designated number of transfers. Both addresses are 24-bit addresses.

Block Transfer Mode: One block of a specified size is transferred per request. A designated number of block transfers are executed. At the end of each block transfer, one address is restored to its initial value. When the designated number of blocks have been transferred, a CPU interrupt can be requested. Both addresses are 24-bit addresses.

### 8.4.2 I/O Mode

I/O mode can be selected independently for each channel.
One byte or word is transferred at each transfer request in I/O mode. A designated number of these transfers are executed. One address is specified in the memory address register (MAR), the other in the I/O address register (IOAR). The direction of transfer is determined automatically from the activation source. The transfer is from the address specified in IOAR to the address specified in MAR if activated by an SCI channel 0 receive-data-full interrupt, and from the address specified in MAR to the address specified in IOAR otherwise.

Table 8.6 indicates the register functions in I/O mode.
Table 8.6 Register Functions in I/O Mode


## Legend

MAR: Memory address register
IOAR: I/O address register
ETCR: Execute transfer count register

MAR and IOAR specify the source and destination addresses. MAR specifies a 24 -bit source or destination address, which is incremented or decremented as each byte or word is transferred. IOAR specifies the lower 8 bits of a fixed address. The upper 16 bits are all 1 s . IOAR is not incremented or decremented.

Figure 8.2 illustrates how I/O mode operates.

Rev. 2.0, 03/01, page 208 of 822


Figure 8.2 Operation in I/O Mode
The transfer count is specified as a 16-bit value in ETCR. The ETCR value is decremented by 1 at each transfer. When the ETCR value reaches H'0000, the DTE bit is cleared and the transfer ends. If the DTIE bit is set to 1 , a CPU interrupt is requested at this time. The maximum transfer count is 65,536 , obtained by setting ETCR to $\mathrm{H}^{\prime} 0000$.

Transfers can be requested (activated) by compare match/input capture A interrupts from ITU channels 0 to 3 , transmit-data-empty and receive-data-full interrupts from SCI channel 0 , and external request signals.

For the detailed settings see section 8.2.4, Data Transfer Control Registers (DTCR).

Figure 8.3 shows a sample setup procedure for I/O mode.


Figure 8.3 I/O Mode Setup Procedure (Example)

### 8.4.3 Idle Mode

Idle mode can be selected independently for each channel.
One byte or word is transferred at each transfer request in idle mode. A designated number of these transfers are executed. One address is specified in the memory address register (MAR), the other in the I/O address register (IOAR). The direction of transfer is determined automatically from the activation source. The transfer is from the address specified in IOAR to the address specified in MAR if activated by an SCI channel 0 receive-data-full interrupt, and from the address specified in MAR to the address specified in IOAR otherwise.

Table 8.7 indicates the register functions in idle mode.

Table 8.7 Register Functions in Idle Mode
Function

| Register |  | Activated by SCI 0 Receive-Data-Full Interrupt | Other Activation | Initial Setting | Operation |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 23 | 0 | Destination address register | Source address register | Destination or source address | Held fixed |
| MAR |  |  |  |  |  |
| 23 | $7 \quad 0$ | Source address register | Destination address register | Source or destination address | Held fixed |
| All 1s | IOAR |  |  |  |  |
| 15 OTCR |  | Transfer counter | Transfer counter | Number of transfers | Decremented once per transfer until $\mathrm{H}^{\prime} 0000$ is reached and transfer ends |
|  |  |  |  |  |  |
|  |  |  |  |  |  |

## Legend

MAR: Memory address register
IOAR: I/O address register
ETCR: Execute transfer count register

MAR and IOAR specify the source and destination addresses. MAR specifies a 24-bit source or destination address. IOAR specifies the lower 8 bits of a fixed address. The upper 16 bits are all 1 s . MAR and IOAR are not incremented or decremented.

Figure 8.4 illustrates how idle mode operates.


Figure 8.4 Operation in Idle Mode

The transfer count is specified as a 16-bit value in ETCR. The ETCR value is decremented by 1 at each transfer. When the ETCR value reaches $\mathrm{H}^{\prime} 0000$, the DTE bit is cleared, the transfer ends, and a CPU interrupt is requested. The maximum transfer count is 65,536 , obtained by setting ETCR to $\mathrm{H}^{\prime} 0000$.

Transfers can be requested (activated) by compare match/input capture A interrupts from ITU channels 0 to 3 , transmit-data-empty and receive-data-full interrupts from SCI channel 0 , and external request signals.

For the detailed settings see section 8.2.4, Data Transfer Control Registers (DTCR).
Figure 8.5 shows a sample setup procedure for idle mode.


Figure 8.5 Idle Mode Setup Procedure (Example)

### 8.4.4 Repeat Mode

Repeat mode is useful for cyclically transferring a bit pattern from a table to the programmable timing pattern controller (TPC) in synchronization, for example, with ITU compare match. Repeat mode can be selected for each channel independently.

One byte or word is transferred per request in repeat mode, as in I/O mode. A designated number of these transfers are executed. One address is specified in the memory address register (MAR), the other in the I/O address register (IOAR). At the end of the designated number of transfers, MAR and ETCR are restored to their original values and operation continues. The direction of transfer is determined automatically from the activation source. The transfer is from the address specified in IOAR to the address specified in MAR if activated by an SCI channel 0 receive-datafull interrupt, and from the address specified in MAR to the address specified in IOAR otherwise.

Table 8.8 indicates the register functions in repeat mode.

Table 8.8 Register Functions in Repeat Mode
Function

| Register |  | Activated by SCI 0 Receive-Data-Full Interrupt | Other Activation | Initial Setting | Operation |
| :---: | :---: | :---: | :---: | :---: | :---: |
|    <br> 23   <br>  MAR  |  | Destination address register | Source address register | Destination or source address | Incremented or decremented at each transfer until H'0000, then restored to initial value |
|  |  |  |  |  |  |
|  |  |  |  |  |  |
| 23 | $7 \quad 0$ | Source address register | Destination address register | Source or destination address | Held fixed |
| All 1 s | IOAR |  |  |  |  |
|  |  | Transfer counter | Transfer counter | Number of transfers | Decremented once per transfer until $\mathrm{H}^{\prime} 0000$ is reached, then reloaded from ETCRL |
|  | $7 \quad 0$ |  |  |  |  |
|  | ETCRH |  |  |  |  |
|  | 4 |  |  |  |  |
|  |  |  |  |  |  |
| $7 \quad 0$ <br> ETCRL |  | Initial transfer count | Initial transfe count | Number of transfers | Held fixed |
|  |  |  |  |  |  |

Legend
MAR: Memory address register
IOAR: I/O address register
ETCR: Execute transfer count register

In repeat mode ETCRH is used as the transfer counter while ETCRL holds the initial transfer count. ETCRH is decremented by 1 at each transfer until it reaches $\mathrm{H}^{\prime} 00$, then is reloaded from ETCRL. MAR is also restored to its initial value, which is calculated from the DTSZ and DTID bits in DTCR. Specifically, MAR is restored as follows:

$$
\text { MAR } \leftarrow \operatorname{MAR}-(-1)^{\text {DTID }} \cdot 2^{\text {DTSZ }} \cdot \text { ETCRL }
$$

ETCRH and ETCRL should be initially set to the same value.
In repeat mode transfers continue until the CPU clears the DTE bit to 0 . After DTE is cleared to 0 , if the CPU sets DTE to 1 again, transfers resume from the state at which DTE was cleared. No CPU interrupt is requested.

As in I/O mode, MAR and IOAR specify the source and destination addresses. MAR specifies a 24-bit source or destination address. IOAR specifies the lower 8 bits of a fixed address. The upper 16 bits are all 1 s . IOAR is not incremented or decremented.

Figure 8.6 illustrates how repeat mode operates.
Address

Figure 8.6 Operation in Repeat Mode
The transfer count is specified as an 8-bit value in ETCRH and ETCRL. The maximum transfer count is 255 , obtained by setting both ETCRH and ETCRL to H'FF.

Transfers can be requested (activated) by compare match/input capture A interrupts from ITU channels 0 to 3 , transmit-data-empty and receive-data-full interrupts from SCI channel 0 , and external request signals.

For the detailed settings see section 8.2.4, Data Transfer Control Registers (DTCR).
Figure 8.7 shows a sample setup procedure for repeat mode.


1. Set the source and destination addresses in MAR and IOAR. The transfer direction is determined automatically from the activation source.
2. Set the transfer count in both ETCRH and ETCRL.
3. Read DTCR while the DTE bit is cleared to 0.
4. Set the DTCR bits as follows.

- Select the DMAC activation source with bits DTS2 to DTS0.
- Clear the DTIE bit to 0 and set the RPE bit to 1 to select repeat mode.
- Select MAR increment or decrement with the DTID bit.
- Select byte size or word size with the DTSZ bit.
- Set the DTE bit to 1 to enable the transfer.

Figure 8.7 Repeat Mode Setup Procedure (Example)

### 8.4.5 Normal Mode

In normal mode the A and B channels are combined. One byte or word is transferred per request. A designated number of these transfers are executed. Addresses are specified in MARA and MARB. Table 8.9 indicates the register functions in I/O mode.

Table 8.9 Register Functions in Normal Mode

| Register |  | Function | Initial Setting | Operation |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  |  | 0 | Source address <br> register | Source address | Incremented or <br> decremented once per <br> transfer, or held fixed |
|  | MARA |  |  |  | Incremented or <br> decremented once per <br> transfer, or held fixed |
| 23 |  | 0 | Destination |  |  |
|  | MARB |  | address register | Destination <br> address | Decremented once per <br> transfer |
|  | 15 |  | 0 | Transfer counter | Number of <br> transfers |
|  | ETCRA |  |  |  |  |

## Legend

MARA: Memory address register A
MARB: Memory address register B
ETCRA: Execute transfer count register A

The source and destination addresses are both 24-bit addresses. MARA specifies the source address. MARB specifies the destination address. MARA and MARB can be independently incremented, decremented, or held fixed as data is transferred.

The transfer count is specified as a 16-bit value in ETCRA. The ETCRA value is decremented by 1 at each transfer. When the ETCRA value reaches $\mathrm{H}^{\prime} 0000$, the DTE bit is cleared and the transfer ends. If the DTIE bit is set, a CPU interrupt is requested at this time. The maximum transfer count is 65,536 , obtained by setting ETCRA to $\mathrm{H}^{\prime} 0000$.

Figure 8.8 illustrates how normal mode operates.


Legend
$\mathrm{L}_{\mathrm{A}}=$ initial setting of MARA
$L_{B}=$ initial setting of MARB
$N=$ initial setting of ETCRA
$\mathrm{T}_{\mathrm{A}}=\mathrm{L}_{\mathrm{A}}$
$B_{A}=L_{A}+$ SAIDE $\cdot(-1)$ SAID $\cdot(2$ DTSZ $\cdot N-1)$
$\mathrm{T}_{\mathrm{B}}=\mathrm{L}_{\mathrm{B}}$
$B_{B}=L_{B}+$ DAIDE $\cdot(-1)^{\text {DAID }} \cdot\left(2^{\text {DTSZ }} \cdot \mathrm{N}-1\right)$

## Figure 8.8 Operation in Normal Mode

Transfers can be requested (activated) by an external request or auto-request. An auto-requested transfer is activated by the register settings alone. The designated number of transfers are executed automatically. Either cycle-steal or burst mode can be selected. In cycle-steal mode the DMAC releases the bus temporarily after each transfer. In burst mode the DMAC keeps the bus until the transfers are completed, unless there is a bus request from a higher-priority bus master.

For the detailed settings see section 8.3.4, Data Transfer Control Registers (DTCR).

Figure 8.9 shows a sample setup procedure for normal mode.


1. Set the initial source address in MARA.
2. Set the initial destination address in MARB.
3. Set the transfer count in ETCRA.
4. Set the DTCRB bits as follows.

- Clear the DTME bit to 0.
- Set the DAID and DAIDE bits to select whether
- MARB is incremented, decremented, or held fixed.
- Select the DMAC activation source with bits
- DTS2B to DTS0B.

5. Set the DTCRA bits as follows.

- Clear the DTE bit to 0 .
- Select byte or word size with the DTSZ bit.
- Set the SAID and SAIDE bits to select whether MARA is incremented, decremented, or held fixed.
- Set or clear the DTIE bit to enable or disable the CPU interrupt at the end of the transfer.
- Clear the DTSOA bit to 0 and set the DTS2A and DTS1A bits to 1 to select normal mode.

6. Read DTCRB with DTME cleared to 0 .
7. Set the DTME bit to 1 in DTCRB.
8. Read DTCRA with DTE cleared to 0 .
9. Set the DTE bit to 1 in DTCRA to enable the transfer.

Normal mode
Note: Carry out settings 1 to 9 with the DEND interrupt masked in the CPU.
If an NMI interrupt occurs during the setup procedure, it may clear the DTME bit to 0 , in which case the transfer will not start.

Figure 8.9 Normal Mode Setup Procedure (Example)

### 8.4.6 Block Transfer Mode

In block transfer mode the A and B channels are combined. One block of a specified size is transferred per request. A designated number of block transfers are executed. Addresses are specified in MARA and MARB. The block area address can be either held fixed or cycled.

Table 8.10 indicates the register functions in block transfer mode.

Table 8.10 Register Functions in Block Transfer Mode


Legend
MARA: Memory address register A
MARB: Memory address register $B$
ETCRA: Execute transfer count register A
ETCRB: Execute transfer count register B

The source and destination addresses are both 24-bit addresses. MARA specifies the source address. MARB specifies the destination address. MARA and MARB can be independently incremented, decremented, or held fixed as data is transferred. One of these registers operates as a block area register: even if it is incremented or decremented, it is restored to its initial value at the end of each block transfer. The TMS bit in DTCRB selects whether the block area is the source or destination.

If $\mathrm{M}(1$ to 255$)$ is the size of the block transferred at each request and $\mathrm{N}(1$ to 65,536$)$ is the number of blocks to be transferred, then ETCRAH and ETCRAL should initially be set to M and ETCRB should initially be set to N .

Figure 8.10 illustrates how block transfer mode operates. In this figure, bit TMS is cleared to 0 , meaning the block area is the destination.


Figure 8.10 Operation in Block Transfer Mode

When activated by a transfer request, the DMAC executes a burst transfer. During the transfer MARA and MARB are updated according to the DTCR settings, and ETCRAH is decremented. When ETCRAH reaches $H^{\prime} 00$, it is reloaded from ETCRAL to restore the initial value. The memory address register of the block area is also restored to its initial value, and ETCRB is decremented. If ETCRB is not $\mathrm{H}^{\prime} 0000$, the DMAC then waits for the next transfer request. ETCRAH and ETCRAL should be initially set to the same value.

The above operation is repeated until ETCRB reaches H'0000, at which point the DTE bit is cleared to 0 and the transfer ends. If the DTIE bit is set to 1 , a CPU interrupt is requested at this time.

Figure 8.11 shows examples of a block transfer with byte data size when the block area is the destination. In (a) the block area address is cycled. In (b) the block area address is held fixed.

Transfers can be requested (activated) by compare match/input capture A interrupts from ITU channels 0 to 3 , and by external request signals.

For the detailed settings see section 8.3.4, Data Transfer Control Registers (DTCR).


b. $\begin{aligned} & \text { DTSZ }=\text { TMS }=0 \\ & \text { SAID }=0 \\ & \text { SAIDE }=1 \\ & \text { DAIDE }=0\end{aligned}$

SAID = DAID $=0$
SAIDE = DAIDE = 1

Figure 8.12 shows a sample setup procedure for block transfer mode.


Figure 8.12 Block Transfer Mode Setup Procedure (Example)

### 8.4.7 DMAC Activation

The DMAC can be activated by an internal interrupt, external request, or auto-request. The available activation sources differ depending on the transfer mode and channel as indicated in table 8.11.

Table 8.11 DMAC Activation Sources

| Activation Source |  | Short Address Mode |  | Full Address Mode |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Channels 0 A and 1A | Channels 0B and 1B | Normal | Block |
| Internal interrupts | IMIA0 | Yes | Yes | No | Yes |
|  | IMIA1 | Yes | Yes | No | Yes |
|  | IMIA2 | Yes | Yes | No | Yes |
|  | IMIA3 | Yes | Yes | No | Yes |
|  | TXIO | Yes | Yes | No | No |
|  | RXIO | Yes | Yes | No | No |
| External requests | Falling edge of DREQ | No | Yes | Yes | Yes |
|  | Low input at DREQ | No | Yes | Yes | No |
| Auto-request |  | No | No | Yes | No |

Activation by Internal Interrupts: When an interrupt request is selected as a DMAC activation source and the DTE bit is set to 1 , that interrupt request is not sent to the CPU. It is not possible for an interrupt request to activate the DMAC and simultaneously generate a CPU interrupt.

When the DMAC is activated by an interrupt request, the interrupt request flag is cleared automatically. If the same interrupt is selected to activate two or more channels, the interrupt request flag is cleared when the highest-priority channel is activated, but the transfer request is held pending on the other channels in the DMAC, which are activated in their priority order.

Activation by External Request: If an external request ( $\overline{\mathrm{DREQ}} \mathrm{pin}$ ) is selected as an activation source, the $\overline{\mathrm{DREQ}}$ pin becomes an input pin and the corresponding $\overline{\text { TEND }}$ pin becomes an output pin, regardless of the port data direction register (DDR) settings. The $\overline{\mathrm{DREQ}}$ input can be levelsensitive or edge-sensitive.

In short address mode and normal mode, an external request operates as follows. If edge sensing is selected, one byte or word is transferred each time a high-to-low transition of the $\overline{\text { DREQ }}$ input is detected. If the next edge is input before the transfer is completed, the next transfer may not be executed. If level sensing is selected, the transfer continues while $\overline{\mathrm{DREQ}}$ is low, until the transfer is completed. The bus is released temporarily after each byte or word has been transferred, however. If the $\overline{\mathrm{DREQ}}$ input goes high during a transfer, the transfer is suspended after the current byte or word has been transferred. When $\overline{\mathrm{DREQ}}$ goes low, the request is held internally until one byte or word has been transferred. The $\overline{\text { TEND }}$ signal goes low during the last write cycle.

In block transfer mode, an external request operates as follows. Only edge-sensitive transfer requests are possible in block transfer mode. Each time a high-to-low transition of the $\overline{\mathrm{DREQ}}$ input is detected, a block of the specified size is transferred. The $\overline{\text { TEND }}$ signal goes low during the last write cycle in each block.

Activation by Auto-Request: The transfer starts as soon as enabled by register setup, and continues until completed. Cycle-steal mode or burst mode can be selected.

In cycle-steal mode the DMAC releases the bus temporarily after transferring each byte or word. Normally, DMAC cycles alternate with CPU cycles.

In burst mode the DMAC keeps the bus until the transfer is completed, unless there is a higherpriority bus request. If there is a higher-priority bus request, the bus is released after the current byte or word has been transferred.

### 8.4.8 DMAC Bus Cycle

Figure 8.13 shows an example of the timing of the basic DMAC bus cycle. This example shows a word-size transfer from a 16 -bit two-state access area to an 8 -bit three-state access area. When the DMAC gets the bus from the CPU, after one dead cycle (Td), it reads from the source address and writes to the destination address. During these read and write operations the bus is not released even if there is another bus request. DMAC cycles comply with bus controller settings in the same way as CPU cycles.


Figure 8.13 DMA Transfer Bus Timing (Example)

Figure 8.14 shows the timing when the DMAC is activated by low input at a $\overline{\mathrm{DREQ}}$ pin. This example shows a word-size transfer from a 16-bit two-state access area to another 16-bit two-state access area. The DMAC continues the transfer while the $\overline{\mathrm{DREQ}}$ pin is held low.


Figure 8.14 Bus Timing of DMA Transfer Requested by Low $\overline{\text { DREQ }}$ Input

Figure 8.15 shows an auto-requested burst-mode transfer. This example shows a transfer of three words from a 16 -bit two-state access area to another 16 -bit two-state access area.


Figure 8.15 Burst DMA Bus Timing
When the DMAC is activated from a $\overline{\text { DREQ }}$ pin there is a minimum interval of four states from when the transfer is requested until the DMAC starts operating. The $\overline{\mathrm{DREQ}}$ pin is not sampled during the time between the transfer request and the start of the transfer. In short address mode and normal mode, the pin is next sampled at the end of the read cycle. In block transfer mode, the pin is next sampled at the end of one block transfer.

Figure 8.16 shows the timing when the DMAC is activated by the falling edge of $\overline{\overline{D R E Q}}$ in normal mode.


Figure 8.16 Timing of DMAC Activation by Falling Edge of $\overline{\text { DREQ }}$ in Normal Mode

Figure 8.17 shows the timing when the DMAC is activated by level-sensitive low $\overline{\mathrm{DREQ}}$ input in normal mode.


Figure 8.17 Timing of DMAC Activation by Low $\overline{\text { DREQ }}$ Level in Normal Mode

Figure 8.18 shows the timing when the DMAC is activated by the falling edge of $\overline{\mathrm{DREQ}}$ in block transfer mode.


Figure 8.18 Timing of DMAC Activation by Falling Edge of $\overline{\overline{D R E Q}}$ in Block Transfer Mode

### 8.4.9 DMAC Multiple-Channel Operation

The DMAC channel priority order is: channel $0>$ channel 1 and channel $\mathrm{A}>$ channel B .
Table 8.12 shows the complete priority order.
Table 8.12 Channel Priority Order

| Short Address Mode | Full Address Mode | Priority |
| :--- | :--- | :--- |
| Channel 0A | Channel 0 | High |
| Channel 0B |  | $\uparrow$ |
| Channel 1A | Channel 1 | Low |
| Channel 1B |  |  |

If transfers are requested on two or more channels simultaneously, or if a transfer on one channel is requested during a transfer on another channel, the DMAC operates as follows.

1. When a transfer is requested, the DMAC requests the bus right. When it gets the bus right, it starts a transfer on the highest-priority channel at that time.
2. Once a transfer starts on one channel, requests to other channels are held pending until that channel releases the bus.
3. After each transfer in short address mode, and each externally-requested or cycle-steal transfer in normal mode, the DMAC releases the bus and returns to step 1. After releasing the bus, if there is a transfer request for another channel, the DMAC requests the bus again.
4. After completion of a burst-mode transfer, or after transfer of one block in block transfer mode, the DMAC releases the bus and returns to step 1 . If there is a transfer request for a higher-priority channel or a bus request from a higher-priority bus master, however, the DMAC releases the bus after completing the transfer of the current byte or word. After releasing the bus, if there is a transfer request for another channel, the DMAC requests the bus again.

Figure 8.19 shows the timing when channel 0 A is set up for I/O mode and channel 1 for burst mode, and a transfer request for channel 0 A is received while channel 1 is active.


Figure 8.19 Timing of Multiple-Channel Operations

### 8.4.10 External Bus Requests, Refresh Controller, and DMAC

During a DMA transfer, if the bus right is requested by an external bus request signal ( $\overline{\mathrm{BREQ}})$ or by the refresh controller, the DMAC releases the bus after completing the transfer of the current byte or word. If there is a transfer request at this point, the DMAC requests the bus right again. Figure 8.20 shows an example of the timing of insertion of a refresh cycle during a burst transfer on channel 0 .


Figure 8.20 Bus Timing of Refresh Controller and DMAC

### 8.4.11 NMI Interrupts and DMAC

NMI interrupts do not affect DMAC operations in short address mode.
If an NMI interrupt occurs during a transfer in full address mode, the DMAC suspends operations. In full address mode, a channel is enabled when its DTE and DTME bits are both set to 1 . NMI input clears the DTME bit to 0 . After transferring the current byte or word, the DMAC releases the bus to the CPU. In normal mode, the suspended transfer resumes when the CPU sets the DTME bit to 1 again. Check that the DTE bit is set to 1 and the DTME bit is cleared to 0 before setting the DTME bit to 1 .

Figure 8.21 shows the procedure for resuming a DMA transfer in normal mode on channel 0 after the transfer was halted by NMI input.


Figure 8.21 Procedure for Resuming a DMA Transfer Halted by NMI (Example)
For information about NMI interrupts in block transfer mode, see section 8.6.6, NMI Interrupts and Block Transfer Mode.

### 8.4.12 Aborting a DMA Transfer

When the DTE bit in an active channel is cleared to 0 , the DMAC halts after transferring the current byte or word. The DMAC starts again when the DTE bit is set to 1 . In full address mode, the DTME bit can be used for the same purpose. Figure 8.22 shows the procedure for aborting a DMA transfer by software.


1. Clear the DTE bit to 0 in DTCR. To avoid generating an interrupt when aborting a DMA transfer, clear the DTIE bit to 0 simultaneously.

Figure 8.22 Procedure for Aborting a DMA Transfer

### 8.4.13 Exiting Full Address Mode

Figure 8.23 shows the procedure for exiting full address mode and initializing the pair of channels. To set the channels up in another mode after exiting full address mode, follow the setup procedure for the relevant mode.


Figure 8.23 Procedure for Exiting Full Address Mode (Example)

### 8.4.14 DMAC States in Reset State, Standby Modes, and Sleep Mode

When the chip is reset or enters hardware or software standby mode, the DMAC is initialized and halts. DMAC operations continue in sleep mode. Figure 8.24 shows the timing of a cycle-steal transfer in sleep mode.


Figure 8.24 Timing of Cycle-Steal Transfer in Sleep Mode

### 8.5 Interrupts

The DMAC generates only DMA-end interrupts. Table 8.13 lists the interrupts and their priority.
Table 8.13 DMAC Interrupts

|  | Description |  | Interrupt |
| :--- | :--- | :--- | :--- |
| Interrupt | Short Address Mode | Full Address Mode | Priority |
| DEND0A | End of transfer on channel 0A | End of transfer on channel 0 | High |
| DEND0B | End of transfer on channel 0B | - |  |
| DEND1A | End of transfer on channel 1A | End of transfer on channel 1 |  |
| DEND1B | End of transfer on channel 1B | - | Low |

Each interrupt is enabled or disabled by the DTIE bit in the corresponding data transfer control register (DTCR). Separate interrupt signals are sent to the interrupt controller.

The interrupt priority order among channels is channel $0>$ channel 1 and channel $\mathrm{A}>$ channel B .
Figure 8.25 shows the DMA-end interrupt logic. An interrupt is requested whenever DTE $=0$ and DTIE $=1$.


## Figure 8.25 DMA-End Interrupt Logic

The DMA-end interrupt for the B channels (DENDB) is unavailable in full address mode. The DTME bit does not affect interrupt operations.

### 8.6 Usage Notes

### 8.6.1 Note on Word Data Transfer

Word data cannot be accessed starting at an odd address. When word-size transfer is selected, set even values in the memory and I/O address registers (MAR and IOAR).

### 8.6.2 DMAC Self-Access

The DMAC itself cannot be accessed during a DMAC cycle. DMAC registers cannot be specified as source or destination addresses.

### 8.6.3 Longword Access to Memory Address Registers

A memory address register can be accessed as longword data at the MARR address.

## Example

MOV.L \#LBL, ERO
MOV.L ERO, @MARR
Four byte accesses are performed. Note that the CPU may release the bus between the second byte (MARE) and third byte (MARH).

Memory address registers should be written and read only when the DMAC is halted.

### 8.6.4 Note on Full Address Mode Setup

Full address mode is controlled by two registers: DTCRA and DTCRB. Care must be taken to prevent the B channel from operating in short address mode during the register setup. The enable bits (DTE and DTME) should not be set to 1 until the end of the setup procedure.

### 8.6.5 Note on Activating DMAC by Internal Interrupts

When using an internal interrupt to activate the DMAC, make sure that the interrupt selected as the activating source does not occur during the interval after it has been selected but before the DMAC has been enabled. The on-chip supporting module that will generate the interrupt should not be activated until the DMAC has been enabled. If the DMAC must be enabled while the onchip supporting module is active, follow the procedure in figure 8.26.


Figure 8.26 Procedure for Enabling DMAC while On-Chip Supporting Module is Operating (Example)

If the DTE bit is set to 1 but the DTME bit is cleared to 0 , the DMAC is halted and the selected activating source cannot generate a CPU interrupt. If the DMAC is halted by an NMI interrupt, for example, the selected activating source cannot generate CPU interrupts. To terminate DMAC operations in this state, clear the DTE bit to 0 to allow CPU interrupts to be requested. To continue DMAC operations, carry out steps 2 and 4 in figure 8.26 before and after setting the DTME bit to 1.

When an ITU interrupt activates the DMAC, make sure the next interrupt does not occur before the DMA transfer ends. If one ITU interrupt activates two or more channels, make sure the next interrupt does not occur before the DMA transfers end on all the activated channels. If the next interrupt occurs before a transfer ends, the channel or channels for which that interrupt was selected may fail to accept further activation requests.

### 8.6.6 NMI Interrupts and Block Transfer Mode

If an NMI interrupt occurs in block transfer mode, the DMAC operates as follows.

- When the NMI interrupt occurs, the DMAC finishes transferring the current byte or word, then clears the DTME bit to 0 and halts. The halt may occur in the middle of a block.
It is possible to find whether a transfer was halted in the middle of a block by checking the block size counter. If the block size counter does not have its initial value, the transfer was halted in the middle of a block.
- If the transfer is halted in the middle of a block, the activating interrupt flag is cleared to 0 . The activation request is not held pending.
- While the DTE bit is set to 1 and the DTME bit is cleared to 0 , the DMAC is halted and does not accept activating interrupt requests. If an activating interrupt occurs in this state, the DMAC does not operate and does not hold the transfer request pending internally. Neither is a CPU interrupt requested.
For this reason, before setting the DTME bit to 1 , first clear the enable bit of the activating interrupt to 0 . Then, after setting the DTME bit to 1 , set the interrupt enable bit to 1 again. See section 8.6.5, Note on Activating DMAC by Internal Interrupts.
- When the DTME bit is set to 1 , the DMAC waits for the next transfer request. If it was halted in the middle of a block transfer, the rest of the block is transferred when the next transfer request occurs. Otherwise, the next block is transferred when the next transfer request occurs.


### 8.6.7 Memory and I/O Address Register Values

Table 8.14 indicates the address ranges that can be specified in the memory and I/O address registers (MAR and IOAR).

Table 8.14 Address Ranges Specifiable in MAR and IOAR
1-Mbyte Mode

| MAR | H'00000 to H'FFFFF <br> $(0$ to 1048575$)$ | H'000000 to H'FFFFFF <br> $(0$ to 16777215) |
| :--- | :--- | :--- |
| IOAR | H'FFF00 to H'FFFFF | H'FFFF00 to H'FFFFFF |
|  | $(1048320$ to 1048575$)$ | $(16776960$ to 16777215) |

MAR bits 23 to 20 are ignored in 1-Mbyte mode.

### 8.6.8 Bus Cycle when Transfer is Aborted

When a transfer is aborted by clearing the DTE bit or suspended by an NMI that clears the DTME bit, if this halts a channel for which the DMAC has a transfer request pending internally, a dead cycle may occur. This dead cycle does not update the halted channel's address register or counter value. Figure 8.27 shows an example in which an auto-requested transfer in cycle-steal mode on channel 0 is aborted by clearing the DTE bit in channel 0 .


Figure 8.27 Bus Timing at Abort of DMA Transfer in Cycle-Steal Mode

## Section 9 I/O Ports

### 9.1 Overview

The H8/3052F has 10 input/output ports (ports $1,2,3,4,5,6,8,9$, A, and B) and one input port (port 7). Table 9.1 summarizes the port functions. The pins in each port are multiplexed as shown in table 9.1.

Each port has a data direction register (DDR) for selecting input or output, and a data register (DR) for storing output data. In addition to these registers, ports 2, 4, and 5 have an input pull-up MOS control register (PCR) for switching input pull-up MOS transistors on and off.

Ports 1 to 6 and port 8 can drive one TTL load and a $90-\mathrm{pF}$ capacitive load. Ports 9 , A, and B can drive one TTL load and a $30-\mathrm{pF}$ capacitive load. Ports 1 to 6 and 8 to B can drive a darlington pair. Ports $1,2,5$, and B can drive LEDs (with $10-\mathrm{mA}$ current sink). $\mathrm{Pins} \mathrm{P8}_{2}$ to $\mathrm{P8}_{0}, \mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$, and $\mathrm{PB}_{3}$ to $\mathrm{PB}_{0}$ have Schmitt-trigger input circuits.

For block diagrams of the ports see appendix C, I/O Port Block Diagrams.

Table 9.1 Port Functions

| Port | Description | Pins | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Port 1 | - 8-bit I/O port <br> - Can drive LEDs | $\begin{aligned} & \mathrm{P} 1_{7} \text { to } \mathrm{P} 1_{d} \\ & \mathrm{~A}_{7} \text { to } \mathrm{A}_{0} \end{aligned}$ | Address output pins ( $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ ) |  |  |  | Address output ( $\mathrm{A}_{7}$ to $A_{0}$ ) and generic input <br> DDR = 0: generic input <br> DDR = 1: address output |  | Generic input/ output |
| Port 2 | - 8-bit I/O port <br> - Input pull-up MOS <br> - Can drive LEDs | $\begin{aligned} & \mathrm{P}_{7} \text { to } \mathrm{P} 2_{d} / \\ & \mathrm{A}_{15} \text { to } \mathrm{A}_{8} \end{aligned}$ | Address output pins ( $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ ) |  |  |  | Address output ( $\mathrm{A}_{15}$ to $A_{8}$ ) and generic input DDR = 0: generic input DDR = 1: address output |  | Generic input/ output |
| Port 3 | - 8-bit I/O port | $\begin{aligned} & \mathrm{P}_{7} \text { to } \mathrm{P} 3_{d} / \\ & \mathrm{D}_{15} \text { to } \mathrm{D}_{8} \end{aligned}$ | Data input/output ( $\mathrm{D}_{15}$ to $\left.\mathrm{D}_{8}\right)$ |  |  |  |  |  | Generic input/ output |
| Port 4 | - 8-bit I/O port <br> - Input pull-up MOS | $\begin{aligned} & \mathrm{P} 4_{7} \text { to } \mathrm{P} 4_{0} / \\ & \mathrm{D}_{7} \text { to } \mathrm{D}_{0} \end{aligned}$ | Data input/output ( $\mathrm{D}_{7}$ to $\mathrm{D}_{0}$ ) and 8-bit generic input/output 8 -bit bus mode: generic input/output 16-bit bus mode: data input/output |  |  |  |  |  | Generic input/ output |
| Port 5 | - 4-bit I/O port <br> - Input pull-up MOS <br> - Can drive LEDs | $\begin{aligned} & P 5_{3} \text { to } P 5_{d} / \\ & A_{19} \text { to } A_{16} \end{aligned}$ | Address output ( $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ ) |  |  |  | Address to $\mathrm{A}_{16}$ ) and generic in DDR $=0$ : input DDR = 1: output | $\begin{aligned} & \text { utput }\left(\mathrm{A}_{19}\right. \\ & \text { 4-bit } \\ & \text { put } \end{aligned}$ <br> generic <br> address | Generic input/ output |
| Port 6 | - 7-bit I/O port | $\begin{aligned} & \mathrm{Pb}_{6} / \overline{\mathrm{LWR}}, \\ & \mathrm{P6} / \mathrm{HWR}, \\ & \mathrm{P6} 6_{4} / \overline{\mathrm{RD}}, \\ & \mathrm{~Pb}_{3} / \overline{\mathrm{AS}} \end{aligned}$ | Bus control signal output ( $\overline{\mathrm{LWR}}, \overline{\mathrm{HWR}}, \overline{\mathrm{RD}}, \overline{\mathrm{AS}}$ ) |  |  |  |  |  | Generic input/ output |
|  |  | $\begin{aligned} & \mathrm{P} 6_{2} / \overline{\mathrm{BACK}}, \\ & \mathrm{P} 6_{1} / \overline{\mathrm{BREQ}}, \\ & \mathrm{P} 6_{0} / \overline{\mathrm{WAIT}} \\ & \hline \end{aligned}$ | Bus control signal input/output ( $\overline{\mathrm{BACK}}, \overline{\mathrm{BREQ}}, \overline{\mathrm{WAIT}}$ ) and 3bit generic input/output |  |  |  |  |  |  |
| Port 7 | - 8-bit I/O port | $\begin{aligned} & \mathrm{P}_{7}^{7} / \mathrm{AN}_{7} / \mathrm{DA}_{1}, \\ & \mathrm{P}_{7} / \mathrm{AN}_{6} / \mathrm{DA}_{0} \end{aligned}$ | Analog input $\left(\mathrm{AN}_{7}, \mathrm{AN}_{6}\right)$ to $\mathrm{A} / \mathrm{D}$ converter, analog output $\left(\mathrm{DA}_{1}, \mathrm{DA}_{0}\right)$ from D/A converter, and generic input |  |  |  |  |  |  |
|  |  | $\begin{aligned} & P 7_{5} \text { to } P 7_{d} \\ & A N_{5} \text { to } A N_{0} \end{aligned}$ | Analog input ( $\mathrm{AN}_{5}$ to $\left.\mathrm{AN} \mathrm{N}_{0}\right)$ to $\mathrm{A} / \mathrm{D}$ converter, and generic input |  |  |  |  |  |  |


| Port | Description | Pins | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Port 8 | - 5-bit I/O port <br> - $\mathrm{P}_{2}$ to P 8 o have Schmitt inputs | $\mathrm{P}_{4} / \overline{\mathrm{CS}}$ | DDR $=0$ : generic input <br> DDR $=1$ (reset value): $\overline{\mathrm{CS}}_{0}$ output |  |  |  |  |  | Generic input/ output |
|  |  |  | $\overline{\mathrm{IRQ}}_{3}$ to $\overline{\mathrm{IRQ}}_{1}$ input, $\overline{\mathrm{CS}}_{1}$ to $\overline{\mathrm{CS}}_{3}$ output, and generic input DDR $=0$ (reset value): generic input <br> DDR $=1: \overline{\mathrm{CS}}$, to $\overline{\mathrm{CS}_{3}}$ output |  |  |  |  |  | $\mathrm{IRQ}_{3}$ to $I R Q_{0}$ input and generic input/ output |
|  |  | $\mathrm{P} 8 / \overline{\mathrm{RFSH}} / \overline{\mathrm{IRQ}}_{0}$ | $\overline{\mathrm{IRQ}}_{0}$ input, $\overline{\mathrm{RFSH}}^{\text {a }}$ output, and generic input/output |  |  |  |  |  |  |
| Port 9 | - 6-bit I/O port | $\begin{aligned} & \mathrm{Pg}_{5} / \mathrm{SCK}_{1} / / \overline{\mathrm{RQ}_{5}}, \\ & \mathrm{Pg}_{4} / \mathrm{SCK}_{6} / \overline{\mathrm{RQ}}{ }_{4}, \\ & \mathrm{Pg}_{3} / \mathrm{RxD}_{1}, \\ & \mathrm{Pg}_{2} / \mathrm{RxD}_{0}, \\ & \mathrm{Pg}_{4} / \mathrm{TxD}_{1}, \\ & \mathrm{Pg}_{0} / \mathrm{TxD}_{0} \end{aligned}$ | Input and output ( SCK $_{1}, \mathrm{SCK}_{0}, \mathrm{RxD}_{1}, \mathrm{RxD}_{0}, \mathrm{TxD}_{1}, \mathrm{TxD}_{0}$ ) for serial communication interfaces 1 and $0(\mathrm{SCl1} / 0), \overline{\mathrm{RQ}}_{5}$ and $\overline{\mathrm{IRQ}}_{4}$ input, and 6bit generic input/output |  |  |  |  |  |  |
| Port A | - 8-bit I/O port <br> - Schmitt inputs output | $\begin{aligned} & \mathrm{PA}_{7} / \mathrm{TP}_{7} / \\ & \mathrm{TIOCB}_{2} / \mathrm{A}_{20} \end{aligned}$ | Output $\left(\mathrm{TP}_{7}\right)$ from programmable timing pattern controller (TPC), input or output $\left(\mathrm{TIOCB}_{2}\right)$ for 16-bit integrated timer unit (ITU), and generic input/output |  | Address output$\left(\mathrm{A}_{20}\right)$ |  | TPC output ( $\mathrm{TP}_{7}$ ), ITU input or output $\left(\mathrm{TIOCB}_{2}\right)$, and generic input/ output | Address output <br> ( $\mathrm{A}_{20}$ ) | TPC output ( $\mathrm{TP}_{7}$ ), ITU input or output ( $\mathrm{TIOCB}_{2}$ ), and generic input/ output |
|  |  | $\begin{aligned} & \mathrm{PA}_{6} / \mathrm{TP}_{6} / \\ & \mathrm{TIOCA}_{2} / \mathrm{A}_{21} /{\overline{\mathrm{CS}_{4}}}^{\mathrm{PA}_{5} / \mathrm{TP}_{5} /} \\ & \mathrm{TIOCB}_{1} / \mathrm{A}_{22} / \overline{\mathrm{CS}}_{5} \\ & \mathrm{PA}_{4} / \mathrm{TP}_{4} / \\ & \mathrm{TIOCA}_{1} / \mathrm{A}_{23} /{\overline{\mathrm{CS}_{6}}}^{2} \end{aligned}$ | TPC output $\left(\mathrm{TP}_{6}\right.$ to $\mathrm{TP}_{4}$ ), ITU input and output ( TIOCA $_{2}$, TIOCB $_{1}$, TIOCA $_{1}$ ), $\overline{\mathrm{CS}}_{4}$ to $\overline{\mathrm{CS}}_{6}$ output, and generic input/ output |  | TPC output ( TP $_{6}$ to TP $_{4}$ ), ITU input and output $\left(\mathrm{TIOCA}_{2}\right.$, TIOCB $_{1}$, TIOCA $_{1}$ ), address output ( $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ ), $\overline{\mathrm{CS}}_{4}$ to $\overline{\mathrm{CS}}_{6}$ output, and generic input/output |  | TPC <br> output <br> $\left(T_{6}\right.$ to <br> $T P_{4}$ ), ITU <br> input and <br> output <br> (TIOCA ${ }_{2}$, <br> TIOCB $_{1}$, <br> TIOCA $_{1}$ ), <br> address <br> output <br> ( $\mathrm{A}_{13}$ to <br> $\left.\mathrm{A}_{11}\right), \overline{\mathrm{CS}_{4}}$ <br> to $\overline{\mathrm{CS}}_{6}$ <br> output, <br> and <br> generic <br> input/ <br> output | TPC <br> output <br> ( $\mathrm{TP}_{6}$ to <br> $\mathrm{TP}_{4}$ ), ITU <br> input and <br> output <br> (TIOCA ${ }_{2}$, <br> TIOCB $_{1}$, <br> TIOCA $_{1}$ ), <br> address <br> output <br> ( $A_{13}$ to <br> $\left.A_{11}\right), \overline{C_{S}}$ <br> to $\overline{\mathrm{CS}}_{6}$ <br> output, <br> and <br> generic <br> input/out <br> put | TPC output $\left(T P_{6}\right.$ to $\mathrm{TP}_{4}$ ), ITU input and output (TIOCA ${ }_{2}$, TIOCB $_{1}$, TIOCA $_{1}$ ), and generic input/ output |


| Port | Description | Pins | Mode 1 | Mode 2 | Mode 3 | Mode 4 | Mode 5 | Mode 6 | Mode 7 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Port A | - 8-bit I/O port <br> - Schmitt inputs output | $\begin{aligned} & \mathrm{PA}_{3} / \mathrm{TP}_{3} / \\ & \mathrm{TIOCB}_{0} / \\ & \mathrm{TCLKD}^{2}, \\ & \mathrm{PA}_{2} / \mathrm{TP}_{2} / \\ & \mathrm{TIOCA}_{0} / \\ & \mathrm{TCLKC}^{2}, \\ & \mathrm{PA}_{1} / \mathrm{TP}_{1} / \\ & \mathrm{TEND}_{0} / \mathrm{TCLKB}, \\ & \mathrm{PA}_{0} / \mathrm{TP}_{0} / \\ & \mathrm{TEND}_{0} / \mathrm{TCLKA} \end{aligned}$ | TPC output ( $\mathrm{TP}_{3}$ to $\mathrm{TP}_{0}$ ), output ( $\overline{\mathrm{TEND}}_{1}, \mathrm{TEND}_{0}$ ) from DMA controller (DMAC), ITU input and output (TCLKD, TCLKC, TCLKB, TCLKA, TIOCB $_{0}$, TIOCA $_{0}$ ), and generic input/output |  |  |  |  |  |  |
| Port B | - 8-bit I/O port <br> - Can drive LEDs <br> - $\mathrm{PB}_{3}$ to $\mathrm{PB}_{0}$ have Schmitt inputs | $\mid$ | TPC output $\left(\mathrm{TP}_{15}\right)$, DMAC input $\left(\overline{\mathrm{DREQ}}_{1}\right)$, trigger input ( $\left.\overline{\mathrm{ADTRG}}\right)$ to A/D converter, and generic input/output |  |  |  |  |  |  |
|  |  | $\begin{aligned} & \mathrm{PB}_{6} / \mathrm{TP}_{14} / \\ & \mathrm{DREQ}_{0}, / \mathrm{CS}_{7} \end{aligned}$ | TPC output $\left(\mathrm{TP}_{14}\right)$, DMAC input ( $\left.\overline{\mathrm{DREQ}}_{0}\right), \overline{\mathrm{CS}}_{7}$ output, and generic input/output |  |  |  |  |  |  |
|  |  | $\begin{aligned} & \mathrm{PB}_{5} \mathrm{TP}_{13} / \\ & \mathrm{TOCXB}_{4}, \\ & \mathrm{~PB}_{4} / \mathrm{TP}_{12} / \\ & \mathrm{TOCXA}_{4}, \\ & \mathrm{~PB}_{3} / \mathrm{TP}_{11} / \\ & \mathrm{TIOCB}_{4}, \\ & \mathrm{~PB}_{2} / \mathrm{TP}_{10} / \\ & \mathrm{TIOCA}_{4}, \\ & \mathrm{~PB}_{3} / \mathrm{TP}_{9} / \\ & \mathrm{TIOCB}_{3}, \\ & \mathrm{~PB}_{0} / \mathrm{TP}_{8} / \\ & \mathrm{TIOCA}_{3} \end{aligned}$ | TPC output $\left(\mathrm{TP}_{13}\right.$ to $\left.\mathrm{TP}_{8}\right)$, ITU input and output $\left(\mathrm{TOCXB}_{4}\right.$, TOCXA $_{4}$, $\mathrm{TIOCB}_{4}$, TIOCA $_{4}$, TIOCB $_{3}$, TIOCA $_{3}$ ), and generic input/output |  |  |  |  |  |  |

### 9.2 Port 1

### 9.2.1 Overview

Port 1 is an 8-bit input/output port with the pin configuration shown in figure 9.1. The pin functions differ between the expanded modes with on-chip ROM disabled, expanded modes with on-chip ROM enabled, and single-chip mode. In modes 1 to 4 (expanded modes with on-chip ROM disabled), they are address bus output pins ( $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ ).

In modes 5 and 6 (expanded modes with on-chip ROM enabled), settings in the port 1 data direction register (P1DDR) can designate pins for address bus output ( $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ ) or generic input. In mode 7 (single-chip mode), port 1 is a generic input/output port.

When DRAM is connected to area $3, \mathrm{~A}_{7}$ to $\mathrm{A}_{0}$ output row and column addresses in read and write cycles. For details see section 7, Refresh Controller.

Pins in port 1 can drive one TTL load and a 90-pF capacitive load. They can also drive a darlington transistor pair.


Figure 9.1 Port 1 Pin Configuration

### 9.2.2 Register Configuration

Table 9.2 summarizes the registers of port 1 .
Table 9.2 Port 1 Registers

|  |  |  | Initial Value |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Address* | Name | Abbreviation | R/W | Modes 1 to 4 | Modes 5 to 7 |
| H'FFC0 | Port 1 data direction <br> register | P1DDR | W | H'FF | H'00 |
| H'FFC2 | Port 1 data register | P1DR | R/W | H'00 | H'00 |

Note: * Lower 16 bits of the address.

## Port 1 Data Direction Register (P1DDR)

P1DDR is an 8-bit write-only register that can select input or output for each pin in port 1.


Port 1 data direction 7 to 0
These bits select input or output for port 1 pins

- Modes 1 to 4 (Expanded Modes with On-Chip ROM Disabled)

P1DDR values are fixed at 1 and cannot be modified. Port 1 functions as an address bus.

- Modes 5 and 6 (Expanded Modes with On-Chip ROM Enabled)

A pin in port 1 becomes an address output pin if the corresponding P1DDR bit is set to 1 , and a generic input pin if this bit is cleared to 0 .

- Mode 7 (Single-Chip Mode)

Port 1 functions as an input/output port. A pin in port 1 becomes an output pin if the corresponding P1DDR bit is set to 1 , and an input pin if this bit is cleared to 0 .

In modes 5 to 7, P1DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P1DDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P1DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 1 Data Register (P1DR)

P1DR is an 8-bit readable/writable register that stores port 1 output data. When this register is read, the pin logic level of a pin is read for bits for which the P1DDR setting is 0 , and the P1DR value is read for bits for which the P1DDR setting is 1 .

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P17 | P16 | P15 | P14 | P13 | P12 | P1 ${ }_{1}$ | P10 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  | Port 1 data 7 to 0 <br> These bits store data for port 1 pins |  |  |  |  |  |

P1DR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 9.3 Port 2

### 9.3.1 Overview

Port 2 is an 8 -bit input/output port with the pin configuration shown in figure 9.2. The pin functions differ according to the operating mode.

In modes 1 to 4 (expanded modes with on-chip ROM disabled), port 2 consists of address bus output pins ( $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ ). In modes 5 and 6 (expanded modes with on-chip ROM enabled), settings in the port 2 data direction register (P2DDR) can designate pins for address bus output ( $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ ) or generic input. In mode 7 (single-chip mode), port 2 is a generic input/output port.

When DRAM is connected to area $3, \mathrm{~A}_{9}$ and $\mathrm{A}_{8}$ output row and column addresses in read and write cycles. For details see section 7, Refresh Controller.

Port 2 has software-programmable built-in pull-up MOS. Pins in port 2 can drive one TTL load and a $90-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair.


Figure 9.2 Port 2 Pin Configuration

### 9.3.2 Register Configuration

Table 9.3 summarizes the registers of port 2 .
Table 9.3 Port 2 Registers

| Address* | Name | Abbreviation | R/W | Modes 1 to 4 | Modes 5 to 7 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| H'FFC1 | Port 2 data direction <br> register | P2DDR | W | H'FF | H'00 $^{\prime}$ |

Note: * Lower 16 bits of the address.

## Port 2 Data Direction Register (P2DDR)

P2DDR is an 8-bit write-only register that can select input or output for each pin in port 2.

| Bit |  | 7 | 6 5 |  | 4 | 32 |  | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\mathrm{P} 2{ }_{7}$ DDR | $\mathrm{P} 2{ }_{6} \mathrm{DDR}$ | $\mathrm{P} 2{ }_{5} \mathrm{DDR}$ | P2 ${ }_{4}$ DDR | $\mathrm{P} 2{ }_{3} \mathrm{DDR}$ | $\mathrm{P} 22_{2} \mathrm{DDR}$ | $\mathrm{P} 2{ }_{1}$ DDR | $\mathrm{P} 2{ }_{0}$ DDR |
| Modes | Initial value | - 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| 1 to 4 | Read/Write | - | - | - | - | - | - | - | - |
| Modes | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 5 to 7 | Read/Write | W | W | W | W | W | W | W | W |

## Port 2 data direction 7 to 0

These bits select input or output for port 2 pins

- Modes 1 to 4 (Expanded Modes with On-Chip ROM Disabled)

P2DDR values are fixed at 1 and cannot be modified. Port 2 functions as an address bus.

- Modes 5 and 6 (Expanded Modes with On-Chip ROM Enabled)

Following a reset, port 2 is an input port. A pin in port 2 becomes an address output pin if the corresponding P2DDR bit is set to 1 , and a generic input port if this bit is cleared to 0 .

- Mode 7 (Single-Chip Mode)

Port 2 functions as an input/output port. A pin in port 2 becomes an output port if the corresponding P2DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

In modes 5 to 7, P2DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P2DDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P2DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 2 Data Register (P2DR)

P 2 DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P} 2_{7}$ to $\mathrm{P} 2_{0}$. When a bit in P2DDR is set to 1 , if port 2 is read the value of the corresponding P2DR bit is returned. When a bit in P2DDR is cleared to 0 , if port 2 is read the corresponding pin level is read.

Bit

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{P} 2_{7}$ | $\mathrm{P} 2_{6}$ | $\mathrm{P}_{5}$ | $\mathrm{P} 2_{4}$ | $\mathrm{P} 2_{3}$ | $\mathrm{P} 2_{2}$ | $\mathrm{P} 2_{1}$ | $\mathrm{P}_{2}{ }_{0}$ |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Port 2 data 7 to 0
These bits store data for port 2 pins

P2DR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

## Port 2 Input Pull-Up MOS Control Register (P2PCR)

P2PCR is an 8-bit readable/writable register that controls the MOS input pull-up transistors in port 2.

Bit
Initial value
Read/Write

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{P} 2{ }_{7} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{6} \mathrm{PCR}$ | $\mathrm{P}{ }_{5} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{4} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{3} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{2} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{1} \mathrm{PCR}$ | $\mathrm{P} 2_{0} \mathrm{PCR}$ |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Port 2 input pull-up MOS control 7 to 0
These bits control input pull-up transistors built into port 2

In modes 5 to 7 , when a P2DDR bit is cleared to 0 (selecting generic input), if the corresponding bit from $\mathrm{P} 2_{7} \mathrm{PCR}$ to $\mathrm{P}_{0} \mathrm{PCR}$ is set to 1 , the input pull-up MOS is turned on.

P2PCR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.4 summarizes the states of the input pull-up transistors.
Table 9.4 Input Pull-Up MOS States (Port 2)

| Mode | Reset | Hardware <br> Standby Mode | Software <br> Standby Mode | Other Modes |
| :--- | :--- | :--- | :--- | :--- |
| 1 | Off | Off | Off | Off |
| 2 |  |  |  |  |
| 3 |  |  |  |  |
| 4 |  |  | On/off | On/off |
| 5 |  |  |  |  |
| 6 |  |  |  |  |

## Legend

Off: The input pull-up MOS is always off.
On/off: The input pull-up MOS is on if P2PCR $=1$ and $\operatorname{P2DDR}=0$. Otherwise, it is off.

### 9.4 Port 3

### 9.4.1 Overview

Port 3 is an 8-bit input/output port with the pin configuration shown in figure 9.3. Port 3 is a data bus in modes 1 to 6 (expanded modes) and a generic input/output port in mode 7 (single-chip mode).

Pins in port 3 can drive one TTL load and a 90-pF capacitive load. They can also drive a darlington transistor pair.

|  | Port 3 pins | Modes 1 to 6 | Mode 7 |
| :---: | :---: | :---: | :---: |
| Port 3 | $\longrightarrow \mathrm{P}_{7} / \mathrm{D}_{15}$ | $\mathrm{D}_{15}$ (input/output) | $\mathrm{P}_{3}$ (input/output) |
|  | $\longrightarrow P 3_{6} / \mathrm{D}_{14}$ | $\mathrm{D}_{14}$ (input/output) | $\mathrm{P}_{6}$ (input/output) |
|  | $\longrightarrow \mathrm{P}_{5} / \mathrm{D}_{13}$ | $\mathrm{D}_{13}$ (input/output) | $\mathrm{P}_{5}$ (input/output) |
|  | $\longrightarrow P 3_{4} / D_{12}$ | $\mathrm{D}_{12}$ (input/output) | $\mathrm{P}_{4}$ (input/output) |
|  | $\longrightarrow \mathrm{P}_{3} / \mathrm{D}_{11}$ | $\mathrm{D}_{11}$ (input/output) | $\mathrm{P}_{3}$ (input/output) |
|  | $\longrightarrow P 3_{2} / \mathrm{D}_{10}$ | $\mathrm{D}_{10}$ (input/output) | $\mathrm{P}_{2}$ (input/output) |
|  | $\longrightarrow P 3_{1} / \mathrm{D}_{9}$ | $\mathrm{D}_{9}$ (input/output) | $\mathrm{P}_{3}$ (input/output) |
|  | $\longleftrightarrow P 3_{0} / \mathrm{D}_{8}$ | $\mathrm{D}_{8}$ (input/output) | $\mathrm{P} 3_{0}$ (input/output) |

Figure 9.3 Port 3 Pin Configuration

### 9.4.2 Register Configuration

Table 9.5 summarizes the registers of port 3 .
Table 9.5 Port 3 Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFC4 | Port 3 data direction register | P3DDR | W | H'00 |
| H'FFC6 | Port 3 data register | P3DR | R/W | H'00 |

Note: * Lower 16 bits of the address.

## Port 3 Data Direction Register (P3DDR)

P3DDR is an 8-bit write-only register that can select input or output for each pin in port 3 .

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P37 ${ }_{7}$ DR | $\mathrm{P}_{6}$ DDR | P35 ${ }_{5}$ DR | P3 ${ }_{4}$ DDR | P3 $3_{3}$ DDR | P32 DDR | P3, DDR | P3 ${ }_{0}$ DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | w | W | W | w | W | W |
| Port 3 data direction 7 to 0 |  |  |  |  |  |  |  |  |

- Modes 1 to 6 (Expanded Modes)

Port 3 functions as a data bus. P3DDR is ignored.

- Mode 7 (Single-Chip Mode)

Port 3 functions as an input/output port. A pin in port 3 becomes an output port if the corresponding P3DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

P3DDR is a write-only register. Its value cannot be read. All bits return 1 when read.
P3DDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P3DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 3 Data Register (P3DR)

P3DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P}_{7}$ to $\mathrm{P}_{0}$. When a bit in P3DDR is set to 1 , if port 3 is read the value of the corresponding P3DR bit is returned. When a bit in P3DDR is cleared to 0 , if port 3 is read the corresponding pin level is read.


P3DR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 9.5 Port 4

### 9.5.1 Overview

Port 4 is an 8 -bit input/output port with the pin configuration shown in figure 9.4. The pin functions differ according to the operating mode.

In modes 1 to 6 (expanded modes), when the bus width control register (ABWCR) designates areas 0 to 7 all as 8 -bit-access areas, the chip operates in 8 -bit bus mode and port 4 is a generic input/output port. When at least one of areas 0 to 7 is designated as a 16 -bit-access area, the chip operates in 16 -bit bus mode and port 4 becomes part of the data bus. In mode 7 (single-chip mode), port 4 is a generic input/output port.

Port 4 has software-programmable built-in pull-up MOS.
Pins in port 4 can drive one TTL load and a $90-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair.

|  | Port 4 pins | Modes 1 to 6 | Mode 7 |
| :---: | :---: | :---: | :---: |
| Port 4 | $\longleftrightarrow \mathrm{P} 4_{7} / \mathrm{D}_{7}$ | $\mathrm{P} 4_{7}$ (input/output)/ $\mathrm{D}_{7}$ (input/output) | $\mathrm{P}_{7}$ (input/output) |
|  | $\longrightarrow P 4_{6} / \mathrm{D}_{6}$ | $\mathrm{P} 4_{6}$ (input/output)/ $\mathrm{D}_{6}$ (input/output) | P4 $6_{6}$ (input/output) |
|  | $\longrightarrow \mathrm{P4}_{5} / \mathrm{D}_{5}$ | P 45 (input/output)/ $\mathrm{D}_{5}$ (input/output) | $\mathrm{P4}_{5}$ (input/output) |
|  | $\longrightarrow P 4_{4} / \mathrm{D}_{4}$ | $\mathrm{P} 4_{4}$ (input/output)/ $\mathrm{D}_{4}$ (input/output) | $\mathrm{P}_{4}$ (input/output) |
|  | $\longrightarrow P 4_{3} / \mathrm{D}_{3}$ | $\mathrm{P} 4_{3}$ (input/output)/ $\mathrm{D}_{3}$ (input/output) | $\mathrm{P}_{3}$ (input/output) |
|  | $\longrightarrow P 4_{2} / D_{2}$ | $\mathrm{P} 4_{2}$ (input/output)/ $\mathrm{D}_{2}$ (input/output) | $\mathrm{P}_{2}$ ( input/output) |
|  | $\longrightarrow P 4_{1} / D_{1}$ | $\mathrm{P} 4_{1}$ (input/output)/ $\mathrm{D}_{1}$ (input/output) | P 41 (input/output) |
|  | $\longrightarrow P 4_{0} / \mathrm{D}_{0}$ | P 40 (input/output)/ $\mathrm{D}_{0}$ (input/output) | P 40 (input/output) |

Figure 9.4 Port 4 Pin Configuration

### 9.5.2 Register Configuration

Table 9.6 summarizes the registers of port 4 .
Table 9.6 Port 4 Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFC5 | Port 4 data direction register | P4DDR | W | H'00 $^{\prime}$ |
| H'FFC7 | Port 4 data register | P4DR | R/W | H'00 $^{\prime}$ |
| H'FFDA | Port 4 input pull-up MOS control <br> register | P4PCR | R/W | H'00 |

Note: * Lower 16 bits of the address.

## Port 4 Data Direction Register (P4DDR)

P4DDR is an 8-bit write-only register that can select input or output for each pin in port 4.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{P} 4{ }_{7}$ DDR | $\mathrm{P} 4{ }_{6}$ DDR | $\mathrm{P} 4{ }_{5}$ DDR | P 44 DDR | P4 ${ }_{3}$ DDR | $\mathrm{P} 4{ }_{2}$ DDR | P4 ${ }_{1}$ DDR | P40 DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | W | W | W | W | W | W |
|  |  |  | Port 4 data direction 7 to 0 <br> These bits select input or output for port 4 pins |  |  |  |  |  |

- Modes 1 to 6 (Expanded Modes)

When all areas are designated as 8 -bit-access areas, selecting 8 -bit bus mode, port 4 functions as a generic input/output port. A pin in port 4 becomes an output port if the corresponding P4DDR bit is set to 1 , and an input port if this bit is cleared to 0 .
When at least one area is designated as a 16-bit-access area, selecting 16-bit bus mode, port 4 functions as part of the data bus.

- Mode 7 (Single-Chip Mode)

Port 4 functions as an input/output port. A pin in port 4 becomes an output port if the corresponding P4DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

P4DDR is a write-only register. Its value cannot be read. All bits return 1 when read.
P4DDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

ABWCR and P4DDR are not initialized in software standby mode. When port 4 functions as a generic input/output port, if a P4DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 4 Data Register (P4DR)

P 4 DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P} 4{ }_{7}$ to $\mathrm{P} 44_{0}$. When a bit in P4DDR is set to 1 , if port 4 is read the value of the corresponding P4DR bit is returned. When a bit in P4DDR is cleared to 0 , if port 4 is read the corresponding pin level is read.

Bit

|  | 7 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{P} 4_{7}$ | $\mathrm{P} 4_{6}$ | $\mathrm{P} 4_{5}$ | $\mathrm{P} 4_{4}$ | $\mathrm{P} 4_{3}$ | $\mathrm{P} 4_{2}$ | $\mathrm{P} 4_{1}$ | $\mathrm{P} 4_{0}$ |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Port 4 data 7 to 0
These bits store data for port 4 pins

P4DR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

## Port 4 Input Pull-Up MOS Control Register (P4PCR)

P4PCR is an 8-bit readable/writable register that controls the MOS input pull-up transistors in port 4.

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{P} 4_{7} \mathrm{PCR}$ | $\mathrm{P}_{6} \mathrm{PCR}$ | $\mathrm{P} 4_{5} \mathrm{PCR}$ | $\mathrm{P} 4_{4} \mathrm{PCR}$ | $\mathrm{P} 4_{3} \mathrm{PCR}$ | $\mathrm{P} 4_{2} \mathrm{PCR}$ | $\mathrm{P} 4_{1} \mathrm{PCR}$ | $\mathrm{P} 4_{0} \mathrm{PCR}$ |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Port 4 input pull-up MOS control 7 to 0
These bits control input pull-up MOS transistors built into port 4

In mode 7 (single-chip mode), and in 8 -bit bus mode in modes 1 to 6 (expanded modes), when a P4DDR bit is cleared to 0 (selecting generic input), if the corresponding P4PCR bit is set to 1 , the input pull-up MOS transistor is turned on.

P4PCR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.7 summarizes the states of the input pull-ups MOS in the 8 -bit and 16 -bit bus modes.
Table 9.7 Input Pull-Up MOS Transistor States (Port 4)

| Mode | Reset | Hardware <br> Standby Mode | Software <br> Standby Mode | Other Modes |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 1 to 6 | 8-bit bus mode | Off | Off | On/off | On/off |
|  |  |  |  | Off | Off |
| 7 |  |  | On/off | On/off |  |

Legend
Off: The input pull-up MOS transistor is always off.
On/off: The input pull-up MOS transistor is on if $\mathrm{P} 4 \mathrm{PCR}=1$ and $\mathrm{P} 4 \mathrm{DDR}=0$. Otherwise, it is off.

### 9.6 Port 5

### 9.6.1 Overview

Port 5 is a 4-bit input/output port with the pin configuration shown in figure 9.5. The pin functions differ depending on the operating mode.

In modes 1 to 4 (expanded modes with on-chip ROM disabled), port 5 consists of address output pins $\left(\mathrm{A}_{19}\right.$ to $\left.\mathrm{A}_{16}\right)$. In modes 5 and 6 (expanded modes with on-chip ROM enabled), settings in the port 5 data direction register ( P 5 DDR ) designate pins for address bus output $\left(\mathrm{A}_{19}\right.$ to $\mathrm{A}_{16}$ ) or generic input. In mode 7 (single-chip mode), port 5 is a generic input/output port.

Port 5 has software-programmable built-in pull-up MOS transistors.
Pins in port 5 can drive one TTL load and a 90-pF capacitive load. They can also drive an LED or a darlington transistor pair.


Figure 9.5 Port 5 Pin Configuration

### 9.6.2 Register Configuration

Table 9.8 summarizes the registers of port 5 .
Table 9.8 Port 5 Registers

| Address* | Name | Abbreviation | R/W | Modes 1 to 4 | Modes 5 to 7 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| H'FFC8 | Port 5 data direction <br> register | P5DDR | W | H'FF | H'F0 |
| H'FFCA | Port 5 data register | P5DR | R/W | H'F0 | H'F0 |
| H'FFDB | Port 5 input pull-up MOS <br> control register | P5PCR | R/W | H'F0 | H'F0 |

Note: * Lower 16 bits of the address.

## Port 5 Data Direction Register (P5DDR)

P5DDR is an 8-bit write-only register that can select input or output for each pin in port 5.


- Modes 1 to 4 (Expanded Modes with On-Chip ROM Disabled) P5DDR values are fixed at 1 and cannot be modified. Port 5 functions as an address bus. The reserved bits $\left(\mathrm{P} 5_{7} \mathrm{DDR}\right.$ to $\left.\mathrm{P} 5_{4} \mathrm{DDR}\right)$ are also fixed at 1.
- Modes 5 and 6 (Expanded Modes with On-Chip ROM Enabled)

Following a reset, port 5 is an input port. A pin in port 5 becomes an address output pin if the corresponding P5DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

- Mode 7 (Single-Chip Mode)

Port 5 functions as an input/output port. A pin in port 5 becomes an output port if the corresponding P5DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

P5DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P5DDR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting, so if a P5DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 5 Data Register (P5DR)

P5DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P5}_{3}$ to $\mathrm{P} 5_{0}$. When a bit in P5DDR is set to 1 , if port 5 is read the value of the corresponding P5DR bit is returned. When a bit in P5DDR is cleared to 0 , if port 5 is read the corresponding pin level is read.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | P5 3 | P5 2 | P5 1 | P50 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
|  | Reserved bits |  |  |  |  | Port 5 data 3 to 0 These bits store data for port 5 pins |  |  |

Bits 7 to 4 are reserved. They cannot be modified and are always read as 1 .

P5DR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

## Port 5 Input Pull-Up MOS Control Register (P5PCR)

Bit

Initial value
Read/Write


Reserved bits

| 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: |
|  | $\mathrm{P} 5_{3} \mathrm{PCR}$ | $\mathrm{P} 5_{2} \mathrm{PCR}$ | $\mathrm{P} 5_{1} \mathrm{PCR}$ |
| $\mathrm{P} 5{ }_{0} \mathrm{PCR}$ |  |  |  |
| 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W |

Port 5 input pull-up MOS control 3 to 0 These bits control input pull-up MOS transistors built into port 5

P5PCR is an 8-bit readable/writable register that controls the MOS input pull-up MOS transistors in port 5 .

In modes 5 to 7, when a P5DDR bit is cleared (selecting the input port function), if the corresponding bit in P5PCR is set up 1, the input pull-up MOS transistor is turned on.

P5PCR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.9 summarizes the states of the input pull-ups MOS in each mode.
Table 9.9 Input Pull-Up MOS Transistor States (Port 5)

| Mode | Reset | Hardware <br> Standby Mode | Software <br> Standby Mode | Other Modes |
| :--- | :--- | :--- | :--- | :--- |
| 1 | Off | Off | Off | Off |
| 2 |  |  |  |  |
| 3 |  |  |  |  |
| 4 |  | Off |  |  |
| 5 |  |  |  | On/off |
| 6 |  |  |  |  |

Off: The input pull-up MOS transistor is always off.
On/off: The input pull-up MOS transistor is on if P5PCR $=1$ and P5DDR $=0$. Otherwise, it is off.

## $9.7 \quad$ Port 6

### 9.7.1 Overview

Port 6 is a 7-bit input/output port that is also used for input and output of bus control signals $(\overline{\mathrm{LWR}}, \overline{\mathrm{HWR}}, \overline{\mathrm{RD}}, \overline{\mathrm{AS}}, \overline{\mathrm{BACK}}, \overline{\mathrm{BREQ}}$, and $\overline{\mathrm{WAIT}}$ ). When DRAM is connected to area $3, \overline{\mathrm{LWR}}$, $\overline{\mathrm{HWR}}$, and $\overline{\mathrm{RD}}$ also function as $\overline{\mathrm{LW}}, \overline{\mathrm{UW}}$, and $\overline{\mathrm{CAS}}$, or $\overline{\mathrm{LCAS}}, \overline{\mathrm{UCAS}}$, and $\overline{\mathrm{WE}}$, respectively. For details see section 7, Refresh Controller.

Figure 9.6 shows the pin configuration of port 6. In modes 1 to 6 (expanded modes) the pin functions are $\overline{\mathrm{LWR}}, \overline{\mathrm{HWR}}, \overline{\mathrm{RD}}, \overline{\mathrm{AS}}, \mathrm{P} 6_{2} / \overline{\mathrm{BACK}}, \mathrm{P} 6_{1} / \overline{\mathrm{BREQ}}$, and $\mathrm{P}_{0} / \overline{\mathrm{WAIT}}$. See table 9.11 for the method of selecting the pin states. In mode 7 (single-chip mode) port 6 is a generic input/output port.

Pins in port 6 can drive one TTL load and a $30-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair.


Figure 9.6 Port 6 Pin Configuration

### 9.7.2 Register Configuration

Table 9.10 summarizes the registers of port 6 .
Table 9.10 Port 6 Registers

|  |  |  | Initial Value |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Address* | Name | Abbreviation | R/W | Mode 1 to 5 | Mode 6, 7 |
| H'FFC9 | Port 6 data direction <br> register | P6DDR | W | H'F8 | H'80 |
| H'FFCB | Port 6 data register | P6DR | R/W | H'80 | H'80 |

Note: * Lower 16 bits of the address.

## Port 6 Data Direction Register (P6DDR)

P6DDR is an 8-bit write-only register that can select input or output for each pin in port 6.


- Modes 1 to 6 (Expanded Modes)
$\mathrm{P6}_{6}$ to $\mathrm{P6}_{3}$ function as bus control output pins ( $\left.\overline{\mathrm{LWR}}, \overline{\mathrm{HWR}}, \overline{\mathrm{RD}}, \overline{\mathrm{AS}}\right) . \mathrm{P6}_{2}$ to $\mathrm{P6}_{0}$ are generic input/output pins, functioning as output port when bits $\mathrm{P} 6_{2} \mathrm{DDR}$ to $\mathrm{P} 6_{0} \mathrm{DDR}$ are set to 1 and input port when these bits are cleared to 0 .
- Mode 7 (Single-Chip Mode)

Port 6 is a generic input/output port. A pin in port 6 becomes an output port if the corresponding P6DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

Bit 7 is reserved.

P6DDR is a write-only register. Its value cannot be read. All bits return 1 when read.
P6DDR is initialized to H'80 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P6DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

Port 6 Data Register (P6DR)


P6DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P}_{6}$ to $\mathrm{P}_{6}$. When this register is read, the pin logic level is read for a bit with the corresponding P6DDR bit cleared to 0 , and the P6DR value is read for a bit with the corresponding P6DDR bit set to 1 .

Bit 7 is reserved, cannot be modified, and always read as 1 .
P6DR is initialized to $\mathrm{H}^{\prime} 80$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.11 Port 6 Pin Functions in Modes 1 to 6

| Pin | Pin Functions and Selection Method |  |  |  |  |
| :--- | :--- | :--- | :--- | :---: | :---: |
| $\mathrm{P} 6_{6} / \overline{\mathrm{LWR}}$ | Functions as follows regardless of $\mathrm{P} 6_{6} \mathrm{DDR}$ <br>  |  |  |  | $\mathrm{P} 6_{6} \mathrm{DDR}$ 0 1 <br> Pin function $\overline{\mathrm{LWR}}$ output  |

## $\mathrm{P} 6_{5} / \overline{\mathrm{HWR}}$

Functions as follows regardless of $\mathrm{P}_{5} \mathrm{DDR}$

| $\mathrm{P}_{5} \mathrm{DDR}$ | 0 |  |
| :--- | :---: | :---: |
| Pin function | $\overline{\text { HWR }}$ output |  |

$\overline{\mathrm{P} 6_{4} / \overline{\mathrm{RD}}}$
Functions as follows regardless of $\mathrm{P}_{4} \mathrm{DDR}$

| $\mathrm{P}_{4} \mathrm{DDR}$ | 0 |  | 1 |
| :--- | :--- | :--- | :--- |
| Pin function | $\overline{\mathrm{RD}}$ output |  |  |

## $\mathrm{Pb}_{3} / \overline{\mathrm{AS}}$

Functions as follows regardless of $\mathrm{P}_{3} \mathrm{DDR}$

| $\mathrm{P}_{3} \mathrm{DDR}$ | 0 |  | 1 |
| :--- | :--- | :--- | :--- |
| Pin function | $\overline{\mathrm{AS}}$ output |  |  |


| $\overline{\text { P6/ }} \overline{\text { BACK }}$ | Bit BRLE in BRCR and bit $\mathrm{P}_{2}$ DDR select the pin function as follows |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | BRLE | 0 |  | 1 |
|  | P6, DDR | 0 | 1 | - |
|  | Pin function | P6, input | P6, output | BACK output |
| $\overline{\text { P6, } / \overline{\text { BREQ }}}$ | Bit BRLE in BRCR and bit P6, DDR select the pin function as follows |  |  |  |
|  | BRLE | 0 |  | 1 |
|  | P6,DDR | 0 | 1 | - |
|  | Pin function | P6, input | P6, output | $\overline{\text { BREQ }}$ input |

P6 $\overline{\text { WAIT }}$
Bits WCE7 to WCEO in WCER, bit WMS1 in WCR, and bit P6 ${ }_{0}$ DDR select the pin function as follows

| WCER | All 1s |  |  | Not all 1s |
| :--- | :---: | :---: | :---: | :---: |
| WMS1 | 0 |  | 1 | - |
| P6 $0_{0}$ DDR | 0 | 1 | $0^{*}$ | $0^{*}$ |
| Pin function | $\mathrm{P6}_{0}$ input | $\mathrm{P6}_{0}$ output | $\overline{\text { WAIT input }}$ |  |

## 9.8 <br> Port 7

### 9.8.1 Overview

Port 7 is an 8-bit input port that is also used for analog input to the A/D converter and analog output from the D/A converter. The pin functions are the same in all operating modes. Figure 9.7 shows the pin configuration of port 7 .

## Port 7 pins



Figure 9.7 Port 7 Pin Configuration

### 9.8.2 Register Configuration

Table 9.12 summarizes the port 7 register. Port 7 is an input-only port, so it has no data direction register.

Table 9.12 Port 7 Data Register

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFCE | Port 7 data register | P7DR | R | Undetermined |

Note: * Lower 16 bits of the address.

## Port 7 Data Register (P7DR)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P 77 | P76 | P 75 | P74 | P 73 | P 72 | P7 ${ }_{1}$ | P 70 |
| Initial value | —* | —* | —* | —* | -* | —* | -* | -* |
| Read/Write | R | R | R | R | R | R | R | R |

Note: * Determined by pins $\mathrm{P} 7_{7}$ to $\mathrm{P} 7_{0}$.

When port 7 is read, the pin levels are always read.

### 9.9.1 Overview

Port 8 is a 5-bit input/output port that is also used for $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{0}$ output, $\overline{\mathrm{RFSH}}$ output, and $\overline{\mathrm{IRQ}}_{3}$ to $\overline{\mathrm{IRQ}}_{0}$ input. Figure 9.8 shows the pin configuration of port 8.

In modes 1 to 6 (expanded modes), port 8 can provide $\overline{\mathrm{CS}}_{3}$ to $\overline{\mathrm{CS}}_{0}$ output, RFSH output, and $\overline{\mathrm{IRQ}}_{3}$ to $\overline{\mathrm{IRQ}}_{0}$ input. See table 9.14 for the selection of pin functions in expanded modes.

In mode 7 (single-chip mode), port 8 can provide $\overline{\operatorname{IRQ}}_{3}$ to $\overline{\mathrm{IRQ}}_{0}$ input. See table 9.15 for the selection of pin functions in single-chip mode.

The $\overline{\mathrm{IRQ}}_{3}$ to $\overline{\mathrm{IRQ}}_{0}$ functions are selected by IER settings, regardless of whether the pin is used for input or output. For details see section 5, Interrupt Controller.

Pins in port 8 can drive one TTL load and a $90-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair.

Pins $\mathrm{P}_{2}$ to $\mathrm{P}_{0}$ have Schmitt-trigger inputs.


Figure 9.8 Port 8 Pin Configuration
Rev. 2.0, 03/01, page 272 of 822

### 9.9.2 Register Configuration

Table 9.13 summarizes the registers of port 8 .
Table 9.13 Port 8 Registers

| Address* | Name | Abbreviation | R/W | Mode 1 to 4 | Mode 5 to 7 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| H'FFCD | Port 8 data direction <br> register | P8DDR | W | H'F0 | H'E0 |
| H'FFCF | Port 8 data register | P8DR | R/W | H'E0 | H'E0 |

Note: * Lower 16 bits of the address.

## Port 8 Data Direction Register (P8DDR)

P8DDR is an 8-bit write-only register that can select input or output for each pin in port 8.


- Modes 1 to 6 (Expanded Modes)

When bits in P8DDR bit are set to $1, \mathrm{P8}_{4}$ to $\mathrm{P}_{1}$ become $\overline{\mathrm{CS}}_{0}$ to $\overline{\mathrm{CS}}_{3}$ output pins. When bits in P8DDR are cleared to 0 , the corresponding pins become input ports. In modes 1 to 4 (expanded modes with on-chip ROM disabled), following a reset only $\overline{\mathrm{CS}}_{0}$ is output. The other three pins are input ports. In modes 5 and 6 (expanded modes with on-chip ROM enabled), following a reset all four pins are input ports.
When the refresh controller is enabled, $\mathrm{P8}_{0}$ is used unconditionally for $\overline{\mathrm{RFSH}}$ output. When the refresh controller is disabled, $\mathrm{P8}_{0}$ becomes a generic input/output port according to the P8DDR setting. For details see table 9.15 .

- Mode 7 (Single-Chip Mode)

Port 8 is a generic input/output port. A pin in port 8 becomes an output port if the corresponding P8DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

P8DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P8DDR is initialized to $\mathrm{H}^{\prime} \mathrm{E} 0$ or $\mathrm{H}^{\prime} \mathrm{F} 0$ by a reset and in hardware standby mode. The reset value depends on the operating mode. In software standby mode P8DDR retains its previous setting. If a P8DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 8 Data Register (P8DR)

P8DR is an 8-bit readable/writable register that stores output data for pins $\mathrm{P}_{4}$ to P 8 . When a bit in P8DDR is set to 1 , if port 8 is read the value of the corresponding P8DR bit is returned. When a bit in P 8 DDR is cleared to 0 , if port 8 is read the corresponding pin level is read.


Bits 7 to 5 are reserved. They cannot be modified and always are read as 1 .
P8DR is initialized to H'E0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.14 Port 8 Pin Functions in Modes 1 to 6
Pin Pin Functions and Selection Method
$\mathrm{P}_{4} / \overline{\mathrm{CS}}$ 0
Bit $\mathrm{P} 8_{4}$ DDR selects the pin function as follows

| $\mathrm{P} 8_{4}$ DDR | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{4}$ input | $\overline{\mathrm{CS}}_{0}$ output |

$\mathrm{P}_{3} / \overline{\mathrm{CS}} / / \overline{\mathrm{IRQ}}_{3} \quad$ Bit $\mathrm{P}_{3} \mathrm{DDR}$ selects the pin function as follows

| $\mathrm{P}_{3}$ DDR | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{3}$ input | $\overline{\mathrm{CS}}_{1}$ output |
|  |  |  |
|  | $\overline{\mathrm{RQQ}}_{3}$ input |  |

$\overline{\mathrm{P}} \mathrm{B}_{2} / \overline{\mathrm{CS}}_{2} / \overline{\mathrm{RQ}}_{2} \quad$ Bit $\mathrm{P} 8_{2} \mathrm{DDR}$ selects the pin function as follows

| $\mathrm{P} 8_{2} \mathrm{DDR}$ | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{2}$ input | $\overline{\mathrm{CS}}_{2}$ output |
|  | $\overline{\mathrm{IRQ}}_{2}$ input |  |

$\mathrm{P}_{1} / \overline{\mathrm{CS}}_{3} / \overline{\mathrm{IRQ}}_{1} \quad$ Bit P8,DDR selects the pin function as follows

| P8,DDR | 0 | 1 |  |
| :--- | :---: | :---: | :---: |
| Pin function | $\mathrm{P}_{1}$ input | $\overline{\mathrm{CS}}_{3}$ output |  |
|  | $\overline{\mathrm{IRQ}}_{1}$ input |  |  |

$\mathrm{P} 8 / \overline{\mathrm{RFSH}} / \overline{\mathrm{RQQ}}_{0} \quad$ Bit RFSHE in RFSHCR and bit $\mathrm{P}_{0} \mathrm{DDR}$ select the pin function as follows

| RFSHE | 0 |  | 1 |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| P88DDR | 0 | 1 | - |  |  |
| Pin function | $\mathrm{P}_{0}$ input | $\mathrm{P}_{0}$ output | $\overline{\text { RFSH }}$ output |  |  |
|  |  |  |  |  | $\overline{\mathrm{IRQ}}_{0}$ input |

Table 9.15 Port 8 Pin Functions in Mode 7
Pin Pin Functions and Selection Method
$\mathrm{P}_{4} \quad$ Bit $\mathrm{P}_{4}$ DDR selects the pin function as follows

| $\mathrm{P} 8_{4}$ DDR | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{4}$ input | $\mathrm{P} 8_{4}$ output |

$\mathrm{P}_{3} / \overline{\mathrm{RQ}}_{3}$
Bit $\mathrm{P}_{3} \mathrm{DDR}$ selects the pin function as follows

| $\mathrm{P} 8_{3}$ DDR | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{3}$ input | $\mathrm{P}_{3}$ output |
|  |  |  |
|  | $\overline{\mathrm{IRQ}}_{3}$ input |  |

$\mathrm{P} 82 / \overline{\mathrm{RQ}}_{2} \quad$ Bit P82DDR selects the pin function as follows

| P82DDR | 0 | 1 |
| :--- | :---: | :---: |
| Pin function | $\mathrm{P8}_{2}$ input | $\mathrm{P} 8_{2}$ output |
|  |  |  |
|  | $\overline{\mathrm{IRQ}}_{2}$ input |  |

$\mathrm{P} 8_{1} / \overline{\mathrm{RQ}}$,
Bit P8,DDR selects the pin function as follows

| P8, DDR | 0 | 1 |  |
| :--- | :---: | :---: | :---: |
| Pin function | $\mathrm{P} 8_{1}$ input | $\mathrm{P} 8_{1}$ output |  |
|  | $\overline{\mathrm{IRQ}}{ }_{1}$ input |  |  |

$\mathrm{P} 8 / \overline{\mathrm{IRQ}}_{0} \quad$ Bit $\mathrm{P} 8_{0} \mathrm{DDR}$ select the pin function as follows

| P8 $8_{0}$ DDR | 0 | 1 |  |  |
| :--- | :---: | :---: | :---: | :---: |
| Pin function | $\mathrm{P}_{0}$ input | $\mathrm{P} 8_{0}$ output |  |  |
|  |  |  |  | $\overline{\mathrm{IRQ}}_{0}$ input |

## $9.10 \quad$ Port 9

### 9.10.1 Overview

Port 9 is a 6-bit input/output port that is also used for input and output $\left(\mathrm{TxD}_{0}, \operatorname{TxD}_{1}, \mathrm{RxD}_{0}, \mathrm{RxD}_{1}\right.$, $\mathrm{SCK}_{0}, \mathrm{SCK}_{1}$ ) by serial communication interface channels 0 and 1 (SCI0 and SCI1), and for $\overline{\mathrm{IRQ}}_{5}$ and $\overline{\mathrm{IRQ}}_{4}$ input. See table 9.17 for the selection of pin functions.

The $\overline{\mathrm{IRQ}}_{5}$ and $\overline{\mathrm{IRQ}}_{4}$ functions are selected by IER settings, regardless of whether the pin is used for input or output. For details see section 5, Interrupt Controller.

Port 9 has the same set of pin functions in all operating modes. Figure 9.9 shows the pin configuration of port 9 .

Pins in port 9 can drive one TTL load and a $30-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair.


Figure 9.9 Port 9 Pin Configuration

### 9.10.2 Register Configuration

Table 9.16 summarizes the registers of port 9 .
Table 9.16 Port 9 Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFD0 | Port 9 data direction register | P9DDR | W | H'C0 |
| H'FFD2 | Port 9 data register | P9DR | R/W | H'C0 |

Note: * Lower 16 bits of the address.

## Port 9 Data Direction Register (P9DDR)

P9DDR is an 8-bit write-only register that can select input or output for each pin in port 9.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | P9 ${ }_{5}$ DDR | P9 ${ }_{4}$ DDR | P93 ${ }_{3}$ DDR | P9 ${ }_{2}$ DDR | P9 ${ }_{1}$ DDR | P9 ${ }_{0}$ DDR |
| Initial value | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | W | W | W | W | W | W |
|  | Reserved bits |  |  | Port 9 data direction 5 to 0 These bits select input or output for port 9 pins |  |  |  |  |

A pin in port 9 becomes an output port if the corresponding P9DDR bit is set to 1 , and an input port if this bit is cleared to 0 .

P9DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P9DDR is initialized to $\mathrm{H}^{\prime} \mathrm{C} 0$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P9DDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port 9 Data Register (P9DR)

P9DR is an 8 -bit readable/writable register that stores output data for pins $\mathrm{P}_{5}$ to $\mathrm{P} 9{ }_{0}$. When a bit in P9DDR is set to 1 , if port 9 is read the value of the corresponding P9DR bit is returned. When a bit in P9DDR is cleared to 0 , if port 9 is read the corresponding pin level is read.


Bits 7 and 6 are reserved. They cannot be modified and are always read as 1 .
P9DR is initialized to $\mathrm{H}^{\prime} \mathrm{C} 0$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

Table 9.17 Port 9 Pin Functions
Pin
Pin Functions and Selection Method
$\mathrm{P9}_{5} / \mathrm{SCK}_{1} / \overline{\mathrm{IRQ}}_{5}$
Bit C//̄ in SMR of SCl1, bits CKE0 and CKE1 in SCR of SCI1, and bit P9 ${ }_{5}$ DDR select the pin function as follows

| CKE1 | 0 |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| C/ $\overline{\mathrm{A}}$ | 0 |  |  |  | 1 |  |
| CKE0 | 0 |  | 1 | - | - |  |
| $\mathrm{P9}_{5}$ DDR | 0 | 1 | - | - | - |  |
| Pin function | $\mathrm{P9}_{5}$ <br> input | $\mathrm{P9}_{5}$ <br> output | $\mathrm{SCK}_{1}$ <br> output | $\mathrm{SCK}_{1}$ <br> output | $\mathrm{SCK}_{1}$ <br> input |  |
|  | $\overline{\mathrm{TRQ}}_{5}$ input |  |  |  |  |  |

$\mathrm{P9}_{4} / \mathrm{SCK}_{0} / \overline{\mathrm{IRQ}}_{4} \quad \mathrm{Bit} \mathrm{C} / \overline{\mathrm{A}}$ in SMR of SCIO, bits CKE0 and CKE1 in SCR of SCIO, and bit $\mathrm{P9}_{4} \mathrm{DDR}$ select the pin function as follows

| CKE1 | 0 |  |  |  | 1 |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| C//̄ | 0 |  |  |  | 1 |  |
| CKE0 | 0 |  | - | - |  |  |
| $\mathrm{P9}_{4}$ DDR | 0 | 1 | - | - | - |  |
| Pin function | $\mathrm{P9}_{4}$ <br> input | $\mathrm{P9}_{4}$ <br> output | $\mathrm{SCK}_{0}$ <br> output | $\mathrm{SCK}_{0}$ <br> output | $\mathrm{SCK}_{0}$ <br> input |  |
|  | $\overline{\mathrm{RQ}}_{4}$ input |  |  |  |  |  |

$\mathrm{Pg}_{3} / \mathrm{RxD} \mathrm{D}_{1}$
Bit RE in SCR of SCI1 and bit $\mathrm{P9}_{3}$ DDR select the pin function as follows

| RE | 0 |  | 1 |
| :--- | :---: | :---: | :---: |
| $\mathrm{P9}_{3}$ DDR | 0 | 1 | - |
| Pin function | $\mathrm{P}_{3}$ input | $\mathrm{P9}_{3}$ output | RxD 1 input |

P9 ${ }_{2} / R x D_{0}$
Bit RE in SCR of SCIO, bit SMIF in SCMR, and bit P9 ${ }_{2}$ DDR select the pin function as follows

| SMIF | 0 |  |  | 1 |
| :--- | :---: | :---: | :---: | :---: |
| RE | 0 |  | 1 | - |
| $\mathrm{Pg}_{2} \mathrm{DDR}$ | 0 | 1 | - | - |
| Pin function | $\mathrm{P9}_{2}$ input | $\mathrm{P9}_{2}$ output | $\mathrm{RxD}_{0}$ input | $\mathrm{RxD}_{0}$ input |

Bit TE in SCR of SCI1 and bit P9, DDR select the pin function as follows

| TE | 0 |  | 1 |
| :--- | :---: | :---: | :---: |
| P9,DDR | 0 | 1 | - |
| Pin function | $\mathrm{P}_{1}$ input | $\mathrm{P}_{1}$ output | $\mathrm{TxD}_{1}$ output |

$\mathrm{Pg}_{0} / \mathrm{TxD}_{0} \quad$ Bit TE in SCR of SCIO, bit SMIF in SCMR, and bit P9 ${ }_{0}$ DDR select the pin function as follows

| SMIF | 0 |  |  | 1 |
| :--- | :---: | :---: | :---: | :---: |
| TE | 0 |  | 1 | - |
| $\mathrm{P9}_{0}$ DDR | 0 | 1 | - | - |
| Pin function | $\mathrm{P9}_{0}$ input | $\mathrm{P9}_{0}$ output | $\mathrm{TxD}_{0}$ output | $\mathrm{TxD}_{0}$ output* |

Note: * Functions as the $\mathrm{TxD}_{0}$ output pin, but there are two states: one in which the pin is driven, and another in which the pin is at high-impedance.

### 9.11.1 Overview

Port A is an 8-bit input/output port that is also used for output $\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{0}\right)$ from the programmable timing pattern controller (TPC), input and output ( $\mathrm{TIOCB}_{2}$, TIOCA $_{2}$, TIOCB $_{1}$, TIOCA $_{1}$, TIOCB $_{0}$, TIOCA $_{0}$, TCLKD, TCLKC, TCLKB, TCLKA) by the 16-bit integrated timer unit (ITU), output $\left(\overline{\mathrm{TEND}}_{1}, \overline{\mathrm{TEND}}_{0}\right.$ ) from the DMA controller (DMAC), $\overline{\mathrm{CS}}_{4}$ to $\overline{\mathrm{CS}}_{6}$ output, and address output ( $\mathrm{A}_{23}$ to $\mathrm{A}_{20}$ ). A reset or hardware standby leaves port A as an input port, except that in modes 3,4 , and 6, one pin is always used for $\mathrm{A}_{20}$ output. Usage of pins for TPC, ITU, and DMAC input and output is described in the sections on those modules. For output of address bits $\mathrm{A}_{23}$ to $\mathrm{A}_{21}$ in modes 3, 4, and 6 , see section 6.2.5, Bus Release Control Register (BRCR). For output of $\overline{\mathrm{CS}}_{4}$ to $\overline{\mathrm{CS}}_{6}$ in modes 1 to 6 , see section 6.3.2, Chip Select Signals. Pins not assigned to any of these functions are available for generic input/output. Figure 9.10 shows the pin configuration of port A.

Pins in port A can drive one TTL load and a $30-\mathrm{pF}$ capacitive load. They can also drive a darlington transistor pair. Port A has Schmitt-trigger inputs.

## Port A pins



## Pin functions in modes 1,2 , and 5

$\mathrm{PA}_{7}$ (input/output)/TP ${ }_{7}$ (output)/TIOCB 2 (input/output)
$\mathrm{PA}_{6}$ (input/output)/TP ${ }_{6}$ (output)/TIOCA 2 (input/output) $/ \overline{C S}_{4}$ (output)
$\mathrm{PA}_{5}$ (input/output)/TP 5 (output)/TIOCB ${ }_{1}$ (input/output)/ $\overline{C S}_{5}$ (output)
$\mathrm{PA}_{4}$ (input/output)/TP 4 (output)/TIOCA ${ }_{1}$ (input/output)/ $\overline{C S}_{6}$ (output)
$\mathrm{PA}_{3}$ (input/output)/TP 3 (output)/TIOCB ${ }_{0}$ (input/output)/TCLKD (input)
$\mathrm{PA}_{2}$ (input/output)/TP 2 (output)/TIOCA ${ }_{0}$ (input/output)/TCLKC (input)
$\mathrm{PA}_{1}$ (input/output)/TP ${ }_{1}$ (output) $/ \overline{T E N D}_{1}$ (output)/TCLKB (input)
$\mathrm{PA}_{0}$ (input/output) $/ T \mathrm{P}_{0}$ (output)/TEND 0 (output)/TCLKA (input)
Pin functions in modes 3, 4, and 6
$\mathrm{A}_{20}$ (output)
$\mathrm{PA}_{6}$ (input/output) $/ T \mathrm{TP}_{6}$ (output)/TIOCA ${ }_{2}$ (input/output) $/ \mathrm{A}_{21}$ (output) $/ \overline{C S}_{4}$ (output)
$\mathrm{PA}_{5}$ (input/output) $/ T \mathrm{P}_{5}$ (output)/TIOCB ${ }_{1}$ (input/output) $/ \mathrm{A}_{22}$ (output) $/ \overline{C S}_{5}$ (output)
$\mathrm{PA}_{4}$ (input/output)/TP 4 (output)/TIOCA ${ }_{1}$ (input/output) $/ \mathrm{A}_{23}$ (output) $/ \overline{C S}_{6}$ (output)
$\mathrm{PA}_{3}$ (input/output)/TP 3 (output)/TIOCB ${ }_{0}$ (input/output)/TCLKD (input)
$\mathrm{PA}_{2}$ (input/output)/TP 2 (output)/TIOCA ${ }_{0}$ (input/output)/TCLKC (input)
$\mathrm{PA}_{1}$ (input/output) $/ T P_{1}$ (output) $/ \mathrm{TEND}_{1}$ (output)/TCLKB (input)
$\mathrm{PA}_{0}$ (input/output) $/ T \mathrm{P}_{0}$ (output) $/ \mathrm{TEND}_{0}$ (output)/TCLKA (input)

## Pin functions in mode 7

$\mathrm{PA}_{7}$ (input/output)/TP ${ }_{7}$ (output)/TIOCB ${ }_{2}$ (input/output)
$\mathrm{PA}_{6}$ (input/output)/TP ${ }_{6}$ (output)/TIOCA ${ }_{2}$ (input/output)
$\mathrm{PA}_{5}$ (input/output)/TP ${ }_{5}$ (output)/TIOCB ${ }_{1}$ (input/output)
$\mathrm{PA}_{4}$ (input/output)/TP ${ }_{4}$ (output)/TIOCA ${ }_{1}$ (input/output)
$\mathrm{PA}_{3}$ (input/output)/TP 3 (output) $/$ TIOCB ${ }_{0}$ (input/output)/TCLKD (input)
$\mathrm{PA}_{2}$ (input/output)/TP 2 (output)/TIOCA ${ }_{0}$ (input/output)/TCLKC (input)
$\mathrm{PA}_{1}$ (input/output)/TP ${ }_{1}$ (output) $/ \mathrm{TEND}_{1}$ (output)/TCLKB (input)
$\mathrm{PA}_{0}$ (input/output)/TP ${ }_{0}$ (output)/TEND 0 (output)/TCLKA (input)

Figure 9.10 Port A Pin Configuration

### 9.11.2 Register Configuration

Table 9.18 summarizes the registers of port A .
Table 9.18 Port A Registers

| Address* | Name | Abbreviation | R/W | Initial Value |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Modes <br> 1, 2, 5 and 7 | Modes 3, 4, and 6 |
| H'FFD1 | Port A data direction register | PADDR | W | H'00 | H'80 |
| H'FFD3 | Port A data register | PADR | R/W | $\mathrm{H}^{\prime} \mathrm{OO}$ | $\mathrm{H}^{\prime} \mathrm{OO}$ |

Note: * Lower 16 bits of the address.

## Port A Data Direction Register (PADDR)

PADDR is an 8-bit write-only register that can select input or output for each pin in port A. When pins are used for TPC output, the corresponding PADDR bits must also be set.

| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\mathrm{PA}_{7}$ DDR | $\mathrm{PA}_{6}$ DDR | PA5 ${ }_{5}$ DR | $\mathrm{PA}_{4}$ DDR | $\mathrm{PA}_{3}$ DDR | $\mathrm{PA}_{2}$ DDR | PA ${ }_{1}$ DDR | PA ${ }_{0}$ DDR |
| Modes $3,4,$ | Initial value | - 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| and 6 | Read/Write | e | W | W | W | W | W | W | W |
| Modes | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| $\begin{aligned} & 1,2,5, \\ & \text { and } 7 \end{aligned}$ | Read/Write | W | W | W | W | W | W | W | W |

Port A data direction 7 to 0
These bits select input or output for port A pins

A pin in port A becomes an output pin if the corresponding PADDR bit is set to 1 , and an input pin if this bit is cleared to 0 . In modes 3,4 , and $6, \mathrm{PA}_{7} \mathrm{DDR}$ is fixed at 1 and $\mathrm{PA}_{7}$ functions as an address output pin.

PADDR is a write-only register. Its value cannot be read. All bits return 1 when read.
PADDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode in modes $1,2,5$, and 7 .

It is initialized to $\mathrm{H}^{\prime} 80$ by a reset and in hardware standby mode in modes 3,4 , and 6 . In software standby mode it retains its previous setting. If a PADDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port A Data Register (PADR)

PADR is an 8-bit readable/writable register that stores output data for pins $\mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$. When a bit in PADDR is set to 1 , if port A is read the value of the corresponding PADR bit is returned. When a bit in PADDR is cleared to 0 , if port A is read the corresponding pin level is read.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{PA}_{7}$ | $\mathrm{PA}_{6}$ | $\mathrm{PA}_{5}$ | $\mathrm{PA}_{4}$ | $\mathrm{PA}_{3}$ | $\mathrm{PA}_{2}$ | PA ${ }_{1}$ | PA ${ }_{0}$ |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Port A data 7 to 0 |  |  |  |  |  |  |  |  |

PADR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 9.11.3 Pin Functions

Table 9.19 describes the selection of pin functions.
Table 9.19 Port A Pin Functions
Pin Pin Functions and Selection Method
$\mathrm{PA}_{7} / \mathrm{TP}_{7} / \quad$ The mode setting, ITU channel 2 settings (bit PWM2 in TMDR and bits IOB2 to IOB0 in $\mathrm{TIOCB}_{2} / \mathrm{A}_{20}$ TIOR2), bit NDER7 in NDERA, and bit PA DDR in PADDR select the pin function as follows

| Mode | 1, 2, 5, 7 |  |  |  | 3, 4, 6 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ITU channel 2 settings | (1) in table below |  |  | (2) in table below | - |
| PA ${ }_{7}$ DDR | - | 0 | 1 | 1 | - |
| NDER7 | - | - | 0 | 1 | - |
| Pin function | TIOCB $_{2}$ output | PA ${ }_{7}$ input | $\mathrm{PA}_{7}$ output | TP ${ }_{7}$ output | $\mathrm{A}_{20}$ output |
|  |  | TIOCB $_{2}$ input* |  |  |  |

Note: $* \mathrm{TIOCB}_{2}$ input when IOB2 $=1$ and PWM2 $=0$.

| ITU <br> channel 2 <br> settings | (2) | (1) |  | (2) |
| :--- | :---: | :---: | :---: | :---: |
| IOB2 |  |  |  |  |
| IOB1 | 0 | 0 | 1 | 1 |
| IOB0 | 0 | 1 | - | - |

$\mathrm{PA}_{6} / \mathrm{TP}_{6} / \quad$ The mode setting, bit $\mathrm{A}_{21} \mathrm{E}$ in BRCR, bit CS4E in CSCR, ITU channel 2 settings (bit TIOCA $_{2} /$ PWM2 in TMDR and bits IOA2 to IOA0 in TIOR2), bit NDER6 in NDERA, and bit $\mathrm{A}_{21} / \overline{\mathrm{CS}}_{4}$ $\mathrm{PA}_{6}$ DDR in PADDR select the pin function as follows

| Mode | 1, 2, 5 |  |  |  |  | 3, 4, 6 |  |  |  |  |  | 7 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CS4E | 0 |  |  |  | 1 | 0 |  |  |  |  | 1 | - |  |  |  |
| $\mathrm{A}_{21} \mathrm{E}$ | - |  |  |  | - | 1 |  |  |  | 0 | - | - |  |  |  |
| ITU channel 2 settings | (1) in table below | (2) in table below |  |  | - | (1) in table below | (2) in table below |  |  | - | - | (1) in table below | (2) in table below |  |  |
| PA ${ }_{6}$ DDR | - | 0 | 1 | 1 | - | - | 0 | 1 | 1 | - | - | - | 0 | 1 | 1 |
| NDER6 | - | - | 0 | 1 | - | - | - | 0 | 1 | - | - | - | - | 0 | 1 |
| Pin <br> function | $\mathrm{TIOCA}_{2}$ output | $\mathrm{PA}_{6}$ input | $\mathrm{PA}_{6}$ <br> out- <br> put | $\mathrm{TP}_{6}$ <br> out- <br> put | $\overline{\mathrm{CS}}_{4}$ <br> out- <br> put | TIOCA $_{2}$ output | $\mathrm{PA}_{6}$ input | $\mathrm{PA}_{6}$ <br> out- <br> put | $\mathrm{TP}_{6}$ <br> out- <br> put | $\mathrm{A}_{21}$ <br> out- <br> put | $\overline{\mathrm{CS}}_{4}$ <br> out- <br> put | TIOCA $_{2}$ output | $\begin{gathered} \mathrm{PA}_{6} \\ \text { input } \end{gathered}$ | $\mathrm{PA}_{6}$ <br> out- <br> put | $\mathrm{TP}_{6}$ <br> out- <br> put |
|  |  | TIOCA $_{2}$ input* |  |  |  |  | TIOCA $_{2}$ input* |  |  |  |  |  | TIOCA $_{2}$ input* |  |  |

Note: $*$ TIOCA $_{2}$ input when IOA2 $=1$.

| ITU <br> channel 2 <br> settings | (2) | (2) | (1) |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| PWM2 | 0 |  |  |  | 1 |
| IOA2 | 0 | 0 | 1 | - |  |
| IOA1 | 0 | 1 | - | - | - |
| IOA0 | 0 | - | - |  |  |

$\mathrm{PA}_{5} / \mathrm{TP}_{5}$ The mode setting, bit $\mathrm{A}_{22} \mathrm{E}$ in BRCR, bit CS5E in CSCR, ITU channel 1 settings (bit TIOCB/ $/$ $\mathrm{A}_{22} / \mathrm{CS}_{5}$ PWM1 in TMDR and bits IOB2 to IOB0 in TIOR1), bit NDER5 in NDERA, and bit $\mathrm{PA}_{5}$ DDR in PADDR select the pin function as follows

| Mode | 1, 2, 5 |  |  |  |  | 3, 4, 6 |  |  |  |  |  | 7 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CS5E | 0 |  |  |  | 1 | 0 |  |  |  |  | 1 | - |  |  |  |
| $\mathrm{A}_{22} \mathrm{E}$ | - |  |  |  | - | 1 |  |  |  | 0 | - | - |  |  |  |
| ITU <br> channel 1 settings | (1) in <br> table <br> below | (2) in table below |  |  | - | (1) in <br> table <br> below | (2) in table below |  |  | - | - | (1) in table below | (2) in table below |  |  |
| PA ${ }_{5}$ DDR | - | 0 | 1 | 1 | - | - | 0 | 1 | 1 | - | - | - | 0 | 1 | 1 |
| NDER5 | - | - | 0 | 1 | - | - | - | 0 | 1 | - | - | - | - | 0 | 1 |
| Pin function | TIOCB ${ }_{1}$ output | $\begin{gathered} \mathrm{PA}_{5} \\ \text { input } \end{gathered}$ | $\mathrm{PA}_{5}$ <br> out- <br> put | $\mathrm{TP}_{5}$ <br> out- <br> put | $\mathrm{CS}_{5}$ <br> out- <br> put | TIOCB $_{1}$ output | $P_{5}$ input | $P_{5}$ <br> out- <br> put | $\mathrm{TP}_{5}$ <br> out- <br> put | $\mathrm{A}_{22}$ <br> out- <br> put | $\mathrm{CS}_{5}$ <br> out- <br> put | TIOCB $_{1}$ output | $\begin{aligned} & \mathrm{PA}_{5} \\ & \text { input } \end{aligned}$ | $\mathrm{PA}_{5}$ <br> out- <br> put | $\mathrm{TP}_{5}$ output |
|  |  | TIOCB $_{1}$ input* |  |  |  |  | TIOCB $_{1}$ input* |  |  |  |  |  | TIOCB $_{1}$ input* |  |  |

Note: * TIOCB $_{1}$ input when IOB2 $=1$ and PWM1 $=0$.

| ITU <br> channel 1 <br> settings | (2) | (1) |  | (2) |
| :--- | :---: | :---: | :---: | :---: |
| IOB2 |  | 0 | 1 | 1 |
| IOB1 | 0 | 0 | - |  |
| IOB0 | 0 | 1 | - | - |

$\mathrm{PA}_{4} / \mathrm{TP}_{4} /$ The mode setting, bit $\mathrm{A}_{23} \mathrm{E}$ in BRCR, bit CS6E in CSCR, ITU channel 1 settings (bit TIOCA $/$ PWM1 in TMDR and bits IOA2 to IOA0 in TIOR1), bit NDER4 in NDERA, and bit $\mathrm{A}_{23} / \mathrm{CS}_{6}$ $\mathrm{PA}_{4} \mathrm{DDR}$ in PADDR select the pin function as follows

| Mode | 1, 2, 5 |  |  |  |  | 3, 4, 6 |  |  |  |  |  | 7 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CS6E | 0 |  |  |  | 1 | 0 |  |  |  |  | 1 | - |  |  |  |
| $\mathrm{A}_{23} \mathrm{E}$ | - |  |  |  | - | 1 |  |  |  | 0 | - | - |  |  |  |
| ITU <br> channel 2 <br> settings | (1) in <br> table <br> below | (2) in table below |  |  | - | (1) in <br> table <br> below | (2) in table below |  |  | - | - | (1) in table below | (2) in table below |  |  |
| $\mathrm{PA}_{4}$ DDR | - | 0 | 1 | 1 | - | - | 0 | 1 | 1 | - | - | - | 0 | 1 | 1 |
| NDER4 | - | - | 0 | 1 | - | - | - | 0 | 1 | - | - | - | - | 0 | 1 |
| Pin <br> function | TIOCA $_{1}$ output | $\begin{aligned} & \mathrm{PA}_{4} \\ & \text { input } \end{aligned}$ | $\mathrm{PA}_{4}$ <br> out- <br> put | $\mathrm{TP}_{4}$ <br> out- <br> put | $\overline{\mathrm{CS}}_{6}$ <br> out- <br> put | TIOCA $_{1}$ output | $\begin{gathered} \mathrm{PA}_{4} \\ \text { input } \end{gathered}$ | $\mathrm{PA}_{4}$ <br> out- <br> put | $\mathrm{TP}_{4}$ <br> out- <br> put | $\mathrm{A}_{23}$ <br> out- <br> put | $\overline{\mathrm{CS}}_{6}$ <br> out- <br> put | TIOCA $_{1}$ output | $\begin{gathered} \mathrm{PA}_{4} \\ \text { input } \end{gathered}$ | $\mathrm{PA}_{4}$ <br> out- <br> put | $\mathrm{TP}_{4}$ <br> out- <br> put |
|  |  | TIOCA $_{1}$ input* |  |  |  |  | TIOCA $_{1}$ input* |  |  |  |  |  | TIOCA ${ }_{1}$ input* |  |  |

Note: * TIOCA1 input when IOA2 $=1$.

| ITU <br> channel 1 <br> settings | (2) | (2) | (1) |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| PWM1 | 0 |  |  |  |  |
| IOA2 | 0 | 0 | 1 | - |  |
| IOA1 | 0 | 1 | 1 | - | - |
| IOA0 | 0 | - | - | - |  |

$\mathrm{PA}_{3} / \mathrm{TP}_{3} /$
TIOCB ${ }^{\prime}$ TCLKD

ITU channel 0 settings (bit PWMO in TMDR and bits IOB2 to IOBO in TIORO), bits TPSC2 to TPSC0 in TCR4 to TCR0, bit NDER3 in NDERA, and bit PA ${ }_{3}$ DDR in PADDR select the pin function as follows

| ITU channel 0 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| PA3 DDR | - | 0 | 1 | 1 |  |
| NDER3 | - | - | 0 | 1 |  |
| Pin function | TIOCB $_{0}$ output | $\mathrm{PA}_{3}$ input | $\mathrm{PA}_{3}$ output | $\mathrm{TP}_{3}$ output |  |
|  |  | TIOCB $_{0}$ input*1 |  |  |  |
|  | TCLKD input $^{* 2}$ |  |  |  |  |

Notes: 1. $\mathrm{TIOCB}_{0}$ input when IOB2 $=1$ and $\mathrm{PWM} 0=0$.
2. TCLKD input when TPSC2 $=$ TPSC1 $=$ TPSC0 $=1$ in any of TCR4 to TCRO.

| ITU channel 0 <br> settings | (2) | (1) |  | (2) |
| :--- | :---: | :---: | :---: | :---: |
| IOB2 | 0 | 1 |  |  |
| IOB1 | 0 | 0 | 1 | - |
| IOB0 | 0 | 1 | - | - |

$\mathrm{PA}_{2} / \mathrm{TP}_{2} /$
TIOCA ${ }_{0} /$ TCLKC

ITU channel 0 settings (bit PWM0 in TMDR and bits IOA2 to IOA0 in TIORO), bits TPSC2 to TPSC0 in TCR4 to TCR0, bit NDER2 in NDERA, and bit PA ${ }_{2}$ DDR in PADDR select the pin function as follows

| ITU channel 0 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |
| :--- | :---: | :---: | :---: | :---: |
| PA DDR | - | 0 | 1 | 1 |
| NDER2 | - | - | 0 | 1 |
| Pin function | IOCA $_{0}$ output | $\mathrm{PA}_{2}$ input | $\mathrm{PA}_{2}$ output | $\mathrm{TP}_{2}$ output |
|  |  | TIOCA $_{0}$ input*1 |  |  |
|  | TCLKC input* $^{*}$ |  |  |  |

Notes: 1. TIOCA $_{0}$ input when IOA2 $=1$.
2. TCLKC input when $\operatorname{TPSC} 2=\operatorname{TPSC} 1=1$ and $\operatorname{TPSC} 0=0$ in any of TCR4 to TCRO.

| ITU channel 0 <br> settings | (2) | (1) |  | (2) | (1) |
| :--- | :---: | :---: | :---: | :---: | :---: |
| PWM0 | 0 |  |  |  |  |
| IOA2 | 0 | 1 | - | - |  |
| IOA1 | 0 | 0 | 1 | - |  |
| IOA0 | 0 | 1 | - | - | - |

DMAC channel 1 settings (bits DTS2/1/0A and DTS2/1/0B in DTCR1A and DTCR1B), TCLKB/ bit NDER1 in NDERA, and bit PA,DDR in PADDR select the pin function as follows
TEND ${ }_{1}$

| DMAC <br> channel 1 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| PA,DDR | - | 0 | 1 | 1 |  |
| NDER1 | - | - | 0 | 1 |  |
| Pin function | $\overline{T E N D}_{1}$ output | PA, input $^{3}$ | PA $_{1}$ output | TP $_{1}$ output |  |
|  | TCLKB input* |  |  |  |  |

Note: *TCLKB input when MDF $=1$ in TMDR, or when TPSC2 $=1$, TPSC1 $=0$, and TPSC0 $=1$ in any of TCR4 to TCR0.

| DMAC <br> channel 1 <br> settings | (2) |  | (1) | (2) | (1) | (2) | (1) |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTS2A, <br> DTS1A | Not both 1 |  |  | Both 1 |  |  |  |  |
| DTS0A | - |  |  | 0 | 0 | 1 | 1 | 1 |
| DTS2B | 0 | 1 | 1 | 0 | 1 | 0 | 1 | 1 |
| DTS1B | - | 0 | 1 | - | - | - | 0 | 1 |

DMAC channel 0 settings (bits DTS2/1/0A and DTS2/1/0B in DTCROA and DTCROB), TCLKA/ TEND。

| DMAC <br> channel 0 <br> settings | (1) in table below | (2) in table below |  |  |
| :---: | :---: | :---: | :---: | :---: |
| PA ${ }_{0}$ DDR | - | 0 | 1 | 1 |
| NDER0 | - | - | 0 | 1 |
| Pin function | $\overline{\mathrm{TEND}}{ }_{0}$ output | PA ${ }_{0}$ input | PA ${ }_{0}$ output | TP ${ }_{0}$ output |
|  |  | TCLKA input* |  |  |

Note: *TCLKA input when MDF $=1$ in TMDR, or when TPSC2 $=1$ and TPSC1 $=0$ in any of TCR4 to TCRO.

| DMAC <br> channel 0 <br> settings | (2) |  | (1) | (2) | (1) | (2) | (1) |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTS2A, <br> DTS1A | Not both 1 |  |  | Both 1 |  |  |  |  |
| DTS0A | - |  |  | 0 | 0 | 1 | 1 | 1 |
| DTS2B | 0 | 1 | 1 | 0 | 1 | 0 | 1 | 1 |
| DTS1B | - | 0 | 1 | - | - | - | 0 | 1 |

### 9.12 Port B

### 9.12.1 Overview

Port B is an 8-bit input/output port that is also used for output $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{8}\right)$ from the programmable timing pattern controller (TPC), input/output $\left(\mathrm{TIOCB}_{4}, \mathrm{TIOCB}_{3}\right.$, TIOCA $_{4}$, , $^{2}$ DOCA $\left._{3}\right)$ and output $\left(\mathrm{TOCXB}_{4}, \mathrm{TOCXA}_{4}\right)$ by the 16 -bit integrated timer unit (ITU), input $\left(\overline{\mathrm{DREQ}}_{1}, \overline{\mathrm{DREQ}}_{0}\right)$ to the DMA controller (DMAC), $\overline{\mathrm{ADTRG}}$ input to the A/D converter, and $\overline{\mathrm{CS}}_{7}$ output. A reset or hardware standby leaves port B as an input port. Usage of pins for TPC, ITU, DMAC, and A/D converter input and output is described in the sections on those modules. For output of $\overline{\mathrm{CS}}_{7}$ in modes 1 to 6, see section 6.3.2, Chip Select Signals. Pins not assigned to any of these functions are available for generic input/output. Figure 9.11 shows the pin configuration of port B.

Pins in port B can drive one TTL load and a $30-\mathrm{pF}$ capacitive load. They can also drive an LED or darlington transistor pair. $\mathrm{Pins}_{3} \mathrm{~PB}_{3}$ to $\mathrm{PB}_{0}$ have Schmitt-trigger inputs.


## Pin functions in modes 1 to 6

$\mathrm{PB}_{7}$ (input/output)/TP 15 (output)/ $/ \overline{\mathrm{DREQ}}_{1}$ (input)/ $\overline{\text { ADTRG }}$ (input)
$\mathrm{PB}_{6}$ (input/output)/TP 14 (output)/DREQ ${ }_{0}$ (input) $/ \overline{\mathrm{CS}}_{7}$ (output)
$\mathrm{PB}_{5}$ (input/output)/TP ${ }_{13}$ (output)/TOCXB ${ }_{4}$ (output)
$\mathrm{PB}_{4}$ (input/output)/TP ${ }_{12}$ (output)/TOCXA 4 (output)
$\mathrm{PB}_{3}$ (input/output)/TP ${ }_{11}$ (output)/TIOCB ${ }_{4}$ (input/output)
$\mathrm{PB}_{2}$ (input/output)/TP ${ }_{10}$ (output)/TIOCA 4 (input/output)
$\mathrm{PB}_{1}$ (input/output)/TP 9 (output) $/ \mathrm{TIOCB}_{3}$ (input/output)
$\mathrm{PB}_{0}$ (input/output)/TP 8 (output)/TIOCA 3 (input/output)

## Pin functions in mode 7

$\mathrm{PB}_{7}$ (input/output)/TP 15 (output)/ $/ \overline{\mathrm{DREQ}}_{1}$ (input)/ $\overline{\mathrm{ADTRG}}$ (input)
$\mathrm{PB}_{6}$ (input/output)/TP ${ }_{14}$ (output)/ $\overline{\mathrm{DREQ}}_{0}$ (input)
$\mathrm{PB}_{5}$ (input/output)/TP ${ }_{13}$ (output)/TOCXB ${ }_{4}$ (output)
$\mathrm{PB}_{4}$ (input/output)/TP ${ }_{12}$ (output)/TOCXA 4 (output)
$\mathrm{PB}_{3}$ (input/output)/TP ${ }_{11}$ (output)/TIOCB ${ }_{4}$ (input/output)
$\mathrm{PB}_{2}$ (input/output)/TP ${ }_{10}$ (output)/TIOCA 4 (input/output)
$\mathrm{PB}_{1}$ (input/output)/TP 9 (output)/TIOCB 3 (input/output)
$\mathrm{PB}_{0}$ (input/output)/TP 8 (output)/TIOCA 3 (input/output)

Figure 9.11 Port B Pin Configuration

### 9.12.2 Register Configuration

Table 9.20 summarizes the registers of port $B$.
Table 9.20 Port B Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFD4 | Port B data direction register | PBDDR | W | H'00 |
| H'FFD6 | Port B data register | PBDR | R/W | H'00 |

Note: * Lower 16 bits of the address.

## Port B Data Direction Register (PBDDR)

PBDDR is an 8-bit write-only register that can select input or output for each pin in port B . When pins are used for TPC output, the corresponding PBDDR bits must also be set.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{PB}_{7}$ DDR | $\mathrm{PB}_{6}$ DDR | $\mathrm{PB}_{5}$ DDR | $\mathrm{PB}_{4}$ DDR | $\mathrm{PB}_{3}$ DDR | $\mathrm{PB}_{2}$ DDR | $\mathrm{PB}_{1}$ DDR | $\mathrm{PB}_{0}$ DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | W | W | W | W | W | W |
|  |  |  | Port B data direction 7 to 0 <br> These bits select input or output for port B pins |  |  |  |  |  |

A pin in port B becomes an output pin if the corresponding PBDDR bit is set to 1 , and an input pin if this bit is cleared to 0 .

PBDDR is a write-only register. Its value cannot be read. All bits return 1 when read.
PBDDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a PBDDR bit is set to 1 , the corresponding pin maintains its output state in software standby mode.

## Port B Data Register (PBDR)

PBDR is an 8-bit readable/writable register that stores output data for pins PB7 to PB0. When a bit in PBDDR is set to 1 , if port B is read the value of the corresponding PBDR bit is returned. When a bit in PBDDR is cleared to 0 , if port B is read the corresponding pin level is read.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{PB}_{7}$ | PB6 | $\mathrm{PB}_{5}$ | $\mathrm{PB}_{4}$ | $\mathrm{PB}_{3}$ | $\mathrm{PB}_{2}$ | $\mathrm{PB}_{1}$ | $\mathrm{PB}_{0}$ |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

These bits store data for port B pins

PBDR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 9.12.3 Pin Functions

Table 9.21 describes the selection of pin functions.

## Table 9.21 Port B Pin Functions

Pin Pin Functions and Selection Method
$\mathrm{PB}_{7} / \mathrm{TP}_{15} / \mathrm{DMAC}$ channel 1 settings (bits DTS2/1/0A and DTS2/1/0B in DTCR1A and DTCR1B), $\overline{\text { DREQ }}_{1} /$ bit TRGE in ADCR, bit NDER15 in NDERB, and bit PB ${ }_{7}$ DDR in PBDDR select the pin $\overline{\text { ADTRG }}$ function as follows

| PB7DDR | 0 | 1 | 1 |
| :---: | :---: | :---: | :---: |
| NDER15 | - | 0 | 1 |
| Pin function | $\mathrm{PB}_{7}$ input | $\mathrm{PB}_{7}$ output | TP ${ }_{15}$ output |
|  | $\overline{\text { DREQ }}_{1}$ input*1 |  |  |
|  | ADTRG input*2 |  |  |

Notes: 1. $\overline{\mathrm{DREQ}}_{1}$ input under DMAC channel 1 settings (1) in the table below.
2. $\overline{\text { ADTRG }}$ input when $\mathrm{TRGE}=1$.

| DMAC <br> channel <br> 1 settings | (2) |  | (1) | (2) | (1) | (2) | (1) |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTS2A, DTS1A | Not both 1 |  |  | Both 1 |  |  |  |  |
| DTS0A | - |  |  | 0 | 0 | 1 | 1 |  |
| DTS2B | 0 | 1 | 1 | 0 | 1 | 0 | 1 |  |
| DTS1B | - | 0 | 1 | - | - | - | 0 |  | Bit CS7E in CSCR, DMAC channel 0 settings (bits DTS2/1/0A and DTS2/1/0B in $\overline{\text { DREQ }}_{0} /$ DTCROA and DTCROB), bit NDER14 in NDERB, and bit PB ${ }_{6}$ DDR in PBDDR select the $\overline{\mathrm{CS}}_{7}$ pin function as follows


| $\mathrm{PB}_{6} \mathrm{DDR}$ | 0 | 1 | 1 | - |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{CS7E}$ | 0 | 0 | 0 | 1 |
| NDER14 | - | 0 | 1 | - |
| Pin function | $\mathrm{PB}_{6}$ input | $\mathrm{PB}_{6}$ output | $\mathrm{TP}_{14}$ output | - |
|  | $\overline{\mathrm{DREQ}}_{6}$ input* |  |  |  |

Note: * $\overline{\mathrm{DREQ}}_{0}$ input under DMAC channel 0 settings (1) in the table below.

| DMAC <br> channel 0 <br> settings | (2) |  | (1) | (2) | (1) | (2) | (1) |  |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTS2A, DTS1A | Not both 1 |  |  |  |  |  |  |  |  | Both 1 |  |  |  |  |
| DTS0A | - |  |  | 0 | 0 | 1 | 1 | 1 |  |  |  |  |  |  |
| DTS2B | 0 | 1 | 1 | 0 | 1 | 0 | 1 | 1 |  |  |  |  |  |  |
| DTS1B | - | 0 | 1 | - | - | - | 0 | 1 |  |  |  |  |  |  |

$\mathrm{PB}_{5} / \mathrm{TP}_{13} /$ ITU channel 4 settings (bit CMD1 in TFCR and bit EXB4 in TOER), bit NDER13 in TOCXB $_{4} \quad$ NDERB, and bit $\mathrm{PB}_{5}$ DDR in PBDDR select the pin function as follows

| EXB4, CMD1 | Not both 1 |  |  | Both 1 |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{PB}_{5} \mathrm{DDR}$ | 0 | 1 | 1 | - |
| NDER13 | - | 0 | 1 | - |
| Pin function | $\mathrm{PB}_{5}$ input | $\mathrm{PB}_{5}$ output | $\mathrm{TP}_{13}$ output | $\mathrm{TOCXB}_{4}$ output |

$\mathrm{PB}_{4} / \mathrm{TP}_{12} /$ TOCXA

ITU channel 4 settings (bit CMD1 in TFCR and bit EXA4 in TOER), bit NDER12 in
NDERB, and bit $\mathrm{PB}_{4}$ DDR in PBDDR select the pin function as follows

| EXA4, CMD1 | Not both 1 |  |  | Both 1 |
| :--- | :---: | :---: | :---: | :---: |
| PB ${ }_{4}$ DDR | 0 | 1 | 1 | - |
| NDER12 | - | 0 | 1 | - |
| Pin function | $\mathrm{PB}_{4}$ input | $\mathrm{PB}_{4}$ output | $\mathrm{TP}_{12}$ output | TOCXA $_{4}$ output |

$\mathrm{PB}_{3} / \mathrm{TP}_{11} /$ $\mathrm{TIOCB}_{4}$

ITU channel 4 settings (bit PWM4 in TMDR, bit CMD1 in TFCR, bit EB4 in TOER, and bits IOB2 to IOB0 in TIOR4), bit NDER11 in NDERB, and bit PB ${ }_{3}$ DDR in PBDDR select the pin function as follows

| ITU channel 4 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{PB}_{3} \mathrm{DDR}$ | - | 0 | 1 | 1 |
| NDER11 $^{2}$ | - | - | 0 | 1 |
| Pin function | $\mathrm{TIOCB}_{4}$ output | $\mathrm{PB}_{3}$ input | $\mathrm{PB}_{3}$ output | $\mathrm{TP}_{11}$ output |
|  |  | $\mathrm{TIOCB}_{4}$ input* |  |  |

Note: * TIOCB $_{4}$ input when CMD1 $=$ PWM4 $=0$ and IOB2 $=1$.

| ITU channel 4 <br> settings | (2) | (2) | (1) |  |  |  |  |  | (2) | (1) |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EB4 | 0 | 1 |  |  |  |  |  |  |  |  |
| CMD1 | - | 0 |  |  |  |  |  | 1 |  |  |
| IOB2 | - | 0 | 0 | 0 | 1 | - |  |  |  |  |
| IOB1 | - | 0 | 0 | 1 | - | - |  |  |  |  |
| IOB0 | - | 0 | 1 | - | - | - |  |  |  |  |

ITU channel 4 settings (bit CMD1 in TFCR, bit EA4 in TOER, bit PWM4 in TMDR, and TIOCA bits IOA2 to IOA0 in TIOR4), bit NDER10 in NDERB, and bit PB2DDR in PBDDR select the pin function as follows

| ITU channel 4 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{PB}_{2} \mathrm{DDR}$ | - | 0 | 1 | 1 |
| NDER10 | - | - | 0 | 1 |
| Pin function | TIOCA $_{4}$ output | $\mathrm{PB}_{2}$ input | $\mathrm{PB}_{2}$ output | $\mathrm{TP}_{10}$ output |
|  |  | $\mathrm{TIOCA}_{4}$ input* |  |  |

Note: * TIOCA $_{4}$ input when CMD1 $=$ PWM4 $=0$ and IOA2 $=1$.

| ITU channel 4 <br> settings | (2) | (2) | (1) |  | (2) | (1) |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EA4 | 0 | 1 |  |  |  |  |  |  |  |  |
| CMD1 | - | 0 |  |  |  |  |  |  | 1 |  |
| PWM4 | - | 0 |  |  |  |  |  |  |  | - |
| IOA2 | - | 0 | 0 | 0 | 1 | - | - |  |  |  |
| IOA1 | - | 0 | 0 | 1 | - | - | - |  |  |  |
| IOA0 | - | 0 | 1 | - | - | - | - |  |  |  |

$\mathrm{PB}_{\mathrm{I}} /$ TP $\mathrm{g}_{\mathrm{g}} /$ $\mathrm{TIOCB}_{3}$

ITU channel 3 settings (bit PWM3 in TMDR, bit CMD1 in TFCR, bit EB3 in TOER, and bits IOB2 to IOB0 in TIOR3), bit NDER9 in NDERB, and bit PB,DDR in PBDDR select the pin function as follows

| ITU channel 3 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |
| :--- | :---: | :---: | :---: | :---: |
| PB, DDR | - | 0 | 1 | 1 |
| NDER9 | - | - | 0 | 1 |
| Pin function | $\mathrm{TIOCB}_{3}$ output | $\mathrm{PB}_{1}$ input | $\mathrm{PB}_{1}$ output | $\mathrm{TP}_{9}$ output |
|  |  | $\mathrm{TIOCB}_{3}$ input* |  |  |

Note: * TIOCB $_{3}$ input when CMD1 $=$ PWM3 $=0$ and IOB2 $=1$.

| ITU channel 3 <br> settings | (2) | (2) | (1) |  | (2) | (1) |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EB3 | 0 | 1 |  |  |  |  |  |  |
| CMD1 | - | 0 |  |  |  |  |  | 1 |
| IOB2 | - | 0 | 0 | 0 | 1 | - |  |  |
| IOB1 | - | 0 | 0 | 1 | - | - |  |  |
| IOB0 | - | 0 | 1 | - | - | - |  |  |

ITU channel 3 settings (bit CMD1 in TFCR, bit EA3 in TOER, bit PWM3 in TMDR, and $\mathrm{TIOCA}_{3}$ bits IOA2 to IOA0 in TIOR3), bit NDER8 in NDERB, and bit PB ${ }_{0}$ DDR in PBDDR select the pin function as follows

| ITU channel 3 <br> settings | (1) in table <br> below | (2) in table <br> below |  |  |
| :--- | :---: | :---: | :---: | :---: |
| PBoDDR | - | 0 | 1 | 1 |
| NDER8 | - | - | 0 | 1 |
| Pin function | $\mathrm{TIOCA}_{3}$ output | $\mathrm{PB}_{0}$ input | $\mathrm{PB}_{0}$ output | $\mathrm{TP}_{8}$ output |
|  |  | $\mathrm{TIOCA}_{3}$ input* |  |  |

Note: $*$ TIOCA $_{3}$ input when CMD1 $=$ PWM $3=0$ and IOA2 $=1$.

| ITU channel 3 <br> settings | (2) | (2) | (1) |  | (2) | (1) |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EA3 | 0 | 1 |  |  |  |  |  |  |  |  |
| CMD1 | - | 0 |  |  |  |  |  |  | 1 |  |
| PWM3 | - | 0 |  |  |  |  |  |  |  | - |
| IOA2 | - | 0 | 0 | 0 | 1 | - | - |  |  |  |
| IOA1 | - | 0 | 0 | 1 | - | - | - |  |  |  |
| IOA0 | - | 0 | 1 | - | - | - | - |  |  |  |

## Section 10 16-Bit Integrated Timer Unit (ITU)

### 10.1 Overview

The H8/3052F has a built-in 16-bit integrated timer unit (ITU) with five 16-bit timer channels.

When the ITU is not used, it can be independently halted to conserve power. For details see section 20.6, Module Standby Function.

### 10.1.1 Features

ITU features are listed below.

- Capability to process up to 12 pulse outputs or 10 pulse inputs
- Ten general registers (GRs, two per channel) with independently-assignable output compare or input capture functions
- Selection of eight counter clock sources for each channel:

Internal clocks: $\phi, \phi / 2, \phi / 4, \phi / 8$
External clocks: TCLKA, TCLKB, TCLKC, TCLKD

- Five operating modes selectable in all channels:
- Waveform output by compare match

Selection of 0 output, 1 output, or toggle output (only 0 or 1 output in channel 2)

- Input capture function

Rising edge, falling edge, or both edges (selectable)

- Counter clearing function

Counters can be cleared by compare match or input capture

- Synchronization

Two or more timer counters (TCNTs) can be preset simultaneously, or cleared simultaneously by compare match or input capture. Counter synchronization enables synchronous register input and output.

- PWM mode

PWM output can be provided with an arbitrary duty cycle. With synchronization, up to five-phase PWM output is possible

- Phase counting mode selectable in channel 2

Two-phase encoder output can be counted automatically.

- Three additional modes selectable in channels 3 and 4
- Reset-synchronized PWM mode

If channels 3 and 4 are combined, three-phase PWM output is possible with three pairs of complementary waveforms.

- Complementary PWM mode

If channels 3 and 4 are combined, three-phase PWM output is possible with three pairs of non-overlapping complementary waveforms.

- Buffering

Input capture registers can be double-buffered. Output compare registers can be updated automatically.

- High-speed access via internal 16-bit bus

The 16-bit timer counters, general registers, and buffer registers can be accessed at high speed via a 16-bit bus.

- Fifteen interrupt sources

Each channel has two compare match/input capture interrupts and an overflow interrupt. All interrupts can be requested independently.

- Activation of DMA controller (DMAC)

Four of the compare match/input capture interrupts from channels 0 to 3 can start the DMAC.

- Output triggering of programmable timing pattern controller (TPC)

Compare match/input capture signals from channels 0 to 3 can be used as TPC output triggers.

Table 10.1 summarizes the ITU functions.

## Table 10.1 ITU Functions

| Item | Channel 0 | Channel 1 | Channel 2 | Channel 3 | Channel 4 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Clock sources | Internal clocks: $\phi, \phi / 2, \phi / 4, \phi / 8$ |  |  |  |  |
|  | External clocks: TCLKA, TCLKB, TCLKC, TCLKD, selectable independently |  |  |  |  |
| General registers (output compare/input capture registers) | GRA0, GRB0 | GRA1, GRB1 | GRA2, GRB2 | GRA3, GRB3 | GRA4, GRB4 |
| Buffer registers | - | - | - | BRA3, BRB3 | BRA4, BRB4 |
| Input/output pins | $\begin{aligned} & \text { TIOCA }_{0}, \\ & \text { TIOCB }_{0} \end{aligned}$ | TIOCA $_{1}$, TIOCB $_{1}$ | $\begin{aligned} & \mathrm{TIOCA}_{2} \\ & \mathrm{TIOCB}_{2} \end{aligned}$ | $\begin{aligned} & \text { TIOCA }_{3}, \\ & \text { TIOCB }_{3} \end{aligned}$ | $\begin{aligned} & \mathrm{TIOCA}_{4} \\ & \mathrm{TIOCB}_{4} \end{aligned}$ |
| Output pins | - | - | - | - | $\begin{aligned} & \text { TOCXA }_{4}, \\ & \text { TOCXB }_{4} \end{aligned}$ |
| Counter clearing function | GRAO/GRB0 <br> compare match or input capture | GRA1/GRB1 compare match or input capture | GRA2/GRB2 <br> compare match or input capture | GRA3/GRB3 <br> compare match or input capture | GRA4/GRB4 <br> compare match or input capture |
| Compare match output | O | O | O | O | O |
|  | 0 | 0 | 0 | 0 | 0 |
|  | O | O | - | 0 | 0 |
| Input capture function | O | O | 0 | 0 | 0 |
| Synchronization | 0 | O | 0 | 0 | 0 |
| PWM mode | 0 | 0 | 0 | 0 | 0 |
| Reset-synchronized PWM mode | - | - | - | O | 0 |
| Complementary PWM mode | - | - | - | 0 | 0 |
| Phase counting mode | - | - | 0 | - | - |
| Buffering | - | - | - | 0 | 0 |
| DMAC activation | GRA0 compare match or input capture | GRA1 compare match or input capture | GRA2 compare match or input capture | GRA3 compare match or input capture | - |
| Interrupt sources | Three sources <br> - Compare match/input capture A0 <br> - Compare match/input capture B0 <br> - Overflow | Three sources <br> - Compare match/input capture A1 <br> - Compare match/input capture B1 <br> - Overflow | Three sources <br> - Compare match/input capture A2 <br> - Compare match/input capture B2 <br> - Overflow | Three sources <br> - Compare match/input capture A3 <br> - Compare match/input capture B3 <br> - Overflow | Three sources <br> - Compare match/input capture A4 <br> - Compare match/input capture B4 <br> - Overflow |

## Legend

o: Available
-: Not available

### 10.1.2 Block Diagrams

ITU Block Diagram (Overall): Figure 10.1 is a block diagram of the ITU.


Figure 10.1 ITU Block Diagram (Overall)

Block Diagram of Channels 0 and 1: ITU channels 0 and 1 are functionally identical. Both have the structure shown in figure 10.2.


Figure 10.2 Block Diagram of Channels 0 and 1 (for Channel 0)

Block Diagram of Channel 2: Figure 10.3 is a block diagram of channel 2. This is the channel that provides only 0 output and 1 output.

TCLKA to TCLKD
$\phi, \phi / 2, \phi / 4, \phi / 8$


Legend
TCNT2: $\quad$ Timer counter 2 (16 bits)
GRA2, GRB2: General registers A2 and B2 (input capture/output compare registers) (16 bits $\times 2$ )
TCR2: $\quad$ Timer control register 2 (8 bits)
TIOR2: Timer I/O control register 2 ( 8 bits)
TIER2: $\quad$ Timer interrupt enable register 2 ( 8 bits)
TSR2: $\quad$ Timer status register $2(8$ bits)

Figure 10.3 Block Diagram of Channel 2

Block Diagrams of Channels 3 and 4: Figure 10.4 is a block diagram of channel 3. Figure 10.5 is a block diagram of channel 4.


Legend
TCNT3: $\quad$ Timer counter 3 ( 16 bits)
GRA3, GRB3: General registers A3 and B3 (input capture/output compare registers) (16 bits $\times 2$ )
BRA3, BRB3: Buffer registers A3 and B3 (input capture/output compare buffer registers) (16 bits $\times 2$ )
TCR3: $\quad$ Timer control register 3 (8 bits)
TIOR3: $\quad$ Timer I/O control register 3 ( 8 bits)
TIER3: Timer interrupt enable register 3 (8 bits)
TSR3: Timer status register 3 (8 bits)
Figure 10.4 Block Diagram of Channel 3

TCLKA to TCLKD $\phi, \phi / 2$, $\phi / 4, \phi / 8$


Legend
TCNT4: $\quad$ Timer counter 4 (16 bits)
GRA4, GRB4: General registers A4 and B4 (input capture/output compare registers) (16 bits $\times 2$ )
BRA4, BRB4: Buffer registers A4 and B4 (input capture/output compare buffer registers) (16 bits $\times 2$ )
TCR4: $\quad$ Timer control register 4 (8 bits)
TIOR4: $\quad$ Timer I/O control register 4 ( 8 bits)
TIER4: Timer interrupt enable register 4 (8 bits)
TSR4: $\quad$ Timer status register 4 ( 8 bits)

Figure 10.5 Block Diagram of Channel 4

### 10.1.3 Pin Configuration

Table 10.2 summarizes the ITU pins.

## Table 10.2 ITU Pins

| Channel | Name | Abbreviation | Input/ Output | Function |
| :---: | :---: | :---: | :---: | :---: |
| Common | Clock input A | TCLKA | Input | External clock A input pin (phase-A input pin in phase counting mode) |
|  | Clock input B | TCLKB | Input | External clock B input pin (phase-B input pin in phase counting mode) |
|  | Clock input C | TCLKC | Input | External clock C input pin |
|  | Clock input D | TCLKD | Input | External clock D input pin |
| 0 | Input capture/output compare A0 | TIOCA ${ }_{0}$ | Input/ output | GRAO output compare or input capture pin PWM output pin in PWM mode |
|  | Input capture/output compare B0 | TIOCB ${ }_{0}$ | Input/ output | GRB0 output compare or input capture pin |
| 1 | Input capture/output compare A1 | TIOCA ${ }_{1}$ | Input/ output | GRA1 output compare or input capture pin PWM output pin in PWM mode |
|  | Input capture/output compare B1 | $\mathrm{TIOCB}_{1}$ | Input/ output | GRB1 output compare or input capture pin |
| 2 | Input capture/output compare A2 | $\mathrm{TIOCA}_{2}$ | Input/ output | GRA2 output compare or input capture pin PWM output pin in PWM mode |
|  | Input capture/output compare B2 | $\mathrm{TIOCB}_{2}$ | Input/ output | GRB2 output compare or input capture pin |
| 3 | Input capture/output compare A3 | $\mathrm{TIOCA}_{3}$ | Input/ output | GRA3 output compare or input capture pin PWM output pin in PWM mode, complementary PWM mode, or resetsynchronized PWM mode |
|  | Input capture/output compare B3 | $\mathrm{TIOCB}_{3}$ | Input/ output | GRB3 output compare or input capture pin PWM output pin in complementary PWM mode or reset-synchronized PWM mode |
| 4 | Input capture/output compare A4 | $\mathrm{TIOCA}_{4}$ | Input/ output | GRA4 output compare or input capture pin PWM output pin in PWM mode, complementary PWM mode, or resetsynchronized PWM mode |
|  | Input capture/output compare B4 | $\mathrm{TIOCB}_{4}$ | Input/ output | GRB4 output compare or input capture pin PWM output pin in complementary PWM mode or reset-synchronized PWM mode |
|  | Output compare XA4 | $\mathrm{TOCXA}_{4}$ | Output | PWM output pin in complementary PWM mode or reset-synchronized PWM mode |
|  | Output compare XB4 | $\mathrm{TOCXB}_{4}$ | Output | PWM output pin in complementary PWM mode or reset-synchronized PWM mode |

### 10.1.4 Register Configuration

Table 10.3 summarizes the ITU registers.

## Table 10.3 ITU Registers

| Channel | Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Common | H'FF60 | Timer start register | TSTR | R/W | H'E0 |
|  | H'FF61 | Timer synchro register | TSNC | R/W | H'E0 |
|  | H'FF62 | Timer mode register | TMDR | R/W | H'80 |
|  | H'FF63 | Timer function control register | TFCR | R/W | H'C0 |
|  | H'FF90 | Timer output master enable register | TOER | R/W | H'FF |
|  | H'FF91 | Timer output control register | TOCR | R/W | H'FF |
| 0 | H'FF64 | Timer control register 0 | TCR0 | R/W | H'80 |
|  | H'FF65 | Timer l/O control register 0 | TIOR0 | R/W | H'88 |
|  | H'FF66 | Timer interrupt enable register 0 | TIER0 | R/W | H'F8 |
|  | H'FF67 | Timer status register 0 | TSR0 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'F8 |
|  | H'FF68 | Timer counter 0 (high) | TCNTOH | R/W | H'00 |
|  | H'FF69 | Timer counter 0 (low) | TCNTOL | R/W | H'00 |
|  | H'FF6A | General register A0 (high) | GRAOH | R/W | H'FF |
|  | H'FF6B | General register A0 (low) | GRAOL | R/W | H'FF |
|  | H'FF6C | General register B0 (high) | GRBOH | R/W | H'FF |
|  | H'FF6D | General register B0 (low) | GRB0L | R/W | H'FF |
| 1 | H'FF6E | Timer control register 1 | TCR1 | R/W | H'80 |
|  | H'FF6F | Timer l/O control register 1 | TIOR1 | R/W | H'88 |
|  | H'FF70 | Timer interrupt enable register 1 | TIER1 | R/W | H'F8 |
|  | H'FF71 | Timer status register 1 | TSR1 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'F8 |
|  | H'FF72 | Timer counter 1 (high) | TCNT1H | R/W | H'00 |
|  | H'FF73 | Timer counter 1 (low) | TCNT1L | R/W | H'00 |
|  | H'FF74 | General register A1 (high) | GRA1H | R/W | H'FF |
|  | H'FF75 | General register A1 (low) | GRA1L | R/W | H'FF |
|  | H'FF76 | General register B1 (high) | GRB1H | R/W | H'FF |
|  | H'FF77 | General register B1 (low) | GRB1L | R/W | H'FF |

Rev. 2.0, 03/01, page 314 of 822

| Channel | Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 2 | H'FF78 | Timer control register 2 | TCR2 | R/W | H'80 |
|  | H'FF79 | Timer I/O control register 2 | TIOR2 | R/W | H'88 |
|  | H'FF7A | Timer interrupt enable register 2 | TIER2 | R/W | H'F8 |
|  | H'FF7B | Timer status register 2 | TSR2 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'F8 |
|  | H'FF7C | Timer counter 2 (high) | TCNT2H | R/W | H'00 |
|  | H'FF7D | Timer counter 2 (low) | TCNT2L | R/W | H'00 |
|  | H'FF7E | General register A2 (high) | GRA2H | R/W | H'FF |
|  | H'FF7F | General register A2 (low) | GRA2L | R/W | H'FF |
|  | H'FF80 | General register B2 (high) | GRB2H | R/W | H'FF |
|  | H'FF81 | General register B2 (low) | GRB2L | R/W | H'FF |
| 3 | H'FF82 | Timer control register 3 | TCR3 | R/W | H'80 |
|  | H'FF83 | Timer I/O control register 3 | TIOR3 | R/W | H'88 |
|  | H'FF84 | Timer interrupt enable register 3 | TIER3 | R/W | H'F8 |
|  | H'FF85 | Timer status register 3 | TSR3 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'F8 |
|  | H'FF86 | Timer counter 3 (high) | TCNT3H | R/W | H'00 |
|  | H'FF87 | Timer counter 3 (low) | TCNT3L | R/W | H'00 |
|  | H'FF88 | General register A3 (high) | GRA3H | R/W | H'FF |
|  | H'FF89 | General register A3 (low) | GRA3L | R/W | H'FF |
|  | H'FF8A | General register B3 (high) | GRB3H | R/W | H'FF |
|  | H'FF8B | General register B3 (low) | GRB3L | R/W | H'FF |
|  | H'FF8C | Buffer register A3 (high) | BRA3H | R/W | H'FF |
|  | H'FF8D | Buffer register A3 (low) | BRA3L | R/W | H'FF |
|  | H'FF8E | Buffer register B3 (high) | BRB3H | R/W | H'FF |
|  | H'FF8F | Buffer register B3 (low) | BRB3L | R/W | H'FF |


| Channel | Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 4 | H'FF92 | Timer control register 4 | TCR4 | R/W | H'80 |
|  | H'FF93 | Timer I/O control register 4 | TIOR4 | R/W | H'88 |
|  | H'FF94 | Timer interrupt enable register 4 | TIER4 | R/W | H'F8 |
|  | H'FF95 | Timer status register 4 | TSR4 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'F8 |
|  | H'FF96 | Timer counter 4 (high) | TCNT4H | R/W | H'00 |
|  | H'FF97 | Timer counter 4 (low) | TCNT4L | R/W | H'00 |
|  | H'FF98 | General register A4 (high) | GRA4H | R/W | H'FF |
|  | H'FF99 | General register A4 (low) | GRA4L | R/W | H'FF |
|  | H'FF9A | General register B4 (high) | GRB4H | R/W | H'FF |
|  | H'FF9B | General register B4 (low) | GRB4L | R/W | H'FF |
|  | H'FF9C | Buffer register A4 (high) | BRA4H | R/W | H'FF |
|  | H'FF9D | Buffer register A4 (low) | BRA4L | R/W | H'FF |
|  | H'FF9E | Buffer register B4 (high) | BRB4H | R/W | H'FF |
|  | H'FF9F | Buffer register B4 (low) | BRB4L | R/W | H'FF |

Notes: 1. The lower 16 bits of the address are indicated.
2. Only 0 can be written, to clear flags.

### 10.2 Register Descriptions

### 10.2.1 Timer Start Register (TSTR)

TSTR is an 8-bit readable/writable register that starts and stops the timer counter (TCNT) in channels 0 to 4 .

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | STR4 | STR3 | STR2 | STR1 | STR0 |
| Initial value | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | R/W | R/W | R/W | R/W | R/W |
|  | Reserved bits |  |  |  | Counter start 4 to 0 These bits start and stop TCNT4 to TCNT0 |  |  |  |

TSTR is initialized to H'E0 by a reset and in standby mode.
Bits 7 to 5—Reserved: Read-only bits, always read as 1.
Bit 4—Counter Start 4 (STR4): Starts and stops timer counter 4 (TCNT4).

| Bit 4: STR4 | Description |  |
| :--- | :--- | :--- |
| 0 | TCNT4 is halted | (Initial value) |
| 1 | TCNT4 is counting |  |

Bit 3—Counter Start 3 (STR3): Starts and stops timer counter 3 (TCNT3).

| Bit 3: STR3 | Description |  |
| :--- | :--- | ---: |
| 0 | TCNT3 is halted | (Initial value) |
| 1 | TCNT3 is counting |  |

Bit 2—Counter Start 2 (STR2): Starts and stops timer counter 2 (TCNT2).

| Bit 2: STR2 | Description |  |
| :--- | :--- | ---: |
| 0 | TCNT2 is halted | (Initial value) |
| 1 | TCNT2 is counting |  |

Bit 1—Counter Start 1 (STR1): Starts and stops timer counter 1 (TCNT1).

## Bit 1: STR1 Description

| 0 | TCNT1 is halted | (Initial value) |
| :--- | :--- | :--- |
| 1 | TCNT1 is counting |  |

Bit 0—Counter Start 0 (STR0): Starts and stops timer counter 0 (TCNT0).

| Bit 0: STR0 | Description |  |
| :--- | :--- | :--- |
| 0 | TCNT0 is halted | (Initial value) |
| 1 | TCNT0 is counting |  |

### 10.2.2 Timer Synchro Register (TSNC)

TSNC is an 8-bit readable/writable register that selects whether channels 0 to 4 operate independently or synchronously. Channels are synchronized by setting the corresponding bits to 1 .


TSNC is initialized to H'E0 by a reset and in standby mode.
Bits 7 to 5—Reserved: Read-only bits, always read as 1.
Bit 4—Timer Sync 4 (SYNC4): Selects whether channel 4 operates independently or synchronously.

## Bit 4: SYNC4 Description

| 0 | Channel 4's timer counter (TCNT4) operates independently | (Initial value) |
| :--- | :--- | :--- |
|  | TCNT4 is preset and cleared independently of other channels |  |
| 1 | Channel 4 operates synchronously |  |
|  | TCNT4 can be synchronously preset and cleared |  |

Bit 3-Timer Sync 3 (SYNC3): Selects whether channel 3 operates independently or synchronously.

## Bit 3: SYNC3 Description

| 0 | Channel 3's timer counter (TCNT3) operates independently | (Initial value) |
| :--- | :--- | :--- |
|  | TCNT3 is preset and cleared independently of other channels |  |
| 1 | Channel 3 operates synchronously |  |
|  | TCNT3 can be synchronously preset and cleared |  |

Bit 2-Timer Sync 2 (SYNC2): Selects whether channel 2 operates independently or synchronously.
Bit 2: SYNC2 Description

| 0 | Channel 2's timer counter (TCNT2) operates independently | (Initial value) |
| :--- | :--- | :--- |
|  | TCNT2 is preset and cleared independently of other channels |  |
| 1 | Channel 2 operates synchronously |  |
|  | TCNT2 can be synchronously preset and cleared |  |

Bit 1—Timer Sync 1 (SYNC1): Selects whether channel 1 operates independently or synchronously.

| Bit 1: SYNC1 | Description |  |
| :--- | :--- | :--- |
| 0 | Channel 1's timer counter (TCNT1) operates independently | (Initial value) |
|  | TCNT1 is preset and cleared independently of other channels |  |
| 1 | Channel 1 operates synchronously |  |
|  | TCNT1 can be synchronously preset and cleared |  |

Bit 0-Timer Sync 0 (SYNC0): Selects whether channel 0 operates independently or synchronously.

| Bit 0: SYNCO | Description |  |
| :--- | :--- | :--- |
| 0 | Channel 0's timer counter (TCNT0) operates independently | (Initial value) |
|  | TCNT0 is preset and cleared independently of other channels |  |
| 1 | Channel 0 operates synchronously |  |
|  | TCNT0 can be synchronously preset and cleared |  |

### 10.2.3 Timer Mode Register (TMDR)

TMDR is an 8-bit readable/writable register that selects PWM mode for channels 0 to 4 . It also selects phase counting mode and the overflow flag (OVF) setting conditions for channel 2.

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | MDF | FDIR | PWM4 | PWM3 | PWM2 | PWM1 | PWM0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
|  |  | R/W |  | R/W | R/W | R/W | R/W |

Flag direction
Selects the setting condition for the overflow flag (OVF) in timer status register 2 (TSR2)
Phase counting mode flag
Selects phase counting mode for channel 2
Reserved bit

TMDR is initialized to H'80 by a reset and in standby mode.
Bit 7—Reserved: Read-only bit, always read as 1.
Bit 6—Phase Counting Mode Flag (MDF): Selects whether channel 2 operates normally or in phase counting mode.
Bit 6: MDF Description

| 0 | Channel 2 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel 2 operates in phase counting mode |  |

When MDF is set to 1 to select phase counting mode, TCNT2 operates as an up/down-counter and pins TCLKA and TCLKB become counter clock input pins. TCNT2 counts both rising and falling edges of TCLKA and TCLKB, and counts up or down as follows.

| Counting Direction | Down-Counting |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| TCLKA pin | $\uparrow$ | High | $\downarrow$ | Low | $\uparrow$ | Low | $\downarrow$ | High |
| TCLKB pin | Low | $\uparrow$ | High | $\downarrow$ | High | $\uparrow$ | Low | $\downarrow$ |

In phase counting mode channel 2 operates as above regardless of the external clock edges selected by bits CKEG1 and CKEG0 and the clock source selected by bits TPSC2 to TPSC0 in TCR2. Phase counting mode takes precedence over these settings.

The counter clearing condition selected by the CCLR1 and CCLR0 bits in TCR2 and the compare match/input capture settings and interrupt functions of TIOR2, TIER2, and TSR2 remain effective in phase counting mode.

Bit 5—Flag Direction (FDIR): Designates the setting condition for the OVF flag in TSR2. The FDIR designation is valid in all modes in channel 2.
Bit 5: FDIR Description

| 0 | OVF is set to 1 in TSR2 when TCNT2 overflows or underflows | (Initial value) |
| :--- | :--- | :--- |
| 1 | OVF is set to 1 in TSR2 when TCNT2 overflows |  |

Bit 4—PWM Mode 4 (PWM4): Selects whether channel 4 operates normally or in PWM mode.

| Bit 4: PWM4 | Description |  |
| :--- | :--- | :--- |
| 0 | Channel 4 operates normally | (Initial value) |
| 1 | Channel 4 operates in PWM mode |  |

When bit PWM4 is set to 1 to select PWM mode, pin TIOCA $_{4}$ becomes a PWM output pin. The output goes to 1 at compare match with GRA4, and to 0 at compare match with GRB4.

If complementary PWM mode or reset-synchronized PWM mode is selected by bits CMD1 and CMD0 in TFCR, the CMD1 and CMD0 setting takes precedence and the PWM4 setting is ignored.

Bit 3—PWM Mode 3 (PWM3): Selects whether channel 3 operates normally or in PWM mode.
Bit 3: PWM3 Description

| 0 | Channel 3 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel 3 operates in PWM mode |  |

When bit PWM3 is set to 1 to select PWM mode, pin TIOCA ${ }_{3}$ becomes a PWM output pin. The output goes to 1 at compare match with GRA3, and to 0 at compare match with GRB3.

If complementary PWM mode or reset-synchronized PWM mode is selected by bits CMD1 and CMD0 in TFCR, the CMD1 and CMD0 setting takes precedence and the PWM3 setting is ignored.

Bit 2—PWM Mode 2 (PWM2): Selects whether channel 2 operates normally or in PWM mode.
Bit 2: PWM2 Description

| 0 | Channel 2 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel 2 operates in PWM mode |  |

When bit PWM2 is set to 1 to select PWM mode, pin TIOCA ${ }_{2}$ becomes a PWM output pin. The output goes to 1 at compare match with GRA2, and to 0 at compare match with GRB2.

Bit 1—PWM Mode 1 (PWM1): Selects whether channel 1 operates normally or in PWM mode.
Bit 1: PWM1 Description

| 0 | Channel 1 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel 1 operates in PWM mode |  |

When bit PWM1 is set to 1 to select PWM mode, pin TIOCA ${ }_{1}$ becomes a PWM output pin. The output goes to 1 at compare match with GRA1, and to 0 at compare match with GRB1.

Bit 0—PWM Mode 0 (PWM0): Selects whether channel 0 operates normally or in PWM mode.
Bit 0: PWMO Description

| 0 | Channel 0 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel 0 operates in PWM mode |  |

When bit PWM0 is set to 1 to select PWM mode, pin TIOCA ${ }_{0}$ becomes a PWM output pin. The output goes to 1 at compare match with GRA0, and to 0 at compare match with GRB0.

### 10.2.4 Timer Function Control Register (TFCR)

TFCR is an 8-bit readable/writable register that selects complementary PWM mode, resetsynchronized PWM mode, and buffering for channels 3 and 4.


TFCR is initialized to $\mathrm{H}^{\prime} \mathrm{C} 0$ by a reset and in standby mode.

Bits 7 and 6-Reserved: Read-only bits, always read as 1.

Bits 5 and 4-Combination Mode 1 and 0 (CMD1, CMD0): These bits select whether channels 3 and 4 operate in normal mode, complementary PWM mode, or reset-synchronized PWM mode.

| Bit 5: CMD1 | Bit 4: CMD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | Channels 3 and 4 operate normally |
| 1 | 0 | Channels 3 and 4 operate together in complementary <br>  <br>  <br> 1 |
|  | PWM mode |  |
|  | Channels 3 and 4 operate together in reset-synchronized <br> PWM mode |  |

Before selecting reset-synchronized PWM mode or complementary PWM mode, halt the timer counter or counters that will be used in these modes.

When these bits select complementary PWM mode or reset-synchronized PWM mode, they take precedence over the setting of the PWM mode bits (PWM4 and PWM3) in TMDR. Settings of timer sync bits SYNC4 and SYNC3 in TSNC are valid in complementary PWM mode and resetsynchronized PWM mode, however. When complementary PWM mode is selected, channels 3 and 4 must not be synchronized (do not set bits SYNC3 and SYNC4 both to 1 in TSNC).

Bit 3—Buffer Mode B4 (BFB4): Selects whether GRB4 operates normally in channel 4, or whether GRB4 is buffered by BRB4.

| Bit 3: BFB4 | Description |  |
| :--- | :--- | :--- |
| 0 | GRB4 operates normally | (Initial value) |
| 1 | GRB4 is buffered by BRB4 |  |

Bit 2—Buffer Mode A4 (BFA4): Selects whether GRA4 operates normally in channel 4, or whether GRA4 is buffered by BRA4.
Bit 2: BFA4 Description

| 0 | GRA4 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | GRA4 is buffered by BRA4 |  |

Bit 1—Buffer Mode B3 (BFB3): Selects whether GRB3 operates normally in channel 3, or whether GRB3 is buffered by BRB3.
Bit 1: BFB3 Description

| 0 | GRB3 operates normally | (Initial value) |
| :--- | :--- | :---: |
| 1 | GRB3 is buffered by BRB3 |  |

Bit 0—Buffer Mode A3 (BFA3): Selects whether GRA3 operates normally in channel 3, or whether GRA3 is buffered by BRA3.

| Bit 0: BFA3 | Description |  |
| :--- | :--- | :--- |
| 0 | GRA3 operates normally | (Initial value) |
| 1 | GRA3 is buffered by BRA3 |  |

### 10.2.5 Timer Output Master Enable Register (TOER)

TOER is an 8-bit readable/writable register that enables or disables output settings for channels 3 and 4.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | EXB4 | EXA4 | EB3 | EB4 | EA4 | EA3 |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | - | - | R/W | R/W | R/W | R/W | R/W | R/W |
| Reserved bits |  |  |  |  |  |  |  |  |
| Master enable TOCXA 4 , TOCXB $_{4}$ These bits enable or disable output settings for pins TOCXA $_{4}$ and TOCXB $_{4}$ |  |  |  |  |  |  |  |  |
| Master enable TIOCA $_{3}$, , TIOCB $_{3}$, ,IOCA $_{4}$, TIOCB $_{4}$ These bits enable or disable output settings for pins $\mathrm{TIOCA}_{3}, \mathrm{TIOCB}_{3}, \mathrm{TIOCA}_{4}$, and $\mathrm{TIOCB}_{4}$ |  |  |  |  |  |  |  |  |

TOER is initialized to H'FF by a reset and in standby mode.
Bits 7 and 6-Reserved: Read-only bits, always read as 1.
Bit 5—Master Enable TOCXB4 (EXB4): Enables or disables ITU output at pin TOCXB ${ }_{4}$.

| Bit 5: EXB4 | Description |
| :--- | :--- |
| 0 | TOCXB $_{4}$ output is disabled regardless of TFCR settings (TOCXB <br> a generic input/output pin). <br> If XTGD $^{2}=0$, EXB4 is cleared to 0 when input capture A occurs in channel 1. |
| 1 | $\mathrm{TOCXB}_{4}$ is enabled for output according to TFCR settings |

Bit 4—Master Enable TOCXA4 (EXA4): Enables or disables ITU output at pin TOCXA ${ }_{4}$.

| Bit 4: EXA4 | Description |
| :---: | :---: |
| 0 | TOCXA ${ }_{4}$ output is disabled regardless of TFCR settings ( TOCXA $_{4}$ operates as a generic input/output pin). |
|  | If $\mathrm{XTGD}=0$, EXA4 is cleared to 0 when input capture A occurs in channel 1. |
| 1 | TOCXA ${ }_{4}$ is enabled for output according to TFCR settings (Initial value) |

Bit 3—Master Enable TIOCB3 (EB3): Enables or disables ITU output at pin TIOCB ${ }_{3}$.

| Bit 3: EB3 | Description |
| :--- | :--- |
| 0 | $\mathrm{TIOCB}_{3}$ output is disabled regardless of TIOR3 and TFCR settings $\left(\mathrm{TIOCB}_{3}\right.$ <br> operates as a generic input/output pin). <br> If XTGD $=0$, EB3 is cleared to 0 when input capture A occurs in channel 1. <br> 1 |
| $\mathrm{TIOCB}_{3}$ is enabled for output according to TIOR3 and TFCR settings |  |
| (Initial value) |  |

Bit 2-Master Enable TIOCB4 (EB4): Enables or disables ITU output at pin TIOCB ${ }_{4}$.

| Bit 2: EB4 | Description |
| :--- | :--- |
| 0 | $\mathrm{TIOCB}_{4}$ output is disabled regardless of TIOR4 and TFCR settings $\left(\mathrm{TIOCB}_{4}\right.$ <br> operates as a generic input/output pin). <br> If XTGD $=0$, EB4 is cleared to 0 when input capture A occurs in channel 1. |
| 1 | $\mathrm{TIOCB}_{4}$ is enabled for output according to TIOR4 and TFCR settings |
| (Initial value) |  |

Bit 1—Master Enable TIOCA4 (EA4): Enables or disables ITU output at pin TIOCA ${ }_{4}$.

| Bit 1: EA4 | Description |
| :--- | :--- |
| 0 | TIOCA $_{4}$ output is disabled regardless of TIOR4, TMDR, and TFCR settings <br>  <br>  <br>  <br>  <br>  <br> 1 <br>  $\mathrm{TIOCA}_{4}$ operates as a generic input/output pin). |
|  |  |

Bit 0-Master Enable TIOCA3 (EA3): Enables or disables ITU output at pin TIOCA ${ }_{3}$.

| Bit 0: EA3 | Description |
| :--- | :--- |
| 0 | $\mathrm{TIOCA}_{3}$ output is disabled regardless of TIOR3, TMDR, and TFCR settings <br> (TIOCA <br>  <br>  <br>  <br> If XTGD $=0$, EA3 is cleared to 0 when input capture A occurs in channel 1. |
| 1 | $\mathrm{TIOCA}_{3}$ is enabled for output according to TIOR3, TMDR, and TFCR settings |
| (Initial value) |  |

Rev. 2.0, 03/01, page 326 of 822

### 10.2.6 Timer Output Control Register (TOCR)

TOCR is an 8-bit readable/writable register that selects externally triggered disabling of output in complementary PWM mode and reset-synchronized PWM mode, and inverts the output levels.

External trigger disable
Selects externally triggered disabling of output in complementary PWM mode and reset-synchronized PWM mode

The settings of the XTGD, OLS4, and OLS3 bits are valid only in complementary PWM mode and reset-synchronized PWM mode. These settings do not affect other modes.

TOCR is initialized to H'FF by a reset and in standby mode.

Bits 7 to 5—Reserved: Read-only bits, always read as 1.
Bit 4-External Trigger Disable (XTGD): Selects externally triggered disabling of ITU output in complementary PWM mode and reset-synchronized PWM mode.

| Bit 4: XTGD | Description |
| :--- | :--- |
| 0 | Input capture A in channel 1 is used as an external trigger signal in <br> complementary PWM mode and reset-synchronized PWM mode. <br> When an external trigger occurs, bits 5 to 0 in TOER are cleared to 0, disabling <br> ITU output. |
| External triggering is disabled | (Initial value) |

Bits 3 and 2—Reserved: Read-only bits, always read as 1.

Bit 1—Output Level Select 4 (OLS4): Selects output levels in complementary PWM mode and reset-synchronized PWM mode.

| Bit 1: OLS4 | Description |  |
| :--- | :--- | ---: |
| 0 | $\mathrm{TIOCA}_{3}, \mathrm{TIOCA}_{4}$, and $\mathrm{TIOCB}_{4}$ outputs are inverted |  |
| 1 | $\mathrm{TIOCA}_{3}, \mathrm{TIOCA}_{4}$, and $\mathrm{TIOCB}_{4}$ outputs are not inverted | (Initial value) |

Bit 0-Output Level Select 3 (OLS3): Selects output levels in complementary PWM mode and reset-synchronized PWM mode.

Bit 0: OLS3
Description

| 0 | $\mathrm{TIOCB}_{3}, \mathrm{TOCXA}_{4}$, and $\mathrm{TOCXB}_{4}$ outputs are inverted |  |
| :--- | :--- | :--- | :--- |
| 1 | $\mathrm{TIOCB}_{3}, \mathrm{TOCXA}_{4}$, and $\mathrm{TOCXB}_{4}$ outputs are not inverted | (Initial value) |

### 10.2.7 Timer Counters (TCNT)

TCNT is a 16 -bit counter. The ITU has five TCNTs, one for each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | TCNT0 | Up-counter |
| 1 | TCNT1 |  |
| 2 | TCNT2 | Phase counting mode: up/down-counter <br> Other modes: up-counter |
| 3 | TCNT3 | Complementary PWM mode: up/down-counter |
| 4 | TCNT4 | Other modes: up-counter |

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Each TCNT is a 16 -bit readable/writable register that counts pulse inputs from a clock source. The clock source is selected by bits TPSC2 to TPSC0 in TCR.

TCNT0 and TCNT1 are up-counters. TCNT2 is an up/down-counter in phase counting mode and an up-counter in other modes. TCNT3 and TCNT4 are up/down-counters in complementary PWM mode and up-counters in other modes.

TCNT can be cleared to H'0000 by compare match with GRA or GRB or by input capture to GRA or GRB (counter clearing function) in the same channel.

When TCNT overflows (changes from H'FFFF to H'0000), the OVF flag is set to 1 in TSR of the corresponding channel.

When TCNT underflows (changes from H'0000 to H'FFFF), the OVF flag is set to 1 in TSR of the corresponding channel.

The TCNTs are linked to the CPU by an internal 16-bit bus and can be written or read by either word access or byte access.

Each TCNT is initialized to $\mathrm{H}^{\prime} 0000$ by a reset and in standby mode.

### 10.2.8 General Registers (GRA, GRB)

The general registers are 16-bit registers. The ITU has 10 general registers, two in each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | GRA0, GRB0 | Output compare/input capture register |
| 1 | GRA1, GRB1 |  |
| 2 | GRA2, GRB2 |  |
| 3 | GRA3, GRB3 | Output compare/input capture register; can be buffered by |
| 4 | GRA4, GRB4 | buffer registers BRA and BRB |

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

A general register is a 16-bit readable/writable register that can function as either an output compare register or an input capture register. The function is selected by settings in TIOR.

When a general register is used as an output compare register, its value is constantly compared with the TCNT value. When the two values match (compare match), the IMFA or IMFB flag is set to 1 in TSR. Compare match output can be selected in TIOR.

When a general register is used as an input capture register, rising edges, falling edges, or both edges of an external input capture signal are detected and the current TCNT value is stored in the general register. The corresponding IMFA or IMFB flag in TSR is set to 1 at the same time. The valid edge or edges of the input capture signal are selected in TIOR.

TIOR settings are ignored in PWM mode, complementary PWM mode, and reset-synchronized PWM mode.

General registers are linked to the CPU by an internal 16 -bit bus and can be written or read by either word access or byte access.

General registers are initialized to the output compare function (with no output signal) by a reset and in standby mode. The initial value is H'FFFF.

### 10.2.9 Buffer Registers (BRA, BRB)

The buffer registers are 16 -bit registers. The ITU has four buffer registers, two each in channels 3 and 4.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 3 | BRA3, BRB3 | Used for buffering |
| 4 | BRA4, BRB4 | -When the corresponding GRA or GRB functions as an <br> output compare register, BRA or BRB can function as an <br> output compare buffer register: the BRA or BRB value is <br> automatically transferred to GRA or GRB at compare match |
|  | -When the corresponding GRA or GRB functions as an input <br> capture register, BRA or BRB can function as an input <br> capture buffer register: the GRA or GRB value is <br> automatically transferred to BRA or BRB at input capture |  |
|  |  |  |

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write

A buffer register is a 16-bit readable/writable register that is used when buffering is selected. Buffering can be selected independently by bits BFB4, BFA4, BFB3, and BFA3 in TFCR.

The buffer register and general register operate as a pair. When the general register functions as an output compare register, the buffer register functions as an output compare buffer register. When the general register functions as an input capture register, the buffer register functions as an input capture buffer register.

The buffer registers are linked to the CPU by an internal 16 -bit bus and can be written or read by either word or byte access.

Buffer registers are initialized to H'FFFF by a reset and in standby mode.
Rev. 2.0, 03/01, page 330 of 822

### 10.2.10 Timer Control Registers (TCR)

TCR is an 8-bit register. The ITU has five TCRs, one in each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | TCR0 | TCR controls the timer counter. The TCRs in all channels are |
| 1 | TCR1 | functionally identical. When phase counting mode is selected in |
| 2 | TCR2 | channel 2, the settings of bits CKEG1 and CKEG0 and TPSC2 |
| 3 | to TPSC0 in TCR2 are ignored. |  |

Bit
Bit
Initial value
Read/Write

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | CCLR1 | CCLR0 | CKEG1 | CKEGO | TPSC2 | TPSC1 | TPSC0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| - | R/W | R/W | R/W | R/W | R/W R/W |  | R/W |
|  | Counte These | clear 1/0 <br> s select | Clock <br> These <br> e counte | dge $1 / 0$ <br> bits select <br> clear so | Timer <br> These count <br> external <br> rce | prescale bits selec clock <br> ock edge | 2 to 0 the |
| ser |  |  |  |  |  |  |  |

Each TCR is an 8-bit readable/writable register that selects the timer counter clock source, selects the edge or edges of external clock sources, and selects how the counter is cleared.

TCR is initialized to H'80 by a reset and in standby mode.

Bit 7—Reserved: Read-only bit, always read as 1.

Bits 6 and 5—Counter Clear 1/0 (CCLR1, CCLR0): These bits select how TCNT is cleared.
Bit 6: CCLR1 Bit 5: CCLR0 Description

| 0 | 0 | TCNT is not cleared |
| :--- | :--- | :--- |
|  | 1 | TCNT is cleared by GRA compare match or input <br> capture* |
| 1 | 0 | TCNT is cleared by GRB compare match or input <br> capture* |
|  | Synchronous clear: TCNT is cleared in synchronization <br> with other synchronized timers*2 |  |

Notes: 1. TCNT is cleared by compare match when the general register functions as an output compare register, and by input capture when the general register functions as an input capture register.
2. Selected in TSNC.

Bits 4 and 3—Clock Edge 1/0 (CKEG1, CKEG0): These bits select external clock input edges when an external clock source is used.

Bit 4: CKEG1 Bit 3: CKEG0 Description

| 0 | 0 | Count rising edges | (Initial value) |
| :--- | :--- | :--- | :--- |
|  | 1 | Count falling edges |  |
| 1 | - | Count both edges |  |

When channel 2 is set to phase counting mode, bits CKEG1 and CKEG0 in TCR2 are ignored.
Phase counting takes precedence.
Bits 2 to 0—Timer Prescaler 2 to 0 (TPSC2 to TPSC0): These bits select the counter clock source.

| Bit 2: TPSC2 | Bit 1: TPSC1 | Bit 0: TPSC0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Internal clock: $\phi$ |
|  |  | 1 | Internal clock: $\phi / 2$ |
| 1 | $\frac{0}{1}$ | Internal clock: $\phi / 4$ |  |
| 1 | 0 | 0 | Internal clock: $\phi / 8$ |
|  |  | 1 | External clock $A:$ TCLKA input |
| 1 | 0 | External clock B: TCLKB input |  |

When bit TPSC2 is cleared to 0 an internal clock source is selected, and the timer counts only falling edges. When bit TPSC2 is set to 1 an external clock source is selected, and the timer counts the edge or edges selected by bits CKEG1 and CKEG0.

When channel 2 is set to phase counting mode (MDF = 1 in TMDR), the settings of bits TPSC2 to TPSC0 in TCR2 are ignored. Phase counting takes precedence.

### 10.2.11 Timer I/O Control Register (TIOR)

TIOR is an 8-bit register. The ITU has five TIORs, one in each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | TIOR0 | TIOR controls the general registers. Some functions differ in |
| 1 | TIOR1 | PWM mode. TIOR3 and TIOR4 settings are ignored when |
| 2 | TIOR2 | complementary PWM mode or reset-synchronized PWM mode |
| 3 | TIOR3 |  |

Bit


Each TIOR is an 8-bit readable/writable register that selects the output compare or input capture function for GRA and GRB, and specifies the functions of the TIOCA and TIOCB pins. If the output compare function is selected, TIOR also selects the type of output. If input capture is selected, TIOR also selects the edge or edges of the input capture signal.

TIOR is initialized to $\mathrm{H}^{\prime} 88$ by a reset and in standby mode.
Bit 7—Reserved: Read-only bit, always read as 1.

Bits 6 to 4-I/O Control B2 to B0 (IOB2 to IOB0): These bits select the GRB function.

| Bit 6: <br> IOB2 | Bit 5: <br> IOB1 | Bit 4: <br> IOBO | Description |  |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | GRB is an output compare register | No output at compare match (Initial value) |
|  |  | 1 |  | 0 output at GRB compare match*1 |
|  | 1 | 0 |  | 1 output at GRB compare match*1 |
|  |  | 1 |  | Output toggles at GRB compare match (1 output in channel 2)*1, *2 |
| 1 | 0 | 0 | GRB is an input capture register | GRB captures rising edge of input |
|  |  | 1 |  | GRB captures falling edge of input |
|  | 1 | 0 |  | GRB captures both edges of input |
|  |  | 1 |  |  |

Notes: 1. After a reset, the output is 0 until the first compare match.
2. Channel 2 output cannot be toggled by compare match. This setting selects 1 output instead.

Bit 3—Reserved: Read-only bit, always read as 1.
Bits 2 to 0-I/O Control A2 to A0 (IOA2 to IOA0): These bits select the GRA function.

| Bit 2: IOA2 | Bit 1: IOA1 | Bit 0: IOAO | Description |  |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | GRA is an output compare register | No output at compare match (Initial value) |
|  |  | 1 |  | 0 output at GRA compare match*1 |
|  | 1 | 0 |  | 1 output at GRA compare match*1 |
|  |  | 1 |  | Output toggles at GRA compare match (1 output in channel 2) |
| 1 | 0 | 0 | GRA is an input capture register | GRA captures rising edge of input |
|  |  | 1 |  | GRA captures falling edge of input |
|  | 1 | 0 |  | GRA captures both edges of input |
|  |  | 1 |  |  |

Notes: 1. After a reset, the output is 0 until the first compare match.
2. Channel 2 output cannot be toggled by compare match. This setting selects 1 output instead.

### 10.2.12 Timer Status Register (TSR)

TSR is an 8-bit register. The ITU has five TSRs, one in each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | TSR0 |  |
| 1 | TSR1 |  |
| 2 | TSR2 |  |
| 3 | TSR3 |  |
| 4 | TSR4 |  |

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | OVF | IMFB | IMFA |
| 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| - | - | - | - | - | $\underline{R /(W) *}$ | $\underline{\mathrm{R} /(\mathrm{W})^{*}}$ | $\underline{\mathrm{R} /(\mathrm{W})^{*}}$ |
| Reserved bits |  |  |  |  |  |  |  |
|  |  |  |  | Ove <br> Stat over | $w$ flag ag indic or und |  |  |
|  |  |  | Input capture/compare match flag B Status flag indicating GRB compare match or input capture |  |  |  |  |

Input capture/compare match flag A Status flag indicating GRA compare match or input capture

Note: * Only 0 can be written, to clear the flag.

Each TSR is an 8-bit readable/writable register containing flags that indicate TCNT overflow or underflow and GRA or GRB compare match or input capture. These flags are interrupt sources and generate CPU interrupts if enabled by corresponding bits in TIER.

TSR is initialized to H'F8 by a reset and in standby mode.
Bits 7 to 3—Reserved: Read-only bits, always read as 1.

Bit 2—Overflow Flag (OVF): This status flag indicates TCNT overflow or underflow.
Bit 2: OVF Description

Read OVF when OVF $=1$, then write 0 in OVF
[Setting condition]
TCNT overflowed from H'FFFF to H'0000, or underflowed from H'0000 to H'FFFF*
Notes: * TCNT underflow occurs when TCNT operates as an up/down-counter. Underflow occurs only under the following conditions:

1. Channel 2 operates in phase counting mode (MDF $=1$ in TMDR)
2. Channels 3 and 4 operate in complementary PWM mode (CMD1 $=1$ and CMD0 $=0$ in TFCR)

Bit 1—Input Capture/Compare Match Flag B (IMFB): This status flag indicates GRB compare match or input capture events.

| Bit 1: IMFB | Description |
| :---: | :---: |
| 0 | [Clearing condition] (Initial value) |
|  | Read IMFB when IMFB $=1$, then write 0 in IMFB |
| 1 | [Setting conditions] |
|  | TCNT = GRB when GRB functions as an output compare register. |
|  | TCNT value is transferred to GRB by an input capture signal, when GRB functions as an input capture register. |

Bit 0—Input Capture/Compare Match Flag A (IMFA): This status flag indicates GRA compare match or input capture events.

| Bit 0: IMFA | Description |
| :--- | :--- |
| 0 | [Clearing condition] |
|  | Read IMFA when IMFA = 1, then write 0 in IMFA. |
|  | DMAC activated by IMIA interrupt (channels 0 to 3 only). |
| 1 | [Setting conditions] |
|  | TCNT = GRA when GRA functions as an output compare register. |
|  | TCNT value is transferred to GRA by an input capture signal, when GRA |
|  | functions as an input capture register. |

### 10.2.13 Timer Interrupt Enable Register (TIER)

TIER is an 8-bit register. The ITU has five TIERs, one in each channel.

| Channel | Abbreviation | Function |
| :--- | :--- | :--- |
| 0 | TIER0 | Enables or disables interrupt requests. |
| 1 | TIER1 |  |
| 2 | TIER2 |  |
| 3 | TIER3 |  |
| 4 | TIER4 |  |



Each TIER is an 8-bit readable/writable register that enables and disables overflow interrupt requests and general register input capture and compare match interrupt requests.

TIER is initialized to H'F8 by a reset and in standby mode.

Bits 7 to 3-Reserved: Read-only bits, always read as 1.

Bit 2-Overflow Interrupt Enable (OVIE): Enables or disables the interrupt requested by the OVF flag in TSR when OVF is set to 1 .

| Bit 2: OVIE | Description |  |
| :--- | :--- | :--- |
| 0 | OVI interrupt requested by OVF is disabled | (Initial value) |
| 1 | OVI interrupt requested by OVF is enabled |  |

Bit 1—Input Capture/Compare Match Interrupt Enable B (IMIEB): Enables or disables the interrupt requested by the IMFB flag in TSR when IMFB is set to 1 .

| Bit 1: IMIEB | Description |  |
| :--- | :--- | :--- |
| 0 | IMIB interrupt requested by IMFB is disabled | (Initial value) |
| 1 | IMIB interrupt requested by IMFB is enabled |  |

Bit 0—Input Capture/Compare Match Interrupt Enable A (IMIEA): Enables or disables the interrupt requested by the IMFA flag in TSR when IMFA is set to 1 .

| Bit 0: IMIEA | Description |  |
| :--- | :--- | :--- |
| 0 | IMIA interrupt requested by IMFA is disabled | (Initial value) |
| 1 | IMIA interrupt requested by IMFA is enabled |  |

### 10.3 CPU Interface

### 10.3.1 16-Bit Accessible Registers

The timer counters (TCNTs), general registers $A$ and $B$ (GRAs and GRBs), and buffer registers $A$ and B (BRAs and BRBs) are 16-bit registers, and are linked to the CPU by an internal 16-bit data bus. These registers can be written or read a word at a time, or a byte at a time.

Figures 10.6 and 10.7 show examples of word access to a timer counter (TCNT). Figures 10.8, $10.9,10.10$, and 10.11 show examples of byte access to TCNTH and TCNTL.


Figure 10.6 Access to Timer Counter (CPU Writes to TCNT, Word)


Figure 10.7 Access to Timer Counter (CPU Reads TCNT, Word)

## On-chip data bus



Figure 10.8 Access to Timer Counter (CPU Writes to TCNT, Upper Byte)


Figure 10.9 Access to Timer Counter (CPU Writes to TCNT, Lower Byte)

On-chip data bus


Figure 10.10 Access to Timer Counter (CPU Reads TCNT, Upper Byte)

On-chip data bus


Figure 10.11 Access to Timer Counter (CPU Reads TCNT, Lower Byte)

### 10.3.2 8-Bit Accessible Registers

The registers other than the timer counters (TCNTS), general registers A and B (GRAs and GRBs), and buffer registers A and B (BRAs and BRBs) are 8-bit registers. These registers are linked to the CPU by an internal 8 -bit data bus.

Figures 10.12 and 10.13 show examples of byte read and write access to a TCR.
If a word-size data transfer instruction is executed, two byte transfers are performed.


Figure 10.12 Access to Timer Counter (CPU Writes to TCR)


Figure 10.13 Access to Timer Counter (CPU Reads TCR)

### 10.4 Operation

### 10.4.1 Overview

A summary of operations in the various modes is given below.

Normal Operation: Each channel has a timer counter and general registers. The timer counter counts up, and can operate as a free-running counter, periodic counter, or external event counter. General registers A and B can be used for input capture or output compare.

Synchronous Operation: The timer counters in designated channels are preset synchronously. Data written to the timer counter in any one of these channels is simultaneously written to the timer counters in the other channels as well. The timer counters can also be cleared synchronously if so designated by the CCLR1 and CCLR0 bits in the TCRs.

PWM Mode: A PWM waveform is output from the TIOCA pin. The output goes to 1 at compare match A and to 0 at compare match B . The duty cycle can be varied from $0 \%$ to $100 \%$ depending on the settings of GRA and GRB. When a channel is set to PWM mode, its GRA and GRB automatically become output compare registers.

Reset-Synchronized PWM Mode: Channels 3 and 4 are paired for three-phase PWM output with complementary waveforms. (The three phases are related by having a common transition point.) When reset-synchronized PWM mode is selected GRA3, GRB3, GRA4, and GRB4 automatically function as output compare registers, TIOCA $_{3}$, TIOCB $_{3}$, TIOCA $_{4}$, TOCXA $_{4}$, TIOCB $_{4}$, and $\mathrm{TOCXB}_{4}$ function as PWM output pins, and TCNT3 operates as an up-counter. TCNT4 operates independently, and is not compared with GRA4 or GRB4.

Complementary PWM Mode: Channels 3 and 4 are paired for three-phase PWM output with non-overlapping complementary waveforms. When complementary PWM mode is selected GRA3, GRB3, GRA4, and GRB4 automatically function as output compare registers, and TIOCA $_{3}$, TIOCB $_{3}$, TIOCA $_{4}$, TOCXA $_{4}$, TIOCB $_{4}$, and TOCXB $_{4}$ function as PWM output pins. TCNT3 and TCNT4 operate as up/down-counters.

Phase Counting Mode: The phase relationship between two clock signals input at TCLKA and TCLKB is detected and TCNT2 counts up or down accordingly. When phase counting mode is selected TCLKA and TCLKB become clock input pins and TCNT2 operates as an up/downcounter.

## Buffering

- If the general register is an output compare register When compare match occurs the buffer register value is transferred to the general register.
- If the general register is an input capture register

When input capture occurs the TCNT value is transferred to the general register, and the previous general register value is transferred to the buffer register.

- Complementary PWM mode

The buffer register value is transferred to the general register when TCNT3 and TCNT4 change counting direction.

- Reset-synchronized PWM mode

The buffer register value is transferred to the general register at GRA3 compare match.

### 10.4.2 Basic Functions

Counter Operation: When one of bits STR0 to STR4 is set to 1 in the timer start register (TSTR), the timer counter (TCNT) in the corresponding channel starts counting. The counting can be freerunning or periodic.

- Sample setup procedure for counter

Figure 10.14 shows a sample procedure for setting up a counter.


1. Set bits TPSC2 to TPSC0 in TCR to select the counter clock source. If an external clock source is selected, set bits CKEG1 and CKEG0 in TCR to select the desired edge(s) of the external clock signal.
2. For periodic counting, set CCLR1 and CCLR0 in TCR to have TCNT cleared at GRA compare match or GRB compare match.
3. Set TIOR to select the output compare function of GRA or GRB, whichever was selected in step 2.
4. Write the count period in GRA or GRB, whichever was selected in step 2.
5. Set the STR bit to 1 in TSTR to start the timer counter.

Figure 10.14 Counter Setup Procedure (Example)

- Free-running and periodic counter operation

A reset leaves the counters (TCNTs) in ITU channels 0 to 4 all set as free-running counters. A free-running counter starts counting up when the corresponding bit in TSTR is set to 1 . When the count overflows from H'FFFF to $\mathrm{H}^{\prime} 0000$, the OVF flag is set to 1 in TSR. If the corresponding OVIE bit is set to 1 in TIER, a CPU interrupt is requested. After the overflow, the counter continues counting up from H'0000. Figure 10.15 illustrates free-running counting.


Figure 10.15 Free-Running Counter Operation
When a channel is set to have its counter cleared by compare match, in that channel TCNT operates as a periodic counter. Select the output compare function of GRA or GRB, set bit CCLR1 or CCLR0 in TCR to have the counter cleared by compare match, and set the count period in GRA or GRB. After these settings, the counter starts counting up as a periodic counter when the corresponding bit is set to 1 in TSTR. When the count matches GRA or GRB, the IMFA or IMFB flag is set to 1 in TSR and the counter is cleared to H'0000. If the corresponding IMIEA or IMIEB bit is set to 1 in TIER, a CPU interrupt is requested at this time. After the compare match, TCNT continues counting up from H'0000. Figure 10.16 illustrates periodic counting.


Figure 10.16 Periodic Counter Operation

- TCNT count timing
- Internal clock source

Bits TPSC2 to TPSC0 in TCR select the system clock $(\phi)$ or one of three internal clock sources obtained by prescaling the system clock $(\phi / 2, \phi / 4, \phi / 8)$.
Figure 10.17 shows the timing.


Figure 10.17 Count Timing for Internal Clock Sources

- External clock source

Bits TPSC2 to TPSC0 in TCR select an external clock input pin (TCLKA to TCLKD), and its valid edge or edges are selected by bits CKEG1 and CKEG0. The rising edge, falling edge, or both edges can be selected.
The pulse width of the external clock signal must be at least 1.5 system clocks when a single edge is selected, and at least 2.5 system clocks when both edges are selected. Shorter pulses will not be counted correctly.
Figure 10.18 shows the timing when both edges are detected.


Figure 10.18 Count Timing for External Clock Sources (when Both Edges are Detected)
Waveform Output by Compare Match: In ITU channels $0,1,3$, and 4 , compare match A or B can cause the output at the TIOCA or TIOCB pin to go to 0 , go to 1 , or toggle. In channel 2 the output can only go to 0 or go to 1 .

- Sample setup procedure for waveform output by compare match

Figure 10.19 shows a sample procedure for setting up waveform output by compare match.


Figure 10.19 Setup Procedure for Waveform Output by Compare Match (Example)

- Examples of waveform output

Figure 10.20 shows examples of 0 and 1 output. TCNT operates as a free-running counter, 0 output is selected for compare match A , and 1 output is selected for compare match B . When the pin is already at the selected output level, the pin level does not change.


Figure 10.200 and 1 Output (Examples)
Figure 10.21 shows examples of toggle output. TCNT operates as a periodic counter, cleared by compare match $B$. Toggle output is selected for both compare match A and B.


Figure 10.21 Toggle Output (Example)

- Output compare timing

The compare match signal is generated in the last state in which TCNT and the general register match (when TCNT changes from the matching value to the next value). When the compare match signal is generated, the output value selected in TIOR is output at the output compare pin (TIOCA or TIOCB). When TCNT matches a general register, the compare match signal is not generated until the next counter clock pulse.
Figure 10.22 shows the output compare timing.


Figure 10.22 Output Compare Timing
Input Capture Function: The TCNT value can be captured into a general register when a transition occurs at an input capture/output compare pin (TIOCA or TIOCB). Capture can take place on the rising edge, falling edge, or both edges. The input capture function can be used to measure pulse width or period.

- Sample setup procedure for input capture

Figure 10.23 shows a sample procedure for setting up input capture.


Figure 10.23 Setup Procedure for Input Capture (Example)

- Examples of input capture

Figure 10.24 illustrates input capture when the falling edge of TIOCB and both edges of TIOCA are selected as capture edges. TCNT is cleared by input capture into GRB.


Figure 10.24 Input Capture (Example)

- Input capture signal timing

Input capture on the rising edge, falling edge, or both edges can be selected by settings in TIOR. Figure 10.25 shows the timing when the rising edge is selected. The pulse width of the input capture signal must be at least 1.5 system clocks for single-edge capture, and 2.5 system clocks for capture of both edges.


Input-capture input


Internal input
capture signal

TCNT

$\qquad$


Figure 10.25 Input Capture Signal Timing

### 10.4.3 Synchronization

The synchronization function enables two or more timer counters to be synchronized by writing the same data to them simultaneously (synchronous preset). With appropriate TCR settings, two or more timer counters can also be cleared simultaneously (synchronous clear). Synchronization enables additional general registers to be associated with a single time base. Synchronization can be selected for all channels ( 0 to 4 ).

Sample Setup Procedure for Synchronization: Figure 10.26 shows a sample procedure for setting up synchronization.


1. Set the SYNC bits to 1 in TSNC for the channels to be synchronized.
2. When a value is written in TCNT in one of the synchronized channels, the same value is simultaneously written in TCNT in the other channels (synchronized preset).
3. Set the CCLR1 or CCLR0 bit in TCR to have the counter cleared by compare match or input capture.
4. Set the CCLR1 and CCLR0 bits in TCR to have the counter cleared synchronously.
5. Set the STR bits in TSTR to 1 to start the synchronized counters.

Figure 10.26 Setup Procedure for Synchronization (Example)

Example of Synchronization: Figure 10.27 shows an example of synchronization. Channels 0,1 , and 2 are synchronized, and are set to operate in PWM mode. Channel 0 is set for counter clearing by compare match with GRB0. Channels 1 and 2 are set for synchronous counter clearing. The timer counters in channels 0,1 , and 2 are synchronously preset, and are synchronously cleared by compare match with GRB0. A three-phase PWM waveform is output from pins TIOCA ${ }_{0}$, TIOCA $_{1}$, and TIOCA ${ }_{2}$. For further information on PWM mode, see section 10.4.4, PWM Mode.


Figure 10.27 Synchronization (Example)

### 10.4.4 PWM Mode

In PWM mode GRA and GRB are paired and a PWM waveform is output from the TIOCA pin. GRA specifies the time at which the PWM output changes to 1 . GRB specifies the time at which the PWM output changes to 0 . If either GRA or GRB is selected as the counter clear source, a PWM waveform with a duty cycle from $0 \%$ to $100 \%$ is output at the TIOCA pin. PWM mode can be selected in all channels (0 to 4 ).

Table 10.4 summarizes the PWM output pins and corresponding registers. If the same value is set in GRA and GRB, the output does not change when compare match occurs.

Table 10.4 PWM Output Pins and Registers

| Channel | Output Pin | $\mathbf{1}$ Output | 0 Output |
| :--- | :--- | :--- | :--- |
| 0 | $\mathrm{TIOCA}_{0}$ | GRA0 | GRB0 |
| 1 | $\mathrm{TIOCA}_{1}$ | GRA1 | GRB1 |
| 2 | $\mathrm{TIOCA}_{2}$ | GRA2 | GRB2 |
| 3 | $\mathrm{TIOCA}_{3}$ | GRA3 | GRB3 |
| 4 | $\mathrm{TIOCA}_{4}$ | GRA4 | GRB4 |

Sample Setup Procedure for PWM Mode: Figure 10.28 shows a sample procedure for setting up PWM mode.


Figure 10.28 Setup Procedure for PWM Mode (Example)

Examples of PWM Mode: Figure 10.29 shows examples of operation in PWM mode. In PWM mode TIOCA becomes an output pin. The output goes to 1 at compare match with GRA, and to 0 at compare match with GRB.

In the examples shown, TCNT is cleared by compare match with GRA or GRB. Synchronized operation and free-running counting are also possible.


Figure 10.29 PWM Mode (Example 1)

Figure 10.30 shows examples of the output of PWM waveforms with duty cycles of $0 \%$ and $100 \%$. If the counter is cleared by compare match with GRB, and GRA is set to a higher value than GRB, the duty cycle is $0 \%$. If the counter is cleared by compare match with GRA, and GRB is set to a higher value than GRA, the duty cycle is $100 \%$.


Figure 10.30 PWM Mode (Example 2)

### 10.4.5 Reset-Synchronized PWM Mode

In reset-synchronized PWM mode channels 3 and 4 are combined to produce three pairs of complementary PWM waveforms, all having one waveform transition point in common.

When reset-synchronized PWM mode is selected $\mathrm{TIOCA}_{3}, \mathrm{TIOCB}_{3}, \mathrm{TIOCA}_{4}, \mathrm{TOCXA}_{4}, \mathrm{TIOCB}_{4}$, and $\mathrm{TOCXB}_{4}$ automatically become PWM output pins, and TCNT3 functions as an up-counter.

Table 10.5 lists the PWM output pins. Table 10.6 summarizes the register settings.
Table 10.5 Output Pins in Reset-Synchronized PWM Mode

| Channel | Output Pin | Description |
| :--- | :--- | :--- |
| 3 | TIOCA $_{3}$ | PWM output 1 |
| $\mathrm{TIOCB}_{3}$ | PWM output 1' (complementary waveform to PWM output 1) |  |
| 4 | $\mathrm{TIOCA}_{4}$ | PWM output 2 |
| $\mathrm{TOCXA}_{4}$ | PWM output 2' (complementary waveform to PWM output 2) |  |
|  | TIOCB $_{4}$ | PWM output 3 |
| $\mathrm{TOCXB}_{4}$ | PWM output 3' (complementary waveform to PWM output 3) |  |

Table 10.6 Register Settings in Reset-Synchronized PWM Mode

| Register | Setting |
| :--- | :--- |
| TCNT3 | Initially set to H'0000 |
| TCNT4 | Not used (operates independently) |
| GRA3 | Specifies the count period of TCNT3 |
| GRB3 | Specifies a transition point of PWM waveforms output from TIOCA $_{3}$ and TIOCB $_{3}$ |
| GRA4 | Specifies a transition point of PWM waveforms output from TIOCA $_{4}$ and TOCXA $_{4}$ |
| GRB4 | Specifies a transition point of PWM waveforms output from TIOCB $_{4}$ and TOCXB $_{4}$ |

Sample Setup Procedure for Reset-Synchronized PWM Mode: Figure 10.31 shows a sample procedure for setting up reset-synchronized PWM mode.


Figure 10.31 Setup Procedure for Reset-Synchronized PWM Mode (Example)

Example of Reset-Synchronized PWM Mode: Figure 10.32 shows an example of operation in reset-synchronized PWM mode. TCNT3 operates as an up-counter in this mode. TCNT4 operates independently, detached from GRA4 and GRB4. When TCNT3 matches GRA3, TCNT3 is cleared and resumes counting from $\mathrm{H}^{\prime} 0000$. The PWM outputs toggle at compare match of TCNT3 with GRB3, GRA4, and GRB4 respectively, and all toggle when the counter is cleared.


Figure 10.32 Operation in Reset-Synchronized PWM Mode (Example) $($ when OLS3 $=$ OLS4 $=1$ )

For the settings and operation when reset-synchronized PWM mode and buffer mode are both selected, see section 10.4.8, Buffering.

### 10.4.6 Complementary PWM Mode

In complementary PWM mode channels 3 and 4 are combined to output three pairs of complementary, non-overlapping PWM waveforms.

When complementary PWM mode is selected $\mathrm{TIOCA}_{3}, \mathrm{TIOCB}_{3}, \mathrm{TIOCA}_{4}, \mathrm{TOCXA}_{4}, \mathrm{TIOCB}_{4}$, and $\mathrm{TOCXB}_{4}$ automatically become PWM output pins, and TCNT3 and TCNT4 function as up/downcounters.

Table 10.7 lists the PWM output pins. Table 10.8 summarizes the register settings.

| Table 10.7 |  |  |
| :---: | :---: | :---: |
| Channel | Output Pin | Description |
| 3 | $\mathrm{TIOCA}_{3}$ | PWM output 1 |
|  | $\mathrm{TIOCB}_{3}$ | PWM output 1' (non-overlapping complementary waveform to PWM output 1) |
| 4 | $\mathrm{TIOCA}_{4}$ | PWM output 2 |
|  | TOCXA $_{4}$ | PWM output 2' (non-overlapping complementary waveform to PWM output 2) |
|  | $\mathrm{TIOCB}_{4}$ | PWM output 3 |
|  | $\mathrm{TOCXB}_{4}$ | PWM output 3' (non-overlapping complementary waveform to PWM output 3) |

## Table 10.8 Register Settings in Complementary PWM Mode

| Register | Setting |
| :--- | :--- |
| TCNT3 | Initially specifies the non-overlap margin (difference to TCNT4) |
| TCNT4 | Initially set to H'0000 |
| GRA3 | Specifies the upper limit value of TCNT3 minus 1 |
| GRB3 | Specifies a transition point of PWM waveforms output from TIOCA $_{3}$ and TIOCB $_{3}$ |
| GRA4 | Specifies a transition point of PWM waveforms output from TIOCA $_{4}$ and TOCXA |
| GRB4 | Specifies a transition point of PWM waveforms output from TIOCB $_{4}$ and TOCXB ${ }_{4}$ |

Setup Procedure for Complementary PWM Mode: Figure 10.33 shows a sample procedure for setting up complementary PWM mode.


1. Clear bits STR3 and STR4 to 0 in TSTR to halt the timer counters. Complementary PWM mode must be set up while TCNT3 and TCNT4 are halted.
2. Set bits TPSC2 to TPSC0 in TCR to select the same counter clock source for channels 3 and 4. If an external clock source is selected, select the external clock edge(s) with bits CKEG1 and CKEG0 in TCR. Do not select any counter clear source with bits CCLR1 and CCLRO in TCR.
3. Set bits CMD1 and CMD0 in TFCR to select complementary PWM mode. TIOCA $_{3}$, TIOCB $_{3}$, TIOCA $_{4}$, $\mathrm{TIOCB}_{4}$, TOCXA $_{4}$, and TOCXB 4 automatically become PWM output pins.
4. Clear TCNT4 to H'0000. Set the nonoverlap margin in TCNT3. Do not set TCNT3 and TCNT4 to the same value.
5. GRA3 is the waveform period register. Set the upper limit value of TCNT3 minus 1 in GRA3. Set transition times of the PWM output waveforms in GRB3, GRA4, and GRB4. Set times within the compare match range of TCNT3 and TCNT4. $T \leq X$ ( X : initial setting of GRB3, GRA4, or GRB4. T: initial setting of TCNT3)
6. Set bits STR3 and STR4 in TSTR to 1 to start TCNT3 and TCNT4.

Note: After exiting complementary PWM mode, to resume operating in complementary PWM mode, follow the entire setup procedure from step 1 again.

Figure 10.33 Setup Procedure for Complementary PWM Mode (Example)

Clearing Procedure for Complementary PWM Mode: Figure 10.34 shows the steps to clear complementary PWM mode.


1. Clear the CMD1 bit of TFCR to 0 to set channels 3 and 4 to normal operating mode.
2. After setting channels 3 and 4 to normal operating mode, wait at least one counter clock period, then clear bits STR3 and STR4 of TSTR to 0 to stop counter operation of TCNT3 and TCNT4.

Figure 10.34 Clearing Procedure for Complementary PWM Mode

Examples of Complementary PWM Mode: Figure 10.35 shows an example of operation in complementary PWM mode. TCNT3 and TCNT4 operate as up/down-counters, counting down from compare match between TCNT3 and GRA3 and counting up from the point at which TCNT4 underflows. During each up-and-down counting cycle, PWM waveforms are generated by compare match with general registers GRB3, GRA4, and GRB4. Since TCNT3 is initially set to a higher value than TCNT4, compare match events occur in the sequence TCNT3, TCNT4, TCNT4, TCNT3.


Figure 10.35 Operation in Complementary PWM Mode (Example 1, OLS3 = OLS4 = 1 )

Figure 10.36 shows examples of waveforms with $0 \%$ and $100 \%$ duty cycles (in one phase) in complementary PWM mode. In this example the outputs change at compare match with GRB3, so waveforms with duty cycles of $0 \%$ or $100 \%$ can be output by setting GRB3 to a value larger than GRA3. The duty cycle can be changed easily during operation by use of the buffer registers. For further information see section 10.4.8, Buffering.


Figure 10.36 Operation in Complementary PWM Mode (Example 2, OLS3 = OLS4 = 1)

In complementary PWM mode, TCNT3 and TCNT4 overshoot and undershoot at the transitions between up-counting and down-counting. The setting conditions for the IMFA bit in channel 3 and the OVF bit in channel 4 differ from the usual conditions. In buffered operation the buffer transfer conditions also differ. Timing diagrams are shown in figures 10.37 and 10.38 .


Figure 10.37 Overshoot Timing


Figure 10.38 Undershoot Timing
In channel 3, IMFA is set to 1 only during up-counting. In channel 4, OVF is set to 1 only when an underflow occurs. When buffering is selected, buffer register contents are transferred to the general register at compare match A3 during up-counting, and when TCNT4 underflows.

General Register Settings in Complementary PWM Mode: When setting up general registers for complementary PWM mode or changing their settings during operation, note the following points.

- Initial settings

Do not set values from $\mathrm{H}^{\prime} 0000$ to $\mathrm{T}-1$ (where T is the initial value of TCNT3). After the counters start and the first compare match A3 event has occurred, however, settings in this range also become possible.

- Changing settings

Use the buffer registers. Correct waveform output may not be obtained if a general register is written to directly.

- Cautions on changes of general register settings



## Figure 10.39 Changing a General Register Setting by Buffer Transfer (Example 1)

- Buffer transfer at transition from up-counting to down-counting

If the general register value is in the range from GRA3 $-T+1$ to GRA3, do not transfer a buffer register value outside this range. Conversely, if the general register value is outside this range, do not transfer a value within this range. See figure 10.40.


## Figure 10.40 Changing a General Register Setting by Buffer Transfer (Caution 1)

- Buffer transfer at transition from down-counting to up-counting

If the general register value is in the range from $\mathrm{H}^{\prime} 0000$ to $\mathrm{T}-1$, do not transfer a buffer register value outside this range. Conversely, when a general register value is outside this range, do not transfer a value within this range. See figure 10.41.


Figure 10.41 Changing a General Register Setting by Buffer Transfer (Caution 2)

- General register settings outside the counting range (H'0000 to GRA3)

Waveforms with a duty cycle of $0 \%$ or $100 \%$ can be output by setting a general register to a value outside the counting range. When a buffer register is set to a value outside the counting range, then later restored to a value within the counting range, the counting direction (up or down) must be the same both times. See figure 10.42.


Figure 10.42 Changing a General Register Setting by Buffer Transfer (Example 2)
Settings can be made in this way by detecting GRA3 compare match or TCNT4 underflow before writing to the buffer register. They can also be made by using GRA3 compare match to activate the DMAC.

### 10.4.7 Phase Counting Mode

In phase counting mode the phase difference between two external clock inputs (at the TCLKA and TCLKB pins) is detected, and TCNT2 counts up or down accordingly.

In phase counting mode, the TCLKA and TCLKB pins automatically function as external clock input pins and TCNT2 becomes an up/down-counter, regardless of the settings of bits TPSC2 to TPSC0, CKEG1, and CKEG0 in TCR2. Settings of bits CCLR1, CCLR0 in TCR2, and settings in TIOR2, TIER2, TSR2, GRA2, and GRB2 are valid. The input capture and output compare functions can be used, and interrupts can be generated.

Phase counting is available only in channel 2 .
Sample Setup Procedure for Phase Counting Mode: Figure 10.43 shows a sample procedure for setting up phase counting mode.


Figure 10.43 Setup Procedure for Phase Counting Mode (Example)

Example of Phase Counting Mode: Figure 10.44 shows an example of operations in phase counting mode. Table 10.9 lists the up-counting and down-counting conditions for TCNT2.

In phase counting mode both the rising and falling edges of TCLKA and TCLKB are counted. The phase difference between TCLKA and TCLKB must be at least 1.5 states, the phase overlap must also be at least 1.5 states, and the pulse width must be at least 2.5 states. See figure 10.45 .


Figure 10.44 Operation in Phase Counting Mode (Example)
Table 10.9 Up/Down Counting Conditions

| Counting Direction | Up-Counting |  |  |  |  |  |  |  |  | Down-Counting |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| TCLKB | $\uparrow$ | High | $\downarrow$ | Low | High | $\downarrow$ | Low | $\uparrow$ |  |  |  |  |  |  |
| TCLKA | Low | $\uparrow$ | High | $\downarrow$ | $\downarrow$ | Low | $\uparrow$ | High |  |  |  |  |  |  |



Figure 10.45 Phase Difference, Overlap, and Pulse Width in Phase Counting Mode

### 10.4.8 Buffering

Buffering operates differently depending on whether a general register is an output compare register or an input capture register, with further differences in reset-synchronized PWM mode and complementary PWM mode. Buffering is available only in channels 3 and 4. Buffering operations under the conditions mentioned above are described next.

- General register used for output compare

The buffer register value is transferred to the general register at compare match.
See figure 10.46.


Figure 10.46 Compare Match Buffering

- General register used for input capture

The TCNT value is transferred to the general register at input capture. The previous general register value is transferred to the buffer register.
See figure 10.47.


Figure 10.47 Input Capture Buffering

- Complementary PWM mode

The buffer register value is transferred to the general register when TCNT3 and TCNT4 change counting direction. This occurs at the following two times:

- When TCNT3 compare matches GRA3
- When TCNT4 underflows
- Reset-synchronized PWM mode

The buffer register value is transferred to the general register at compare match A3.
Sample Buffering Setup Procedure: Figure 10.48 shows a sample buffering setup procedure.


Figure 10.48 Buffering Setup Procedure (Example)

Examples of Buffering: Figure 10.49 shows an example in which GRA is set to function as an output compare register buffered by BRA, TCNT is set to operate as a periodic counter cleared by GRB compare match, and TIOCA and TIOCB are set to toggle at compare match A and B. Because of the buffer setting, when TIOCA toggles at compare match A, the BRA value is simultaneously transferred to GRA. This operation is repeated each time compare match A occurs. Figure 10.50 shows the transfer timing.


Figure 10.49 Register Buffering (Example 1: Buffering of Output Compare Register)


Compare match signal


Buffer transfer signal

BR


Figure 10.50 Compare Match and Buffer Transfer Timing (Example)

Figure 10.51 shows an example in which GRA is set to function as an input capture register buffered by BRA, and TCNT is cleared by input capture B. The falling edge is selected as the input capture edge at TIOCB. Both edges are selected as input capture edges at TIOCA. Because of the buffer setting, when the TCNT value is captured into GRA at input capture A, the previous GRA value is simultaneously transferred to BRA. Figure 10.52 shows the transfer timing.


Figure 10.51 Register Buffering (Example 2: Buffering of Input Capture Register)


Figure 10.52 Input Capture and Buffer Transfer Timing

Figure 10.53 shows an example in which GRB3 is buffered by BRB3 in complementary PWM mode. Buffering is used to set GRB3 to a higher value than GRA3, generating a PWM waveform with $0 \%$ duty cycle. The BRB3 value is transferred to GRB3 when TCNT3 matches GRA3, and when TCNT4 underflows.


Figure 10.53 Register Buffering (Example 3: Buffering in Complementary PWM Mode)

### 10.4.9 ITU Output Timing

The ITU outputs from channels 3 and 4 can be disabled by bit settings in TOER or by an external trigger, or inverted by bit settings in TOCR.

Timing of Enabling and Disabling of ITU Output by TOER: In this example an ITU output is disabled by clearing a master enable bit to 0 in TOER. An arbitrary value can be output by appropriate settings of the data register (DR) and data direction register (DDR) of the corresponding input/output port. Figure 10.54 illustrates the timing of the enabling and disabling of ITU output by TOER.


Figure 10.54 Timing of Disabling of ITU Output by Writing to TOER (Example)

Timing of Disabling of ITU Output by External Trigger: If the XTGD bit is cleared to 0 in TOCR in reset-synchronized PWM mode or complementary PWM mode, when an input capture A signal occurs in channel 1 , the master enable bits are cleared to 0 in TOER, disabling ITU output. Figure 10.55 shows the timing.


Figure 10.55 Timing of Disabling of ITU Output by External Trigger (Example)
Timing of Output Inversion by TOCR: The output levels in reset-synchronized PWM mode and complementary PWM mode can be inverted by inverting the output level select bits (OLS4 and OLS3) in TOCR. Figure 10.56 shows the timing.


Figure 10.56 Timing of Inverting of ITU Output Level by Writing to TOCR (Example)

### 10.5 Interrupts

The ITU has two types of interrupts: input capture/compare match interrupts, and overflow interrupts.

### 10.5.1 Setting of Status Flags

Timing of Setting of IMFA and IMFB at Compare Match: IMFA and IMFB are set to 1 by a compare match signal generated when TCNT matches a general register (GR). The compare match signal is generated in the last state in which the values match (when TCNT is updated from the matching count to the next count). Therefore, when TCNT matches a general register, the compare match signal is not generated until the next timer clock input. Figure 10.57 shows the timing of the setting of IMFA and IMFB.


Figure 10.57 Timing of Setting of IMFA and IMFB by Compare Match

Timing of Setting of IMFA and IMFB by Input Capture: IMFA and IMFB are set to 1 by an input capture signal. The TCNT contents are simultaneously transferred to the corresponding general register. Figure 10.58 shows the timing.


Figure 10.58 Timing of Setting of IMFA and IMFB by Input Capture

Timing of Setting of Overflow Flag (OVF): OVF is set to 1 when TCNT overflows from H'FFFF to H'0000 or underflows from H'0000 to H'FFFF. Figure 10.59 shows the timing.


Figure 10.59 Timing of Setting of OVF

### 10.5.2 Clearing of Status Flags

If the CPU reads a status flag while it is set to 1 , then writes 0 in the status flag, the status flag is cleared. Figure 10.60 shows the timing.


Figure 10.60 Timing of Clearing of Status Flags

### 10.5.3 Interrupt Sources and DMA Controller Activation

Each ITU channel can generate a compare match/input capture A interrupt, a compare match/input capture B interrupt, and an overflow interrupt. In total there are 15 interrupt sources, all independently vectored. An interrupt is requested when the interrupt request flag and interrupt enable bit are both set to 1 .

The priority order of the channels can be modified in interrupt priority registers A and B (IPRA and IPRB). For details see section 5, Interrupt Controller.

Compare match/input capture A interrupts in channels 0 to 3 can activate the DMA controller (DMAC). When the DMAC is activated a CPU interrupt is not requested.

Table 10.10 lists the interrupt sources.
Table 10.10 ITU Interrupt Sources

| Channel | Interrupt Source | Description | DMAC <br> Activatable | Priority* |
| :---: | :---: | :---: | :---: | :---: |
| 0 | IMIA0 | Compare match/input capture A0 | Yes | High |
|  | IMIB0 | Compare match/input capture B0 | No |  |
|  | OVIO | Overflow 0 | No |  |
| 1 | IMIA1 | Compare match/input capture A1 | Yes |  |
|  | IMIB1 | Compare match/input capture B1 | No |  |
|  | OVI1 | Overflow 1 | No |  |
| 2 | IMIA2 | Compare match/input capture A2 | Yes |  |
|  | IMIB2 | Compare match/input capture B2 | No |  |
|  | OVI2 | Overflow 2 | No |  |
| 3 | IMIA3 | Compare match/input capture A3 | Yes |  |
|  | IMIB3 | Compare match/input capture B3 | No |  |
|  | OVI3 | Overflow 3 | No |  |
| 4 | IMIA4 | Compare match/input capture A4 | No |  |
|  | IMIB4 | Compare match/input capture B4 | No |  |
|  | OVI4 | Overflow 4 | No |  |

Note: * The priority immediately after a reset is indicated. Inter-channel priorities can be changed by settings in IPRA and IPRB.

### 10.6 Usage Notes

This section describes contention and other matters requiring special attention during ITU operations.

Contention between TCNT Write and Clear: If a counter clear signal occurs in the $\mathrm{T}_{3}$ state of a TCNT write cycle, clearing of the counter takes priority and the write is not performed. See figure 10.61 .


Figure 10.61 Contention between TCNT Write and Clear

Contention between TCNT Word Write and Increment: If an increment pulse occurs in the $\mathrm{T}_{3}$ state of a TCNT word write cycle, writing takes priority and TCNT is not incremented. See figure 10.62 .


Figure 10.62 Contention between TCNT Word Write and Increment

Contention between TCNT Byte Write and Increment: If an increment pulse occurs in the $\mathrm{T}_{2}$ or $\mathrm{T}_{3}$ state of a TCNT byte write cycle, writing takes priority and TCNT is not incremented. The TCNT byte that was not written retains its previous value. See figure 10.63 , which shows an increment pulse occurring in the $\mathrm{T}_{2}$ state of a byte write to TCNTH.


Figure 10.63 Contention between TCNT Byte Write and Increment

Contention between General Register Write and Compare Match: If a compare match occurs in the $\mathrm{T}_{3}$ state of a general register write cycle, writing takes priority and the compare match signal is inhibited. See figure 10.64.
Address bus
Internal write signal
GCNT

Figure 10.64 Contention between General Register Write and Compare Match

Contention between TCNT Write and Overflow or Underflow: If an overflow occurs in the $\mathrm{T}_{3}$ state of a TCNT write cycle, writing takes priority and the counter is not incremented. OVF is set to 1.The same holds for underflow. See figure 10.65.


Figure 10.65 Contention between TCNT Write and Overflow

Contention between General Register Read and Input Capture: If an input capture signal occurs during the $\mathrm{T}_{3}$ state of a general register read cycle, the value before input capture is read. See figure 10.66.


Figure 10.66 Contention between General Register Read and Input Capture

Contention between Counter Clearing by Input Capture and Counter Increment: If an input capture signal and counter increment signal occur simultaneously, the counter is cleared according to the input capture signal. The counter is not incremented by the increment signal. The value before the counter is cleared is transferred to the general register. See figure 10.67.


Figure 10.67 Contention between Counter Clearing by Input Capture and Counter Increment

Contention between General Register Write and Input Capture: If an input capture signal occurs in the $\mathrm{T}_{3}$ state of a general register write cycle, input capture takes priority and the write to the general register is not performed. See figure 10.68.


Figure 10.68 Contention between General Register Write and Input Capture
Note on Waveform Period Setting: When a counter is cleared by compare match, the counter is cleared in the last state at which the TCNT value matches the general register value, at the time when this value would normally be updated to the next count. The actual counter frequency is therefore given by the following formula:
$f=\frac{\phi}{(N+1)}$
(f: counter frequency. $\phi$ : system clock frequency. N : value set in general register.)

Contention between Buffer Register Write and Input Capture: If a buffer register is used for input capture buffering and an input capture signal occurs in the $\mathrm{T}_{3}$ state of a write cycle, input capture takes priority and the write to the buffer register is not performed. See figure 10.69.


Figure 10.69 Contention between Buffer Register Write and Input Capture

Note on Write Operations when Using Synchronous Operation: When channels are synchronized, if a TCNT value is modified by byte write access, all 16 bits of all synchronized counters assume the same value as the counter that was addressed.

Example: When channels 2 and 3 are synchronized

- Byte write to channel 2 or byte write to channel 3

Write A to upper byte

TCNT2

| $W$ | X |
| :--- | :--- |



TCNT3 $\square$
Upper byte Lower byte


Upper byte Lower byte


TCNT3

| Y | A |
| :--- | :--- |

Upper byte Lower byte

- Word write to channel 2 or word write to channel 3

| TCNT2 | W | X |  | TCNT2 | A | B |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TCNT3 | Y | Z | Write AB word to channel 2 or 3 | TCNT3 | A | B |
| Upper byte Lower byte |  |  |  | Upper byte Lower byte |  |  |

## ITU Operating Modes

Table 10.11 (1) ITU Operating Modes (Channel 0)

| Operating Mode |  | Register Settings |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | TSNC | TMDR |  |  | TFCR |  |  | TOCR |  | toer <br> Master Enable | TIORO |  | TCRO |  |
|  |  | Synchronization | MDF | FDIR | PWM | Complementary PWM | Reset-Synchronized PWM | Buffering | XTGD | Output Level Select |  | IOA | IOB | Clear Select | Clock Select |
| Synchronous preset |  | SYNC0 = 1 | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| PWM mode |  | $\bigcirc$ | - | - | PWM0 $=1$ | - | - | - | - | - | - | - | O* | $\bigcirc$ | $\bigcirc$ |
| Output compare A |  | $\bigcirc$ | - | - | $\mathrm{PWM0}=0$ | - | - | - | - | - | - | $\mathrm{IOA} 2=0$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Output compare B |  | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\mathrm{IOB2}=0$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Input capture A |  | $\bigcirc$ | - | - | PWM0 $=0$ | - | - | - | - | - | - | IOA2 $=1$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Input capture B |  | 0 | - | - | PWM0 $=0$ | - | - | - | - | - | - | $\bigcirc$ | $\text { IOB2 = } 1$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Counter clearing | By compare match/input capture A | $\bigcirc$ | - | - | 0 | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLRO }=1 \end{aligned}$ | $\bigcirc$ |
|  | By compare match/input capture B | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLRO }=0 \end{aligned}$ | $\bigcirc$ |
|  | Synchronous clear | SYNC0 $=1$ | - | - | 0 | - | - | - | - | - | - | $\bigcirc$ | 0 | $\begin{array}{l\|} \hline \text { CCLR1 }=1 \\ \text { CCLR0 }=1 \end{array}$ | $\bigcirc$ |

Legend: O Setting available (valid). - Setting does not affect this mode.
Note: * The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously, the compare match signal is inhibited.

Table 10.11 (2) ITU Operating Modes (Channel 1)

| Operating Mode |  | Register Settings |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | TSNC | TMDR |  |  | TFCR |  |  | TOCR |  | TOER | TIOR1 |  | TCR1 |  |
|  |  | Synchronization | MDF | FDIR | PWM | Complementary PWM | Reset-Synchronized PWM | Buffering | XTGD | Output Level Select | Master <br> Enable | IOA | IOB | Clear <br> Select | Clock <br> Select |
| Synchronous preset |  | SYNC1 = 1 | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| PWM mode |  | $\bigcirc$ | - | - | PWM1 = 1 | - | - | - | - | - | - | - | - ${ }^{1}$ | $\bigcirc$ | $\bigcirc$ |
| Output compare A |  | $\bigcirc$ | - | - | PWM1 = 0 | - | - | - | - | - | - | $\text { IOA2 = } 0$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Output compare B |  | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\text { IOB2 = } 0$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Input capture A |  | $\bigcirc$ | - | - | PWM1 = 0 | - | - | - | $\bigcirc^{* 2}$ | - | - | IOA2 = 1 <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Input capture B |  | $\bigcirc$ | - | - | PWM1 = 0 | - | - | - | - | - | - | $\bigcirc$ | $\text { IOB2 = } 1$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Counter clearing | By compare match/input capture A | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
|  | By compare match/input capture B | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=0 \end{aligned}$ | $\bigcirc$ |
|  | Syn- <br> chronous <br> clear | SYNC1 = 1 | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |

[^0]Notes: 1. The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously, the compare match signal is inhibited. 2. Valid only when channels 3 and 4 are operating in complementary PWM mode or reset-synchronized PWM mode.

Table 10.11 (3) ITU Operating Modes (Channel 2)

| Operating Mode |  | Register Settings |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | TSNC | TMDR |  |  | TFCR |  |  | TOCR |  | TOER | TIOR2 |  | TCR2 |  |
|  |  | Synchronization | MDF | FDIR | PWM | Complementary PWM | Reset-Synchronized PWM | Buffering | XTGD | Output Level <br> Select | Master <br> Enable | IOA | IOB | Clear <br> Select | Clock <br> Select |
| Synchronous preset |  | SYNC2 = 1 | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| PWM mode |  | $\bigcirc$ | - | - | PWM2 = 1 | - | - | - | - | - | - | - | O* | $\bigcirc$ | $\bigcirc$ |
| Output compare A |  | $\bigcirc$ | - | - | PWM2 $=0$ | - | - | - | - | - | - | $\mathrm{IOA} 2=0$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Output compare B |  | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\mathrm{IOB} 2=0$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Input capture A |  | $\bigcirc$ | - | - | PWM2 = 0 | - | - | - | - | - | - | $\text { IOA2 = } 1$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Input capture B |  | $\bigcirc$ | - | - | PWM2 = 0 | - | - | - | - | - | - | $\bigcirc$ | $\text { IOB2 = } 1$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Counter clearing | By compare match/input capture A | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
|  | By compare match/input capture B | $\bigcirc$ | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=0 \end{aligned}$ | $\bigcirc$ |
|  | Synchronous clear | SYNC2 = 1 | - | - | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
| Phase counting mode |  | $\bigcirc$ | MDF $=1$ | $\bigcirc$ | $\bigcirc$ | - | - | - | - | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - |

Legend: O Setting available (valid). - Setting does not affect this mode.
Note: * The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously, the compare match signal is inhibited.

Table 10.11 (4) ITU Operating Modes (Channel 3)

| Operating Mode |  | Register Settings |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | TSNC | TMDR |  |  | TFCR |  |  | TOCR |  | TOER <br> Master Enable | TIOR3 |  | TCR3 |  |
|  |  | Synchronization | MDF | FDIR | PWM | Complementary PWM | Reset-Synchronized PWM | Buffering | XTGD | Output Level Select |  | IOA | IOB | Clear Select | Clock <br> Select |
| Synchronous preset |  | SYNC3 = 1 | - | - | $\bigcirc$ | ○*3 | $\bigcirc$ | $\bigcirc$ | - | - | -*1 | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| PWM mode |  | $\bigcirc$ | - | - | PWM3 = 1 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | - | ${ }^{*}{ }^{2}$ | $\bigcirc$ | $\bigcirc$ |
| Output compare A |  | $\bigcirc$ | - | - | PWM3 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | IOA2 = 0 Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Output compare B |  | $\bigcirc$ | - | - | $\bigcirc$ | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | IOB2 $=0$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Input capture A |  | $\bigcirc$ | - | - | PWM3 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | EA3 ignored Other bits unrestricted | $I O A 2=1$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Input capture B |  | $\bigcirc$ | - | - | PWM3 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | EB3 ignored Other bits unrestricted | $\bigcirc$ | $\mathrm{IOB} 2=1$ <br> Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Counter clearing | By compare match/input capture A | $\bigcirc$ | - | - | $\bigcirc$ | $\begin{aligned} & \text { Illegal setting: } \\ & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | O*4 | $\bigcirc$ | - | - | ○*1 | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
|  | By compare match/input capture B | $\bigcirc$ | - | - | $\bigcirc$ | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc{ }^{* 1}$ | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=0 \end{aligned}$ | $\bigcirc$ |
|  | Synchronous clear | SYNC3 = 1 | - | - | $\bigcirc$ | $\begin{aligned} & \text { Illegal setting: } \\ & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | $\bigcirc$ | $\bigcirc$ | - | - | ○*1 | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
| Complementary PWM mode |  | O*3 | - | - | - | $\begin{array}{\|l} \mid C M D 1=1 \\ \text { CMD0 }=0 \end{array}$ | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | $\bigcirc$ | O*6 | $\bigcirc$ | $\bigcirc$ | - | - | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=0 \end{aligned}$ | ○* |
| Reset-synchronized PWM mode |  | $\bigcirc$ | - | - | - | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=1 \end{aligned}$ | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=1 \end{aligned}$ | $\bigcirc$ | O*6 | $\bigcirc$ | $\bigcirc$ | - | - | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
| Buffering (BRA) |  | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $B F A 3=1$ Other bits unrestricted | - | - | O* | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Buffering (BRB) |  | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | BFB3 $=1$ Other bits unrestricted | - | - | $\bigcirc * 1$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Legend: ○ Setting available (valid). - Setting does not affect this mode. |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| Notes: 1. Master enable bit settings are valid only during waveform output. <br> 2. The input capture function cannot be used in PWM mode. If compare match $A$ and compare match B occur simultaneously, the compare match signal <br> 3. Do not set both channels 3 and 4 for synchronous operation when complementary PWM mode is selected. <br> 4. The counter cannot be cleared by input capture A when reset-synchronized PWM mode is selected. <br> 5. In complementary PWM mode, select the same clock source for channels 3 and 4. <br> 6. Use the input capture A function in channel 1. |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Table 10.11 (5) ITU Operating Modes (Channel 4)

| Operating Mode |  | Register Settings |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | TSNC | TMDR |  |  | TFCR |  |  | TOCR |  | TOER <br> Master <br> Enable | TIOR4 |  | TCR4 |  |
|  |  | Synchronization | MDF | FDIR | PWM | Complementary PWM | Reset-Synchronized PWM | Buffering | XTGD | Output Level Select |  | IOA | IOB | Clear Select | Clock <br> Select |
| Synchronous preset |  | SYNC4 = 1 | - | - | $\bigcirc$ | -*3 | $\bigcirc$ | $\bigcirc$ | - | - | -*1 | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| PWM mode |  | $\bigcirc$ | - | - | PWM4 = 1 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | - | O*2 | $\bigcirc$ | $\bigcirc$ |
| Output compare A |  | $\bigcirc$ | - | - | PWM4 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | IOA2 $=0$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Output compare B |  | $\bigcirc$ | - | - | $\bigcirc$ | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | IOB2 $=0$ Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Input capture A |  | $\bigcirc$ | - | - | PWM4 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | EA4 ignored Other bits unrestricted | IOA2 = 1 Other bits unrestricted | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Input capture B |  | $\bigcirc$ | - | - | PWM4 = 0 | CMD1 $=0$ | CMD1 $=0$ | $\bigcirc$ | - | - | EB4 ignored Other bits unrestricted | $\bigcirc$ | IOB2 = 1 Other bits unrestricted | $\bigcirc$ | $\bigcirc$ |
| Counter clearing | By compare match/input capture A | $\bigcirc$ | - | - | $\bigcirc$ | $\begin{aligned} & \text { Illegal setting: } \\ & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | O*4 | $\bigcirc$ | - | - | ○*1 | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
|  | By compare match/input capture B | $\bigcirc$ | - | - | $\bigcirc$ | $\begin{aligned} & \text { Illegal setting: } \\ & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | ○*4 | $\bigcirc$ | - | - | ○*1 | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=0 \end{aligned}$ | $\bigcirc$ |
|  | Synchronous clear | SYNC4 = 1 | - | - | $\bigcirc$ | $\begin{aligned} & \text { Illegal setting: } \\ & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | O*4 | O | - | - | -*1 | $\bigcirc$ | $\bigcirc$ | $\begin{aligned} & \text { CCLR1 }=1 \\ & \text { CCLR0 }=1 \end{aligned}$ | $\bigcirc$ |
| Complementary PWM mode |  | O*3 | - | - | - | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=0 \end{aligned}$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | - | $\begin{aligned} & \text { CCLR1 }=0 \\ & \text { CCLR0 }=0 \end{aligned}$ | O*5 |
| Reset-synchronized PWM mode |  | $\bigcirc$ | - | - | - | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=1 \end{aligned}$ | $\begin{aligned} & \text { CMD1 }=1 \\ & \text { CMD0 }=1 \end{aligned}$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | - | - | ○*6 | O*6 |
| Buffering (BRA) |  | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | BFA4 = 1 Other bits unrestricted | - | - | ○*1 | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| Buffering (BRB) |  | $\bigcirc$ | - | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | BFB4 = 1 Other bits unrestricted | - | - | $\bigcirc * 1$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |

Legend: $\bigcirc$ Setting available (valid). - Setting does not affect this mode.
Notes: 1. Master enable bit settings are valid only during waveform output.
match A and compare match B ocour simultaneously the compare match signal is inhibited Do not set both channels 3 and 4 for synchronous operation when complementary PWM mode is selected.
4. When reset-synchronized PWM mode is selected, TCNT4 operates independently and the counter clearing function is available. Waveform output is not affected. . In complementary PWM mode, select the same clock source for channels 3 and 4 .
6. TCR4 settings are valid in reset-synchronized PWM mode, but TCNT4 operates independently, without affecting waveform output.

## Section 11 Programmable Timing Pattern Controller

### 11.1 Overview

The H8/3052F has a built-in programmable timing pattern controller (TPC) that provides pulse outputs by using the 16 -bit integrated timer unit (ITU) as a time base. The TPC pulse outputs are divided into 4-bit groups (group 3 to group 0) that can operate simultaneously and independently.

### 11.1.1 Features

TPC features are listed below.

- 16-bit output data

Maximum 16-bit data can be output. TPC output can be enabled on a bit-by-bit basis.

- Four output groups

Output trigger signals can be selected in 4-bit groups to provide up to four different 4-bit outputs.

- Selectable output trigger signals

Output trigger signals can be selected for each group from the compare-match signals of four ITU channels.

- Non-overlap mode

A non-overlap margin can be provided between pulse outputs.

- Can operate together with the DMA controller (DMAC)

The compare-match signals selected as trigger signals can activate the DMAC for sequential output of data without CPU intervention.

### 11.1.2 Block Diagram

Figure 11.1 shows a block diagram of the TPC.


Figure 11.1 TPC Block Diagram

### 11.1.3 Pin Configuration

Table 11.1 summarizes the TPC output pins.
Table 11.1 TPC Pins

| Name | Symbol | I/O | Function |
| :---: | :---: | :---: | :---: |
| TPC output 0 | TP 0 | Output | Group 0 pulse output |
| TPC output 1 | TP ${ }_{1}$ | Output |  |
| TPC output 2 | $\mathrm{TP}_{2}$ | Output |  |
| TPC output 3 | $\mathrm{TP}_{3}$ | Output |  |
| TPC output 4 | $\mathrm{TP}_{4}$ | Output | Group 1 pulse output |
| TPC output 5 | TP ${ }_{5}$ | Output |  |
| TPC output 6 | $\mathrm{TP}_{6}$ | Output |  |
| TPC output 7 | $\mathrm{TP}_{7}$ | Output |  |
| TPC output 8 | $\mathrm{TP}_{8}$ | Output | Group 2 pulse output |
| TPC output 9 | TP 9 | Output |  |
| TPC output 10 | TP ${ }_{10}$ | Output |  |
| TPC output 11 | TP ${ }_{11}$ | Output |  |
| TPC output 12 | TP ${ }_{12}$ | Output | Group 3 pulse output |
| TPC output 13 | $\mathrm{TP}_{13}$ | Output |  |
| TPC output 14 | TP ${ }_{14}$ | Output |  |
| TPC output 15 | TP ${ }_{15}$ | Output |  |

### 11.1.4 Register Configuration

Table 11.2 summarizes the TPC registers.
Table 11.2 TPC Registers

| Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: |
| H'FFD1 | Port A data direction register | PADDR | W | H'00 |
| H'FFD3 | Port A data register | PADR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'00 |
| H'FFD4 | Port B data direction register | PBDDR | W | H'00 |
| H'FFD6 | Port B data register | PBDR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'00 |
| H'FFA0 | TPC output mode register | TPMR | R/W | H'F0 |
| H'FFA1 | TPC output control register | TPCR | R/W | H'FF |
| H'FFA2 | Next data enable register B | NDERB | R/W | H'00 |
| H'FFA3 | Next data enable register A | NDERA | R/W | H'00 |
| H'FFA5/ $H^{\prime} F F A 7 *^{3}$ | Next data register A | NDRA | R/W | H'00 |
| H'FFA4 H'FFA6* ${ }^{3}$ | Next data register B | NDRB | R/W | H'OO |

Notes: 1. Lower 16 bits of the address.
2. Bits used for TPC output cannot be written.
3. The NDRA address is H'FFA5 when the same output trigger is selected for TPC output groups 0 and 1 by settings in TPCR. When the output triggers are different, the NDRA address is H'FFA7 for group 0 and H'FFA5 for group 1. Similarly, the address of NDRB is H'FFA4 when the same output trigger is selected for TPC output groups 2 and 3 by settings in TPCR. When the output triggers are different, the NDRB address is H'FFA6 for group 2 and H'FFA4 for group 3.

### 11.2 Register Descriptions

### 11.2.1 Port A Data Direction Register (PADDR)

PADDR is an 8-bit write-only register that selects input or output for each pin in port A.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{PA}_{7}$ DDR | $\mathrm{PA}_{6}$ DDR | PA ${ }_{5}$ DDR | $\mathrm{PA}_{4} \mathrm{DDR}$ | $\mathrm{PA}_{3}$ DDR | $\mathrm{PA}_{2}$ DDR | PA ${ }_{1}$ DDR | PA ${ }_{0}$ DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | W | W | W | W | W | W |
|  |  |  |  | Port A data direction 7 to 0 These bits select input or output for port A pins |  |  |  |  |

Port A is multiplexed with pins $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$. Bits corresponding to pins used for TPC output must be set to 1. For further information about PADDR, see section 9.11, Port A.

### 11.2.2 Port A Data Register (PADR)

PADR is an 8-bit readable/writable register that stores TPC output data for groups 0 and 1 , when these TPC output groups are used.


Note: * Bits selected for TPC output by NDERA settings become read-only bits.

For further information about PADR, see section 9.11, Port A.

### 11.2.3 Port B Data Direction Register (PBDDR)

PBDDR is an 8-bit write-only register that selects input or output for each pin in port B.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{PB}_{7}$ DDR | $\mathrm{PB}_{6}$ DDR | $\mathrm{PB}_{5}$ DDR | $\mathrm{PB}_{4}$ DDR | $\mathrm{PB}_{3}$ DDR | $\mathrm{PB}_{2}$ DDR | $\mathrm{PB}_{1}$ DDR | $\mathrm{PB}_{0}$ DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | W | W | W | W | W | W |
|  |  |  |  | Port B data direction 7 to 0 These bits select input or output for port B pins |  |  |  |  |

Port B is multiplexed with pins $\mathrm{TP}_{15}$ to $\mathrm{TP}_{8}$. Bits corresponding to pins used for TPC output must be set to 1 . For further information about PBDDR, see section 9.12, Port B.

### 11.2.4 Port B Data Register (PBDR)

PBDR is an 8-bit readable/writable register that stores TPC output data for groups 2 and 3, when these TPC output groups are used.


Note: * Bits selected for TPC output by NDERB settings become read-only bits.

For further information about PBDR, see section 9.12, Port B.

### 11.2.5 Next Data Register A (NDRA)

NDRA is an 8-bit readable/writable register that stores the next output data for TPC output groups 1 and 0 (pins $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$ ). During TPC output, when an ITU compare match event specified in TPCR occurs, NDRA contents are transferred to the corresponding bits in PADR. The address of NDRA differs depending on whether TPC output groups 0 and 1 have the same output trigger or different output triggers.

NDRA is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. It is not initialized in software standby mode.

Same Trigger for TPC Output Groups 0 and 1: If TPC output groups 0 and 1 are triggered by the same compare match event, the NDRA address is H'FFA5. The upper 4 bits belong to group 1 and the lower 4 bits to group 0 . Address H'FFA7 consists entirely of reserved bits that cannot be modified and are always read as 1 .

## Address H'FFA5

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR7 | NDR6 | NDR5 | NDR4 | NDR3 | NDR2 | NDR1 | NDR0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/w | R/W | R/W | R/W | R/W |
|  | These bits store the next output data for TPC output group 1 |  |  |  | These bits store the next output data for TPC output group 0 |  |  |  |

## Address H'FFA7

| 7 <br> Bit |  |  |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | - | - |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | - | - | - | - | - | - | - | - |

Different Triggers for TPC Output Groups 0 and 1: If TPC output groups 0 and 1 are triggered by different compare match events, the address of the upper 4 bits of NDRA (group 1) is H'FFA5 and the address of the lower 4 bits (group 0 ) is H'FFA7. Bits 3 to 0 of address H'FFA5 and bits 7 to 4 of address H'FFA7 are reserved bits that cannot be modified and are always read as 1 .

Address H'FFA5

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR7 | NDR6 | NDR5 | NDR4 | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | - | - | - | - |
|  |  | Next data 7 to 4 <br> These bits store the next output data for TPC output group 1 |  |  |  | Reserved bits |  |  |

Address H'FFA7

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | NDR3 | NDR2 | NDR1 | NDR0 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
|  | Reserved bits |  |  |  | Next data 3 to 0 <br> These bits store the next output data for TPC output group 0 |  |  |  |

### 11.2.6 Next Data Register B (NDRB)

NDRB is an 8-bit readable/writable register that stores the next output data for TPC output groups 3 and 2 (pins TP ${ }_{15}$ to $\mathrm{TP}_{8}$ ). During TPC output, when an ITU compare match event specified in TPCR occurs, NDRB contents are transferred to the corresponding bits in PBDR. The address of NDRB differs depending on whether TPC output groups 2 and 3 have the same output trigger or different output triggers.

NDRB is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. It is not initialized in software standby mode.

Same Trigger for TPC Output Groups 2 and 3: If TPC output groups 2 and 3 are triggered by the same compare match event, the NDRB address is H'FFA4. The upper 4 bits belong to group 3 and the lower 4 bits to group 2 . Address H'FFA6 consists entirely of reserved bits that cannot be modified and are always read as 1 .

Address H'FFA4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR15 | NDR14 | NDR13 | NDR12 | NDR11 | NDR10 | NDR9 | NDR8 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/w | R/w | R/W | R/W | R/W |
| These bits store the next output |  |  |  |  | These bits store the next output data for TPC output group 2 |  |  |  |

Address H'FFA6

| 7 <br> Bit |  |  |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | - | - |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | - | - | - | - | - | - | - | - |

Different Triggers for TPC Output Groups 2 and 3: If TPC output groups 2 and 3 are triggered by different compare match events, the address of the upper 4 bits of NDRB (group 3) is H'FFA4 and the address of the lower 4 bits (group 2) is H'FFA6. Bits 3 to 0 of address H'FFA4 and bits 7 to 4 of address H'FFA6 are reserved bits that cannot be modified and are always read as 1 .

Address H'FFA4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR15 | NDR14 | NDR13 | NDR12 | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | - | - | - | - |
|  |  | Next data 15 to 12 These bits store the next output data for TPC output group 3 |  |  |  | Reserved bits |  |  |

Address H'FFA6


### 11.2.7 Next Data Enable Register A (NDERA)

NDERA is an 8-bit readable/writable register that enables or disables TPC output groups 1 and 0 ( $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$ ) on a bit-by-bit basis.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDER7 | NDER6 | NDER5 | NDER4 | NDER3 | NDER2 | NDER1 | NDERO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  | Next data enable 7 to 0 These bits enable or disable TPC output groups 1 and 0 |  |  |  |  |

If a bit is enabled for TPC output by NDERA, then when the ITU compare match event selected in the TPC output control register (TPCR) occurs, the NDRA value is automatically transferred to the corresponding PADR bit, updating the output value. If TPC output is disabled, the bit value is not transferred from NDRA to PADR and the output value does not change.

NDERA is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 0—Next Data Enable 7 to 0 (NDER7 to NDER0): These bits enable or disable TPC output groups 1 and $0\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{0}\right)$ on a bit-by-bit basis.

Bits 7 to 0 :
NDER7 to NDER0

## Description

0
TPC outputs $\mathrm{TP}_{7}$ to $\mathrm{TP}_{n}$ are disabled
(Initial value)
(NDR7 to NDR0 are not transferred to $\mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$ )
1
TPC outputs $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$ are enabled
(NDR7 to NDR0 are transferred to $\mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$ )

### 11.2.8 Next Data Enable Register B (NDERB)

NDERB is an 8-bit readable/writable register that enables or disables TPC output groups 3 and 2 ( $\mathrm{TP}_{15}$ to $\mathrm{TP}_{8}$ ) on a bit-by-bit basis.


If a bit is enabled for TPC output by NDERB, then when the ITU compare match event selected in the TPC output control register (TPCR) occurs, the NDRB value is automatically transferred to the corresponding PBDR bit, updating the output value. If TPC output is disabled, the bit value is not transferred from NDRB to PBDR and the output value does not change.

NDERB is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 0—Next Data Enable 15 to 8 (NDER15 to NDER8): These bits enable or disable TPC output groups 3 and $2\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{8}\right)$ on a bit-by-bit basis.

| Bits 7 to 0: <br> NDER15 to NDER8 | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | TPC outputs $\mathrm{TP}_{15}$ to $\mathrm{TP}_{8}$ are disabled |  |
|  | (NDR15 to NDR8 are not transferred to $\mathrm{PB}_{7}$ to $\mathrm{PB}_{0}$ ) |  |
| 1 | ${\mathrm{TPC} \mathrm{outputs} \mathrm{TP}_{15} \text { to } \mathrm{TP}_{8} \text { are enabled }}$ (NDR15 to NDR8 are transferred to $\mathrm{PB}_{7}$ to $\mathrm{PB}_{0}$ ) |  |

### 11.2.9 TPC Output Control Register (TPCR)

TPCR is an 8-bit readable/writable register that selects output trigger signals for TPC outputs on a group-by-group basis.


TPCR is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 and 6-Group 3 Compare Match Select 1 and 0 (G3CMS1, G3CMS0): These bits select the compare match event that triggers TPC output group $3\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$.

| Bit 7: G3CMS1 | Bit 6: G3CMS0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | TPC output group $3\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare <br> match in ITU channel 0 |
|  | 1 | TPC output group $3\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare <br> match in ITU channel 1 |
| 1 | 0 | TPC output group $3\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare <br> match in ITU channel 2 |
|  | TPC output group 3 $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare <br> match in ITU channel 3 |  |

Bits 5 and 4-Group 2 Compare Match Select 1 and 0 (G2CMS1, G2CMS0): These bits select the compare match event that triggers TPC output group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$.

| Bit 5: G2CMS1 | Bit 4: G2CMS0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | TPC output group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare <br> match in ITU channel 0 |
|  | 1 | TPC output group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare <br> match in ITU channel 1 |
| 1 | 0 | TPC output group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare <br> match in ITU channel 2 |
|  | TPC output group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare <br> match in ITU channel 3 |  |

Bits 3 and 2-Group 1 Compare Match Select 1 and 0 (G1CMS1, G1CMS0): These bits select the compare match event that triggers TPC output group $1\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$.

| Bit 3: G1CMS1 | Bit 2: G1CMS0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | TPC output group 1 $\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare <br> match in ITU channel 0 |
|  | 1 | TPC output group 1 $\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare <br> match in ITU channel 1 |
| 1 | 0 | TPC output group 1 $\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare <br> match in ITU channel 2 |
|  | TPC output group 1 $\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare <br> match in ITU channel 3 |  |

Bits 1 and 0-Group 0 Compare Match Select 1 and 0 (G0CMS1, G0CMSO): These bits select the compare match event that triggers TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$.

| Bit 1: GOCMS1 | Bit 0: GOCMS0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare <br> match in ITU channel 0 |
|  | 1 | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare <br> match in ITU channel 1 |
| 1 | 0 | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare <br> match in ITU channel 2 |
|  | TPC output group 0 $\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare <br> match in ITU channel 3 |  |

Rev. 2.0, 03/01, page 414 of 822

### 11.2.10 TPC Output Mode Register (TPMR)

TPMR is an 8-bit readable/writable register that selects normal or non-overlapping TPC output for each group.


The output trigger period of a non-overlapping TPC output waveform is set in general register B (GRB) in the ITU channel selected for output triggering. The non-overlap margin is set in general register A (GRA). The output values change at compare match A and B. For details see section 11.3.4, Non-Overlapping TPC Output.

TPMR is initialized to H'F0 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 4—Reserved: Read-only bits, always read as 1 .

Bit 3-Group 3 Non-Overlap (G3NOV): Selects normal or non-overlapping TPC output for group $3\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$.

Bit 3: G3NOV Description
$0 \quad$ Normal TPC output in group 3 (output values change at compare match A in the selected ITU channel)
(Initial value)
1
Non-overlapping TPC output in group 3 (independent 1 and 0 output at compare match A and B in the selected ITU channel)

Bit 2-Group 2 Non-Overlap (G2NOV): Selects normal or non-overlapping TPC output for group $2\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$.

Bit 2: G2NOV Description

| 0 | Normal TPC output in group 2 (output values change at compare match A in <br> the selected ITU channel) |
| :--- | :--- |
| 1 | Non-overlapping TPC output in group 2 (independent value) 1 and 0 output at <br> compare match A and B in the selected ITU channel) |

Bit 1—Group 1 Non-Overlap (G1NOV): Selects normal or non-overlapping TPC output for group 1 ( $\mathrm{TP}_{7}$ to $\mathrm{TP}_{4}$ ).

| Bit 1: G1NOV | Description |
| :--- | :--- |
| 0 | Normal TPC output in group 1 (output values change at compare match A in <br> the selected ITU channel) <br> (Initial value) |
| 1 | Non-overlapping TPC output in group 1 (independent 1 and 0 output at <br> compare match A and B in the selected ITU channel) |

Bit 0—Group 0 Non-Overlap (G0NOV): Selects normal or non-overlapping TPC output for group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$.

| Bit 0: GONOV | Description |
| :--- | :--- |
| 0 | Normal TPC output in group 0 (output values change at compare match A in <br> the selected ITU channel) <br> (Initial value) |
| 1 | Non-overlapping TPC output in group 0 (independent 1 and 0 output at <br> compare match A and B in the selected ITU channel) |

Rev. 2.0, 03/01, page 416 of 822

### 11.3 Operation

### 11.3.1 Overview

When corresponding bits in PADDR or PBDDR and NDERA or NDERB are set to 1, TPC output is enabled. The TPC output initially consists of the corresponding PADR or PBDR contents. When a compare-match event selected in TPCR occurs, the corresponding NDRA or NDRB bit contents are transferred to PADR or PBDR to update the output values.

Figure 11.2 illustrates the TPC output operation. Table 11.3 summarizes the TPC operating conditions.


Figure 11.2 TPC Output Operation
Table 11.3 TPC Operating Conditions

| NDER | DDR | Pin Function |
| :--- | :--- | :--- |
| 0 | 0 | Generic input port |
|  | 1 | Generic output port |
| 1 | 0 | Generic input port (but the DR bit is a read-only bit, and when compare <br> match occurs, the NDR bit value is transferred to the DR bit) |
|  | 1 | TPC pulse output |

Sequential output of up to 16-bit patterns is possible by writing new output data to NDRA and NDRB before the next compare match. For information on non-overlapping operation, see section 11.3.4, Non-Overlapping TPC Output.

### 11.3.2 Output Timing

If TPC output is enabled, NDRA/NDRB contents are transferred to PADR/PBDR and output when the selected compare match event occurs. Figure 11.3 shows the timing of these operations for the case of normal output in groups 2 and 3, triggered by compare match A .


Figure 11.3 Timing of Transfer of Next Data Register Contents and Output (Example)

### 11.3.3 Normal TPC Output

Sample Setup Procedure for Normal TPC Output: Figure 11.4 shows a sample procedure for setting up normal TPC output.


Figure 11.4 Setup Procedure for Normal TPC Output (Example)

Example of Normal TPC Output (Example of Five-Phase Pulse Output): Figure 11.5 shows an example in which the TPC is used for cyclic five-phase pulse output.


1. The ITU channel to be used as the output trigger channel is set up so that GRA is an output compare register and the counter will be cleared by compare match A. The trigger period is set in GRA. The IMIEA bit is set to 1 in TIER to enable the compare match A interrupt.
2. H'F8 is written in PBDDR and NDERB, and bits G3CMS1, G3CMS0, G2CMS1, and G2CMS0 are set in TPCR to select compare match in the ITU channel set up in step 1 as the output trigger. Output data $\mathrm{H}^{\prime} 80$ is written in NDRB.
3. The timer counter in this ITU channel is started. When compare match A occurs, the NDRB contents are transferred to PBDR and output. The compare match/input capture A (IMFA) interrupt service routine writes the next output data ( $\mathrm{H}^{\prime} \mathrm{CO}$ ) in NDRB.
4. Five-phase overlapping pulse output (one or two phases active at a time) can be obtained by writing $\mathrm{H}^{\prime} 40$, $\mathrm{H}^{\prime} 60, \mathrm{H}^{\prime} 20, \mathrm{H}^{\prime} 30, \mathrm{H}^{\prime} 10, \mathrm{H}^{\prime} 18, \mathrm{H}^{\prime} 08, \mathrm{H}^{\prime} 88 \ldots$ at successive IMFA interrupts. If the DMAC is set for activation by this interrupt, pulse output can be obtained without loading the CPU.

Figure 11.5 Normal TPC Output Example (Five-Phase Pulse Output)

### 11.3.4 Non-Overlapping TPC Output

Sample Setup Procedure for Non-Overlapping TPC Output: Figure 11.6 shows a sample procedure for setting up non-overlapping TPC output.


Figure 11.6 Setup Procedure for Non-Overlapping TPC Output (Example)

Example of Non-Overlapping TPC Output (Example of Four-Phase Complementary NonOverlapping Output): Figure 11.7 shows an example of the use of TPC output for four-phase complementary non-overlapping pulse output.


Figure 11.7 Non-Overlapping TPC Output Example (Four-Phase Complementary Non-Overlapping Pulse Output)

This operation example is described below.

- The output trigger ITU channel is set up so that GRA and GRB are output compare registers and the counter will be cleared by compare match B. The TPC output trigger period is set in GRB. The non-overlap margin is set in GRA. The IMIEA bit is set to 1 in TIER to enable IMFA interrupts.
- H'FF is written in PBDDR and NDERB, and bits G3CMS1, G3CMS0, G2CMS1, and G2CMS0 are set in TPCR to select compare match in the ITU channel set up in step 1 as the output trigger. Bits G3NOV and G2NOV are set to 1 in TPMR to select non-overlapping output. Output data H'95 is written in NDRB.
Rev. 2.0, 03/01, page 422 of 822
- The timer counter in this ITU channel is started. When compare match B occurs, outputs change from 1 to 0 . When compare match A occurs, outputs change from 0 to 1 (the change from 0 to 1 is delayed by the value of GRA). The IMFA interrupt service routine writes the next output data (H'65) in NDRB.
- Four-phase complementary non-overlapping pulse output can be obtained by writing H'59, H'56, H'95... at successive IMFA interrupts. If the DMAC is set for activation by this interrupt, pulse output can be obtained without loading the CPU.


### 11.3.5 TPC Output Triggering by Input Capture

TPC output can be triggered by ITU input capture as well as by compare match. If GRA, and GRB functions as an input capture register in the ITU channel selected in TPCR, TPC output will be triggered by the input capture signal. Figure 11.8 shows the timing.


Figure 11.8 TPC Output Triggering by Input Capture (Example)

### 11.4 Usage Notes

### 11.4.1 Operation of TPC Output Pins

$\mathrm{TP}_{0}$ to $\mathrm{TP}_{15}$ are multiplexed with ITU, DMAC, address bus, and other pin functions. When ITU, DMAC, or address output is enabled, the corresponding pins cannot be used for TPC output. The data transfer from NDR bits to DR bits takes place, however, regardless of the usage of the pin.

Pin functions should be changed only under conditions in which the output trigger event will not occur.

### 11.4.2 Note on Non-Overlapping Output

During non-overlapping operation, the transfer of NDR bit values to DR bits takes place as follows.

1. NDR bits are always transferred to DR bits at compare match A.
2. At compare match $B, N D R$ bits are transferred only if their value is 0 . Bits are not transferred if their value is 1 .

Figure 11.9 illustrates the non-overlapping TPC output operation.


Figure 11.9 Non-Overlapping TPC Output

Therefore, 0 data can be transferred ahead of 1 data by making compare match B occur before compare match A. NDR contents should not be altered during the interval from compare match B to compare match A (the non-overlap margin).

This can be accomplished by having the IMFA interrupt service routine write the next data in NDR, or by having the IMFA interrupt activate the DMAC. The next data must be written before the next compare match $B$ occurs.

Figure 11.10 shows the timing relationships.


Figure 11.10 Non-Overlapping Operation and NDR Write Timing

## Section 12 Watchdog Timer

### 12.1 Overview

The H8/3052F has an on-chip watchdog timer (WDT). The WDT has two selectable functions: it can operate as a watchdog timer to supervise system operation, or it can operate as an interval timer. As a watchdog timer, it generates a reset signal for the chip if a system crash allows the timer counter (TCNT) to overflow before being rewritten. In interval timer operation, an interval timer interrupt is requested at each TCNT overflow.

### 12.1.1 Features

WDT features are listed below.

- Selection of eight counter clock sources
$\phi / 2, \phi / 32, \phi / 64, \phi / 128, \phi / 256, \phi / 512, \phi / 2048$, or $\phi / 4096$
- Interval timer option
- Timer counter overflow generates a reset signal or interrupt.

The reset signal is generated in watchdog timer operation. An interval timer interrupt is generated in interval timer operation.

- Watchdog timer reset signal resets the entire chip internally.

The reset signal generated by timer counter overflow during watchdog timer operation resets the entire chip internally.
With the $\mathrm{H} 8 / 3052 \mathrm{~F}$, a reset signal cannot be output externally.

### 12.1.2 Block Diagram

Figure 12.1 shows a block diagram of the WDT.


Figure 12.1 WDT Block Diagram

### 12.1.3 Register Configuration

Table 12.1 summarizes the WDT registers.
Table 12.1 WDT Registers

| Address*1 |  | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Write*2 | Read |  |  |  |  |
| H'FFA8 | H'FFA8 | Timer control/status register | TCSR | $\mathrm{R} /(\mathrm{W})$ * $^{3}$ | H'18 |
|  | H'FFA9 | Timer counter | TCNT | R/W | H'00 |
| H'FFAA | H'FFAB | Reset control/status register | RSTCSR | $\mathrm{R} /(\mathrm{W}){ }^{3}$ | H'3F |

Notes: 1. Lower 16 bits of the address.
2. Write word data starting at this address.
3. Only 0 can be written in bit 7 , to clear the flag.

### 12.2 Register Descriptions

### 12.2.1 Timer Counter (TCNT)

TCNT is an 8-bit readable and writable* up-counter.


When the TME bit is set to 1 in TCSR, TCNT starts counting pulses generated from an internal clock source selected by bits CKS2 to CKS0 in TCSR. When the count overflows (changes from $H^{\prime} F F$ to $\mathrm{H}^{\prime} 00$ ), the OVF bit is set to $1 \mathrm{in} \mathrm{TCSR} .\mathrm{TCNT} \mathrm{is} \mathrm{initialized} \mathrm{to} \mathrm{H}^{\prime} 00$ by a reset and when the TME bit is cleared to 0 .

Note: * TCNT is write-protected by a password. For details see section 12.2.4, Notes on Register Access.

### 12.2.2 Timer Control/Status Register (TCSR)

TCSR is an 8-bit readable and writable ${ }^{* 1}$ register. Its functions include selecting the timer mode and clock source.


Bits 7 to 5 are initialized to 0 by a reset and in standby mode. Bits 2 to 0 are initialized to 0 by a reset. In software standby mode bits 2 to 0 are not initialized, but retain their previous values.

Notes: 1. TCSR differs from other registers in being more difficult to write. For details see section 12.2.4, Notes on Register Access.
2. Only 0 can be written, to clear the flag.

Bit 7—Overflow Flag (OVF): This status flag indicates that the timer counter has overflowed from H'FF to H'00.

| Bit 7: OVF | Description |  |
| :--- | :--- | :--- |
| 0 | [Clearing condition] |  |
|  | Cleared by reading OVF when OVF = 1, then writing 0 in OVF | (Initial value) |
| 1 | [Setting condition] |  |
|  | Set when TCNT changes from H'FF to H'00 |  |

Bit 6-Timer Mode Select (WT/IT): Selects whether to use the WDT as a watchdog timer or interval timer. If used as an interval timer, the WDT generates an interval timer interrupt request when TCNT overflows. If used as a watchdog timer, the WDT generates a reset signal when TCNT overflows.

| Bit 6: WT/IT | Description |  |
| :--- | :--- | ---: |
| 0 | Interval timer: requests interval timer interrupts | (Initial value) |
| 1 | Watchdog timer: generates a reset signal |  |

Bit 5—Timer Enable (TME): Selects whether TCNT runs or is halted.
When $\mathrm{WT} / \overline{\mathrm{IT}}=1$, clear the SYSCR software standby bit (SSBY) to 0 , then set the TME to 1 . When SSBY is set to 1 , clear TME to 0 .

| Bit 5: TME | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | TCNT is initialized to H'00 and halted |  |
| 1 | TCNT is counting and CPU interrupt requests are enabled |  |

Bits 4 and 3-Reserved: Read-only bits, always read as 1.

Bits 2 to 0—Clock Select 2 to 0 (CKS2/1/0): These bits select one of eight internal clock sources, obtained by prescaling the system clock $(\phi)$, for input to TCNT.

| Bit 2: CKS2 | Bit 1: CKS1 | Bit 0: CKSO | Description |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | ¢/2 (Initial value) |
|  |  | 1 | ¢/32 |
|  | 1 | 0 | ¢/64 |
|  |  | 1 | \$/128 |
| 1 | 0 | 0 | \$/256 |
|  |  | 1 | \$/512 |
|  | 1 | 0 | ¢/2048 |
|  |  | 1 | \$/4096 |

### 12.2.3 Reset Control/Status Register (RSTCSR)

RSTCSR is an 8-bit readable/writable* ${ }^{1}$ register that monitors the state of the reset signal generated by watchdog timer overflow.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | WRST | - | - | - | - | - | - | - |
| Initial value | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | $\mathrm{R} /(\mathrm{W})^{* 2}$ | - | - | - | - | - | - | - |
|  |  |  |  |  | Reserved bits |  |  |  |
|  | Reserved bit Must not be set to $1^{* 3}$ |  |  |  |  |  |  |  |

## Watchdog timer reset

Indicates that a reset signal has been generated

Bit 7 is initialized by input of a reset signal at the $\overline{\mathrm{RES}}$ pin. It is not initialized by reset signals generated by watchdog timer overflow.

Notes: 1. RSTCSR differs from other registers in being more difficult to write. For details see section 12.2.4, Notes on Register Access.
2. Only 0 can be written in bit 7 , to clear the flag.
3. Do not set bit 6 to 1 .

Bit 7—Watchdog Timer Reset (WRST): During watchdog timer operation, this bit indicates that TCNT has overflowed and generated a reset signal. This reset signal resets the entire chip internally.

| Bit 7: WRST | Description |
| :--- | :--- |
| 0 | [Clearing conditions] |
|  | Cleared to 0 by reset signal input at $\overline{\text { RES }}$ pin <br>  <br> Cleared by reading WRST when WRST $=1$, then writing 0 in WRST |
| 1 | [Setting condition] <br> Set when TCNT overflow generates a reset signal during watchdog timer <br> operation |

Bit 6—Reserved: Do not set to 1 .

Bits 5 to 0—Reserved: Read-only bits, always read as 1.

Rev. 2.0, 03/01, page 432 of 822

### 12.2.4 Notes on Register Access

The watchdog timer's TCNT, TCSR, and RSTCSR registers differ from other registers in being more difficult to write. The procedures for writing and reading these registers are given below.

Writing to TCNT and TCSR: These registers must be written by a word transfer instruction. They cannot be written by byte instructions. Figure 12.2 shows the format of data written to TCNT and TCSR. TCNT and TCSR both have the same write address. The write data must be contained in the lower byte of the written word. The upper byte must contain H'5A (password for TCNT) or H'A5 (password for TCSR). This transfers the write data from the lower byte to TCNT or TCSR.


Figure 12.2 Format of Data Written to TCNT and TCSR
Writing to RSTCSR: RSTCSR must be written by a word transfer instruction. It cannot be written by byte transfer instructions. Figure 12.3 shows the format of data written to RSTCSR. To write 0 in the WRST bit, the write data must have H'A5 in the upper byte and $\mathrm{H}^{\prime} 00$ in the lower byte. The $\mathrm{H}^{\prime} 00$ in the lower byte clears the WRST bit in RSTCSR to 0 .


Figure 12.3 Format of Data Written to RSTCSR
Reading TCNT, TCSR, and RSTCSR: These registers are read like other registers. Byte access instructions can be used. The read addresses are H'FFA8 for TCSR, H'FFA9 for TCNT, and H'FFAB for RSTCSR, as listed in table 12.2.

Table 12.2 Read Addresses of TCNT, TCSR, and RSTCSR

| Address* | Register |
| :--- | :--- |
| H'FFA8 | TCSR |
| H'FFA9 | TCNT |
| H'FFAB | RSTCSR |

Note: * Lower 16 bits of the address.

### 12.3 Operation

Operations when the WDT is used as a watchdog timer and as an interval timer are described below.

### 12.3.1 Watchdog Timer Operation

Figure 12.4 illustrates watchdog timer operation. To use the WDT as a watchdog timer, set the WT/IT and TME bits to 1 in TCSR. Software must prevent TCNT overflow by rewriting the TCNT value (normally by writing H'00) before overflow occurs. If TCNT fails to be rewritten and overflows due to a system crash etc., the chip is internally reset for a duration of 518 states.

A reset generated by the WDT has the same vector as a reset generated by input at the $\overline{\mathrm{RES}}$ pin. Software can distinguish a $\overline{\mathrm{RES}}$ reset from a watchdog reset by checking the WRST bit in RSTCSR.

If a $\overline{\mathrm{RES}}$ reset and a watchdog reset occur simultaneously, the $\overline{\mathrm{RES}}$ reset takes priority.


Figure 12.4 Watchdog Timer Operation

### 12.3.2 Interval Timer Operation

Figure 12.5 illustrates interval timer operation. To use the WDT as an interval timer, clear bit $\mathrm{WT} / \overline{\mathrm{IT}}$ to 0 and set bit TME to 1 in TCSR. An interval timer interrupt request is generated at each TCNT overflow. This function can be used to generate interval timer interrupts at regular intervals.


Figure 12.5 Interval Timer Operation

### 12.3.3 Timing of Setting of Overflow Flag (OVF)

Figure 12.6 shows the timing of setting of the OVF flag in TCSR. The OVF flag is set to 1 when TCNT overflows. At the same time, a reset signal is generated in watchdog timer operation, or an interval timer interrupt is generated in interval timer operation.


Figure 12.6 Timing of Setting of OVF

### 12.3.4 Timing of Setting of Watchdog Timer Reset Bit (WRST)

The WRST bit in RSTCSR is valid when bits WT/痛 and TME are both set to 1 in TCSR.
Figure 12.7 shows the timing of setting of WRST and the internal reset timing. The WRST bit is set to 1 when TCNT overflows and OVF is set to 1 . At the same time an internal reset signal is generated for the entire chip. This internal reset signal clears OVF to 0 , but the WRST bit remains set to 1 . The reset routine must therefore clear the WRST bit.


Figure 12.7 Timing of Setting of WRST Bit and Internal Reset

### 12.4 Interrupts

During interval timer operation, an overflow generates an interval timer interrupt (WOVI). The interval timer interrupt is requested whenever the OVF bit is set to 1 in TCSR.

### 12.5 Usage Notes

Contention between TCNT Write and Increment: If a timer counter clock pulse is generated during the $\mathrm{T}_{3}$ state of a write cycle to TCNT, the write takes priority and the timer count is not incremented. See figure 12.8.


Figure 12.8 Contention between TCNT Write and Increment
Changing CKS2 to CKS0 Values: Halt TCNT by clearing the TME bit to 0 in TCSR before changing the values of bits CKS2 to CKS0.

## Section 13 Serial Communication Interface

### 13.1 Overview

The H8/3052F has a serial communication interface (SCI) with two independent channels. The two channels are functionally identical. The SCI can communicate in asynchronous or synchronous mode. It also has a multiprocessor communication function for serial communication among two or more processors.

When the SCI is not used, it can be halted to conserve power. Each SCI channel can be halted independently. For details see section 20.6, Module Standby Function.

Channel 0 (SCI0) also has a smart card interface function conforming to the ISO/IEC7816-3 (Identification Card) standard. This function supports serial communication with a smart card. For details, see section 14, Smart Card Interface.

### 13.1.1 Features

SCI features are listed below.

- Selection of asynchronous or synchronous mode for serial communication
- Asynchronous mode

Serial data communication is synchronized one character at a time. The SCI can communicate with a universal asynchronous receiver/transmitter (UART), asynchronous communication interface adapter (ACIA), or other chip that employs standard asynchronous serial communication. It can also communicate with two or more other processors using the multiprocessor communication function. There are twelve selectable serial data communication formats.

- Data length: 7 or 8 bits
- Stop bit length: 1 or 2 bits
- Parity bit: even, odd, or none
- Multiprocessor bit: 1 or 0
- Receive error detection: parity, overrun, and framing errors
- Break detection: by reading the RxD level directly when a framing error occurs
- Synchronous mode

Serial data communication is synchronized with a clock signal. The SCI can communicate with other chips having a synchronous communication function. There is one serial data communication format.

- Data length: 8 bits
- Receive error detection: overrun errors
- Full duplex communication

The transmitting and receiving sections are independent, so the SCI can transmit and receive simultaneously. The transmitting and receiving sections are both double-buffered, so serial data can be transmitted and received continuously.

- Built-in baud rate generator with selectable bit rates
- Selectable transmit/receive clock sources: internal clock from baud rate generator, or external clock from the SCK pin.
- Four types of interrupts

Transmit-data-empty, transmit-end, receive-data-full, and receive-error interrupts are requested independently. The transmit-data-empty and receive-data-full interrupts from SCI0 can activate the DMA controller (DMAC) to transfer data.

### 13.1.2 Block Diagram

Figure 13.1 shows a block diagram of the SCI.


Figure 13.1 SCI Block Diagram

### 13.1.3 Pin Configuration

The SCI has serial pins for each channel as listed in table 13.1.
Table 13.1 SCI Pins

| Channel | Name | Abbreviation | $\mathbf{I} / \mathbf{O}$ | Function |
| :--- | :--- | :--- | :--- | :--- |
| 0 | Serial clock pin | $\mathrm{SCK}_{0}$ | Input/output | $\mathrm{SCI}_{0}$ clock input/output |
|  | Receive data pin | $\mathrm{RxD}_{0}$ | Input | $\mathrm{SCI}_{0}$ receive data input |
|  | Transmit data pin | $\mathrm{TxD}_{0}$ | Output | $\mathrm{SCI}_{0}$ transmit data output |
| 1 | Serial clock pin | $\mathrm{SCK}_{1}$ | Input/output | $\mathrm{SCI}_{1}$ clock input/output |
|  | Receive data pin | $\mathrm{RxD}_{1}$ | Input | $\mathrm{SCI}_{1}$ receive data input |
|  | Transmit data pin | $\mathrm{TxD}_{1}$ | Output | $\mathrm{SCl}_{1}$ transmit data output |

### 13.1.4 Register Configuration

The SCI has internal registers as listed in table 13.2. These registers select asynchronous or synchronous mode, specify the data format and bit rate, and control the transmitter and receiver sections.

Table 13.2 Registers

| Channel | Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | H'FFB0 | Serial mode register | SMR | R/W | H'00 |
|  | H'FFB1 | Bit rate register | BRR | R/W | H'FF |
|  | H'FFB2 | Serial control register | SCR | R/W | H'00 |
|  | H'FFB3 | Transmit data register | TDR | R/W | H'FF |
|  | H'FFB4 | Serial status register | SSR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'84 |
|  | H'FFB5 | Receive data register | RDR | R | H'00 |
| 1 | H'FFB8 | Serial mode register | SMR | R/W | H'00 |
|  | H'FFB9 | Bit rate register | BRR | R/W | H'FF |
|  | H'FFBA | Serial control register | SCR | R/W | H'00 |
|  | H'FFBB | Transmit data register | TDR | R/W | H'FF |
|  | H'FFBC | Serial status register | SSR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'84 |
|  | H'FFBD | Receive data register | RDR | R | H'00 |

Notes: 1. Lower 16 bits of the address.
2. Only 0 can be written, to clear flags.

### 13.2 Register Descriptions

### 13.2.1 Receive Shift Register (RSR)

RSR is the register that receives serial data.

Bit


Read/Write

The SCI loads serial data input at the RxD pin into RSR in the order received, LSB (bit 0) first, thereby converting the data to parallel data. When 1 byte has been received, it is automatically transferred to RDR. The CPU cannot read or write RSR directly.

### 13.2.2 Receive Data Register (RDR)

RDR is the register that stores received serial data.


When the SCI finishes receiving 1 byte of serial data, it transfers the received data from RSR into RDR for storage. RSR is then ready to receive the next data. This double buffering allows data to be received continuously.

RDR is a read-only register. Its contents cannot be modified by the CPU. RDR is initialized to H'00 by a reset and in standby mode.

### 13.2.3 Transmit Shift Register (TSR)

TSR is the register that transmits serial data.

Bit


Read/Write

The SCI loads transmit data from TDR into TSR, then transmits the data serially from the TxD pin, LSB (bit 0) first. After transmitting one data byte, the SCI automatically loads the next transmit data from TDR into TSR and starts transmitting it. If the TDRE flag is set to 1 in SSR, however, the SCI does not load the TDR contents into TSR. The CPU cannot read or write TSR directly.

### 13.2.4 Transmit Data Register (TDR)

TDR is an 8-bit register that stores data for serial transmission.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

When the SCI detects that TSR is empty, it moves transmit data written in TDR from TDR into TSR and starts serial transmission. Continuous serial transmission is possible by writing the next transmit data in TDR during serial transmission from TSR.

The CPU can always read and write TDR. TDR is initialized to H'FF by a reset and in standby mode.

### 13.2.5 Serial Mode Register (SMR)

SMR is an 8-bit register that specifies the SCI serial communication format and selects the clock source for the baud rate generator.


## Communication mode

Selects asynchronous or synchronous mode

The CPU can always read and write SMR. SMR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.

Bit 7—Communication Mode (C/ $\overline{\mathrm{A}})$ : Selects whether the SCI operates in asynchronous or synchronous mode.

Bit 7: C/ $\overline{\mathrm{A}}$

## Description

| 0 | Asynchronous mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | Synchronous mode |  |

Bit 6-Character Length (CHR): Selects 7-bit or 8-bit data length in asynchronous mode. In synchronous mode the data length is 8 bits regardless of the CHR setting.

Bit 6: CHR Description

| 0 | 8-bit data | (Initial value) |
| :--- | :--- | :--- |
| 1 | 7-bit data* |  |

Note: * When 7-bit data is selected, the MSB (bit 7) in TDR is not transmitted.

Bit 5—Parity Enable (PE): In asynchronous mode, this bit enables or disables the addition of a parity bit to transmit data, and the checking of the parity bit in receive data. In synchronous mode the parity bit is neither added nor checked, regardless of the PE setting.

## Bit 5: PE Description

| 0 | Parity bit not added or checked | (Initial value) |
| :--- | :--- | ---: |
| 1 | Parity bit added and checked* |  |
| Note: * When PE is set to 1, an even or odd parity bit is added to transmit data according to the |  |  |
| even or odd parity mode selected by the O/ $\overline{\mathrm{E}}$ bit, and the parity bit in receive data is <br> checked to see that it matches the even or odd mode selected by the $\mathrm{O} / \overline{\mathrm{E}}$ bit. |  |  |

Bit 4—Parity Mode ( $\mathbf{O} / \overline{\mathrm{E}}$ ): Selects even or odd parity. The $\mathrm{O} / \overline{\mathrm{E}}$ bit setting is valid in asynchronous mode when the PE bit is set to 1 to enable the adding and checking of a parity bit. The $\mathrm{O} / \overline{\mathrm{E}}$ setting is ignored in synchronous mode, or when parity adding and checking is disabled in asynchronous mode.

## Bit 4: O/E Description

| 0 | Even parity*1 | (Initial value) |
| :--- | :--- | :--- |
| 1 | Odd parity*2$^{*}$ |  |

Notes: 1. When even parity is selected, the parity bit added to transmit data makes an even number of 1 s in the transmitted character and parity bit combined. Receive data must have an even number of 1 s in the received character and parity bit combined.
2. When odd parity is selected, the parity bit added to transmit data makes an odd number of 1 s in the transmitted character and parity bit combined. Receive data must have an odd number of 1 s in the received character and parity bit combined.

Bit 3—Stop Bit Length (STOP): Selects one or two stop bits in asynchronous mode. This setting is used only in asynchronous mode. In synchronous mode no stop bit is added, so the STOP bit setting is ignored.

Bit 3: STOP Description

| 0 | One stop bit*1 | (Initial value) |
| :--- | :--- | :--- |
| 1 | Two stop bits*2 |  |

Notes: 1. One stop bit (with value 1) is added at the end of each transmitted character. 2. Two stop bits (with value 1) are added at the end of each transmitted character.

In receiving, only the first stop bit is checked, regardless of the STOP bit setting. If the second stop bit is 1 it is treated as a stop bit. If the second stop bit is 0 it is treated as the start bit of the next incoming character.

Bit 2—Multiprocessor Mode (MP): Selects a multiprocessor format. When a multiprocessor format is selected, parity settings made by the PE and $\mathrm{O} / \overline{\mathrm{E}}$ bits are ignored. The MP bit setting is valid only in asynchronous mode. It is ignored in synchronous mode.

For further information on the multiprocessor communication function, see section 13.3.3, Multiprocessor Communication.

| Bit 2: MP | Description |  |
| :--- | :--- | :--- |
| 0 | Multiprocessor function disabled | (Initial value) |
| 1 | Multiprocessor format selected |  |

Bits 1 and 0—Clock Select 1 and 0 (CKS1/0): These bits select the clock source of the on-chip baud rate generator. Four clock sources are available: $\phi, \phi / 4, \phi / 16$, and $\phi / 64$.

For the relationship between the clock source, bit rate register setting, and baud rate, see section 13.2.8, Bit Rate Register (BRR).

| Bit 1: CKS1 | Bit 0: CKS0 | Description |  |
| :--- | :--- | :--- | :--- |
| 0 | 0 | $\phi$ | (Initial value) |
|  | 1 | $\phi / 4$ |  |
| 1 | 0 | $\phi / 16$ |  |
| 1 | $\phi / 64$ |  |  |

### 13.2.6 Serial Control Register (SCR)

SCR enables the SCI transmitter and receiver, enables or disables serial clock output in asynchronous mode, enables or disables interrupts, and selects the transmit/receive clock source.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKEO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  | Rece <br> Enab receiv | Trans Enabl <br> interru <br> or disab rror int | Recei Enabl enab or disa enable receiv upts (E | Multipro nables nterrupt enable or disab <br> the tra <br> data-ful | smit-e bles or interrup ssor in disable <br> the re <br> mitter <br> terrupts | lock ena hese bit CI clock <br> interru ables tr (TEI) <br> rupt en multipro <br> ver <br> XI) and | le $1 / 0$ <br> select the ource <br> enable smit- <br> le sor |
|  | Transmit interrupt enableEnables or disables transmit-data-empty interrupts (TXI) |  |  |  |  |  |  |  |

The CPU can always read and write SCR. SCR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.

Bit 7—Transmit Interrupt Enable (TIE): Enables or disables the transmit-data-empty interrupt (TXI) requested when the TDRE flag in SSR is set to 1 due to transfer of serial transmit data from TDR to TSR.

Bit 7: TIE
Description
0
Transmit-data-empty interrupt request (TXI) is disabled*
(Initial value)
$1 \quad$ Transmit-data-empty interrupt request (TXI) is enabled
Note: *TXI interrupt requests can be cleared by reading the value 1 from the TDRE flag, then clearing it to 0 ; or by clearing the TIE bit to 0 .

Bit 6-Receive Interrupt Enable (RIE): Enables or disables the receive-data-full interrupt (RXI) requested when the RDRF flag is set to 1 in SSR due to transfer of serial receive data from RSR to RDR; also enables or disables the receive-error interrupt (ERI).
Bit 6: RIE Description
$0 \quad$ Receive-data-full (RXI) and receive-error (ERI) interrupt requests are disabled*
(Initial value)
1 Receive-data-full (RXI) and receive-error (ERI) interrupt requests are enabled
$\overline{\text { Note: * RXI and ERI interrupt requests can be cleared by reading the value } 1 \text { from the RDRF, FER, }}$ PER, or ORER flag, then clearing it to 0 ; or by clearing the RIE bit to 0 .

Bit 5-Transmit Enable (TE): Enables or disables the start of SCI serial transmitting operations.
Bit 5: TE Description

| 0 | Transmitting disabled*1 | (Initial value) |
| :--- | :--- | :---: |
| 1 | Transmitting enabled*2 |  |

Notes: 1. The TDRE bit is locked at 1 in SSR.
2. In the enabled state, serial transmitting starts when the TDRE bit in SSR is cleared to 0 after writing of transmit data into TDR. Select the transmit format in SMR before setting the TE bit to 1 .

Bit 4-Receive Enable (RE): Enables or disables the start of SCI serial receiving operations.
Bit 4: RE Description

| 0 | Receiving disabled*1 | (Initial value) |
| :--- | :--- | :--- |
| 1 | Receiving |  |

1 Receiving enabled*2
Notes: 1. Clearing the RE bit to 0 does not affect the RDRF, FER, PER, and ORER flags. These flags retain their previous values.
2. In the enabled state, serial receiving starts when a start bit is detected in asynchronous mode, or serial clock input is detected in synchronous mode. Select the receive format in SMR before setting the RE bit to 1 .

Bit 3—Multiprocessor Interrupt Enable (MPIE): Enables or disables multiprocessor interrupts. The MPIE setting is valid only in asynchronous mode, and only if the MP bit is set to 1 in SMR. The MPIE setting is ignored in synchronous mode or when the MP bit is cleared to 0 .

## Bit 3: MPIE Description

Multiprocessor interrupts are disabled (normal receive operation) (Initial value) [Clearing conditions]
The MPIE bit is cleared to 0 .
MPB $=1$ in received data.
1 Multiprocessor interrupts are enabled*
Receive-data-full interrupts (RXI), receive-error interrupts (ERI), and setting of the RDRF, FER, and ORER status flags in SSR are disabled until data with the multiprocessor bit set to 1 is received.
Note: * The SCI does not transfer receive data from RSR to RDR, does not detect receive errors, and does not set the RDRF, FER, and ORER flags in SSR. When it receives data in which MPB $=1$, the SCI sets the MPB bit to 1 in SSR, automatically clears the MPIE bit to 0 , enables RXI and ERI interrupts (if the RIE bit is set to 1 in SCR), and allows the FER and ORER flags to be set.

Bit 2-Transmit-End Interrupt Enable (TEIE): Enables or disables the transmit-end interrupt (TEI) requested if TDR does not contain new transmit data when the MSB is transmitted.

## Bit 2: TEIE Description

| 0 | Transmit-end interrupt requests (TEI) are disabled* | (Initial value) |
| :--- | :--- | :--- |
| 1 | Transmit-end interrupt requests (TEI) are enabled* |  |
| Note: * TEI interrupt requests can be cleared by reading the value 1 from the TDRE flag in SSR, |  |  |
| then clearing the TDRE flag to 0, thereby also clearing the TEND flag to 0; or by clearing |  |  |
| the TEIE bit to 0. |  |  |

Bits 1 and 0 —Clock Enable 1 and 0 (CKE1/0): These bits select the SCI clock source and enable or disable clock output from the SCK pin. Depending on the settings of CKE1 and CKE0, the SCK pin can be used for generic input/output, serial clock output, or serial clock input.

The CKE0 setting is valid only in asynchronous mode, and only when the SCI is internally clocked $(\mathrm{CKE} 1=0)$. The CKE0 setting is ignored in synchronous mode, or when an external clock source is selected $(\mathrm{CKE} 1=1)$. Select the SCI operating mode in SMR before setting the CKE1 and CKE0 bits. For further details on selection of the SCI clock source, see table 13.9 in section 13.3, Operation.

| Bit 1: <br> CKE1 | Bit 0: <br> CKEO | Description |  |
| :---: | :---: | :---: | :---: |
| 0 | 0 | Asynchronous mode | Internal clock, SCK pin available for generic input/output *1 |
|  |  | Synchronous mode | Internal clock, SCK pin used for serial clock output *1 |
|  | 1 | Asynchronous mode | Internal clock, SCK pin used for clock output*2 |
|  |  | Synchronous mode | Internal clock, SCK pin used for serial clock output |
| 1 | 0 | Asynchronous mode | External clock, SCK pin used for clock input *3 |
|  |  | Synchronous mode | External clock, SCK pin used for serial clock input |
|  | 1 | Asynchronous mode | External clock, SCK pin used for clock input *3 |
|  |  | Synchronous mode | External clock, SCK pin used for serial clock input |

Notes: 1. Initial value
2. The output clock frequency is the same as the bit rate.
3. The input clock frequency is 16 times the bit rate.

### 13.2.7 Serial Status Register (SSR)

SSR is an 8-bit register containing multiprocessor bit values, and status flags that indicate SCI operating status.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TDRE | RDRF | ORER | FER | PER | TEND | MPB | MPBT |
| Initial value | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| Read/Write | R/(W)* | R/(W)* | R/(W)* | R/(W)* | R/(W)* | R | R | R/W |
|  |  |  |  |  |  |  |  | processor ransfer of multiessor bit to ansmitted <br> sor bit ceived or bit value <br> end of |

## Parity error

Status flag indicating detection of a receive parity error

## Framing error

Status flag indicating detection of a receive framing error

## Overrun error

Status flag indicating detection of a receive overrun error

## Receive data register full

Status flag indicating that data has been received and stored in RDR

## Transmit data register empty

Status flag indicating that transmit data has been transferred from TDR into TSR and new data can be written in TDR

Note: * Only 0 can be written, to clear the flag.

The CPU can always read and write SSR, but cannot write 1 in the TDRE, RDRF, ORER, PER, and FER flags. These flags can be cleared to 0 only if they have first been read while set to 1 . The TEND and MPB flags are read-only bits that cannot be written.

SSR is initialized to $\mathrm{H}^{\prime} 84$ by a reset and in standby mode.

Bit 7—Transmit Data Register Empty (TDRE): Indicates that the SCI has loaded transmit data from TDR into TSR and the next serial transmit data can be written in TDR.

| Bit 7: TDRE | Description |
| :--- | :--- |
| 0 | TDR contains valid transmit data |
|  | [Clearing conditions] |
|  | Software reads TDRE while it is set to 1, then writes 0. |
|  | The DMAC writes data in TDR. |
| 1 | TDR does not contain valid transmit data |
|  | [Setting conditions] |
|  | The chip is reset or enters standby mode. |
|  | The TE bit in SCR is cleared to 0. |
|  | TDR contents are loaded into TSR, so new datae) can be written in TDR. |

Bit 6—Receive Data Register Full (RDRF): Indicates that RDR contains new receive data.

| Bit 6: RDRF | Description |  |
| :--- | :--- | :--- |
| 0 | RDR does not contain new receive data | (Initial value) |
|  | [Clearing conditions] |  |

The chip is reset or enters standby mode.
Software reads RDRF while it is set to 1 , then writes 0 .
The DMAC reads data from RDR.
1
RDR contains new receive data
[Setting condition]
When serial data is received normally and transferred from RSR to RDR.
Note: The RDR contents and RDRF flag are not affected by detection of receive errors or by clearing of the RE bit to 0 in SCR. They retain their previous values. If the RDRF flag is still set to 1 when reception of the next data ends, an overrun error occurs and receive data is lost.

Bit 5-Overrun Error (ORER): Indicates that data reception ended abnormally due to an overrun error.

## Bit 5: ORER Description

Receiving is in progress or has ended normally
[Clearing conditions]
The chip is reset or enters standby mode.
Software reads ORER while it is set to 1 , then writes 0 .
A receive overrun error occurred*2
[Setting condition]
Reception of the next serial data ends when RDRF $=1$.
Notes: 1. Clearing the RE bit to 0 in SCR does not affect the ORER flag, which retains its previous value.
2. RDR continues to hold the receive data before the overrun error, so subsequent receive data is lost. Serial receiving cannot continue while the ORER flag is set to 1 . In synchronous mode, serial transmitting is also disabled.

Bit 4-Framing Error (FER): Indicates that data reception ended abnormally due to a framing error in asynchronous mode.

| Bit 4: FER | Description |
| :---: | :---: |
| 0 | Receiving is in progress or has ended normally (Initial value)*1 |
|  | [Clearing conditions] |
|  | The chip is reset or enters standby mode. |
|  | Software reads FER while it is set to 1 , then writes 0 . |
| 1 | A receive framing error occurred*2 |
|  | [Setting condition] |
|  | The stop bit at the end of receive data is checked and found to be 0 . |

Notes: 1. Clearing the RE bit to 0 in SCR does not affect the FER flag, which retains its previous value.
2. When the stop bit length is 2 bits, only the first bit is checked. The second stop bit is not checked. When a framing error occurs the SCI transfers the receive data into RDR but does not set the RDRF flag. Serial receiving cannot continue while the FER flag is set to 1 . In synchronous mode, serial transmitting is also disabled.

Bit 3-Parity Error (PER): Indicates that data reception ended abnormally due to a parity error in asynchronous mode.

Bit 3: PER Description
0
Receiving is in progress or has ended normally*1
(Initial value)
[Clearing conditions]
The chip is reset or enters standby mode.
Software reads PER while it is set to 1 , then writes 0 .
A receive parity error occurred*2
[Setting condition]
The number of 1 s in receive data, including the parity bit, does not match the even or odd parity setting of $O / E$ in SMR.
Notes: 1. Clearing the RE bit to 0 in SCR does not affect the PER flag, which retains its previous value.
2. When a parity error occurs the SCI transfers the receive data into RDR but does not set the RDRF flag. Serial receiving cannot continue while the PER flag is set to 1. In synchronous mode, serial transmitting is also disabled.

Bit 2—Transmit End (TEND): Indicates that when the last bit of a serial character was transmitted TDR did not contain new transmit data, so transmission has ended. The TEND flag is a read-only bit and cannot be written.
Bit 2: TEND Description

| 0 | Transmission is in progress |
| :--- | :--- |
| [Clearing conditions] |  |
|  | Software reads TDRE while it is set to 1 , then writes 0 in the TDRE flag. |
|  | The DMAC writes data in TDR. |
| 1 | End of transmission |
| [Setting conditions] |  |
|  | The chip is reset or enters standby mode. |
|  | The TE bit is cleared to 0 in SCR. |
|  | TDRE is 1 when the last bit of a serial character is transmitted. |

Bit 1—Multiprocessor Bit (MPB): Stores the value of the multiprocessor bit in receive data when a multiprocessor format is used in asynchronous mode. MPB is a read-only bit and cannot be written.

Bit 1: MPB
Description

| 0 | Multiprocessor bit value in receive data is $0^{*}$ | (Initial value) |
| :--- | :--- | :--- |
| 1 | Multiprocessor bit value in receive data is 1 |  |

Note: * If the RE bit is cleared to 0 when a multiprocessor format is selected, MPB retains its previous value.

Bit 0—Multiprocessor Bit Transfer (MPBT): Stores the value of the multiprocessor bit added to transmit data when a multiprocessor format is selected for transmitting in asynchronous mode. The MPBT setting is ignored in synchronous mode, when a multiprocessor format is not selected, or when the SCI is not transmitting.

Bit 0: MPBT Description

| 0 | Multiprocessor bit value in transmit data is 0 | (Initial value) |
| :--- | :--- | :--- |
| 1 | Multiprocessor bit value in transmit data is 1 |  |

### 13.2.8 Bit Rate Register (BRR)

BRR is an 8-bit register that, together with the CKS1 and CKS0 bits in SMR that select the baud rate generator clock source, determines the serial communication bit rate.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

The CPU can always read and write BRR. BRR is initialized to H'FF by a reset and in standby mode. The two SCI channels have independent baud rate generator control, so different values can be set in the two channels.

Table 13.3 shows examples of BRR settings in asynchronous mode. Table 13.4 shows examples of BRR settings in synchronous mode.

Table 13.3 Examples of Bit Rates and BRR Settings in Asynchronous Mode
$\phi(\mathrm{MHz})$

| Bit Rate (bits/s) | 2 |  |  | 2.097152 |  |  | 2.4576 |  |  | 3 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ |
| 110 | 1 | 141 | 0.03 | 1 | 148 | -0.04 | 1 | 174 | -0.26 | 1 | 212 | 0.03 |
| 150 | 1 | 103 | 0.16 | 1 | 108 | 0.21 | 1 | 127 | 0 | 1 | 155 | 0.16 |
| 300 | 0 | 207 | 0.16 | 0 | 217 | 0.21 | 0 | 255 | 0 | 1 | 77 | 0.16 |
| 600 | 0 | 103 | 0.16 | 0 | 108 | 0.21 | 0 | 127 | 0 | 0 | 155 | 0.16 |
| 1200 | 0 | 51 | 0.16 | 0 | 54 | -0.70 | 0 | 63 | 0 | 0 | 77 | 0.16 |
| 2400 | 0 | 25 | 0.16 | 0 | 26 | 1.14 | 0 | 31 | 0 | 0 | 38 | 0.16 |
| 4800 | 0 | 12 | 0.16 | 0 | 13 | -2.48 | 0 | 15 | 0 | 0 | 19 | -2.34 |
| 9600 | 0 | 6 | -6.99 | 0 | 6 | -2.48 | 0 | 7 | 0 | 0 | 9 | -2.34 |
| 19200 | 0 | 2 | 8.51 | 0 | 2 | 13.78 | 0 | 3 | 0 | 0 | 4 | -2.34 |
| 31250 | 0 | 1 | 0 | 0 | 1 | 4.86 | 0 | 1 | 22.88 | 0 | 2 | 0 |
| 38400 | 0 | 1 | -18.62 | 0 | 1 | -14.67 | 0 | 1 | 0 | - | - | - |


| Bit Rate (bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 3.6864 |  |  | 4 |  |  | 4.9152 |  |  | 5 |  |  |
|  | n | N | Error <br> (\%) | n | N | Error <br> (\%) | n | N | Error <br> (\%) | n | N | Error <br> (\%) |
| 110 | 2 | 64 | 0.70 | 2 | 70 | 0.03 | 2 | 86 | 0.31 | 2 | 88 | -0.25 |
| 150 | 1 | 191 | 0 | 1 | 207 | 0.16 | 1 | 255 | 0 | 2 | 64 | 0.16 |
| 300 | 1 | 95 | 0 | 1 | 103 | 0.16 | 1 | 127 | 0 | 1 | 129 | 0.16 |
| 600 | 0 | 191 | 0 | 0 | 207 | 0.16 | 0 | 255 | 0 | 1 | 64 | 0.16 |
| 1200 | 0 | 95 | 0 | 0 | 103 | 0.16 | 0 | 127 | 0 | 0 | 129 | 0.16 |
| 2400 | 0 | 47 | 0 | 0 | 51 | 0.16 | 0 | 63 | 0 | 0 | 64 | 0.16 |
| 4800 | 0 | 23 | 0 | 0 | 25 | 0.16 | 0 | 31 | 0 | 0 | 32 | -1.36 |
| 9600 | 0 | 11 | 0 | 0 | 12 | 0.16 | 0 | 15 | 0 | 0 | 15 | 1.73 |
| 19200 | 0 | 5 | 0 | 0 | 6 | -6.99 | 0 | 7 | 0 | 0 | 7 | 1.73 |
| 31250 | - | - | - | 0 | 3 | 0 | 0 | 4 | -1.70 | 0 | 4 | 0 |
| 38400 | 0 | 2 | 0 | 0 | 2 | 8.51 | 0 | 3 | 0 | 0 | 3 | 1.73 |


| Bit Rate (bits/s) | 6 |  |  | 6.144 |  |  | 7.3728 |  |  | 8 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \\ & \hline \end{aligned}$ |
| 110 | 2 | 106 | -0.44 | 2 | 108 | 0.08 | 2 | 130 | -0.07 | 2 | 141 | 0.03 |
| 150 | 2 | 77 | 0.16 | 2 | 79 | 0 | 2 | 95 | 0 | 2 | 103 | 0.16 |
| 300 | 1 | 155 | 0.16 | 1 | 159 | 0 | 1 | 191 | 0 | 1 | 207 | 0.16 |
| 600 | 1 | 77 | 0.16 | 1 | 79 | 0 | 1 | 95 | 0 | 1 | 103 | 0.16 |
| 1200 | 0 | 155 | 0.16 | 0 | 159 | 0 | 0 | 191 | 0 | 0 | 207 | 0.16 |
| 2400 | 0 | 77 | 0.16 | 0 | 79 | 0 | 0 | 95 | 0 | 0 | 103 | 0.16 |
| 4800 | 0 | 38 | 0.16 | 0 | 39 | 0 | 0 | 47 | 0 | 0 | 51 | 0.16 |
| 9600 | 0 | 19 | -2.34 | 0 | 19 | 0 | 0 | 23 | 0 | 0 | 25 | 0.16 |
| 19200 | 0 | 9 | -2.34 | 0 | 9 | 0 | 0 | 11 | 0 | 0 | 12 | 0.16 |
| 31250 | 0 | 5 | 0 | 0 | 5 | 2.40 | 0 | 6 | 5.33 | 0 | 7 | 0 |
| 38400 | 0 | 4 | -2.34 | 0 | 4 | 0 | 0 | 5 | 0 | 0 | 6 | -6.99 |

$\phi(\mathrm{MHz})$

| Bit Rate (bits/s) | 9.8304 |  |  | 10 |  |  | 12 |  |  | 12.288 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | n | N | Error (\%) | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 2 | 174 | -0.26 | 2 | 177 | -0.25 | 2 | 212 | 0.03 | 2 | 217 | 0.08 |
| 150 | 2 | 127 | 0 | 2 | 129 | 0.16 | 2 | 155 | 0.16 | 2 | 159 | 0 |
| 300 | 1 | 255 | 0 | 2 | 64 | 0.16 | 2 | 77 | 0.16 | 2 | 79 | 0 |
| 600 | 1 | 127 | 0 | 1 | 129 | 0.16 | 1 | 155 | 0.16 | 1 | 159 | 0 |
| 1200 | 0 | 255 | 0 | 1 | 64 | 0.16 | 1 | 77 | 0.16 | 1 | 79 | 0 |
| 2400 | 0 | 127 | 0 | 0 | 129 | 0.16 | 0 | 155 | 0.16 | 0 | 159 | 0 |
| 4800 | 0 | 63 | 0 | 0 | 64 | 0.16 | 0 | 77 | 0.16 | 0 | 79 | 0 |
| 9600 | 0 | 31 | 0 | 0 | 32 | -1.36 | 0 | 38 | 0.16 | 0 | 39 | 0 |
| 19200 | 0 | 15 | 0 | 0 | 15 | 1.73 | 0 | 19 | -2.34 | 0 | 19 | 0 |
| 31250 | 0 | 9 | -1.70 | 0 | 9 | 0 | 0 | 11 | 0 | 0 | 11 | 2.40 |
| 38400 | 0 | 7 | 0 | 0 | 7 | 1.73 | 0 | 9 | -2.34 | 0 | 9 | 0 |

Rev. 2.0, 03/01, page 458 of 822

| Bit Rate (bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 13 |  |  | 14 |  |  | 14.7456 |  |  | 16 |  |  |
|  | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | Error (\%) |
| 110 | 2 | 230 | -0.08 | 2 | 248 | -0.17 | 3 | 64 | 0.70 | 3 | 70 | 0.03 |
| 150 | 2 | 168 | 0.16 | 2 | 181 | 0.16 | 2 | 191 | 0 | 2 | 207 | 0.16 |
| 300 | 2 | 84 | -0.43 | 2 | 90 | 0.16 | 2 | 95 | 0 | 2 | 103 | 0.16 |
| 600 | 1 | 168 | 0.16 | 1 | 181 | 0.16 | 1 | 191 | 0 | 1 | 207 | 0.16 |
| 1200 | 1 | 84 | -0.43 | 1 | 90 | 0.16 | 1 | 95 | 0 | 1 | 103 | 0.16 |
| 2400 | 0 | 168 | 0.16 | 0 | 181 | 0.16 | 0 | 191 | 0 | 0 | 207 | 0.16 |
| 4800 | 0 | 84 | -0.43 | 0 | 90 | 0.16 | 0 | 95 | 0 | 0 | 103 | 0.16 |
| 9600 | 0 | 41 | 0.76 | 0 | 45 | -0.93 | 0 | 47 | 0 | 0 | 51 | 0.16 |
| 19200 | 0 | 20 | 0.76 | 0 | 22 | -0.93 | 0 | 23 | 0 | 0 | 25 | 0.16 |
| 31250 | 0 | 12 | 0.00 | 0 | 13 | 0 | 0 | 14 | -1.70 | 0 | 15 | 0 |
| 38400 | 0 | 10 | -3.82 | 0 | 10 | 3.57 | 0 | 11 | 0 | 0 | 12 | 0.16 |

$\phi$ (MHz)

| Bit Rate (bits/s) |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 18 |  |  | 20 |  |  | 25 |  |  |
|  | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \end{aligned}$ | n | N | $\begin{aligned} & \text { Error } \\ & \text { (\%) } \\ & \hline \end{aligned}$ | n | N | $\begin{aligned} & \hline \text { Error } \\ & \text { (\%) } \end{aligned}$ |
| 110 | 3 | 79 | -0.12 | 3 | 88 | -0.25 | 3 | 110 | -0.02 |
| 150 | 2 | 233 | 0.16 | 3 | 64 | 0.16 | 3 | 80 | -0.47 |
| 300 | 2 | 116 | 0.16 | 2 | 129 | 0.16 | 2 | 162 | 0.15 |
| 600 | 1 | 233 | 0.16 | 2 | 64 | 0.16 | 2 | 80 | -0.47 |
| 1200 | 1 | 116 | 0.16 | 1 | 129 | 0.16 | 1 | 162 | 0.15 |
| 2400 | 0 | 233 | 0.16 | 1 | 64 | 0.16 | 1 | 80 | -0.47 |
| 4800 | 0 | 116 | 0.16 | 0 | 129 | 0.16 | 0 | 162 | 0.15 |
| 9600 | 0 | 58 | -0.69 | 0 | 64 | 0.16 | 0 | 80 | -0.47 |
| 19200 | 0 | 28 | 1.02 | 0 | 32 | -1.36 | 0 | 40 | -0.76 |
| 31250 | 0 | 17 | 0.00 | 0 | 19 | 0.00 | 0 | 24 | 0.00 |
| 38400 | 0 | 14 | -2.34 | 0 | 15 | 1.73 | 0 | 19 | 1.73 |

Table 13.4 Examples of Bit Rates and BRR Settings in Synchronous Mode

| Bit Rate (bits/s) | $\phi(\mathrm{MHz})$ |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 2 |  | 4 |  | 8 |  | 10 |  | 13 |  |
|  | n | N | n | N | n | N | n | N | n | N |
| 110 | 3 | 70 | - | - | - | - | - | - | - | - |
| 250 | 2 | 124 | 2 | 249 | 3 | 124 | - | - | 3 | 202 |
| 500 | 1 | 249 | 2 | 124 | 2 | 249 | - | - | 3 | 101 |
| 1 k | 1 | 124 | 1 | 249 | 2 | 124 | - | - | 2 | 202 |
| 2.5 k | 0 | 199 | 1 | 99 | 1 | 199 | 1 | 249 | 2 | 80 |
| 5 k | 0 | 99 | 0 | 199 | 1 | 99 | 1 | 124 | 1 | 162 |
| 10 k | 0 | 49 | 0 | 99 | 0 | 199 | 0 | 249 | 1 | 80 |
| 25 k | 0 | 19 | 0 | 39 | 0 | 79 | 0 | 99 | 0 | 129 |
| 50 k | 0 | 9 | 0 | 19 | 0 | 39 | 0 | 49 | 0 | 64 |
| 100 k | 0 | 4 | 0 | 9 | 0 | 19 | 0 | 24 | - | - |
| 250 k | 0 | 1 | 0 | 3 | 0 | 7 | 0 | 9 | 0 | 12 |
| 500 k | 0 | 0* | 0 | 1 | 0 | 3 | 0 | 4 | - | - |
| 1 M |  |  | 0 | 0* | 0 | 1 | - | - | - | - |
| 2 M |  |  |  |  | 0 | 0* | - | - | - | - |
| 2.5 M |  |  |  |  | - | - | 0 | 0* | - | - |
| 4 M |  |  |  |  |  |  |  |  |  |  |

Note: Settings with an error of $1 \%$ or less are recommended.
Legend
Blank: No setting available
-: Setting possible, but error occurs
*: Continuous transmit/receive not possible

| Bit Rate (bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 16 |  | 18 |  | 20 |  | 25 |  |
|  | n | N | n | N | n | N | n | N |
| 110 | - | - | - | - | - | - | - | - |
| 250 | 3 | 249 | - | - | - | - | - | - |
| 500 | 3 | 124 | 3 | 140 | 3 | 155 | - | - |
| 1 k | 2 | 249 | 3 | 69 | 3 | 77 | 3 | 97 |
| 2.5 k | 2 | 99 | 2 | 112 | 2 | 124 | 2 | 155 |
| 5 k | 1 | 199 | 1 | 224 | 1 | 249 | 2 | 77 |
| 10 k | 1 | 99 | 1 | 112 | 1 | 124 | 1 | 155 |
| 25 k | 0 | 159 | 0 | 179 | 0 | 199 | 0 | 249 |
| 50 k | 0 | 79 | 0 | 89 | 0 | 99 | 0 | 124 |
| 100 k | 0 | 39 | 0 | 44 | 0 | 49 | 0 | 62 |
| 250 k | 0 | 15 | 0 | 17 | 0 | 19 | 0 | 24 |
| 500 k | 0 | 7 | 0 | 8 | 0 | 9 | - | - |
| 1 M | 0 | 3 | 0 | 4 | 0 | 4 | - | - |
| 2 M | 0 | 1 | - | - | - | - | - | - |
| 2.5 M | - | - | - | - | - | - | - | - |
| 4 M | 0 | 0* | - | - | - | - | - | - |

Note: Settings with an error of $1 \%$ or less are recommended.
Legend
Blank: No setting available
-: Setting possible, but error occurs
*: Continuous transmit/receive not possible

The BRR setting is calculated as follows:
Asynchronous mode:

$$
N=\frac{\phi}{64 \times 2^{2 n-1} \times B} \times 10^{6}-1
$$

Synchronous mode:

$$
N=\frac{\phi}{8 \times 2^{2 n-1} \times B} \times 10^{6}-1
$$

B: Bit rate (bits/s)
N : BRR setting for baud rate generator $(0 \leq \mathrm{N} \leq 255)$
$\phi$ : System clock frequency $(\mathrm{MHz})$
n : Baud rate generator clock source ( $\mathrm{n}=0,1,2,3$ )
(For the clock sources and values of $n$, see the following table.)
SMR Settings

| $\mathbf{n}$ | Clock Source | CKS1 | CKS0 |
| :--- | :--- | :--- | :--- |
| 0 | $\phi$ | 0 | 0 |
| 1 | $\phi / 4$ | 0 | 1 |
| 2 | $\phi / 16$ | 1 | 0 |
| 3 | $\phi / 64$ | 1 | 1 |

The bit rate error in asynchronous mode is calculated as follows.

$$
\text { Error }(\%)=\left\{\frac{\phi \times 10^{6}}{(N+1) \times B \times 64 \times 2^{2 n-1}}-1\right\} \times 100
$$

Table 13.5 indicates the maximum bit rates in asynchronous mode for various system clock frequencies. Tables 13.6 and 13.7 indicate the maximum bit rates with external clock input.

Table 13.5 Maximum Bit Rates for Various Frequencies (Asynchronous Mode)

| $\phi(\mathrm{MHz})$ | Maximum Bit Rate (bits/s) | Settings |  |
| :---: | :---: | :---: | :---: |
|  |  | n | N |
| 2 | 62500 | 0 | 0 |
| 2.097152 | 65536 | 0 | 0 |
| 2.4576 | 76800 | 0 | 0 |
| 3 | 93750 | 0 | 0 |
| 3.6864 | 115200 | 0 | 0 |
| 4 | 125000 | 0 | 0 |
| 4.9152 | 153600 | 0 | 0 |
| 5 | 156250 | 0 | 0 |
| 6 | 187500 | 0 | 0 |
| 6.144 | 192000 | 0 | 0 |
| 7.3728 | 230400 | 0 | 0 |
| 8 | 250000 | 0 | 0 |
| 9.8304 | 307200 | 0 | 0 |
| 10 | 312500 | 0 | 0 |
| 12 | 375000 | 0 | 0 |
| 12.288 | 384000 | 0 | 0 |
| 14 | 437500 | 0 | 0 |
| 14.7456 | 460800 | 0 | 0 |
| 16 | 500000 | 0 | 0 |
| 17.2032 | 537600 | 0 | 0 |
| 18 | 562500 | 0 | 0 |
| 20 | 625000 | 0 | 0 |
| 25 | 781250 | 0 | 0 |

Table 13.6 Maximum Bit Rates with External Clock Input (Asynchronous Mode)

| $\phi(\mathrm{MHz})$ | External Input Clock (MHz) | Maximum Bit Rate (bits/s) |
| :---: | :---: | :---: |
| 2 | 0.5000 | 31250 |
| 2.097152 | 0.5243 | 32768 |
| 2.4576 | 0.6144 | 38400 |
| 3 | 0.7500 | 46875 |
| 3.6864 | 0.9216 | 57600 |
| 4 | 1.0000 | 62500 |
| 4.9152 | 1.2288 | 76800 |
| 5 | 1.2500 | 78125 |
| 6 | 1.5000 | 93750 |
| 6.144 | 1.5360 | 96000 |
| 7.3728 | 1.8432 | 115200 |
| 8 | 2.0000 | 125000 |
| 9.8304 | 2.4576 | 153600 |
| 10 | 2.5000 | 156250 |
| 12 | 3.0000 | 187500 |
| 12.288 | 3.0720 | 192000 |
| 14 | 3.5000 | 218750 |
| 14.7456 | 3.6864 | 230400 |
| 16 | 4.0000 | 250000 |
| 17.2032 | 4.3008 | 268800 |
| 18 | 4.5000 | 281250 |
| 20 | 5.0000 | 312500 |
| 25 | 6.2500 | 390625 |

Rev. 2.0, 03/01, page 464 of 822

Table 13.7 Maximum Bit Rates with External Clock Input (Synchronous Mode)

| $\boldsymbol{\phi}(\mathrm{MHz})$ | External Input Clock (MHz) | Maximum Bit Rate (bits/s) |
| :--- | :--- | :--- |
| 2 | 0.3333 | 333333.3 |
| 4 | 0.6667 | 666666.7 |
| 6 | 1.0000 | 1000000.0 |
| 8 | 1.3333 | 1333333.3 |
| 10 | 1.6667 | 1666666.7 |
| 12 | 2.0000 | 2000000.0 |
| 14 | 2.3333 | 2333333.3 |
| 16 | 2.6667 | 2666666.7 |
| 18 | 3.0000 | 3000000.0 |
| 20 | 3.3333 | 3333333.3 |
| 25 | 4.1667 | 4166666.7 |

### 13.3 Operation

### 13.3.1 Overview

The SCI has an asynchronous mode in which characters are synchronized individually, and a synchronous mode in which communication is synchronized with clock pulses. Serial communication is possible in either mode. Asynchronous or synchronous mode and the communication format are selected in SMR, as shown in table 13.8. The SCI clock source is selected by the C/A bit in SMR and the CKE1 and CKE0 bits in SCR, as shown in table 13.9.

## Asynchronous Mode

- Data length is selectable: 7 or 8 bits.
- Parity and multiprocessor bits are selectable. So is the stop bit length (1 or 2 bits). These selections determine the communication format and character length.
- In receiving, it is possible to detect framing errors, parity errors, overrun errors, and the break state.
- An internal or external clock can be selected as the SCI clock source.
- When an internal clock is selected, the SCI operates using the on-chip baud rate generator, and can output a serial clock signal with a frequency matching the bit rate.
- When an external clock is selected, the external clock input must have a frequency 16 times the bit rate. (The on-chip baud rate generator is not used.)


## Synchronous Mode

- The communication format has a fixed 8 -bit data length.
- In receiving, it is possible to detect overrun errors.
- An internal or external clock can be selected as the SCI clock source.
- When an internal clock is selected, the SCI operates using the on-chip baud rate generator, and outputs a serial clock signal to external devices.
- When an external clock is selected, the SCI operates on the input serial clock. The on-chip baud rate generator is not used.

Table 13.8 SMR Settings and Serial Communication Formats

| SMR Settings |  |  |  |  | Mode | SCI Communication Format |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { Bit } 7: \\ & \text { C// } \end{aligned}$ | Bit 6: <br> CHR | Bit 2: <br> MP | Bit 5: <br> PE | Bit 3: STOP |  | Data <br> Length | Multiprocessor Bit | Parity Bit | Stop <br> Bit <br> Length |
| 0 | 0 | 0 | 0 | 0 | Asynchronous mode | 8-bit data | Absent | Absent | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
|  |  |  | 1 | 0 |  |  |  | Present | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
|  | 1 |  | 0 | 0 |  | 7-bit data |  | Absent | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
|  |  |  | 1 | 0 |  |  |  | Present | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
|  | 0 | 1 | - | 0 | Asynchronous mode (multiprocessor format) | 8-bit data | Present | Absent | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
|  | 1 |  |  | 0 |  | 7-bit data |  |  | 1 bit |
|  |  |  |  | 1 |  |  |  |  | 2 bits |
| 1 | - | - | - | - | Synchronous mode | 8-bit data | Absent |  | None |

Table 13.9 SMR and SCR Settings and SCI Clock Source Selection

| SMR | SCR Settings |  | Mode | SCI Transmit/Receive Clock |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { Bit } 7: \\ & \text { C/ } / \overline{\text { A }} \end{aligned}$ | Bit 1: CKE1 | Bit 0: <br> CKEO |  | Clock Source | SCK Pin Function |
| 0 | 0 | 0 | Asynchronous mode | Internal | SCI does not use the SCK pin |
|  |  | 1 |  |  | Outputs a clock with frequency matching the bit rate |
|  | 1 | 0 |  | External | Inputs a clock with frequency 16 times the bit rate |
|  |  | 1 |  |  |  |
| 1 | 0 | 0 | Synchronous mode | Internal | Outputs the serial clock |
|  |  | 1 |  |  |  |
|  | 1 | 0 |  | External | Inputs the serial clock |
|  |  | 1 |  |  |  |

### 13.3.2 Operation in Asynchronous Mode

In asynchronous mode each transmitted or received character begins with a start bit and ends with a stop bit. Serial communication is synchronized one character at a time.

The transmitting and receiving sections of the SCI are independent, so full duplex communication is possible. The transmitter and receiver are both double buffered, so data can be written and read while transmitting and receiving are in progress, enabling continuous transmitting and receiving.

Figure 13.2 shows the general format of asynchronous serial communication. In asynchronous serial communication the communication line is normally held in the mark (high) state. The SCI monitors the line and starts serial communication when the line goes to the space (low) state, indicating a start bit. One serial character consists of a start bit (low), data (LSB first), parity bit (high or low), and stop bit (high), in that order.

When receiving in asynchronous mode, the SCI synchronizes at the falling edge of the start bit. The SCI samples each data bit on the eighth pulse of a clock with a frequency 16 times the bit rate. Receive data is latched at the center of each bit.


Figure 13.2 Data Format in Asynchronous Communication (Example: 8-Bit Data with Parity and 2 Stop Bits)

Communication Formats: Table 13.10 shows the 12 communication formats that can be selected in asynchronous mode. The format is selected by settings in SMR.

Table 13.10 Serial Communication Formats (Asynchronous Mode)

| SMR Settings |  |  |  | Serial Communication Format and Frame Length |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CHR | PE | MP | STOP | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 |
| 0 | 0 | 0 | 0 | s | 8 -bit data |  |  |  |  |  |  |  | STOP |  |  |
| 0 | 0 | 0 | 1 | S | 8-bit data |  |  |  |  |  |  |  | STOP STOP |  |  |
| 0 | 1 | 0 | 0 | S | 8-bit data |  |  |  |  |  |  |  | STOP |  |  |
| 0 | 1 | 0 | 1 | S | 8-bit data |  |  |  |  |  |  |  | STOP STOP |  |  |
| 1 | 0 | 0 | 0 | S | 7-bit data |  |  |  |  |  |  | STOP |  |  |  |
| 1 | 0 | 0 | 1 | S | 7-bit data |  |  |  |  |  |  | STOP\|STOP |  |  |  |
| 1 | 1 | 0 | 0 | S | 7-bit data |  |  |  |  |  |  | STOP |  |  |  |
| 1 | 1 | 0 | 1 | S | 7-bit data |  |  |  |  |  |  | STOP STOP $^{\text {d }}$ |  |  |  |
| 0 | - | 1 | 0 | S | 8-bit data |  |  |  |  |  |  |  | MPB STOP |  |  |
| 0 | - | 1 | 1 | S | 8-bit data |  |  |  |  |  |  |  | MPB STOP STOP |  |  |
| 1 | - | 1 | 0 | S | 7-bit data |  |  |  |  |  |  | MPB STOP |  |  |  |
| 1 | - | 1 | 1 | S | 7-bit data |  |  |  |  |  |  | MPB STOP STOP |  |  |  |

Legend
S: Start bit
STOP: Stop bit
P: Parity bit
MPB: Multiprocessor bit

Clock: An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected as the SCI transmit/receive clock. The clock source is selected by the C// bit in SMR and bits CKE1 and CKE0 in SCR. See table 13.9.

When an external clock is input at the SCK pin, it must have a frequency equal to 16 times the desired bit rate.

When the SCI operates on an internal clock, it can output a clock signal at the SCK pin. The frequency of this output clock is equal to the bit rate. The phase is aligned as in figure 13.3 so that the rising edge of the clock occurs at the center of each transmit data bit.


Figure 13.3 Phase Relationship between Output Clock and Serial Data (Asynchronous Mode)

## Transmitting and Receiving Data

- SCI Initialization (Asynchronous Mode)

Before transmitting or receiving, clear the TE and RE bits to 0 in SCR, then initialize the SCI as follows.
When changing the communication mode or format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing TE to 0 sets the TDRE flag to 1 and initializes TSR. Clearing RE to 0 , however, does not initialize the RDRF, PER, FER, and ORER flags and RDR, which retain their previous contents.
When an external clock is used, the clock should not be stopped during initialization or subsequent operation. SCI operation becomes unreliable if the clock is stopped.
Figure 13.4 is a sample flowchart for initializing the SCI.


1. Select the clock source in SCR. Clear the RIE, TIE, TEIE, MPIE, TE, and RE bits to 0 . If clock output is selected in asynchronous mode, clock output starts immediately after the setting is made in SCR.
2. Select the communication format in SMR.
3. Write the value corresponding to the bit rate in BRR. This step is not necessary when an external clock is used.
4. Wait for at least the interval required to transmit or receive 1 bit, then set the TE or RE bit to 1 in SCR. Set the RIE, TIE, TEIE, and MPIE bits as necessary. Setting the TE or RE bit enables the SCI to use the TxD or RxD pin.

Figure 13.4 Sample Flowchart for SCI Initialization

- Transmitting Serial Data (Asynchronous Mode)

Figure 13.5 shows a sample flowchart for transmitting serial data and indicates the procedure to follow.


1. SCI initialization: the transmit data output function of the TxD pin is selected automatically. After the TE bit is set to 1 , one frame of 1 is output, then transmission is possible.
2. SCl status check and transmit data write: read SSR, check that the TDRE flag is 1 , then write transmit data in TDR and clear the TDRE flag to 0 .
3. To continue transmitting serial data: after checking that the TDRE flag is 1 , indicating that data can be written, write data in TDR, then clear the TDRE flag to 0 . When the DMAC is activated by a transmit-dataempty interrupt request (TXI) to write data in TDR, the TDRE flag is checked and cleared automatically.
4. To output a break signal at the end of serial transmission: set the DDR bit to 1 and clear the DR bit to 0 (DDR and DR are I/O port registers), then clear the TE bit to 0 in SCR.

Figure 13.5 Sample Flowchart for Transmitting Serial Data

In transmitting serial data, the SCI operates as follows.

1. The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
2. After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit is set to 1 in SCR, the SCI requests a transmit-data-empty interrupt (TXI) at this time.
Serial transmit data is transmitted in the following order from the TxD pin:
a. Start bit: One 0 bit is output.
b. Transmit data: 7 or 8 bits are output, LSB first.
c. Parity bit or multiprocessor bit: One parity bit (even or odd parity) or one multiprocessor bit is output. Formats in which neither a parity bit nor a multiprocessor bit is output can also be selected.
d. Stop bit: One or two 1 bits (stop bits) are output.
e. Mark state: Output of 1 bits continues until the start bit of the next transmit data.
3. The SCI checks the TDRE flag when it outputs the stop bit. If the TDRE flag is 0 , the SCI loads new data from TDR into TSR, outputs the stop bit, then begins serial transmission of the next frame. If the TDRE flag is 1 , the SCI sets the TEND flag to 1 in SSR, outputs the stop bit, then continues output of 1 bits in the mark state. If the TEIE bit is set to 1 in SCR, a transmit-end interrupt (TEI) is requested at this time.

Figure 13.6 shows an example of SCI transmit operation in asynchronous mode.


Figure 13.6 Example of SCI Transmit Operation in Asynchronous Mode (8-Bit Data with Parity and 1 Stop Bit)

- Receiving Serial Data (Asynchronous Mode)

Figure 13.7 shows a sample flowchart for receiving serial data and indicates the procedure to follow.


Figure 13.7 Sample Flowchart for Receiving Serial Data (1)


Figure 13.7 Sample Flowchart for Receiving Serial Data (2)

In receiving, the SCI operates as follows.

1. The SCI monitors the receive data line. When it detects a start bit, the SCI synchronizes internally and starts receiving.
2. Receive data is stored in RSR in order from LSB to MSB.
3. The parity bit and stop bit are received.

After receiving, the SCI makes the following checks:
a. Parity check: The number of 1 s in the receive data must match the even or odd parity setting of the $\mathrm{O} / \overline{\mathrm{E}}$ bit in SMR.
b. Stop bit check: The stop bit value must be 1. If there are two stop bits, only the first stop bit is checked.
c. Status check: The RDRF flag must be 0 so that receive data can be transferred from RSR into RDR.

If these checks all pass, the RDRF flag is set to 1 and the received data is stored in RDR. If one of the checks fails (receive error), the SCI operates as indicated in table 13.11.

Note: When a receive error occurs, further receiving is disabled. In receiving, the RDRF flag is not set to 1 . Be sure to clear the error flags to 0 .
4. When the RDRF flag is set to 1 , if the RIE bit is set to 1 in SCR, a receive-data-full interrupt (RXI) is requested. If the ORER, PER, or FER flag is set to 1 and the RIE bit in SCR is also set to 1 , a receive-error interrupt (ERI) is requested.

Table 13.11 Receive Error Conditions

| Receive Error | Abbreviation | Condition | Data Transfer |
| :--- | :--- | :--- | :--- |
| Overrun error | ORER | Receiving of next data ends <br> while RDRF flag is still set <br> to 1 in SSR | Receive data not transferred <br> from RSR to RDR |
| Framing error | FER | Stop bit is 0 | Receive data transferred from <br> RSR to RDR |
| Parity error | PER | Parity of receive data differs <br> from even/odd parity setting <br> in SMR | Receive data transferred from <br> RSR to RDR |

Figure 13.8 shows an example of SCI receive operation in asynchronous mode.


Figure 13.8 Example of SCI Receive Operation (8-Bit Data with Parity and One Stop Bit)

### 13.3.3 Multiprocessor Communication

The multiprocessor communication function enables several processors to share a single serial communication line. The processors communicate in asynchronous mode using a format with an additional multiprocessor bit (multiprocessor format).

In multiprocessor communication, each receiving processor is addressed by an ID. A serial communication cycle consists of an ID-sending cycle that identifies the receiving processor, and a data-sending cycle. The multiprocessor bit distinguishes ID-sending cycles from data-sending cycles.

The transmitting processor starts by sending the ID of the receiving processor with which it wants to communicate as data with the multiprocessor bit set to 1 . Next the transmitting processor sends transmit data with the multiprocessor bit cleared to 0 .

Receiving processors skip incoming data until they receive data with the multiprocessor bit set to 1. When they receive data with the multiprocessor bit set to 1 , receiving processors compare the data with their IDs. The receiving processor with a matching ID continues to receive further incoming data. Processors with IDs not matching the received data skip further incoming data until they again receive data with the multiprocessor bit set to 1 . Multiple processors can send and receive data in this way.

Figure 13.9 shows an example of communication among different processors using a multiprocessor format.

Communication Formats: Four formats are available. Parity-bit settings are ignored when a multiprocessor format is selected. For details see table 13.10.

Clock: See the description of asynchronous mode.


Serial data


ID-sending cycle: receiving processor address

Legend
MPB: Multiprocessor bit
Figure 13.9 Example of Communication among Processors using Multiprocessor Format (Sending Data H'AA to Receiving Processor A)

## Transmitting and Receiving Data

- Transmitting Multiprocessor Serial Data

Figure 13.10 shows a sample flowchart for transmitting multiprocessor serial data and indicates the procedure to follow.


Figure 13.10 Sample Flowchart for Transmitting Multiprocessor Serial Data

In transmitting serial data, the SCI operates as follows.

1. The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
2. After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit in SCR is set to 1 , the SCI requests a transmit-data-empty interrupt (TXI) at this time.
Serial transmit data is transmitted in the following order from the TxD pin:
a. Start bit: One 0 bit is output.
b. Transmit data: 7 or 8 bits are output, LSB first.
c. Multiprocessor bit: One multiprocessor bit (MPBT value) is output.
d. Stop bit: One or two 1 bits (stop bits) are output.
e. Mark state: Output of 1 bits continues until the start bit of the next transmit data.
3. The SCI checks the TDRE flag when it outputs the stop bit. If the TDRE flag is 0 , the SCI loads data from TDR into TSR, outputs the stop bit, then begins serial transmission of the next frame. If the TDRE flag is 1 , the SCI sets the TEND flag in SSR to 1 , outputs the stop bit, then continues output of 1 bits in the mark state. If the TEIE bit is set to 1 in SCR, a transmit-end interrupt (TEI) is requested at this time.

Figure 13.11 shows an example of SCI transmit operation using a multiprocessor format.


Figure 13.11 Example of SCI Transmit Operation (8-Bit Data with Multiprocessor Bit and One Stop Bit)

- Receiving Multiprocessor Serial Data

Figure 13.12 shows a sample flowchart for receiving multiprocessor serial data and indicates the procedure to follow.


Figure 13.12 Sample Flowchart for Receiving Multiprocessor Serial Data (1)


Figure 13.12 Sample Flowchart for Receiving Multiprocessor Serial Data (2)
Figure 13.13 shows an example of SCI receive operation using a multiprocessor format.


Figure 13.13 Example of SCI Receive Operation (8-Bit Data with Multiprocessor Bit and One Stop Bit)

### 13.3.4 Synchronous Operation

In synchronous mode, the SCI transmits and receives data in synchronization with clock pulses. This mode is suitable for high-speed serial communication.

The SCI transmitter and receiver share the same clock but are otherwise independent, so full duplex communication is possible. The transmitter and receiver are also double buffered, so continuous transmitting or receiving is possible by reading or writing data while transmitting or receiving is in progress.

Figure 13.14 shows the general format in synchronous serial communication.


Figure 13.14 Data Format in Synchronous Communication
In synchronous serial communication, each data bit is placed on the communication line from one falling edge of the serial clock to the next. Data is guaranteed valid at the rise of the serial clock. In each character, the serial data bits are transmitted in order from LSB (first) to MSB (last). After output of the MSB, the communication line remains in the state of the MSB. In synchronous mode the SCI receives data by synchronizing with the rise of the serial clock.

Communication Format: The data length is fixed at 8 bits. No parity bit or multiprocessor bit can be added.

Clock: An internal clock generated by the on-chip baud rate generator or an external serial clock input from the SCK pin can be selected by means of the C/A bit in SMR and bits CKE1 and CKE0 in SCR. See table 13.9 for details of SCI clock source selection.

When the SCI operates on an internal clock, the serial clock is output from the SCK pin.
Eight serial clock pulses are output per transmitted or received character. When the SCI is not transmitting or receiving, the clock signal remains in the high state. However, in a receive-only operation, the serial clock is output until an overrun error occurs or the RE bit is cleared to 0 . For character-by-character reception, an external clock should be selected as the clock source.

Rev. 2.0, 03/01, page 484 of 822

## Transmitting and Receiving Data

- SCI Initialization (Synchronous Mode)

Before transmitting or receiving, clear the TE and RE bits to 0 in SCR, then initialize the SCI as follows.

When changing the communication mode or format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing the TE bit to 0 sets the TDRE flag to 1 and initializes TSR. Clearing the RE bit to 0 , however, does not initialize the RDRF, PER, FER, and ORE flags and RDR, which retain their previous contents.
Figure 13.15 is a sample flowchart for initializing the SCI.


Figure 13.15 Sample Flowchart for SCI Initialization

- Transmitting Serial Data (Synchronous Mode)

Figure 13.16 shows a sample flowchart for transmitting serial data and indicates the procedure to follow.


1. SCI initialization: the transmit data output function of the TxD pin is selected automatically. After setting TE bit to 1 , output 1 from frame one transmission is possible.
2. SCI status check and transmit data write: read SSR, check that the TDRE flag is 1 , then write transmit data in TDR and clear the TDRE flag to 0 .
3. To continue transmitting serial data: after checking that the TDRE flag is 1 , indicating that data can be written, write data in TDR, then clear the TDRE flag to 0 . When the DMAC is activated by a transmit-data-empty interrupt request (TXI) to write data in TDR, the TDRE flag is checked and cleared automatically.

Figure 13.16 Sample Flowchart for Serial Transmitting

In transmitting serial data, the SCI operates as follows.

1. The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
2. After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit is set to 1 in SCR, the SCI requests a transmit-data-empty interrupt (TXI) at this time.
If clock output is selected, the SCI outputs eight serial clock pulses. If an external clock source is selected, the SCI outputs data in synchronization with the input clock. Data is output from the TxD pin in order from LSB (bit 0) to MSB (bit 7).
3. The SCI checks the TDRE flag when it outputs the MSB (bit 7). If the TDRE flag is 0 , the SCI loads data from TDR into TSR and begins serial transmission of the next frame. If the TDRE flag is 1 , the SCI sets the TEND flag to 1 in SSR, and after transmitting the MSB, holds the TxD pin in the MSB state. If the TEIE bit in SCR is set to 1 , a transmit-end interrupt (TEI) is requested at this time.
4. After the end of serial transmission, the SCK pin is held in a constant state.

Figure 13.17 shows an example of SCI transmit operation.


Figure 13.17 Example of SCI Transmit Operation

- Receiving Serial Data

Figure 13.18 shows a sample flowchart for receiving serial data and indicates the procedure to follow. When switching from asynchronous mode to synchronous mode, make sure that the ORER, PER, and FER flags are cleared to 0 . If the FER or PER flag is set to 1 the RDRF flag will not be set and both transmitting and receiving will be disabled.


Figure 13.18 Sample Flowchart for Serial Receiving (1)


Figure 13.18 Sample Flowchart for Serial Receiving (2)
In receiving, the SCI operates as follows.

1. The SCI synchronizes with serial clock input or output and initializes internally.
2. Receive data is stored in RSR in order from LSB to MSB.

After receiving the data, the SCI checks that the RDRF flag is 0 so that receive data can be transferred from RSR to RDR. If this check passes, the RDRF flag is set to 1 and the received data is stored in RDR. If the check does not pass (receive error), the SCI operates as indicated in table 13.11.
When a receive error has been identified in the error check, subsequent transmit and receive operations are disabled.
3. After setting the RDRF flag to 1 , if the RIE bit is set to 1 in SCR, the SCI requests a receive-data-full interrupt (RXI). If the ORER flag is set to 1 and the RIE bit in SCR is also set to 1 , the SCI requests a receive-error interrupt (ERI).

Figure 13.19 shows an example of SCI receive operation.

Serial clock


Serial data


Figure 13.19 Example of SCI Receive Operation

- Transmitting and Receiving Serial Data Simultaneously (Synchronous Mode)

Figure 13.20 shows a sample flowchart for transmitting and receiving serial data simultaneously and indicates the procedure to follow.


1. SCI initialization: the transmit data output function of the TxD pin and receive data input function of the RxD pin are selected, enabling simultaneous transmitting and receiving.
2. SCl status check and transmit data write: read SSR, check that the TDRE flag is 1 , then write transmit data in TDR and clear the TDRE flag to 0 . Notification that the TDRE flag has changed from 0 to 1 can also be given by the TXI interrupt.
3. Receive error handling: if a receive error occurs, read the ORER flag in SSR, then after executing the necessary error handling, clear the ORER flag to 0 . Neither transmitting nor receiving can resume while the ORER flag remains set to 1 .
4. SCI status check and receive data read: read SSR, check that the RDRF flag is 1 , then read receive data from RDR and clear the RDRF flag to 0 . Notification that the RDRF flag has changed from 0 to 1 can also be given by the RXI interrupt.
5. To continue transmitting and receiving serial data: check the RDRF flag, read RDR, and clear the RDRF flag to 0 before the MSB (bit 7) of the current frame is received. Also check that the TDRE flag is set to 1 , indicating that data can be written, write data in TDR, then clear the TDRE flag to 0 before the MSB (bit 7) of the current frame is transmitted. When the DMAC is activated by a transmit-data-empty interrupt request (TXI) to write data in TDR, the TDRE flag is checked and cleared automatically. When the DMA C is activated by a receive-data-full interrupt request (RXI) to read RDR, the RDRF flag is cleared automatically.

Note: When switching from transmitting or receiving to simultaneous transmitting and receiving, clear both the TE bit and the RE bit to 0 , then set both bits to 1 .

Figure 13.20 Sample Flowchart for Serial Transmitting

### 13.4 SCI Interrupts

The SCI has four interrupt request sources: TEI (transmit-end interrupt), ERI (receive-error interrupt), RXI (receive-data-full interrupt), and TXI (transmit-data-empty interrupt). Table 13.12 lists the interrupt sources and indicates their priority. These interrupts can be enabled and disabled by the TIE, TEIE, and RIE bits in SCR. Each interrupt request is sent separately to the interrupt controller.

The TXI interrupt is requested when the TDRE flag is set to 1 in SSR. The TEI interrupt is requested when the TEND flag is set to 1 in SSR. The TXI interrupt request can activate the DMAC to transfer data. Data transfer by the DMAC automatically clears the TDRE flag to 0 . The TEI interrupt request cannot activate the DMAC.

The RXI interrupt is requested when the RDRF flag is set to 1 in SSR. The ERI interrupt is requested when the ORER, PER, or FER flag is set to 1 in SSR. The RXI interrupt request can activate the DMAC to transfer data. Data transfer by the DMAC automatically clears the RDRF flag to 0 . The ERI interrupt request cannot activate the DMAC.

The DMAC can be activated by interrupts from SCI channel 0 .
Table 13.12 SCI Interrupt Sources

| Interrupt | Description | Priority |
| :--- | :--- | :--- |
| ERI | Receive error (ORER, FER, or PER) | High |
| RXI | Receive data register full (RDRF) |  |
| TXI | Transmit data register empty (TDRE) |  |
| TEI | Transmit end (TEND) | Low |

### 13.5 Usage Notes

Note the following points when using the SCI.
TDR Write and TDRE Flag: The TDRE flag in SSR is a status flag indicating the loading of transmit data from TDR into TSR. The SCI sets the TDRE flag to 1 when it transfers data from TDR to TSR.

Data can be written into TDR regardless of the state of the TDRE flag. If new data is written in TDR when the TDRE flag is 0 , the old data stored in TDR will be lost because this data has not yet been transferred to TSR. Before writing transmit data in TDR, be sure to check that the TDRE flag is set to 1 .

Simultaneous Multiple Receive Errors: Table 13.13 indicates the state of SSR status flags when multiple receive errors occur simultaneously. When an overrun error occurs the RSR contents are not transferred to RDR, so receive data is lost.

Table 13.13 SSR Status Flags and Transfer of Receive Data

|  |  |  | Receive Data <br> Transfer |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| RDRF | ORER | FER | PER | RSR $\rightarrow$ RDR | Receive Errors |
| 1 | 1 | 0 | 0 | $\times$ | Overrun error |
| 0 | 0 | 1 | 0 | O | Framing error |
| 0 | 0 | 0 | 1 | $O$ | Parity error |
| 1 | 1 | 1 | 0 | $\times$ | Overrun error + framing error |
| 1 | 1 | 0 | 1 | $\times$ | Overrun error + parity error |
| 0 | 0 | 1 | 1 | O | Framing error + parity error |
| 1 | 1 | 1 | 1 | $\times$ | Overrun error + framing error + parity error |

Notes: O: Receive data is transferred from RSR to RDR.
$x$ : Receive data is not transferred from RSR to RDR.

Break Detection and Processing: Break signals can be detected by reading the RxD pin directly when a framing error (FER) is detected. In the break state the input from the RxD pin consists of all 0s, so the FER flag is set and the parity error flag (PER) may also be set. In the break state the SCI receiver continues to operate, so if the FER flag is cleared to 0 it will be set to 1 again.

Sending a Break Signal: When the TE bit is cleared to 0 the TxD pin becomes an I/O port, the level and direction (input or output) of which are determined by DR and DDR bits. This feature can be used to send a break signal.

After the serial transmitter is initialized, the DR value substitutes for the mark state until the TE bit is set to 1 (the TxD pin function is not selected until the TE bit is set to 1 ). The DDR and DR bits should therefore both be set to 1 beforehand.

To send a break signal during serial transmission, clear the DR bit to 0 , then clear the TE bit to 0 . When the TE bit is cleared to 0 the transmitter is initialized, regardless of its current state, so the TxD pin becomes an output port outputting the value 0 .

Receive Error Flags and Transmitter Operation (Synchronous Mode Only): When a receive error flag (ORER, PER, or FER) is set to 1 the SCI will not start transmitting, even if the TDRE flag is cleared to 0 . Be sure to clear the receive error flags to 0 when starting to transmit. Note that clearing the RE bit to 0 does not clear the receive error flags to 0 .

Receive Data Sampling Timing in Asynchronous Mode and Receive Margin: In asynchronous mode the SCI operates on a base clock with 16 times the bit rate frequency. In receiving, the SCI synchronizes internally with the fall of the start bit, which it samples on the base clock. Receive data is latched at the rising edge of the eighth base clock pulse. See figure 13.21.


Figure 13.21 Receive Data Sampling Timing in Asynchronous Mode

The receive margin in asynchronous mode can therefore be expressed as in equation (1).

$$
\begin{equation*}
M=\left|\left(0.5-\frac{1}{2 N}\right)-(L-0.5) F-\frac{|D-0.5|}{N}(1+F)\right| \times 100 \% \tag{1}
\end{equation*}
$$

M: Receive margin (\%)
$\mathrm{N}:$ Ratio of clock frequency to bit rate $(\mathrm{N}=16)$
D: Clock duty cycle ( $\mathrm{D}=0$ to 1.0 )
L: Frame length ( $\mathrm{L}=9$ to 12)
F: Absolute deviation of clock frequency
From equation (1), if $\mathrm{F}=0$ and $\mathrm{D}=0.5$ the receive margin is $46.875 \%$, as given by equation (2).

$$
\begin{align*}
D & =0.5, F=0 \\
M & =\{0.5-1 /(2 \times 16)\} \times 100 \% \\
& =46.875 \% \ldots \ldots . . . . . . . . . . . . . . . . . . . . . . . . . . ~ \tag{2}
\end{align*}
$$

This is a theoretical value. A reasonable margin to allow in system designs is $20 \%$ to $30 \%$.
Restrictions on Usage of DMAC: To have the DMAC read RDR, be sure to select the SCI receive-data-full interrupt (RXI) as the activation source with bits DTS2 to DTS0 in DTCR.

Restrictions on Usage of the Serial Clock: When transmitting data using the serial clock as an external clock, after clearing SSR of TDRE, maintain the space between each frame of the lead of the transmission clock (start-up edge) at five states or more (see Figure 13.22). This condition is also needed for continuous transmission. If it is not fulfilled, operational error will occur.


Figure 13.22 Serial Clock Transmission (Example)

Switching SCK Pin to Port Output Pin Function in Synchronous Mode: When the SCK pin is used as the serial clock output in synchronous mode, and is then switched to its output port function at the end of transmission, a low level may be output for one half-cycle. Half-cycle lowlevel output occurs when SCK is switched to its port function with the following settings when $\mathrm{DDR}=1, \mathrm{DR}=1, \mathrm{C} / \overline{\mathrm{A}}=1, \mathrm{CKE} 1=0, \mathrm{CKE} 0=0$, and $\mathrm{TE}=1$.

1. End of serial data transmission
2. TE bit $=0$
3. $\mathrm{C} / \overline{\mathrm{A}}$ bit $=0 \ldots$ switchover to port outpu
4. Occurrence of low-level output (see figure 13.23)


Figure 13.23 Operation when Switching from SCK Pin Function to Port Pin Function

Sample Procedure for Avoiding Low-Level Output
As this sample procedure temporarily places the SCK pin in the input state, the SCK/port pin should be pulled up beforehand with an external circuit.

With $\mathrm{DDR}=1, \mathrm{DR}=1, \mathrm{C} / \overline{\mathrm{A}}=1, \mathrm{CKE} 1=0, \mathrm{CKE} 0=0$, and $\mathrm{TE}=1$, make the following settings in the order shown.

1. End of serial data transmission
2. TE bit $=0$
3. CKE 1 bit $=1$
4. $\mathrm{C} / \overline{\mathrm{A}}$ bit $=0 \ldots$ switchover to port output
5. CKE1 bit $=0$


Figure 13.24 Operation when Switching from SCK Pin Function to Port Pin Function (Example of Preventing Low-Level Output)

## Section 14 Smart Card Interface

### 14.1 Overview

As an extension of its serial communication interface functions, SCI0 supports a smart card (IC card) interface conforming to the ISO/IEC7816-3 (Identification Card) standard. Switchover between normal serial communication and the smart card interface is controlled by a register setting.

### 14.1.1 Features

Features of the smart-card interface supported by the $\mathrm{H} 8 / 3052 \mathrm{~F}$ are listed below.

- Asynchronous communication
— Data length: 8 bits
- Parity bits generated and checked
- Error signal output in receive mode (parity error)
- Error signal detect and automatic data retransmit in transmit mode
- Supports both direct convention and inverse convention
- Built-in baud rate generator with selectable bit rates
- Three types of interrupts

Transmit-data-empty, receive-data-full, and receive-error interrupts are requested independently. The transmit-data-empty and receive-data-full interrupts can activate the DMA controller (DMAC) to transfer data.

### 14.1.2 Block Diagram

Figure 14.1 shows a block diagram of the smart card interface.


Figure 14.1 Smart Card Interface Block Diagram

### 14.1.3 Pin Configuration

Table 14.1 lists the smart card interface pins.
Table 14.1 Smart Card Interface Pins

| Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Serial clock pin | $\mathrm{SCK}_{0}$ | Output | Clock output |
| Receive data pin | $\mathrm{RxD}_{0}$ | Input | Receive data input |
| Transmit data pin | $\mathrm{TxD}_{0}$ | Output | Transmit data output |

### 14.1.4 Register Configuration

The smart card interface has the internal registers listed in table 14.2. BRR, TDR, and RDR have their normal serial communication interface functions, as described in section 13, Serial Communication Interface.

Table 14.2 Registers

| Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: |
| H'FFB0 | Serial mode register | SMR | R/W | H'00 |
| H'FFB1 | Bit rate register | BRR | R/W | H'FF |
| H'FFB2 | Serial control register | SCR | R/W | H'00 |
| H'FFB3 | Transmit data register | TDR | R/W | H'FF |
| H'FFB4 | Serial status register | SSR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | F'84 |
| H'FFB5 | Receive data register | RDR | R | H'00 |
| H'FFB6 | Smart card mode register | SCMR | R/W | H'F2 |

Notes: 1. Lower 16 bits of the address.
2. Only 0 can be written, to clear flags.

### 14.2 Register Descriptions

This section describes the new or modified registers and bit functions in the smart card interface.

### 14.2.1 Smart Card Mode Register (SCMR)

SCMR is an 8-bit readable/writable register that selects smart card interface functions.


SCMR is initialized to H'F2 by a reset and in standby mode.
Bits 7 to 4—Reserved: Read-only bits, always read as 1.
Bit 3—Smart Card Data Transfer Direction (SDIR): Selects the serial/parallel conversion format.

| Bit 3: SDIR | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | TDR contents are transmitted LSB-first |  |
| 1 | Received data is stored LSB-first in RDR |  |
|  | TDR contents are transmitted MSB-first |  |

Bit 2-Smart Card Data Inverter (SINV): Inverts data logic levels. This function is used in combination with bit 3 to communicate with inverse-convention cards. SINV does not affect the logic level of the parity bit. For parity settings, see section 14.3.4, Register Settings.

| Bit 2: SINV | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | Unmodified TDR contents are transmitted |  |
| 1 | Received data is stored unmodified in RDR |  |
|  | Inverted TDR contents are transmitted |  |

Bit 1—Reserved: Read-only bit, always read as 1.
Bit 0—Smart Card Interface Mode Select (SMIF): Enables the smart card interface function.

| Bit 0: SMIF | Description |  |
| :--- | :--- | :--- |
| 0 | Smart card interface function is disabled | (Initial value) |
| 1 | Smart card interface function is enabled |  |

### 14.2.2 Serial Status Register (SSR)

The function of SSR bit 4 is modified in the smart card interface. This change also causes a modification to the setting conditions for bit 2 (TEND).

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TDRE | RDRF | ORER | ERS | PER | TEND | MPB | MPBT |
| Initial value | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| Read/Write | R/(W)* | R/(W)* | R/(W)* | R/(W)* | R/(W)* | R | R | R/W |
|  |  |  |  |  |  | ransmit <br> atus flag <br> d of tran | dicating ission |  |
|  |  |  | Error signal status (ERS) Status flag indicating that an error signal has been received |  |  |  |  |  |

Note: * Only 0 can be written, to clear the flag.

Bits 7 to 5: These bits operate as in normal serial communication. For details see section 13, Serial Communication Interface.

Bit 4-Error Signal Status (ERS): In smart card interface mode, this flag indicates the status of the error signal sent from the receiving device to the transmitting device. The smart card interface does not detect framing errors.

## Bit 4: ERS Description

| 0 | Indicates normal data transmission, with no error signal returned (Initial value) |
| :--- | :--- |
| [Clearing conditions] |  |
|  | The chip is reset or enters standby mode. |
|  | Software reads ERS while it is set to 1, then writes 0. |
| 1 | Indicates that the receiving device sent an error signal reporting a parity error |
|  | [Setting condition] |
| A low error signal was sampled. |  |

Note: Clearing the TE bit to 0 in SCR does not affect the ERS flag, which retains its previous value.

Bits 3 to 0: These bits operate as in normal serial communication. For details see section 13, Serial Communication Interface. The setting conditions for transmit end (TEND, bit 2), however, are modified as follows.

## Bit 2: TEND Description

$0 \quad$ Transmission is in progress
[Clearing conditions]
Software reads TDRE while it is set to 1 , then writes 0 in the TDRE flag.
The DMAC writes data in TDR.

| 1 | End of transmission |
| :--- | :--- |
| [Setting conditions] |  |
|  | The chip is reset or enters standby mode. |
|  | The TE bit and FER/ERS bit are both cleared to 0 in SCR. |
|  | TDRE is 1 and FER/ERS is 0 at a time 2.5 etu after the last bit of a 1 -byte |
| serial character is transmitted (normal transmission) |  |

Note: An etu (elementary time unit) is the time needed to transmit one bit.

### 14.2.3 Serial Mode Register (SMR)

Bit 7 of SMR has a different function in smart card interface mode. The related serial control register (SCR) changes from bit 1 to bit 0 .

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | GM | CHR | PR | O/E | STOP | MP | CKS1 | CKSO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Bit 7—GSM Mode (GM): Set at 0 when using the regular smart card interface. In GSM mode, set to 1 . When transmission is complete, initially the TEND flag set timing appears followed by clock output restriction mode. Clock output restriction mode comprises serial control register bit 1 and bit 0 .

## Bit 7: GM Description

0
Using the regular smart card interface mode
The TEND flag is set 12.5 etu after the beginning of the start bit (Initial value)
Clock output on/off control only
1
Using the GSM mode smart card interface mode
The TEND flag is set 11.0 etu after the beginning of the start bit
Clock output on/off and fixed-high/fixed-low control (set in SCR)

Bits 6 to 0-Operate in the same way as for the normal SCI.
For details, see section 13.2.5, Serial Mode Register (SMR).

### 14.2.4 Serial Control Register (SCR)

Bits 1 and 0 have different functions in smart card interface mode.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKEO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Bits 7 to 2 -Operate in the same way as for the normal SCI.
For details, see section 13.2.6, Serial Control Register (SCR).
Bits 1 and 0—Clock Enable (CKE1, CKE0): Setting enable or disable for the SCI clock selection and clock output from the SCK pin. In smart card interface mode, it is possible to switch between enabling and disabling of the normal clock output, and specify a fixed high level or fixed low level for the clock output.

| SMR | SCR |  | Description |
| :---: | :---: | :---: | :---: |
| Bit 7: GM | Bit 1: <br> CKE1 | Bit 0 : <br> CKEO |  |
| 0 | 0 | 0 | The internal clock/SCK0 pin functions as an I/O port (Initial value) |
| 0 | 0 | 1 | The internal clock/SCK0 pin functions as the clock output |
| 1 | 0 | 0 | The internal clock/SCK0 pin is fixed at low-level output |
| 1 | 0 | 1 | The internal clock/SCK0 pin functions as the clock output |
| 1 | 1 | 0 | The internal clock/SCK0 pin is fixed at high-level output |
| 1 | 1 | 1 | The internal clock/SCK0 pin functions as the clock output |

### 14.3 Operation

### 14.3.1 Overview

The main features of the smart-card interface are as follows.

- One frame consists of eight data bits and a parity bit.
- In transmitting, a guard time of at least two elementary time units (2 etu) is provided between the end of the parity bit and the start of the next frame. (An elementary time unit is the time required to transmit one bit.)
- In receiving, if a parity error is detected, a low error signal is output for 1 etu, beginning 10.5 etu after the start bit.
- In transmitting, if an error signal is received, after at least 2 etu, the same data is automatically transmitted again.
- Only asynchronous communication is supported. There is no synchronous communication function.


### 14.3.2 Pin Connections

Figure 14.2 shows a pin connection diagram for the smart card interface.
In communication with a smart card, data is transmitted and received over the same signal line. The $\mathrm{TxD}_{0}$ and $\mathrm{RxD}_{0}$ pins should both be connected to this line. The data transmission line should be pulled up to $\mathrm{V}_{\mathrm{CC}}$ through a resistor.

If the smart card uses the clock generated by the smart card interface, connect the $\mathrm{SCK}_{0}$ output pin to the card's CLK input. If the card uses its own internal clock, this connection is unnecessary.

The reset signal should be output from one of the H8/3052F' generic ports.
In addition to these pin connections, power and ground connections will normally also be necessary.


Figure 14.2 Smart Card Interface Connection Diagram
Note: A loop-back test can be performed by setting both RE and TE to 1 without connecting a smart card.

### 14.3.3 Data Format

Figure 14.3 shows the data format of the smart card interface. In receive mode, parity is checked once per frame. If a parity error is detected, an error signal is returned to the transmitting device to request retransmission. In transmit mode, the error signal is sampled and the same data is retransmitted if the error signal is low.


Figure 14.3 Smart Card Interface Data Format
The operating sequence is as follows.

1. When not in use, the data line is in the high-impedance state, and is pulled up to the high level through a resistor.
2. To start transmitting a frame of data, the transmitting device transmits a low start bit (Ds), followed by eight data bits (D0 to D7) and a parity bit (Dp).
3. Next, in the smart card interface, the transmitting device returns the data line to the highimpedance state. The data line is pulled up to the high level through a resistor.
4. The receiving device performs a parity check. If there is no parity error, the receiving device waits to receive the next data. If a parity error is present, the receiving device outputs a low error signal (DE) to request retransmission of the data. After outputting the error signal for a designated interval, the receiving device returns the signal line to the high-impedance state. The signal line is pulled back up to the high level through the pull-up resistor.
5. If the transmitting device does not receive an error signal, it proceeds to transmit the next data. If it receives an error signal, it returns to step 2 and transmits the same data again.

### 14.3.4 Register Settings

Table 14.3 shows a bit map of the registers used in the smart card interface. Bits indicated as 0 or 1 should always be set to the indicated value. The settings of the other bits will be described in this section.

Table 14.3 Register Settings in Smart Card Interface

| Register | Address* ${ }^{1}$ | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMR | H'FFB0 | GM | 0 | 1 | O/E | 1 | 0 | CKS1 | CKS0 |
| BRR | H'FFB1 | BRR7 | BRR6 | BRR5 | BRR4 | BRR3 | BRR2 | BRR1 | BRRO |
| SCR | H'FFB2 | TIE | RIE | TE | RE | 0 | 0 | CKE1*2 | CKE0 |
| TDR | H'FFB3 | TDR7 | TDR6 | TDR5 | TDR4 | TDR3 | TDR2 | TDR1 | TDR0 |
| SSR | H'FFB4 | TDRE | RDRF | ORER | ERS | PER | TEND | 0 | 0 |
| RDR | H'FFB5 | RDR7 | RDR6 | RDR5 | RDR4 | RDR3 | RDR2 | RDR1 | RDR0 |
| SCMR | H'FFB6 | - | - | - | - | SDIR | SINV | - | SMIF |

Notes: - Unused bit.

1. Lower 16 bits of the address.
2. When the GM of the SMR is set at 0 , be sure the CKE1 bit is 0 .

Serial Mode Register (SMR) Settings: In regular smart card interface mode, set the GM bit at 0 . In regular smart card mode, clear the GM bit to 0 . In GSM mode, set the GM bit to 1 . Clear the $\mathrm{O} / \overline{\mathrm{E}}$ bit to 0 if the smart card uses the direct convention. Set the O/ $\overline{\mathrm{E}}$ bit to 1 if the smart card uses the inverse convention. Bits CKS1 and CKS0 select the clock source of the built-in baud rate generator. See section 14.3.5, Clock.

Bit Rate Register (BRR) Settings: This register sets the bit rate. Equations for calculating the setting are given in section 14.3.5, Clock.

Serial Control Register (SCR): The TIE, RIE, TE, and RE bits have their normal serial communication functions. For details, see section 13, Serial Communication Interface. The CKE1 and CKEO bits select clock output. When the GM bit of the SMR is cleared to 0 , to disable clock output, clear this bit to 00 . To enable clock output, set this bit to 01 . When the GM bit of the SMR is set to 1 , clock output is enabled. Clock output is fixed at high or low.

Smart Card Mode Register (SCMR): If the smart card follows the direct convention, clear the SDIR and SINV bits to 0 . If the smart card follows the indirect convention, set the SDIR and SINV bits to 1 . To use the smart card interface, set the SMIF bit to 1 .

The register settings and examples of starting character waveforms are shown below for two smart cards, one following the direct convention and one the inverse convention.

- Direct convention (SDIR $=\mathrm{SINV}=\mathrm{O} / \overline{\mathrm{E}}=0)$


In the direct convention, state Z corresponds to logic level 1 , and state A to logic level 0 . Characters are transmitted and received LSB-first. In the example above the first character data is $H^{\prime} 3 \mathrm{~B}$. The parity bit is 1 , following the even parity rule designated for smart cards.

- Inverse convention $(\mathrm{SDIR}=\mathrm{SINV}=\mathrm{O} / \overline{\mathrm{E}}=1)$


In the inverse convention, state A corresponds to the logic level 1 , and state $Z$ to the logic level 0. Characters are transmitted and received MSB-first. In the example above the first character data is $\mathrm{H}^{\prime} 3 \mathrm{~F}$. Following the even parity rule designated for smart cards, the parity bit logic level is 0 , corresponding to state $Z$.

In the $\mathrm{H} 8 / 3052 \mathrm{~F}$, the SINV bit inverts only the data bits D 7 to D 0 . The parity bit is not inverted, so the $\mathrm{O} / \overline{\mathrm{E}}$ bit in SMR must be set to odd parity mode. This applies in both transmitting and receiving.

### 14.3.5 Clock

As its serial communication clock, the smart card interface can use only the internal clock generated by the on-chip baud rate generator. The bit rate can be selected by setting the bit rate register (BRR) and bits CKS1 and CKS0 in the serial mode register (SMR). The bit rate can be calculated from the equation given below. Table 14.5 lists some examples of bit rate settings.

If bit CKE0 is set to 1 , a clock signal with a frequency equal to 372 times the bit rate is output from the $\mathrm{SCK}_{0}$ pin.

$$
B=\frac{\phi}{1488 \times 2^{2 n-1} \times(N+1)} \times 10^{6}
$$

where, N : BRR setting $(0 \leq \mathrm{N} \leq 255)$
B: Bit rate (bits/s)
$\phi$ : System clock frequency $(\mathrm{MHz})^{*}$
n: See table 14.4
Table 14.4 n-Values of CKS1 and CKS0 Settings

| $\mathbf{n}$ | CKS1 | CKSO |
| :--- | :--- | :--- |
| 0 | 0 | 0 |
| 1 | 0 | 1 |
| 2 | 1 | 0 |
| 3 | 1 | 1 |

Note: * If the gear function is used to divide the system clock frequency, use the divided frequency to calculate the bit rate. The equation above applies directly to $1 / 1$ frequency division.

Table 14.5 Bit Rates (bits/s) for Different BRR Settings (when $\mathbf{n}=\mathbf{0}$ )
$\phi(\mathrm{MHz})$

| $\mathbf{N}$ | $\mathbf{7 . 1 4 2 4}$ | $\mathbf{1 0 . 0 0}$ | $\mathbf{1 0 . 7 1 3 6}$ | $\mathbf{1 3 . 0 0}$ | $\mathbf{1 4 . 2 8 4 8}$ | $\mathbf{1 6 . 0 0}$ | $\mathbf{1 8 . 0 0}$ | $\mathbf{2 0 . 0 0}$ | $\mathbf{2 5 . 0 0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 9600.0 | 13440.9 | 14400.0 | 17473.1 | 19200.0 | 21505.4 | 24193.5 | 26881.7 | 33602.2 |
| 1 | 4800.0 | 6720.4 | 7200.0 | 8736.6 | 9600.0 | 10752.7 | 12096.8 | 13440.9 | 16801.1 |
| 2 | 3200.0 | 4480.3 | 4800.0 | 5824.4 | 6400.0 | 7168.5 | 8064.5 | 8960.6 | 11200.7 |

Note: Bit rates are rounded off to one decimal place.

The following equation calculates the bit rate register (BRR) setting from the system clock frequency and bit rate. N is an integer from 0 to 255 , specifying the value with the smaller error.

$$
N=\frac{\phi}{1488 \times 2^{2 n-1} \times B} \times 10^{6}-1
$$

Table 14.6 BRR Settings for Typical Bit Rate (bits/s) (when $\mathbf{n}=0)$

|  | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 7.1424 | 10.00 | 10.7136 | 13.00 | 14.2848 | 16.00 | 18.00 | 20.00 | 25.00 |
| Bit/s | N Error | N Error | N Error | N Error | N Error | N Error | N Error | N Error | N Error |
| 9600 | 00.00 | 130.00 | 125.00 | 18.99 | 10.00 | 112.01 | 215.99 | 26.66 | 312.49 |


| Table 14.7 | Maximum Bit Rates for Various Frequencies (Smart Card Interface) |  |  |
| :--- | :--- | :--- | :--- |
| $\boldsymbol{\phi}(\mathbf{M H z})$ | Maximum Bit Rate (bits/s) | $\mathbf{N}$ | $\mathbf{n}$ |
| 7.1424 | 9600 | 0 | 0 |
| 10 | 13441 | 0 | 0 |
| 10.7136 | 14400 | 0 | 0 |
| 13 | 17473 | 0 | 0 |
| 14.2848 | 19200 | 0 | 0 |
| 16 | 21505 | 0 | 0 |
| 18 | 24194 | 0 | 0 |
| 20 | 26882 | 0 | 0 |
| 25 | 33602 | 0 | 0 |

The bit rate error is calculated from the following equation.

$$
\operatorname{Error}(\%)=\left\{\frac{\phi}{1488 \times 2^{2 n-1} \times B \times(N+1)} \times 10^{6}-1\right\} \times 100
$$

### 14.3.6 Transmitting and Receiving Data

Initialization: Before transmitting or receiving data, initialize the smart card interface by the procedure below. Initialization is also necessary when switching from transmit mode to receive mode or from receive mode to transmit mode.

1. Clear the TE and RE bits to 0 in the serial control register (SCR).
2. Clear the ERS, PER, and ORER error flags to 0 in the serial status register (SSR).
3. Set the parity mode bit ( $\mathrm{O} / \overline{\mathrm{E}}$ ) and baud rate generator clock source select bits (CKS1 and CKSO) as required in the serial mode register (SMR). At the same time, clear the C/ $\overline{\mathrm{A}}, \mathrm{CHR}$, and MP bits to 0 , and set the STOP and PE bits to 1 .
4. Set the SMIF, SDIR, and SINV bits as required in the smart card mode register (SMR). When the SMIF bit is set to 1 , the $\mathrm{TxD}_{0}$ and $\mathrm{RxD}_{0}$ pins switch from their I/O port functions to their serial communication interface functions, and are placed in the high-impedance state.
5. Set a value corresponding to the desired bit rate in the bit rate register (BRR).
6. Set clock enable bit 0 (CKE0) as required in the serial control register (SCR). Write 0 in the TIE, RIE, TE, RE, MPIE, TEIE, and CKE1 bits. If bit CKE0 is set to 1 , a serial clock will be output from the $\mathrm{SCK}_{0}$ pin.
7. Wait for at least the interval required to transmit or receive one bit, then set the TIE, RIE, TE, and RE bits as necessary in SCR. Do not set TE and RE both to 1, except when performing a loop-back test.

Transmitting Serial Data: The transmitting procedure in smart card mode is different from the normal SCI procedure, because of the need to sample the error signal and retransmit. Figure 14.4 shows a flowchart for transmitting, and figure 14.5 shows the relation between a transmit operation and the internal registers.

1. Initialize the smart card interface by the procedure given above in Initialization.
2. Check that the ERS error flag is cleared to 0 in SSR.
3. Check that the TEND flag is set to 1 in SSR. Repeat steps 2 and 3 until this check passes.
4. Write transmit data in TDR and clear the TDRE flag to 0 . The data will be transmitted and the TEND flag will be cleared to 0 .
5. To continue transmitting data, return to step 2 .
6. To terminate transmission, clear the TE bit to 0 .

This procedure may include interrupt handling and DMA transfer.
If the TIE bit is set to 1 to enable interrupt requests, when transmission is completed and the TEND flag is set to 1 , a transmit-data-empty interrupt (TXI) is requested. If the RIE bit is set to 1 to enable interrupt requests, when a transmit error occurs and the ERS flag is set to 1, a transmit/receive-error interrupt (ERI) is requested.

The timing of TEND flag setting depends on the GM bit in SMR. The timing is shown in figure 14.6 .

If the TXI interrupt activates the DMAC, the number of bytes designated in the DMAC can be transmitted automatically, including automatic retransmit.

For details, see Interrupt Operations and Data Transfer by DMAC in this section.


Figure 14.4 Transmit Flowchart (Example)


#### Abstract

TDR TSR (shift register) (1) Data write  (2) Transfer from TDR to TSR  ; Data remains in TDR (3) Serial data output  $\xrightarrow{\text { Data } 1}$ I/O signal line output

In case of normal transmission: TEND flag is set In case of transmit error:

ERS flag is set Steps (2) and (3) above are repeated until the TEND flag is set

Note: When the ERS flag is set, it should be cleared until transfer of the last bit (D7 in LSB-first transmission, D0 in MSB-first transmission) of the next transfer data to be transmitted has been completed.


Figure 14.5 Relation Between Transmit Operation and Internal Registers


Figure 14.6 TEND Flag Occurrence Timing

Receiving Serial Data: The receiving procedure in smart card mode is the same as the normal SCI procedure. Figure 14.7 shows a flowchart for receiving.

1. Initialize the smart card interface by the procedure given in Initialization at the beginning of this section.
2. Check that the ORER and PER error flags are cleared to 0 in SSR. If either flag is set, carry out the necessary error handling, then clear both the ORER and PER flags to 0 .
3. Check that the RDRF flag is set to 1 . Repeat steps 2 and 3 until this check passes.
4. Read receive data from RDR.
5. To continue receiving data, clear the RDRF flag to 0 and return to step 2 .
6. To terminate receiving, clear the RE bit to 0 .


Figure 14.7 Receive Flowchart (Example)

This procedure may include interrupt handling and DMA transfer.

If the RIE bit is set to 1 to enable interrupt requests, when receiving is completed and the RDRF flag is set to 1 , a receive-data-full interrupt (RXI) is requested. If a receive error occurs, either the ORER or PER flag is set to 1 and a transmit/receive-error interrupt (ERI) is requested.

If the RXI interrupt activates the DMAC, the number of bytes designated in the DMAC will be transferred, skipping receive data in which an error occurred.

For details, see Interrupt Operations and Data Transfer by DMAC below.

When a parity error occurs and PER is set to 1 , the receive data is transferred to RDR, so the erroneous data can be read.

Switching Modes: To switch from receive mode to transmit mode, check that receiving operations have completed, then initialize the smart card interface, clearing RE to 0 and setting TE to 1 . Completion of receive operations is indicated by the RDRF, PER, or ORER flag.

To switch from transmit mode to receive mode, check that transmitting operations have completed, then initialize the smart card interface, clearing TE to 0 and setting RE to 1 . Completion of transmit operations can be verified from the TEND flag.

Fixing Clock Output: When the GM bit of the SMR is set to 1 , clock output is fixed by CKE1 and CKE0 of SCR. In this case, the clock pulse can be set at minimum value.

Figure 14.8 shows clock output fixed timing: CKE0 is restricted with $\mathrm{GM}=1$ and $\mathrm{CKE} 1=1$.


Figure 14.8 Clock Output Fixed Timing

Interrupt Operations: The smart card interface has three interrupt sources: transmit-data-empty (TXI), transmit/receive-error (ERI), and receive-data-full (RXI). The transmit-end interrupt request (TEI) is not available in smart card mode.

A TXI interrupt is requested when the TEND flag is set to 1 in SSR. An RXI interrupt is requested when the RDRF flag is set to 1 in SSR. An ERI interrupt is requested when the ORER, PER, or ERS flag is set to 1 in SSR. These relationships are shown in table 14.8.

## Table 14.8 Smart Card Mode Operating States and Interrupt Sources

| Operating State |  | Flag | Mask Bit | Interrupt <br> Source | DMAC <br> Activation |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Transmit mode | Normal operation | TEND | TIE | TXI | Available |
|  | Error | ERS | RIE | ERI | Not available |
| Receive mode | Normal operation | RDRF | RIE | RXI | Available |
|  | Error | PER, ORER | RIE | ERI | Not available |

Data Transfer by DMAC: The DMAC can be used to transmit and receive in smart card mode, as in normal SCI operations. In transmit mode, when the TEND flag is set to 1 in SSR, the TDRE flag is set simultaneously, generating a TXI interrupt. If TXI is designated in advance as a DMAC activation source, the DMAC will be activated by the TXI request and will transfer the next transmit data. This data transfer by the DMAC automatically clears the TDRE and TEND flags to 0 . When an error occurs, the SCI automatically retransmits the same data, keeping TEND cleared to 0 so that the DMAC is not activated. The SCI and DMAC will therefore automatically transmit the designated number of bytes, including retransmission when an error occurs. When an error occurs the ERS flag is not cleared automatically, so the RIE bit should be set to 1 to enable the error to generate an ERI request, and the ERI interrupt handler should clear ERS.

When using the DMAC to transmit or receive, first set up and enable the DMAC, then make SCI settings. DMAC settings are described in section 8, DMA Controller.

In receive operations, when the RDRF flag is set to 1 in SSR, an RXI interrupt is requested. If RXI is designated in advance as a DMAC activation source, the DMAC will be activated by the RXI request and will transfer the received data. This data transfer by the DMAC automatically clears the RDRF flag to 0 . When an error occurs, the RDRF flag is not set and an error flag is set instead. The DMAC is not activated. The ERI interrupt request is directed to the CPU. The ERI interrupt handler should clear the error flags.

Examples of Operation in GSM Mode: When switching between smart card interface mode and software standby mode, use the following procedures to maintain the clock duty cycle.

- Switching from smart card interface mode to software standby mode

1. Set the $\mathrm{P} 9_{4}$ data register (DR) and data direction register (DDR) to the values for the fixed output state in software standby mode.
2. Write 0 to the TE and RE bits in the serial control register (SCR) to stop transmit/receive operations. At the same time, set the CKE1 bit to the value for the fixed output state in software standby mode.
3. Write 0 to the CKE0 bit in SCR to stop the clock.
4. Wait for one serial clock cycle. During this period, the duty cycle is preserved and clock output is fixed at the specified level.
5. Write $\mathrm{H}^{\prime} 00$ to the serial mode register (SMR) and smart card mode register (SCMR).
6. Make the transition to the software standby state.

- Returning from software standby mode to smart card interface mode

1. Clear the software standby state.
2. Set the CKE1 bit in SCR to the value for the fixed output state at the start of software standby (the current P94 pin state).
3. Set smart card interface mode and output the clock. Clock signal generation is started with the normal duty cycle.


Figure 14.9 Procedure for Stopping and Restarting the Clock
Use the following procedure to secure the clock duty cycle after powering on.

1. The initial state is port input and high impedance. Use pull-up or pull-down resistors to fix the potential.
2. Fix at the output specified by the CKE1 bit in SCR.
3. Set SMR and SCMR, and switch to smart card interface mode operation.
4. Set the CKE0 bit in SCR to 1 to start clock output.

### 14.4 Usage Notes

When using the SCI as a smart card interface, note the following points.
Receive Data Sampling Timing in Smart Card Mode and Receive Margin: In smart card mode the SCI operates on a base clock with 372 times the bit rate frequency. In receiving, the SCI synchronizes internally with the fall of the start bit, which it samples on the base clock. Receive data is latched at the rising edge of the 186th base clock pulse. See figure 14.10.


Figure 14.10 Receive Data Sampling Timing in Smart Card Mode

The receive margin can therefore be expressed as follows.
Receive margin in smart card mode:
$M=\left|\left\{0.5-\frac{1}{2 N}\right\}-(L-0.5) F-\frac{|D-0.5|}{N}(1+F)\right| \times 100 \%$
M: Receive margin (\%)
N : Ratio of clock frequency to bit rate $(\mathrm{N}=372)$
D: Clock duty cycle ( $\mathrm{D}=0$ to 1.0)
L: Frame length ( $L=10$ )
F: Absolute deviation of clock frequency
From this equation, if $\mathrm{F}=0$ and $\mathrm{D}=0.5$ the receive margin is as follows.

$$
\begin{aligned}
\mathrm{D} & =0.5, \mathrm{~F}=0 \\
\mathrm{M} & =\{0.5-1 /(2 \times 372)\} \times 100 \% \\
& =49.866 \%
\end{aligned}
$$

Retransmission: Retransmission is described below for the separate cases of transmit mode and receive mode.

- Retransmission when SCI is in Receive Mode (See Figure 14.11)

1. The SCI checks the received parity bit. If it detects an error, it automatically sets the PER flag to 1 . If the RIE bit in SCR is set to the enable state, an ERI interrupt is requested. The PER flag should be cleared to 0 in SSR before the next parity bit sampling timing.
2. The RDRF bit in SSR is not set to 1 for the error frame.
3. If an error is not detected when the parity bit is checked, the PER flag is not set in SSR.
4. If an error is not detected when the parity bit is checked, receiving operations are assumed to have ended normally, and the RDRF bit is automatically set to 1 in SSR. If the RIE bit in SCR is set to the enable state, an RXI interrupt is requested. If RXI is enabled as a DMA transfer activation source, the RDR contents can be read automatically. When the DMAC reads the RDR data, it automatically clears RDRF to 0 .
5. When a normal frame is received, at the error signal transmit timing, the data pin is held in the high-impedance state.


Figure 14.11 Retransmission in SCI Receive Mode

- Retransmission when SCI is in Transmit Mode (See Figure 14.12)

6. After transmitting one frame, if the receiving device returns an error signal, the SCI sets the ERS flag to 1 in SSR. If the RIE bit in SCR is set to the enable state, an ERI interrupt is requested. The ERS flag should be cleared to 0 in SSR before the next parity bit sampling timing.
7. The TEND bit in SSR is not set for the frame in which the error signal was received, indicating an error.
8. If no error signal is returned from the receiving device, the ERS flag is not set in SSR.
9. If no error signal is returned from the receiving device, transmission of the frame, including retransmission, is assumed to be complete, and the TEND bit is set to 1 in SSR. If the TIE bit in SCR is set to the enable state, a TXI interrupt is requested. If TXI is enabled as a DMA transfer activation source, the next data can be written in TDR automatically. When the DMAC writes data in TDR, it automatically clears the TDRE bit to 0 .


Figure 14.12 Retransmission in SCI Transmit Mode

## Section 15 A/D Converter

### 15.1 Overview

The H8/3052F includes a 10-bit successive-approximations A/D converter with a selection of up to eight analog input channels.

When the $\mathrm{A} / \mathrm{D}$ converter is not used, it can be halted independently to conserve power. For details see section 20.6, Module Standby Function.

### 15.1.1 Features

A/D converter features are listed below.

- 10-bit resolution
- Eight input channels
- Selectable analog conversion voltage range

The analog voltage conversion range can be programmed by input of an analog reference voltage at the $V_{\text {REF }}$ pin.

- High-speed conversion

Conversion time: maximum $5.4 \mu$ ser channel (with 25 MHz system clock)

- Two conversion modes

Single mode: A/D conversion of one channel
Scan mode: continuous conversion on one to four channels

- Four 16-bit data registers

A/D conversion results are transferred for storage into data registers corresponding to the channels.

- Sample-and-hold function
- A/D conversion can be externally triggered
- A/D interrupt requested at end of conversion

At the end of $A / D$ conversion, an $A / D$ end interrupt (ADI) can be requested.

### 15.1.2 Block Diagram

Figure 15.1 shows a block diagram of the $\mathrm{A} / \mathrm{D}$ converter.


Figure 15.1 A/D Converter Block Diagram

### 15.1.3 Pin Configuration

Table 15.1 summarizes the $A / D$ converter's input pins. The eight analog input pins are divided into two groups: group $0\left(\mathrm{AN}_{0}\right.$ to $\left.A N_{3}\right)$, and group $1\left(\mathrm{AN}_{4}\right.$ to $\left.A N_{7}\right) . A V_{\mathrm{CC}}$ and $A V_{\mathrm{SS}}$ are the power supply for the analog circuits in the $A / D$ converter. $V_{\text {REF }}$ is the $A / D$ conversion reference voltage.

## Table 15.1 A/D Converter Pins

| Pin Name | Abbreviation | I/O | Function |
| :---: | :---: | :---: | :---: |
| Analog power supply pin | $\mathrm{AV}_{\text {cc }}$ | Input | Analog power supply |
| Analog ground pin | $\mathrm{AV}_{\text {ss }}$ | Input | Analog ground and reference voltage |
| Reference voltage pin | $\mathrm{V}_{\text {REF }}$ | Input | Analog reference voltage |
| Analog input pin 0 | $\mathrm{AN}_{0}$ | Input | Group 0 analog inputs |
| Analog input pin 1 | $\mathrm{AN}_{1}$ | Input |  |
| Analog input pin 2 | $\mathrm{AN}_{2}$ | Input |  |
| Analog input pin 3 | $\mathrm{AN}_{3}$ | Input |  |
| Analog input pin 4 | $\mathrm{AN}_{4}$ | Input | Group 1 analog inputs |
| Analog input pin 5 | $\mathrm{AN}_{5}$ | Input |  |
| Analog input pin 6 | $\mathrm{AN}_{6}$ | Input |  |
| Analog input pin 7 | $\mathrm{AN}_{7}$ | Input |  |
| A/D external trigger input pin | $\overline{\text { ADTRG }}$ | Input | External trigger input for starting A/D conversion |

### 15.1.4 Register Configuration

Table 15.2 summarizes the A/D converter's registers.
Table 15.2 A/D Converter Registers

| Address*1 | Name | Abbreviation | R/W | Initial Value |
| :---: | :---: | :---: | :---: | :---: |
| H'FFE0 | A/D data register A (high) | ADDRAH | R | H'00 |
| H'FFE1 | A/D data register A (low) | ADDRAL | R | H'00 |
| H'FFE2 | A/D data register B (high) | ADDRBH | R | H'00 |
| H'FFE3 | A/D data register B (low) | ADDRBL | R | H'00 |
| H'FFE4 | A/D data register C (high) | ADDRCH | R | H'00 |
| H'FFE5 | A/D data register C (low) | ADDRCL | R | H'00 |
| H'FFE6 | A/D data register D (high) | ADDRDH | R | H'00 |
| H'FFE7 | A/D data register D (low) | ADDRDL | R | H'00 |
| H'FFE8 | A/D control/status register | ADCSR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'00 |
| H'FFE9 | A/D control register | ADCR | R/W | H'7E |

Notes: 1. Lower 16 bits of the address
2. Only 0 can be written in bit 7 , to clear the flag.

### 15.2 Register Descriptions

15.2.1 A/D Data Registers A to D (ADDRA to ADDRD)


The four A/D data registers (ADDRA to ADDRD) are 16-bit read-only registers that store the results of $\mathrm{A} / \mathrm{D}$ conversion.

An A/D conversion produces 10-bit data, which is transferred for storage into the A/D data register corresponding to the selected channel. The upper 8 bits of the result are stored in the upper byte of the A/D data register. The lower 2 bits are stored in the lower byte. Bits 5 to 0 of an A/D data register are reserved bits that are always read as 0 . Table 15.3 indicates the pairings of analog input channels and $\mathrm{A} / \mathrm{D}$ data registers.

The CPU can always read and write the A/D data registers. The upper byte can be read directly, but the lower byte is read through a temporary register (TEMP). For details see section 15.3, CPU Interface.

The A/D data registers are initialized to $\mathrm{H}^{\prime} 0000$ by a reset and in standby mode.
Table 15.3 Analog Input Channels and A/D Data Registers
Analog Input Channel

| Group 0 | Group 1 | A/D Data Register |
| :--- | :--- | :--- |
| AN $_{0}$ | $\mathrm{AN}_{4}$ | ADDRA |
| $\mathrm{AN}_{1}$ | $\mathrm{AN}_{5}$ | ADDRB |
| $\mathrm{AN}_{2}$ | $\mathrm{AN}_{6}$ | ADDRC |
| $\mathrm{AN}_{3}$ | $\mathrm{AN}_{7}$ | ADDRD |

### 15.2.2 A/D Control/Status Register (ADCSR)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ADF | ADIE | ADST | SCAN | CKS | CH 2 | CH1 | CHO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | $\mathrm{R} /(\mathrm{W})^{*}$ | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  | inter ables and | start rts or s t enabl disable | an mod lects sin <br> A/D <br> A/D end | ck se ects th mod versio <br> rrupt |  | l sel bits se anne <br> ion tim | 2 to 0 analog |
|  | end fla dicates end | of $A / D$ | nversio |  |  |  |  |  |

Note: * Only 0 can be written, to clear the flag.

ADCSR is an 8-bit readable/writable register that selects the mode and controls the $\mathrm{A} / \mathrm{D}$ converter. ADCSR is initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.

Bit 7—A/D End Flag (ADF): Indicates the end of A/D conversion.

| Bit 7: ADF | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | [Clearing condition] | Cleared by reading ADF while ADF = 1, then writing 0 in ADF |

Bit 6-A/D Interrupt Enable (ADIE): Enables or disables the interrupt (ADI) requested at the end of $\mathrm{A} / \mathrm{D}$ conversion.

Bit 6: ADIE Description

| 0 | A/D end interrupt request (ADI) is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | A/D end interrupt request (ADI) is enabled |  |

Bit 5-A/D Start (ADST): Starts or stops A/D conversion. The ADST bit remains set to 1 during $\mathrm{A} / \mathrm{D}$ conversion. It can also be set to 1 by external trigger input at the $\overline{\mathrm{ADTRG}}$ pin.

| Bit 5: ADST | Description |
| :--- | :--- |
| 0 | A/D conversion is stopped |
| 1 | Single mode: A/D conversion starts; ADST is automatically cleared to 0 when <br> conversion ends. <br> Scan mode: A/D conversion starts and continues, cycling among the selected <br> channels, until ADST is cleared to 0 by software, by a reset, or by a transition <br> to standby mode. |

Bit 4-Scan Mode (SCAN): Selects single mode or scan mode. For further information on operation in these modes, see section 15.4, Operation. Clear the ADST bit to 0 before switching the conversion mode.

Bit 4: SCAN Description

| 0 | Single mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | Scan mode |  |

Bit 3-Clock Select (CKS): Selects the A/D conversion time. Clear the ADST bit to 0 before switching the conversion time.

| Bit 3: CKS | Description |  |
| :--- | :--- | :--- |
| 0 | Conversion time $=266$ states (maximum) | (Initial value) |
| 1 | Conversion time $=134$ states (maximum) |  |

Bits 2 to 0—Channel Select 2 to 0 ( $\mathbf{C H} 2$ to $\mathbf{C H} 0)$ : These bits and the SCAN bit select the analog input channels. Clear the ADST bit to 0 before changing the channel selection.

| Group Selection | Channel Selection |  | Description |  |
| :---: | :---: | :---: | :---: | :---: |
| CH2 | CH1 | CHO | Single Mode | Scan Mode |
| 0 | 0 | 0 | $\mathrm{AN}_{0}$ (Initial value) | $\mathrm{AN}_{0}$ |
|  |  | 1 | $\mathrm{AN}_{1}$ | $\mathrm{AN}_{0}, \mathrm{AN}_{1}$ |
|  | 1 | 0 | $\mathrm{AN}_{2}$ | $\mathrm{AN}_{0}$ to $\mathrm{AN}_{2}$ |
|  |  | 1 | $\mathrm{AN}_{3}$ | $\mathrm{AN}_{0}$ to $\mathrm{AN}_{3}$ |
| 1 | 0 | 0 | $\mathrm{AN}_{4}$ | $\mathrm{AN}_{4}$ |
|  |  | 1 | $\mathrm{AN}_{5}$ | $\mathrm{AN}_{4}, \mathrm{AN}_{5}$ |
|  | 1 | 0 | $\mathrm{AN}_{6}$ | $\mathrm{AN}_{4}$ to $\mathrm{AN}_{6}$ |
|  |  | 1 | $\mathrm{AN}_{7}$ | $\mathrm{AN}_{4}$ to $\mathrm{AN}_{7}$ |

### 15.2.3 A/D Control Register (ADCR)

Bit
Initial value


Trigger enable
Enables or disables external triggering of A/D conversion

Reserved bit
Must not be set to 1

ADCR is an 8-bit readable/writable register that enables or disables external triggering of $\mathrm{A} / \mathrm{D}$ conversion. ADCR is initialized to H'7E by a reset and in standby mode.

Bit 7—Trigger Enable (TRGE): Enables or disables external triggering of A/D conversion.
Bit 7: TRGE Description
$1 \quad \mathrm{~A} / \mathrm{D}$ conversion starts at the falling edge of the external trigger signal ( $\overline{\mathrm{ADTRG}})$

Bits 6 to 1—Reserved: Read-only bits, always read as 1.
Bit 0—Reserved: Do not set to 1 .

### 15.3 CPU Interface

ADDRA to ADDRD are 16 -bit registers, but they are connected to the CPU by an 8 -bit data bus. Therefore, although the upper byte can be be accessed directly by the CPU, the lower byte is read through an 8-bit temporary register (TEMP).

An A/D data register is read as follows. When the upper byte is read, the upper-byte value is transferred directly to the CPU and the lower-byte value is transferred into TEMP. Next, when the lower byte is read, the TEMP contents are transferred to the CPU.

When reading an $\mathrm{A} / \mathrm{D}$ data register, always read the upper byte before the lower byte. It is possible to read only the upper byte, but if only the lower byte is read, incorrect data may be obtained.

Figure 15.2 shows the data flow for access to an $\mathrm{A} / \mathrm{D}$ data register.


Figure 15.2 A/D Data Register Access Operation (Reading H'AA40)

### 15.4 Operation

The A/D converter operates by successive approximations with 10 -bit resolution. It has two operating modes: single mode and scan mode.

### 15.4.1 Single Mode (SCAN = 0)

Single mode should be selected when only one A/D conversion on one channel is required. A/D conversion starts when the ADST bit is set to 1 by software, or by external trigger input. The ADST bit remains set to 1 during A/D conversion and is automatically cleared to 0 when conversion ends.

When conversion ends the ADF bit is set to 1 . If the ADIE bit is also set to 1 , an ADI interrupt is requested at this time. To clear the ADF flag to 0 , first read ADCSR, then write 0 in ADF.

When the mode or analog input channel must be switched during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1 to start A/D conversion again. The ADST bit can be set at the same time as the mode or channel is changed.

Typical operations when channel $1\left(\mathrm{AN}_{1}\right)$ is selected in single mode are described next.
Figure 15.3 shows a timing diagram for this example.

1. Single mode is selected $(\mathrm{SCAN}=0)$, input channel $\mathrm{AN}_{1}$ is selected $(\mathrm{CH} 2=\mathrm{CH} 1=0$, $\mathrm{CH} 0=1$ ), the $\mathrm{A} / \mathrm{D}$ interrupt is enabled ( $\mathrm{ADIE}=1$ ), and $\mathrm{A} / \mathrm{D}$ conversion is started $(\mathrm{ADST}=1)$.
2. When $\mathrm{A} / \mathrm{D}$ conversion is completed, the result is transferred into ADDRB. At the same time the ADF flag is set to 1 , the ADST bit is cleared to 0 , and the $\mathrm{A} / \mathrm{D}$ converter becomes idle.
3. Since $\mathrm{ADF}=1$ and $\mathrm{ADIE}=1$, an ADI interrupt is requested.
4. The $\mathrm{A} / \mathrm{D}$ interrupt handling routine starts.
5. The routine reads ADCSR, then writes 0 in the ADF flag.
6. The routine reads and processes the conversion result (ADDRB).
7. Execution of the $\mathrm{A} / \mathrm{D}$ interrupt handling routine ends. After that, if the ADST bit is set to 1 , A/D conversion starts again and steps 2 to 7 are repeated.


Figure 15.3 Example of A/D Converter Operation (Single Mode, Channel 1 Selected)

### 15.4.2 Scan Mode (SCAN = 1)

Scan mode is useful for monitoring analog inputs in a group of one or more channels. When the ADST bit is set to 1 by software or external trigger input, $\mathrm{A} / \mathrm{D}$ conversion starts on the first channel in the group $\left(\mathrm{AN}_{0}\right.$ when $\mathrm{CH} 2=0, \mathrm{AN}_{4}$ when $\left.\mathrm{CH} 2=1\right)$. When two or more channels are selected, after conversion of the first channel ends, conversion of the second channel ( $\mathrm{AN}_{1}$ or $\mathrm{AN}_{5}$ ) starts immediately. A/D conversion continues cyclically on the selected channels until the ADST bit is cleared to 0 . The conversion results are transferred for storage into the A/D data registers corresponding to the channels.

When the mode or analog input channel selection must be changed during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1 . A/D conversion will start again from the first channel in the group. The ADST bit can be set at the same time as the mode or channel selection is changed.

Typical operations when three channels in group $0\left(\mathrm{AN}_{0}\right.$ to $\left.\mathrm{AN}_{2}\right)$ are selected in scan mode are described next. Figure 15.4 shows a timing diagram for this example.

1. Scan mode is selected ( $\mathrm{SCAN}=1$ ), scan group 0 is selected $(\mathrm{CH} 2=0)$, analog input channels $\mathrm{AN}_{0}$ to $\mathrm{AN}_{2}$ are selected $(\mathrm{CH1}=1, \mathrm{CH} 0=0)$, and $\mathrm{A} / \mathrm{D}$ conversion is started $(\mathrm{ADST}=1)$.
2. When $A / D$ conversion of the first channel $\left(\mathrm{AN}_{0}\right)$ is completed, the result is transferred into ADDRA. Next, conversion of the second channel $\left(\mathrm{AN}_{1}\right)$ starts automatically.
3. Conversion proceeds in the same way through the third channel $\left(\mathrm{AN}_{2}\right)$.
4. When conversion of all selected channels $\left(\mathrm{AN}_{0}\right.$ to $\left.\mathrm{AN}_{2}\right)$ is completed, the ADF flag is set to 1 and conversion of the first channel $\left(\mathrm{AN}_{0}\right)$ starts again. If the ADIE bit is set to 1 , an ADI interrupt is requested at this time.
5. Steps 2 to 4 are repeated as long as the ADST bit remains set to 1 . When the ADST bit is cleared to $0, \mathrm{~A} / \mathrm{D}$ conversion stops. After that, if the ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion starts again from the first channel $\left(\mathrm{AN}_{0}\right)$.


Figure 15.4 Example of A/D Converter Operation
(Scan Mode, Channels $\mathbf{A N}_{0}$ to $\mathbf{A N}_{2}$ Selected)

### 15.4.3 Input Sampling and A/D Conversion Time

The A/D converter has a built-in sample-and-hold circuit. The A/D converter samples the analog input at a time $t_{D}$ after the ADST bit is set to 1 , then starts conversion. Figure 15.5 shows the A/D conversion timing. Table 15.4 indicates the $\mathrm{A} / \mathrm{D}$ conversion time.

As indicated in figure 15.5, the $\mathrm{A} / \mathrm{D}$ conversion time includes $\mathrm{t}_{\mathrm{D}}$ and the input sampling time. The length of $t_{\mathrm{D}}$ varies depending on the timing of the write access to ADCSR. The total conversion time therefore varies within the ranges indicated in table 15.4.

In scan mode, the values given in table 15.4 apply to the first conversion. In the second and subsequent conversions the conversion time is fixed at 256 states when CKS $=0$ or 128 states when $\mathrm{CKS}=1$.


Legend
(1): ADCSR write cycle
(2): ADCSR address
$t_{D}$ : Synchronization delay
$\mathrm{t}_{\text {sPL }}$ : Input sampling time
$t_{\text {conv: }}$ A/D conversion time
Figure 15.5 A/D Conversion Timing

Table 15.4 A/D Conversion Time (Single Mode)

|  | Symbol | CKS $=0$ |  |  | CKS $=1$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max | Min | Typ | Max |
| Synchronization delay | $\mathrm{t}_{\text {D }}$ | 10 | - | 17 | 6 | - | 9 |
| Input sampling time | $\mathrm{t}_{\text {SPL }}$ | - | 63 | - | - | 31 | - |
| A/D conversion time | $\mathrm{t}_{\text {conv }}$ | 259 | - | 266 | 131 | - | 134 |

Note: Values in the table are numbers of states.

### 15.4.4 External Trigger Input Timing

A/D conversion can be externally triggered. When the TRGE bit is set to 1 in ADCR, external trigger input is enabled at the $\overline{\text { ADTRG }}$ pin. A high-to-low transition at the $\overline{\text { ADTRG }}$ pin sets the ADST bit to 1 in ADCSR, starting A/D conversion. Other operations, in both single and scan modes, are the same as if the ADST bit had been set to 1 by software. Figure 15.6 shows the timing.


Figure 15.6 External Trigger Input Timing

### 15.5 Interrupts

The A/D converter generates an interrupt (ADI) at the end of A/D conversion. The ADI interrupt request can be enabled or disabled by the ADIE bit in ADCSR.

### 15.6 Usage Notes

When using the $\mathrm{A} / \mathrm{D}$ converter, note the following points:

## 1. Note on Board Design

In board layout, separate the digital circuits from the analog circuits as much as possible. Particularly avoid layouts in which the signal lines of digital circuits cross or closely approach the signal lines of analog circuits. Induction and other effects may cause the analog circuits to operate incorrectly, or may adversely affect the accuracy of $\mathrm{A} / \mathrm{D}$ conversion. The analog input signals $\left(\mathrm{AN}_{0}\right.$ to $\left.\mathrm{AN}_{7}\right)$, analog reference voltage ( $\mathrm{V}_{\text {REF }}$ ), and analog supply voltage $\left(\mathrm{AV}_{\mathrm{cc}}\right)$ must be separated from digital circuits by the analog ground $\left(\mathrm{AV}_{\mathrm{ss}}\right)$. The analog ground $\left(\mathrm{AV}_{\mathrm{ss}}\right)$ should be connected to a stable digital ground $\left(\mathrm{V}_{\mathrm{ss}}\right)$ at one point on the board.
2. Note on Noise

To prevent damage from surges and other abnormal voltages at the analog input pins $\left(\mathrm{AN}_{0}\right.$ to $\left.\mathrm{AN}_{7}\right)$ and analog reference voltage pin $\left(\mathrm{V}_{\text {REF }}\right)$, connect a protection circuit like the one in figure 15.7 between $\mathrm{AV}_{\mathrm{CC}}$ and $\mathrm{AV}_{\text {SS }}$. The bypass capacitors connected to $\mathrm{AV}_{\mathrm{CC}}$ and $\mathrm{V}_{\text {REF }}$ and the filter capacitors connected to $\mathrm{AN}_{0}$ to $\mathrm{AN}_{7}$ must be connected to $\mathrm{AV}_{5 S}$. If filter capacitors like the ones in figure 15.7 are connected, the voltage values input to the analog input pins ( $\mathrm{AN}_{0}$ to $\mathrm{AN}_{7}$ ) will be smoothed, which may give rise to error. Error can also occur if A/D conversion is frequently performed in scan mode so that the current that charges and discharges the capacitor in the sample-and-hold circuit of the A/D converter becomes greater than that input to the analog input pins via input impedance $\mathrm{R}_{\mathrm{in}}$. The circuit constants should therefore be selected carefully.


Notes: 1. Numeric values are approximate.

2. $\mathrm{R}_{\mathrm{in}}$ : input impedance

## Figure 15.7 Example of Analog Input Protection Circuit

## Table 15.5 Analog Input Pin Ratings

| Item | Min | Max | Unit |
| :--- | :--- | :--- | :--- |
| Analog input capacitance | - | 20 | pF |
| Allowable signal-source impedance | - | $10^{*}$ | $\mathrm{k} \Omega$ |

Note: * When $\mathrm{V}_{\mathrm{cc}}=4.0 \mathrm{~V}$ to 5.5 V and $\phi \leq 12 \mathrm{MHz}$.For details, refer to section 21, Electrical Characteristics.


Note: Numeric values are approximate.

Figure 15.8 Analog Input Pin Equivalent Circuit
3. A/D Conversion Accuracy Definitions

A/D conversion accuracy in the $\mathrm{H} 8 / 3052 \mathrm{~F}$ is defined as follows:

- Resolution

Digital output code length of A/D converter

- Offset error

Deviation from ideal A/D conversion characteristic of analog input voltage required to raise digital output from minimum voltage value 0000000000 to 0000000001 (figure 15.10)

- Full-scale error

Deviation from ideal A/D conversion characteristic of analog input voltage required to raise digital output from 1111111110 to 1111111111 (figure 15.10)

- Quantization error

Intrinsic error of the A/D converter; 1/2 LSB (figure 15.9)

- Nonlinearity error

Deviation from ideal A/D conversion characteristic in range from zero volts to full scale, exclusive of offset error, full-scale error, and quantization error.

- Absolute accuracy

Deviation of digital value from analog input value, including offset error, full-scale error, quantization error, and nonlinearity error.


Figure 15.9 A/D Converter Accuracy Definitions (1)


Figure 15.10 A/D Converter Accuracy Definitions (2)
4. Allowable Signal-Source Impedance

The analog inputs of the $\mathrm{H} 8 / 3052 \mathrm{~F}$ are designed to assure accurate conversion of input signals with a signal-source impedance not exceeding $10 \mathrm{k} \Omega$. The reason for this rating is that it enables the input capacitor in the sample-and-hold circuit in the A/D converter to charge within the sampling time. If the sensor output impedance exceeds $10 \mathrm{k} \Omega$, charging may be inadequate and the accuracy of A/D conversion cannot be guaranteed.
If a large external capacitor is provided in scan mode, then the internal $10-\mathrm{k} \Omega$ input resistance becomes the only significant load on the input. In this case the impedance of the signal source is not a problem.
A large external capacitor, however, acts as a low-pass filter. This may make it impossible to track analog signals with high dv/dt (e.g. a variation of $5 \mathrm{mV} / \mu \mathrm{s}$ ) (figure 15.11). To convert high-speed analog signals or to use scan mode, insert a low-impedance buffer.
5. Effect on Absolute Accuracy

Attaching an external capacitor creates a coupling with ground, so if there is noise on the ground line, it may degrade absolute accuracy. The capacitor must be connected to an electrically stable ground, such as $\mathrm{AV}_{\mathrm{ss}}$.
If a filter circuit is used, be careful of interference with digital signals on the same board, and make sure the circuit does not act as an antenna.


Figure 15.11 Analog Input Circuit (Example)

## Section 16 D/A Converter

### 16.1 Overview

The H8/3052F includes a D/A converter with two channels.

### 16.1.1 Features

D/A converter features are listed below.

- Eight-bit resolution
- Two output channels
- Conversion time: maximum $10 \mu \mathrm{~s}$ (with $20-\mathrm{pF}$ capacitive load)
- Output voltage: 0 V to $\mathrm{V}_{\text {ref }}$
- D/A outputs can be sustained in software standby mode


### 16.1.2 Block Diagram

Figure 16.1 shows a block diagram of the D/A converter.


Figure 16.1 D/A Converter Block Diagram

### 16.1.3 Pin Configuration

Table 16.1 summarizes the D/A converter's input and output pins.
Table 16.1 D/A Converter Pins

| Pin Name | Abbreviation | $\mathrm{I} / \mathbf{O}$ | Function |
| :--- | :--- | :--- | :--- |
| Analog power supply pin | $\mathrm{AV}_{\mathrm{cc}}$ | Input | Analog power supply |
| Analog ground pin | $\mathrm{AV}_{\mathrm{ss}}$ | Input | Analog ground and reference voltage |
| Analog output pin 0 | $\mathrm{DA}_{0}$ | Output | Analog output, channel 0 |
| Analog output pin 1 | $\mathrm{DA}_{1}$ | Output | Analog output, channel 1 |
| Reference voltage input pin | $\mathrm{V}_{\text {REF }}$ | Input | Analog reference voltage |

### 16.1.4 Register Configuration

Table 16.2 summarizes the D/A converter's registers.
Table 16.2 D/A Converter Registers

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFDC | D/A data register 0 | DADR0 | R/W | H'00 |
| H'FFDD | D/A data register 1 | DADR1 | R/W | H'00 |
| H'FFDE | D/A control register | DACR | R/W | H'1F |
| H'FF5C | D/A standby control register | DASTCR | R/W | H'FE |

Note: * Lower 16 bits of the address

### 16.2 Register Descriptions

### 16.2.1 D/A Data Registers 0 and 1 (DADR0/1)



The D/A data registers (DADR0 and DADR1) are 8-bit readable/writable registers that store the data to be converted. When analog output is enabled, the $\mathrm{D} / \mathrm{A}$ data register values are constantly converted and output at the analog output pins.

The D/A data registers are initialized to $\mathrm{H}^{\prime} 00$ by a reset and in standby mode.

### 16.2.2 D/A Control Register (DACR)

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DAOE1 | DAOE0 | DAE | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | - | - | - | - | - |
|  |  |  | D/A <br> Con | D/A <br> Co <br> ut e <br> D/A | le <br> D/A <br> 0 <br> rsio | an |  |  |
|  |  | D/A ou Contro | D/A con | 1 | ana |  |  |  |

DACR is an 8-bit readable/writable register that controls the operation of the $\mathrm{D} / \mathrm{A}$ converter. DACR is initialized to $\mathrm{H}^{\prime} 1 \mathrm{~F}$ by a reset and in standby mode.

Bit 7—D/A Output Enable 1 (DAOE1): Controls D/A conversion and analog output.
Bit 7: DAOE1 Description

| 0 | DA $_{1}$ analog output is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel-1 D/A conversion and DA, analog output are enabled |  |

Bit 6-D/A Output Enable 0 (DAOE0): Controls D/A conversion and analog output.

## Bit 6: DAOE0 Description

| 0 | $\mathrm{DA}_{0}$ analog output is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | Channel-0 D/A conversion and $\mathrm{DA}_{0}$ analog output are enabled |  |

Bit 5-D/A Enable (DAE): Controls D/A conversion, together with bits DAOE0 and DAOE1. When the DAE bit is cleared to 0 , analog conversion is controlled independently in channels 0 and 1. When the DAE bit is set to 1 , analog conversion is controlled together in channels 0 and 1 . Output of the conversion results is always controlled independently by DAOE0 and DAOE1.

| Bit 7: <br> DAOE1 | Bit 6: <br> DAOE0 | Bit 5: <br> DAE | Description |
| :--- | :--- | :--- | :--- | | 0 | 0 | - | D/A conversion is disabled in channels 0 and 1 |
| :--- | :--- | :--- | :--- |
|  | 1 | 0 | D/A conversion is enabled in channel 0 |
|  |  | 1 | D/A conversion is disabled in channel 1 |
| 1 | 0 | 0 | D/A conversion is enabled in channels 0 and 1 |
|  |  | - | D/A conversion is enabled in channel 1 |
|  |  |  |  |
|  |  |  | D/A conversion is enabled in channels 0 and 1 |

When the DAE bit is set to 1 , even if bits DAOE0 and DAOE1 in DACR and the ADST bit in ADCSR are cleared to 0 , the same current is drawn from the analog power supply as during $\mathrm{A} / \mathrm{D}$ and D/A conversion.

Bits 4 to 0—Reserved: Read-only bits, always read as 1.

### 16.2.3 D/A Standby Control Register (DASTCR)

DASTCR is an 8-bit readable/writable register that enables or disables D/A output in software standby mode.


DASTCR is initialized to H'FE by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 1—Reserved: Read-only bits, always read as 1.
Bit 0—D/A Standby Enable (DASTE): Enables or disables D/A output in software standby mode.

| Bit 0: DASTE | Description |  |
| :--- | :--- | :--- |
| 0 | D/A output is disabled in software standby mode | (Initial value) |
| 1 | D/A output is enabled in software standby mode |  |

### 16.3 Operation

The D/A converter has two built-in D/A conversion circuits that can perform conversion independently.

D/A conversion is performed constantly while enabled in DACR. If the DADR0 or DADR1 value is modified, conversion of the new data begins immediately. The conversion results are output when bits DAOE0 and DAOE1 are set to 1 .

An example of $\mathrm{D} / \mathrm{A}$ conversion on channel 0 is given next. Timing is indicated in figure 16.2.

1. Data to be converted is written in DADR0.
2. Bit DAOE0 is set to 1 in DACR. D/A conversion starts and $\mathrm{DA}_{0}$ becomes an output pin. The converted result is output after the conversion time. The output value is (DADR0 contents $/ 256) \times \mathrm{V}_{\text {REF }}$. Output of this conversion result continues until the value in DADR0 is modified or the DAOE0 bit is cleared to 0 .
3. If the DADR0 value is modified, conversion starts immediately, and the result is output after the conversion time.
4. When the DAOE0 bit is cleared to $0, \mathrm{DA}_{0}$ becomes an input pin.


Figure 16.2 Example of D/A Converter Operation

### 16.4 D/A Output Control

In the $\mathrm{H} 8 / 3052 \mathrm{~F}, \mathrm{D} / \mathrm{A}$ converter output can be enabled or disabled in software standby mode.
When the DASTE bit is set to 1 in DASTCR, D/A converter output is enabled in software standby mode. The D/A converter registers retain the values they held prior to the transition to software standby mode.

When D/A output is enabled in software standby mode, the reference supply current is the same as during normal operation.

## Section 17 RAM

### 17.1 Overview

The H8/3052F has 8 kbytes of high-speed static RAM on-chip. The RAM is connected to the CPU by a 16 -bit data bus. The CPU accesses both byte data and word data in two states, making the RAM useful for rapid data transfer.

The on-chip RAM of the H8/3052F is assigned to addresses H'FDF10 to H'FFF0F in modes 1, 2, 5, and 7, and to addresses H'FFDF10 to H'FFFF0F in modes 3, 4, and 6. The RAM enable bit (RAME) in the system control register (SYSCR) can enable or disable the on-chip RAM.

### 17.1.1 Block Diagram

Figure 17.1 shows a block diagram of the on-chip RAM.


Even addresses Odd addresses
Legend
SYSCR: System control register
Note: * This example is of the H8/3052F operating in mode 7. The lower 20 bits of the address are shown.

Figure 17.1 RAM Block Diagram

### 17.1.2 Register Configuration

The on-chip RAM is controlled by SYSCR. Table 17.1 gives the address and initial value of SYSCR.

Table 17.1 System Control Register

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFF2 | System control register | SYSCR | R/W | H'0B |

Note: * Lower 16 bits of the address.

### 17.2 System Control Register (SYSCR)

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SSBY | STS2 | STS1 | STS0 | UE | NMIEG | - | RAME |
| 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |
| $\frac{\text { R/W }}{\square}$ | R/W | R/W | R/W | R/W | R/W | - | R/W |

RAM enable
Enables or disables on-chip RAM

Reserved bit
NMI edge select
User bit enable
Standby timer select 2 to 0

## Software standby

One function of SYSCR is to enable or disable access to the on-chip RAM. The on-chip RAM is enabled or disabled by the RAME bit in SYSCR. For details about the other bits, see section 3.3, System Control Register (SYSCR).

Bit 0-RAM Enable (RAME): Enables or disables the on-chip RAM. The RAME bit is initialized at the rising edge of the input at the $\overline{\mathrm{RES}}$ pin. It is not initialized in software standby mode.

Bit 0: RAME
Description

| 0 | On-chip RAM is disabled |  |
| :--- | :--- | :--- |
| 1 | On-chip RAM is enabled | (Initial value) |

### 17.3 Operation

When the RAME bit is set to 1 , the on-chip RAM is enabled. Accesses to addresses H'FDF10 to H'FFF0F in the H8/3052F in modes $1,2,5$, and 7, addresses H'FFDF10 to H'FFFF0F in the H8/3052F in modes 3,4 , and 6 are directed to the on-chip RAM. In modes 1 to 6 (expanded modes), when the RAME bit is cleared to 0 , the external address space is accessed. In mode 7 (single-chip mode), when the RAME bit is cleared to 0 , the on-chip RAM is not accessed: read access always results in H'FF data, and write access is ignored.

Since the on-chip RAM is connected to the CPU by an internal 16-bit data bus, it can be written and read by word access. It can also be written and read by byte access. Byte data is accessed in two states using the upper 8 bits of the data bus. Word data starting at an even address is accessed in two states using all 16 bits of the data bus.

## Section 18 ROM

### 18.1 Features

The H8/3052F has 512 kbytes of on-chip flash memory. The features of the flash memory are summarized below.

- Four flash memory operating modes
- Program mode
- Erase mode
- Program-verify mode
- Erase-verify mode
- Programming/erase methods

The flash memory is programmed 128 bytes at a time. Block erase (in single-block units) can be performed. To erase the entire flash memory, each block must be erased in turn. Block erasing can be performed as required on 4 kbytes, 32 kbytes, and 64 kbytes blocks.

- Programming/erase times

The flash memory programming time is 10 ms (typ.) for simultaneous 128-byte programming, equivalent approximately to $80 \mu \mathrm{~s}$ (typ.) per byte, and the erase time is 100 ms (typ.).

- Reprogramming capability

The flash memory can be reprogrammed up to 100 times.

- On-board programming modes

There are two modes in which flash memory can be programmed/erased/verified on-board:

- Boot mode
- User program mode
- Automatic bit rate adjustment

With data transfer in boot mode, the LSI's bit rate can be automatically adjusted to match the transfer bit rate of the host.

- Flash memory emulation in RAM

Flash memory programming can be emulated in real time by overlapping a part of RAM onto flash memory.

- Protect modes

There are two protect modes, hardware and software, which allow protected status to be designated for flash memory program/erase/verify operations.

- PROM mode

Flash memory can be programmed/erased in PROM mode, using a PROM programmer, as well as in on-board programming mode.

### 18.2 Overview

### 18.2.1 Block Diagram



Legend
FLMCR1: Flash memory control register 1
FLMCR2: Flash memory control register 2
EBR1: Erase block register 1
EBR2: Erase block register 2
RAMCR: RAM control register
Figure 18.1 Block Diagram of Flash Memory

### 18.2.2 Mode Transitions

When the mode pins and the FWE pin are set in the reset state and a reset-start is executed, the microcomputer enters an operating mode as shown in figure 18.2. In user mode, flash memory can be read but not programmed or erased.

The boot, user program and PROM modes are provided as modes to write and erase the flash memory.


Notes: Only make a transition between user mode and user program mode when the CPU is not accessing the flash memory.

1. RAM emulation possible
2. The $\mathrm{H} 8 / 3052 \mathrm{~F}$ is placed in PROM mode by means of a dedicated PROM writer.
3. Mode settings are shown in the following table.

| Mode | Pins |  |  |  |
| :--- | :---: | :---: | :---: | :---: |
|  | FWE | $\mathrm{MD}_{2}$ | $\mathrm{MD}_{1}$ | $\mathrm{MD}_{0}$ |
| Mode 1 | 0 | 0 | 0 | 1 |
| Mode 2 |  | 0 | 1 | 0 |
| Mode 3 |  | 0 | 1 | 1 |
| Mode 4 |  | 1 | 0 | 0 |
| Mode 5 |  | 1 | 0 | 1 |
| Mode 6 |  | 1 | 1 | 0 |
| Mode 7 |  | 1 | 1 | 1 |
| Boot mode 5 | 1 | 0 | 0 | 1 |
| Boot mode 6 |  | 0 | 1 | 0 |
| Boot mode 7 |  | 0 | 1 | 1 |
| Setting prohibited |  | 1 | 0 | 0 |
| User program mode 5 |  | 1 | 0 | 1 |
|  |  |  | 1 | 1 |
| User program mode 6 |  | 0 |  |  |
| User program mode 7 |  | 1 | 1 | 1 |

Figure 18.2 Flash Memory State Transitions

State transitions between the normal user mode and on-board programming mode are performed by changing the FWE pin level from high to low or from low to high. To prevent misoperation (erroneous programming or erasing) in these cases, the bits in the flash memory control registers (FLMCR1, FLMCR2) should be cleared to 0 before making such a transition. After the bits are cleared, a wait time is necessary. Normal operation is not guaranteed if this wait time is insufficient.

### 18.2.3 On-Board Programming Modes

## Boot Mode

1. Initial state

The old program version or data remains written in the flash memory. The user should prepare the programming control program and new application program beforehand in the host.

3. Flash memory initialization

The erase program in the boot program area (in RAM) is executed, and the flash memory is initialized (to H'FF). In boot mode, total flash memory erasure is performed, without regard to blocks.

2. Programming control program transfer When boot mode is entered, the boot program in the $\mathrm{H} 8 / 3052 \mathrm{~F}$ (originally incorporated in the chip) is started and the programming control program in the host is transferred to RAM via SCl communication. The boot program required for flash memory erasing is automatically transferred to the RAM boot program area.

4. Writing new application program The programming control program transferred from the host to RAM is executed, and the new application program in the host is written into the flash memory.


Program execution state

## User Program Mode

1. Initial state

The FWE assessment program that confirms that user program mode has been entered, and the program that will transfer the programming/erase control program from flash memory to on-chip RAM should be written into the flash memory by the user beforehand. The programming/erase control program should be prepared in the host or in the flash memory.

3. Flash memory initialization The programming/erase program in RAM is executed, and the flash memory is initialized (to H'FF). Erasing can be performed in block units, but not in byte units.

2. Programming/erase control program transfer When user program mode is entered, user software confirms this fact, executes transfer program in the flash memory, and transfers the programming/erase control program to RAM.

4. Writing new application program Next, the new application program in the host is written into the erased flash memory blocks. Do not write to unerased blocks.


### 18.2.4 Flash Memory Emulation in RAM

In the $\mathrm{H} 8 / 3052 \mathrm{~F}$, flash memory programming can be emulated in real time by overlapping the flash memory with part of RAM ("overlap RAM"). When the emulation block set in RAMCR is accessed while the emulation function is being executed, data written in the overlap RAM is read. Emulation should be performed in user mode or user program mode.


Figure 18.3 Reading Overlap RAM Data in User Mode or User Program Mode

When overlap RAM data is confirmed, clear the RAMS bit to release RAM overlap, and actually perform writes to the flash memory.

When the programming control program is transferred to RAM in on-board programming mode, ensure that the transfer destination and the overlap RAM do not overlap, as this will cause data in the overlap RAM to be rewritten.

Flash memory


Figure 18.4 Writing Overlap RAM Data in User Program Mode
18.2.5 Differences between Boot Mode and User Program Mode

| Item | Boot Mode | User Program Mode |
| :--- | :--- | :--- |
| Total erase | Yes | Yes |
| Block erase | No | Yes |
| Programming control program* | Boot program is initiated, <br> and programming control | Program that controls <br> programming program in flash <br> program is transferred from <br> most to on-chip RAM, and <br> hosery is executed. Program <br> should be written beforehand in |
|  | executed there. | PROM mode and boot mode. |

Note: * To be provided by the user, in accordance with the recommended algorithm.

### 18.2.6 Block Configuration

The flash memory in the H8/3052F is divided into seven 64-kbyte blocks, one 32-kbyte block, and eight 4-kbyte blocks.
Address H'7FFFF $\quad 64$ kbytes

Figure 18.5 Erase Area Block Divisions

### 18.3 Pin Configuration

The flash memory is controlled by means of the pins shown in table 18.1.
Table 18.1 Pin Configuration

| Pin Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Reset | $\overline{R E S}$ | Input | Reset |
| Flash write enable | FWE | Input | Flash program/erase protection by hardware |
| Mode 2 | MD2 | Input | Sets LSI operating mode |
| Mode 1 | MD1 | Input | Sets LSI operating mode |
| Mode 0 | MD0 | Input | Sets LSI operating mode |
| Transmit data | TxD1 | Output | Serial transmit data output |
| Receive data | RxD1 | Input | Serial receive data input |

### 18.4 Register Configuration

The registers $*^{1}$ used to control the on-chip flash memory when enabled are shown in table 18.2.

## Table 18.2 Register Configuration

| Register Name | Abbreviation | R/W | Initial Value | Address* ${ }^{2}$ |
| :---: | :---: | :---: | :---: | :---: |
| Flash memory control register 1 | FLMCR1* ${ }^{6}$ | R/W* ${ }^{3}$ | $\mathrm{H}^{\prime} 00{ }^{* 4}$ | H'FF40 |
| Flash memory control register 2 | FLMCR2* ${ }^{6}$ | R/W* ${ }^{3}$ | H'00 | H'FF41 |
| Erase block register 1 | EBR1* ${ }^{*}$ | R/W* ${ }^{3}$ | $\mathrm{H}^{\prime} 00^{* 5}$ | H'FF42 |
| Erase block register 2 | EBR2*6 | R/W*3 | $\mathrm{H}^{\prime} 00^{* 5}$ | H'FF43 |
| RAM control register | RAMCR** | R/W | H'F0 | H'FF47 |

Notes: 1. Access is prohibited to lower 16 address bits H'FF44 to H'FF46 and H'FF48 to H'FF4F.
2. Lower 16 bits of the address.
3. If the chip is in a mode in which the on-chip flash memory is disabled, a read will return H'00 and writes are invalid. Writes are also invalid when the FWE bit in FLMCR1 is not set to 1.
4. When a high level is input to the FWE pin, the initial value is $\mathrm{H}^{\prime} 80$.
5. When a low level is input to the FWE pin, or if a high level is input and the SWE1 bit in FLMCR1 or SWE2 bit in FLMCR2 is not set, these registers are initialized to H'00.
6. FLMCR1, FLMCR2, EBR1, and EBR2, and RAMCR are 8-bit registers.

Byte access must be used on these registers (do not use word or longword access).

### 18.5 Register Descriptions

### 18.5.1 Flash Memory Control Register 1 (FLMCR1)

FLMCR1 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode for addresses $\mathrm{H}^{\prime} 00000$ to $\mathrm{H}^{\prime} 3 \mathrm{FFFF}$ is entered by setting SWE1 bit to 1 when FWE $=1$, then setting the PV1 or EV1 bit. Program mode for addresses H'00000 to H'3FFFF is entered by setting SWE1 bit to 1 when $\mathrm{FWE}=1$, then setting the PSU1 bit, and finally setting the P1 bit. Erase mode for addresses H'00000 to H'3FFFF is entered by setting SWE1 bit to 1 when FWE $=1$, then setting the ESU1 bit, and finally setting the E1 bit. FLMCR1 is initialized by a power-on reset, and in hardware standby mode and software standby mode. Its initial value is $\mathrm{H}^{\prime} 80$ when a high level is input to the FWE pin, and H'00 when a low level is input. When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

Writes are enabled only in the following cases: Writes to bit SWE1 of FLMCR1 enabled when FWE $=1$, to bits ESU1, PSU1, EV1, and PV1 when FWE $=1$ and SWE1 $=1$, to bit E1 when $\mathrm{FWE}=1, \mathrm{SWE} 1=1$ and ESU1 $=1$, and to bit P 1 when $\mathrm{FWE}=1, \mathrm{SWE} 1=1$, and PSU1 $=1$.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FWE | SWE1 | ESU1 | PSU1 | EV1 | PV1 | E1 | P1 |
| Initial value | 1/0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Bit 7—Flash Write Enable Bit (FWE): Sets hardware protection against flash memory programming/erasing.
Bit 7: FWE Description

| 0 | When a low level is input to the FWE pin (hardware-protected state) |
| :--- | :--- |
| 1 | When a high level is input to the FWE pin |

Bit 6—Software Write Enable Bit 1 (SWE1): Enables or disables flash memory programming and erasing (applicable addresses: H'00000 to H'3FFFF). Set this bit when setting bits 5 to 0 , bits 7 to 0 of EBR1, and bits 3 to 0 of EBR2.

Bit 6: SWE1 Description

| 0 | Writes disabled | (Initial value) |
| :--- | :--- | :---: |
| 1 | Writes enabled* |  |
|  | $[$ Setting condition $]$ |  |
|  | When FWE $=1$ |  |

[^1]Bit 5-Erase Setup Bit 1 (ESU1): Prepares for a transition to erase mode (applicable addresses: $H^{\prime} 00000$ to H'3FFFF). Do not set the SWE1, PSU1, EV1, PV1, E1, or P1 bit at the same time. Set this bit to 1 before setting bit E1 to 1 in FLMCR1.

Bit 5: ESU1 Description

| 0 | Erase setup cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Erase setup |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE1 $=1$ |  |

Bit 4—Program Setup Bit 1 (PSU1): Prepares for a transition to program mode (applicable addresses: $\mathrm{H}^{\prime} 00000$ to $\left.\mathrm{H}^{\prime} 3 F F F F\right)$. Do not set the SWE1, ESU1, EV1, PV1, E1, or P1 bit at the same time. Set this bit to 1 before setting bit P1 to 1 in FLMCR1.

Bit 4: PSU1 Description

| 0 | Program setup cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Program setup |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE1 $=1$ |  |

Bit 3—Erase-Verify 1 (EV1): Selects erase-verify mode transition or clearing (applicable addresses: H'00000 to H'3FFFF). Do not set the SWE1, ESU1, PSU1, PV1, E1, or P1 bit at the same time.

| Bit 3: EV1 | Description |  |
| :--- | :--- | :--- |
| 0 | Erase-verify mode cleared | (Initial value) |
| 1 | Transition to erase-verify mode |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE1 $=1$ |  |

Bit 2—Program-Verify 1 (PV1): Selects program-verify mode transition or clearing (applicable addresses: $\mathrm{H}^{\prime} 00000$ to $\mathrm{H}^{\prime} 3$ FFFF). Do not set the SWE1, ESU1, PSU1, EV1, E1, or P1 bit at the same time.

| Bit 2: PV1 | Description | (Initial value) |
| :--- | :--- | :--- |
| 0 | Program-verify mode cleared |  |
| 1 | Transition to program-verify mode |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE1 $=1$ |  |

Rev. 2.0, 03/01, page 570 of 822

Bit 1—Erase 1 (E1): Selects erase mode transition or clearing (applicable addresses: H'00000 to H'3FFFF). Do not set the SWE1, ESU1, PSU1, EV1, PV1, or P1 bit at the same time.

Bit 1: E1 Description

| 0 | Erase mode cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Transition to erase mode* |  |
|  | [Setting condition] |  |
|  | When FWE $=1$, SWE1 $=1$, and ESU1 $=1$ |  |

Note: * Do not access flash memory while the E1 bit is set to 1 .

Bit 0—Program (P1): Selects program mode transition or clearing (applicable addresses: H'00000 to H'3FFFF). Do not set the SWE1, PSU1, ESU1, EV1, PV1, or E1 bit at the same time.

Bit 0: P1 Description

| 0 | Program mode cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Transition to program mode* |  |
|  | [Setting condition] |  |
|  | When FWE $=1$, SWE1 $=1$, and PSU1 $=1$ |  |

Note: * Do not access flash memory while the P1 bit is set.

### 18.5.2 Flash Memory Control Register 2 (FLMCR2)

FLMCR2 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode for addresses H'40000 to H'7FFFF is entered by setting SWE2 to 1 when FWE $($ FLMCR1 $)=1$, then setting the EV2 or PV2 bit. Program mode for addresses H'40000 to H'7FFFF is entered by setting SWE2 to 1 when FWE (FLMCR1) = 1, then setting the PSU2 bit, and finally setting the P2 bit. Erase mode for addresses H'40000 to H'7FFFF is entered by setting SWE2 to 1 when FWE $($ FLMCR1 $)=1$, then setting the ESU2 bit, and finally setting the E2 bit. FLMCR2 is initialized to H'00 by a power-on reset, in hardware standby mode and software standby mode, when a low level is input to the FWE pin, and when a high level is input to the FWE pin and the SWE2 bit in FLMCR2 is not set (the exception is the FLER bit, which is initialized only by a power-on reset and in hardware standby mode). When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

Writes are enabled only in the following cases: Writes to bit SWE2 of FLMCR2 enabled when FWE $($ FLMCR1 $)=1$, to bits ESU2, PSU2, EV2, and PV2 when FEW (FLMCR1) $=1$ and SWE2 $=1$, to bit E2 when FWE $(\mathrm{FLMCR} 1)=1, \mathrm{SWE} 2=1$, and ESU2 $=1$, to bit P2 when FWE $($ FLMCR1 $)=1$, SWE2 $=1$, and PSU2 $=1$.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FLER | SWE2 | ESU2 | PSU2 | EV2 | PV2 | E2 | P2 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Bit 7—Flash Memory Error (FLER): Indicates that an error has occurred during an operation on flash memory (programming or erasing). When FLER is set to 1 , flash memory goes to the error-protection state.

Bit 7: FLER Description

| 0 | Flash memory is operating normally <br>  <br> Flash memory program/erase protection (error protection) is disabled <br> [Clearing condition] <br>  <br>  <br> Power-on reset or hardware standby mode |
| :--- | :--- |
| 1 | An error has occurred during flash memory programming/erasing |
|  | Flash memory program/erase protection (error protection) is enabled |
|  | [Setting condition] |
|  | See 18.8.3 Error Protection |

Bit 6—Software Write Enable Bit 2 (SWE2): Enables or disables flash memory programming and erasing (applicable addresses: H'40000 to H'7FFFF). Set this bit when setting bits 5 to 0 and bits 7 to 4 of EBR2.

| Bit 6: SWE2 | Description |  |
| :--- | :--- | :--- |
| 0 | Writes disabled | (Initial value) |
| 1 | Writes enabled* |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ |  |

Note: * Do not execute a SLEEP instruction while the SWE2 bit is set to 1.

Bit 5—Erase Setup Bit 2 (ESU2): Prepares for a transition to erase mode (applicable addresses: $H^{\prime} 40000$ to H'7FFFF). Set this bit to 1 before setting bit E2 to 1 in FLMCR2. Do not set the PSU2, EV2, PV2, E2, or P2 bit at the same time.

Bit 5: ESU2
Description

| 0 | Erase setup cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Erase setup |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE2 $=1$ |  |

Bit 4—Program Setup Bit 2 (PSU2): Prepares for a transition to program mode (applicable addresses: H'40000 to H'7FFFF). Set this bit to 1 before setting bit P2 to 1 in FLMCR2. Do not set the ESU2, EV2, PV2, E2, or P2 bit at the same time.

## Bit 4: PSU2 Description

| 0 | Program setup cleared | (Initial value) |
| :--- | :--- | :---: |
| 1 | Program setup |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE2 $=1$ |  |

Bit 3—Erase-Verify 2 (EV2): Selects erase-verify mode transition or clearing (applicable addresses: H'40000 to H'7FFFF). Do not set the ESU2, PSU2, PV2, E2, or P2 bit at the same time.

## Bit 3: EV2 Description

| 0 | Erase-verify mode cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Transition to erase-verify mode |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE2 $=1$ |  |

Bit 2—Program-Verify 2 (PV2): Selects program-verify mode transition or clearing (applicable addresses: H'40000 to H'7FFFF). Do not set the ESU2, PSU2, EV2, E2, or P2 bit at the same time.

Bit 2: PV2
Description

| 0 | Program-verify mode cleared | (Initial value) |
| :--- | :--- | :--- |
| 1 | Transition to program-verify mode |  |
|  | [Setting condition] |  |
|  | When FWE $=1$ and SWE2 $=1$ |  |

Bit 1—Erase 2 (E2): Selects erase mode transition or clearing (applicable addresses: H'40000 to H'7FFFF). Do not set the ESU2, PSU2, EV2, PV2, or P2 bit at the same time.

Bit 1: E2 Description

| 0 | Erase mode cleared | (Initial value) |
| :--- | :--- | :---: |
| 1 | Transition to erase mode* |  |
|  | [Setting condition] |  |
|  | When FWE $=1$, SWE2 $=1$, and ESU2 $=1$ |  |

[^2]Bit 0—Program 2 (P2): Selects program mode transition or clearing (applicable addresses: H'40000 to H'7FFFF). Do not set the ESU2, PSU2, EV2, PV2, or E2 bit at the same time.
Bit 0: P2 Description

| 0 | Program mode cleared | (Initial value) |
| :--- | :--- | :---: |
| 1 | Transition to program mode* |  |
|  | [Setting condition ] |  |
|  | When FWE $=1$, SWE2 $=1$, and PSU2 $=1$ |  |
|  |  |  |

Note: * Do not access flash memory while the P2 bit is set.

### 18.5.3 Erase Block Register 1 (EBR1)

EBR1 is an 8-bit register that specifies the flash memory erase area block by block. EBR1 is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset, in hardware standby mode and software standby mode, when a low level is input to the FWE pin, and when a high level is input to the FWE pin and the SWE1 bit in FLMCR1 is not set. When a bit in EBR1 is set to 1, the corresponding block can be erased. Other blocks are erase-protected. Only one of the bits of EBR1 and EBR2 combined can be set. Do not set more than one bit, as this will cause all the bits in both EBR1 and EBR2 to be automatically cleared to 0 . When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

The flash memory block configuration is shown in table 18.3.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | EB7 | EB6 | EB5 | EB4 | EB3 | EB2 | EB1 | EB0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

### 18.5.4 Erase Block Register 2 (EBR2)

EBR2 is an 8-bit register that specifies the flash memory erase area block by block. EBR2 is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset, in hardware standby mode and software standby mode, when a low level is input to the FWE pin. Bits EB11 to EB8 will be initialized to 0 if bit SWE1 of FLMCR1 is not set, even though a high level is input to pin FWE. Also, bits EB15 to EB12 will be initialized to 0 if bit SWE2 of FLMCR2 is not set. When a bit in EBR2 is set to 1 , the corresponding block can be erased. Other blocks are erase-protected. Only one of the bits of EBR1 and EBR2 combined can be set. Do not set more than one bit, as this will cause all the bits in both EBR1 and EBR2 to be automatically cleared to 0 . When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

The flash memory block configuration is shown in table 18.3.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | EB15 | EB14 | EB13 | EB12 | EB11 | EB10 | EB9 | EB8 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Table 18.3 Flash Memory Erase Blocks

| Block (Size) | Addresses |
| :---: | :---: |
| EB0 (4 kbytes) | H'000000-H'000FFF |
| EB1 (4 kbytes) | H'001000-H'001FFF |
| EB2 (4 kbytes) | H'002000-H'002FFF |
| EB3 (4 kbytes) | H'003000-H'003FFF |
| EB4 (4 kbytes) | H'004000-H'004FFF |
| EB5 (4 kbytes) | H'005000-H'005FFF |
| EB6 (4 kbytes) | H'006000-H'006FFF |
| EB7 (4 kbytes) | H'007000-H'007FFF |
| EB8 (32 kbytes) | H'008000-H'00FFFF |
| EB9 (64 kbytes) | H'010000-H'01FFFF |
| EB10 (64 kbytes) | H'020000-H'02FFFF |
| EB11 (64 kbytes) | H'030000-H'03FFFF |
| EB12 (64 kbytes) | H'040000-H'04FFFF |
| EB13 (64 kbytes) | H'050000-H'05FFFF |
| EB14 (64 kbytes) | H'060000-H'06FFFF |
| EB15 (64 kbytes) | H'070000-H'07FFFF |

### 18.5.5 RAM Control Register (RAMCR)

RAMCR specifies the area of flash memory to be overlapped with part of RAM when emulating real-time flash memory programming. RAMCR initialized to H'F0 by a power-on reset and in hardware standby mode. It is not initialized by a manual reset and in software standby mode. RAMCR settings should be made in user mode or user program mode.

Flash memory area divisions are shown in table 18.4. To ensure correct operation of the emulation function, the ROM for which RAM emulation is performed should not be accessed immediately after this register has been modified. Normal execution of an access immediately after register modification is not guaranteed.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | RAMS | RAM2 | RAM1 | RAM0 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |

Bits $\mathbf{7}$ to 4—Reserved: These bits always read 1 .
Bit 3-RAM Select (RAMS): Specifies selection or non-selection of flash memory emulation in RAM. When RAMS $=1$, all flash memory block are program/erase-protected.

Bit 3: RAMS Description

| 0 | Emulation not selected | (Initial value) |
| :--- | :--- | :--- |
|  | Program/erase-protection of all flash memory blocks is disabled |  |
| 1 | Emulation selected |  |
|  | Program/erase-protection of all flash memory blocks is enabled |  |

Bits 2 to 0—Flash Memory Area Selection: These bits are used together with bit 3 to select the flash memory area to be overlapped with RAM. (See table 18.4.)

Table 18.4 Flash Memory Area Divisions

| Addresses | Block Name | RAMS | RAM1 | RAM1 | RAM0 |
| :--- | :--- | :--- | :--- | :--- | :--- |
| H'FFE000-H'FFEFFF | RAM area 4 kbytes | 0 | $*$ | $*$ | $*$ |
| H'000000-H'000FFF $_{\text {' }^{\prime} 001000-H^{\prime} 001 F F F}$ | EB0 $(4$ kbytes $)$ | 1 | 0 | 0 | 0 |
| H'002000-H'002FFF $^{\text {H'003000-H'003FFF }}$ | EB1 $(4$ kbytes $)$ | 1 | 0 | 0 | 1 |
| H'004000-H'004FFF $_{\text {H}^{\prime} 005000-H^{\prime} 005 F F F}$ | EB2 $(4$ kbytes $)$ | 1 | 0 | 1 | 0 |
| H'006000-H'006FFF $_{\text {H'007000-H'007FFF }}$ | EB3 $(4$ kbytes $)$ | 1 | 0 | 1 | 1 |

*: Don't care

### 18.6 On-Board Programming Modes

When pins are set to on-board programming mode and a reset-start is executed, a transition is made to the on-board programming state in which program/erase/verify operations can be performed on the on-chip flash memory. There are two on-board programming modes: boot mode and user program mode. The pin settings for transition to each of these modes are shown in table 18.5. For a diagram of the transitions to the various flash memory modes, see figure 18.2.

Table 18.5 Setting On-Board Programming Modes

| Mode |  | FWE | MD2 | MD1 | MDO | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Boot mode | Mode 5 | 1* | 0* | 0 | 1 | 0: V ${ }_{\text {IL }}$ |
|  | Mode 6 |  | 0* | 1 | 0 | 1: $\mathrm{V}_{\mathrm{IH}}$ |
|  | Mode 7 |  | 0* | 1 | 1 |  |
| User program mode | Mode 5 |  | 1* | 0 | 1 |  |
|  | Mode 6 |  | 1* | 1 | 0 |  |
|  | Mode 7 |  | 1* | 1 | 1 |  |

Notes: 1. For the high-level application timing, see items 6 and 7 in Notes on Use of Boot Mode.
2. In boot mode, the inverse of the $M D_{2}$ setting should be input.
3. In boot mode, the mode control register (MDCR) can be used to monitor the status of modes 5,6 , and 7 , in the same way as in normal mode.

### 18.6.1 Boot Mode

When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. The SCI channel to be used is set to asynchronous mode.

When a reset-start is executed after the LSI's pins have been set to boot mode, the boot program built into the LSI is started and the programming control program prepared in the host is serially transmitted to the LSI via the SCI. In the LSI, the programming control program received via the SCI is written into the programming control program area in on-chip RAM. After the transfer is completed, control branches to the start address of the programming control program area and the programming control program execution state is entered (flash memory programming is performed).

The transferred programming control program must therefore include coding that follows the programming algorithm given later.

The system configuration in boot mode is shown in figure 18.6, and the boot mode execution procedure in figure 18.7.


Figure 18.6 System Configuration in Boot Mode


Figure 18.7 Boot Mode Execution Procedure

## Automatic SCI Bit Rate Adjustment



When boot mode is initiated, the LSI measures the low period of the asynchronous SCI communication data ( $\mathrm{H}^{\prime} 00$ ) transmitted continuously from the host. The SCI transmit/receive format should be set as follows: 8 -bit data, 1 stop bit, no parity. The LSI calculates the bit rate of the transmission from the host from the measured low period, and transmits one H'00 byte to the host to indicate the end of bit rate adjustment. The host should confirm that this adjustment end indication (H'00) has been received normally, and transmit one H'55 byte to the LSI. If reception cannot be performed normally, initiate boot mode again (reset), and repeat the above operations. Depending on the host's transmission bit rate and the LSI's system clock frequency, there will be a discrepancy between the bit rates of the host and the LSI. Set the host transfer bit rate at 4,800, 9,600 or $19,200 \mathrm{bps}$ to operate the SCI properly.

Table 18.6 shows host transfer bit rates and system clock frequencies for which automatic adjustment of the LSI bit rate is possible. The boot program should be executed within this system clock range.

Table 18.6 System Clock Frequencies for which Automatic Adjustment of LSI Bit Rate is Possible

Host Bit Rate
System Clock Frequency for Which Automatic Adjustment
of LSI Bit Rate is Possible (MHz)

| 4800 bps | 4 to 25 |
| :--- | :--- |
| $9,600 \mathrm{bps}$ | 8 to 25 |
| $19,200 \mathrm{bps}$ | 16 to 25 |

Notes: 1. Use a host bit rate setting of 4800, 9600, or 19200 bps only. No other setting should be used.
2. Although the $\mathrm{H} 8 / 3052 \mathrm{~F}$ may also perform automatic bit rate adjustment with bit rate and system clock combinations other than those shown in table 18.6, a degree of error will arise between the bit rates of the host and the H8/3052F, and subsequent transfer will not be performed normally. Therefore, only combinations of bit rate and system clock within the ranges shown in table 18.6 can be used for boot mode execution.

On-Chip RAM Area Divisions in Boot Mode: In boot mode, the RAM area is divided into an area used by the boot program and an area to which the programming control program is transferred via the SCI, as shown in figure 18.8. The boot program area cannot be used until the execution state in boot mode switches to the programming control program transferred from the host.


Figure 18.8 RAM Areas in Boot Mode
Notes: 1. The boot program area cannot be used until a transition is made to the execution state for the programming control program transferred to RAM. Note also that the boot program remains in this area of the on-chip RAM even after control branches to the programming control program.
2. In flash memory emulation by RAM, part (H'FE000 to H'FEFFF) of the user program transfer area is used as the area in which emulation is performed, and therefore the user program must not be transferred to this area.

## Notes on Using the Boot Mode

1. When this LSI comes out of reset in boot mode, it measures the low period the input at the SCI's RXD ${ }_{1}$ pin. The reset should end with RXD $_{1}$ high. After the reset ends, it takes about 100 states for this LSI to get ready to measure the low period of the $\mathrm{RXD}_{1}$ input.
2. If any data has been written to the flash memory (if all data is not H'FF), all flash memory blocks are erased when this mode is executed. Therefore, boot mode should be used for initial on-board programming, or for forced recovery if the program to be activated in user program mode is accidentally erased and user program mode cannot be executed, for example.
3. Interrupts cannot be used during programming or erasing of flash memory.
4. The RXD $_{1}$ and TXD $_{1}$ pins should be pulled up on the board.
5. This LSI terminates transmit and receive operations by the on-chip SCI(channel 1) (by clearing the RE and TE bits in serial control register (SCR)) before branching to the user program. However, the adjusted bit rate is held in the bit rate register (BRR). At this time, the TXD ${ }_{1}$ pin is in the high level output state (P9DDR P9 ${ }_{1} \mathrm{DDR}=1, \mathrm{P} 9 \mathrm{DR}$ P9 $9_{1} \mathrm{DR}=1$ ).
Before branching to the user program the value of the general registers in the CPU are also undefined. Therefore, the general registers must be initialized immediately after control branches to the user program. Since the stack pointer $(\mathrm{SP})$ is implicitly used during subroutine call, etc., a stack area must be specified for use by the user program.
There are no other internal I/O registers in which the initial value is changed.
6. Transition to the boot mode executes a reset-start of this LSI after setting the $\mathrm{MD}_{0}$ to $\mathrm{MD}_{2}$ and FWE pins according to the mode setting conditions shown in Table 18.5.
At this time, this LSI latches the status of the mode pin inside the microcomputer to maintain the boot mode status at the reset clear (startup with Low $\rightarrow$ High) timing* ${ }^{1}$.
To clear boot mode, it is necessary to drive the FWE pin low during the reset, and then execute reset release ${ }^{*{ }^{1}}$. The following points must be noted:

- Before making a transition from the boot mode to the regular mode, the microcomputer boot mode must be reset by reset input via the $\overline{\mathrm{RES}}$ pin. At this time, the $\overline{\mathrm{RES}}$ pin must be hold at low level for at least 20 system clock. $*^{3}$
- Do not change the input levels at the mode pins $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$ or the FWE pin while in boot mode. When making a mode transition, first enter the reset state by inputting a low level to the $\overline{\mathrm{RES}}$ pin. When a watchdog timer reset was generated in the boot mode, the microcomputer mode is not reset and the on-chip boot program is restarted regardless of the state of the mode pin.
- Do not input low level to the FWE pin while the boot program is executing and when programming/erasing flash memory. $*^{2}$

7. If the mode pin and FWE pin input levels are changed from 0 V to $\mathrm{V}_{\mathrm{CC}}$ or from $\mathrm{V}_{\mathrm{CC}}$ to 0 V during a reset (while a low level is being input to the $\overline{\mathrm{RES}}$ pin), the microcomputer's operating mode will change.
Therefore, since the state of the address dual port and bus control output signals $(\overline{\mathrm{CSn}}, \overline{\mathrm{RD}}$, $\overline{\mathrm{HWR}}, \overline{\mathrm{LWR}}$ ) changes, use of these pins as output signals during reset must be disabled outside the microcomputer.


Figure 18.9 Recommended System Block Diagram
Notes: 1. The mode pin and FWE pin input must satisfy the mode programming setup time ( $\mathrm{t}_{\text {MDS }}$ ) relative to the reset clear timing.
2. For notes on FWE pin High/Low, see section 18.11, Notes on Flash Memory Programming/Erasing.
3. See section 4.2.2, Reset Sequence and 18.11, Notes on Flash Memory Programming/Erasing. The H8/3052F requires a minimum of 20 system clocks.

### 18.6.2 User Program Mode

When set to the user program mode, this LSI can erase and program its flash memory by executing a user program. Therefore, on-chip flash memory on-board programming can be performed by providing a means of controlling FWE and supplying the write data on the board and providing a write program in a part of the program area.

To select this mode, set the LSI to on-chip ROM enable modes 5, 6, and 7 and apply a high level to the FWE pin. In this mode, the peripheral functions, other than flash memory, are performed the same as in modes 5, 6 , and 7 .

Since the flash memory cannot be read while it is being programmed/erased, place a programming program on external memory, or transfer the programming program to RAM area, and execute it in the RAM.

Figure 18.10 shows the procedure for executing when transferred to on-chip RAM. During reset start, starting from the user program mode is possible.


Procedure
The user writes a program that executes steps 3 to 8 in advance as shown below.

1. Sets the mode pin to an on-chip ROM enable mode (mode 5, 6 , or 7 ).
2. Starts the CPU via reset.
(The CPU can also be started from the user program mode by setting the FWE pin to High level during reset; that is, during the period the $\overline{R E S}$ pin is a low level.)
3. Transfers the on-board programming program to RAM.
4. Branches to the program in RAM.
5. Sets the FWE pin to a high level.* (Switches to user program mode.)
6. After confirming that the FWE pin is a high level, executes the on-board programming program in RAM. This reprograms the user application program in flash memory.
7. At the end of reprograming, clears the SWE1 and SWE2 bit, and exits the user program mode by switching the FWE pin from a high level to a low level.*
8. Branches to, and executes, the user application program reprogrammed in flash memory.

Note: * For notes on FWE pin High/Low, see section 18.11, Notes on Flash Memory Programming/Erasing.

Figure 18.10 User Program Mode Execution Procedure (Example)
Note: Normally do not apply a high level to the FWE pin. To prevent erroneous programming or erasing in the event of program runaway, etc., apply a high level to the FWE pin only when programming/erasing flash memory (including flash memory emulation by RAM). If program runaway, etc. causes overprogramming or overerasing of flash memory, the memory cells will not operate normally.
Also, while a high level is applied to the FWE pin, the watchdog timer should be activated to prevent overprogramming or overerasing due to program runaway, etc.

### 18.7 Programming/Erasing Flash Memory

A software method, using the CPU, is employed to program and erase flash memory in the onboard programming modes. There are four flash memory operating modes: program mode, erase mode, program-verify mode, and erase-verify mode. Transitions to these modes are made by setting the PSU1, ESU1, P1, E1, PV1, and EV1 bits in FLMCR1 for addresses H'00000 to H'3FFFF, or the PSU2, ESU2, P2, E2, PV2, and EV2 bits in FLMCR2 for addresses H'40000 to H'7FFFF.

The flash memory cannot be read while it is being written or erased. Install the program to control flash memory programming and erasing (programming control program) in the on-chip RAM, in external memory, or in flash memory outside the address area, and execute the program from there.

See section 18.1, Notes on Flash Memory Programming/Erasing, for points to be noted when programming or erasing the flash memory. In the following operation descriptions, wait times after setting or clearing individual bits in FLMCR1 and FLMCR2 are given as parameters; for details of the wait times, see section 21.2.5, Flash Memory Characteristics.

Notes: 1. Operation is not guaranteed if bits SWE1, ESU1, PSU1, EV1, PV1, E1, and P1 of FLMCR1 and bits SWE2, ESU2, PSU2, EV2, PV2, E2 and P2 of FLMCR2 are set/reset by a program in flash memory in the corresponding address areas.
2. When programming or erasing, set FWE to 1 (programming/erasing will not be executed if FWE $=0$ ).
3. Programming should be performed in the erased state. Do not perform additional programming on previously programmed addresses.
4. Do not program addresses $\mathrm{H}^{\prime} 00000$ to $\mathrm{H}^{\prime} 3$ FFFF and $\mathrm{H}^{\prime} 40000$ to $\mathrm{H}^{\prime} 7 \mathrm{FFFF}$ simultaneously. Operation is not guaranteed if this is done.


Notes: In order to perform a normal read of flash memory, SWE must be cleared to 0 .
Also note that verify-reads can be performed during the programming/erasing process.

1. $\bigcirc$ : Normal mode $\bigcirc$ : On-board programming mode
2. Do not make a state transition by setting or clearing multiple bits simultaneously.
3. After a transition from erase mode to the erase setup state, do not enter erase mode without passing through the software programming enable state.
4. After a transition from program mode to the program setup state, do not enter program mode without passing through the software programming enable state.

Figure 18.11 State Transitions Caused by FLMCR1 and FLMCR2 Bit Settings

### 18.7.1 Program Mode

When writing data or programs to flash memory, the program/program-verify flowchart shown in figure 18.12 should be followed. Performing program operations according to this flowchart will enable data or programs to be written to flash memory without subjecting the device to voltage stress or sacrificing program data reliability. Programming should be carried out 128 bytes at a time.

The wait times after bits are set or cleared in the flash memory control register (FLMCR1, FLMCR2) and the maximum number of programming operations ( N ) are shown in table 21.10 in section 21.2.5, Flash Memory Characteristics.

Following the elapse of (tsswe) $\mu \mathrm{s}$ or more after the SWE1 and SWE2 bits are set to 1 in FLMCR1 and FLMCR2, 128-byte data is written consecutively to the write addresses. The lower 8 bits of the first address written to must be $\mathrm{H}^{\prime} 00$ and $\mathrm{H}^{\prime} 80,128$ consecutive byte data transfers are performed. The program address and program data are latched in the flash memory. A 128-byte data transfer must be performed even if writing fewer than 128 bytes; in this case, H'FF data must be written to the extra addresses.

Next, the watchdog timer (WDT) is set to prevent overprogramming due to program runaway, etc. Set a value greater than (tspsu $+\mathrm{tsp}+\mathrm{tcp}+\mathrm{tcpsu}) \mu \mathrm{s}$ as the WDT overflow period. Preparation for entering program mode (program setup) is performed next by setting the PSU1 and PSU2 bits in FLMCR1 and FLMCR2. The operating mode is then switched to program mode by setting the P1 and P2 bits in FLMCR1 and FLMCR2 after the elapse of at least (tspsu) $\mu$ s. The time during which the P1 and P2 bits are set is the flash memory programming time. Make a program setting so that the time for one programming operation is within the range of (tsp) $\mu \mathrm{s}$.

The wait time after P1 and P2 bits setting must be changed according to the number of reprogramming loops. For details, see section 21.2.5, Flash Memory Characteristics.

### 18.7.2 Program-Verify Mode

In program-verify mode, the data written in program mode is read to check whether it has been correctly written in the flash memory.

After the elapse of the given programming time, clear the P1 and P2 bits in FLMCR1 and FLMCR2, then wait for at least (tcp) $\mu$ s before clearing the PSU1 and PSU2 bits to exit program mode. After exiting program mode, the watchdog timer setting is also cleared. The operating mode is then switched to program-verify mode by setting the PV1 and PV2 bits in FLMCR1 and FLMCR2. Before reading in program-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (tspv ) $\mu \mathrm{s}$ or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (tspvr) $\mu \mathrm{s}$ after the dummy write before performing this read operation. Next, the originally written data is compared with the verify data, and reprogram data is computed (see figure 18.12) and transferred to RAM. After verification of 128 bytes of data has been completed, exit program-verify mode, wait for at least (tcpv) $\mu \mathrm{s}$, then determine whether 128 -byte programming has finished. If reprogramming is necessary, set program mode again, and repeat the program/program-verify sequence as before. The maximum value for repetition of the program/program-verify sequence is indicated by the maximum programming count (N). Leave a wait time of at least (tcswe) $\mu \mathrm{s}$ after clearing SWE1 or SWE2.

### 18.7.3 Notes on Program/Program-Verify Procedure

1. The program/program-verify procedure for the $\mathrm{H} 8 / 3052 \mathrm{~F}$ is a 128 -byte-unit programming algorithm.
In order to perform 128-byte-unit programming, the lower 8 bits of the write start address must be $\mathrm{H}^{\prime} 00$ or $\mathrm{H}^{\prime} 80$.
2. When performing continuous writing of 128 -byte data to flash memory, byte-unit transfer should be used.
128-byte data transfer is necessary even when writing fewer than 128 bytes of data. H'FF data must be written to the extra addresses.
3. Verify data is read in word units.
4. The write pulse is applied and a flash memory write executed while the P1 bit in FLMCR1 or the P2 bit in FLMCR2 is set. In the H8/3052F, write pulses should be applied as follows in the program/program-verify procedure to prevent voltage stress on the device and loss of write data reliability.
a. After write pulse application, perform a verify-read in program-verify mode and apply a write pulse again for any bits read as 1 (reprogramming processing). When all the 0 -write bits in the 128 -byte write data are read as 0 in the verify-read operation, the program/program-verify procedure is completed. In the $\mathrm{H} 8 / 3052 \mathrm{~F}$, the number of loops in reprogramming processing is guaranteed not to exceed the maximum programming count (N).

Rev. 2.0, 03/01, page 588 of 822
b. After write pulse application, a verify-read is performed in program-verify mode, and programming is judged to have been completed for bits read as 0 . The following processing is necessary for programmed bits.
When programming is completed at an early stage in the program/program-verify procedure:
If programming is completed in the 1st to 6th reprogramming processing loop, additional programming should be performed on the relevant bits. Additional programming should only be performed on bits which first return 0 in a verify-read in certain reprogramming processing.
When programming is completed at a late stage in the program/program-verify procedure: If programming is completed in the 7th or later reprogramming processing loop, additional programming is not necessary for the relevant bits.
c. If programming of other bits is incomplete in the 128 bytes, reprogramming process should be executed. If a bit for which programming has been judged to be completed is read as 1 in a subsequent verify-read, a write pulse should again be applied to that bit.
5. The period for which the P1 bit in FLMCR1 or the P2 bit in FLMCR2 is set (the write pulse width) should be changed according to the degree of progress through the program/program-verify procedure. For detailed wait time specifications, see section 21.2.5, Flash Memory Characteristics.

## Table 18.7 Wait Time after P Bit Setting

| Item | Symbol | Conditions | Symbol |
| :--- | :--- | :--- | :--- |
| Wait time after $P$ | $\mathrm{t}_{\mathrm{sp}}$ | When reprogramming loop count $(\mathrm{n})$ is 1 to 6 | $\mathrm{t}_{\mathrm{sp}} 30$ |
| bit setting |  | When reprogramming loop count $(\mathrm{n})$ is 7 or more | $\mathrm{t}_{\mathrm{sp}} 200$ |
|  |  | In case of additional programming processing* | $\mathrm{t}_{\mathrm{sp}} 10$ |

Note: * Additional programming processing is necessary only when the reprogramming loop count $(\mathrm{n})$ is 1 to 6 .
6. The program/program-verify flowchart for the $\mathrm{H} 8 / 3052 \mathrm{~F}$ is shown in figure 18.12 .

To cover the points noted above, bits on which reprogramming processing is to be executed, and bits on which additional programming is to be executed, must be determined as shown below.

Since reprogram data and additional-programming data vary according to the progress of the programming procedure, it is recommended that the following data storage areas ( 128 bytes each) be provided in RAM.

Table 18.8 Reprogram Data Computation Table

|  | Result of Verify-Read <br> after Write Pulse | (X) <br> Result of Operation | Comments |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 1 | Programming completed: reprogramming <br> processing not to be executed |
| 0 | 1 | 0 | Programming incomplete: reprogramming <br> processing to be executed |
| 1 | 0 | 1 | - |
| 1 | 1 | 1 | Still in erased state: no action |

Source data of bits on which programming is executed: (D)
Data of bits on which reprogramming is executed: ( X )

Table 18.9 Additional-Programming Data Computation Table Result of Verify-Read after Write Pulse (Y)
X Application (V) Result of Operation Comments

| 0 | 0 | 0 | Programming by write pulse application <br> judged to be completed: additional <br> programming processing to be executed |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 1 | 1 | Programming by write pulse application <br> incomplete: additional programming <br> processing not to be executed |
| 1 | 0 | 1 | Programming already completed: additional <br> programming processing not to be executed |
| 1 | 1 | 1 | Still in erased state: no action |

Data of bits on which additional programming is executed: (Y)
Data of bits on which reprogramming is executed in a certain reprogramming loop: ( $\mathrm{X}^{\prime}$ )
7. It is necessary to execute additional programming processing during the course of the $\mathrm{H} 8 / 3052 \mathrm{~F}$ program/program-verify procedure. However, once 128 -byte-unit programming is finished, additional programming should not be carried out on the same address area. When executing reprogramming, an erase must be executed first. Note that normal operation of reads, etc., is not guaranteed if additional programming is performed on addresses for which a program/program-verify operation has finished.


Notes: 1. Data transfer is performed by byte transfer. The lower 8 bits of the first address written to must be $\mathrm{H}^{\prime} 00$ or $\mathrm{H}^{\prime} 80$.
A 128-byte data transfer must be performed even if writing fewer than 128 bytes; in this case, H'FF data must be written to the extra addresses.
2. Verity data is read in 16 -bit ( $W$ ) units.
3. Reprogram data is determined by the operation shown in the table below (comparison between the data stored in the program data area and the verify data). Bits fo which the reprogram data is 0 are programmed in the next reprogramming loop. Therefore, even bits for which programming has been completed will be subjected to programming once again if the result of the subsequent verify operation is NG.
4. A 128 -byte area for storing program data and a 128 -byte area for storing reprogram data must be provided in RAM The contents of the reprogram data area are modified as programming proceeds.
5. A write pulse of $30 \mu \mathrm{~s}$ or $200 \mu \mathrm{~s}$ should be applied according to the progress of the programming operation. See Note 6 for the pulse widths. When writing of additional programming data is executed, a $10 \mu$ s write pulse should be applied. Reprogram data $X^{\prime}$ means reprogram data when the write pulse is applied
7. The wait times and value of N are shown in section 21.2.5, Flash Memory


Figure 18.12 Program/Program-Verify Flowchart (128-Byte Programming)

### 18.7.4 Erase Mode

To erase an individual flash memory block, follow the erase/erase-verify flowchart (single-block erase) shown in figure 18.13.

The wait times after bits are set or cleared in the flash memory control register (FLMCR1, FLMCR2) and the maximum number of erase operations ( N ) are shown in table 21.10 in section 21.2.5, Flash Memory Characteristics.

To erase flash memory contents, make a 1-bit setting for the flash memory area to be erased in erase block register 1 and 2 (EBR1, EBR2) at least (tsswe) $\mu \mathrm{s}$ after setting the SWE1 and SWE2 bits to 1 in FLMCR1 and FLMCR2. Next, the watchdog timer (WDT) is set to prevent overerasing due to program runaway, etc. Set a value greater than (tse) $\mathrm{ms}+$ (tsesu + tce + tcesu) $\mu \mathrm{s}$ as the WDT overflow period. Preparation for entering erase mode (erase setup) is performed next by setting the ESU1 and ESU2 bits in FLMCR1 and FLMCR2. The operating mode is then switched to erase mode by setting the E1 and E2 bits in FLMCR1 and FLMCR2 after the elapse of at least (tsesu) $\mu$ s. The time during which the E1 and E2 bits are set is the flash memory erase time. Ensure that the erase time does not exceed (tse) ms.

Note: With flash memory erasing, preprogramming (setting all memory data in the memory to be erased to all 0 ) is not necessary before starting the erase procedure.

### 18.7.5 Erase-Verify Mode

In erase-verify mode, data is read after memory has been erased to check whether it has been correctly erased.

After the elapse of the fixed erase time, clear the E1 and E2 bits in FLMCR1 and FLMCR2, then wait for at least (tce) $\mu \mathrm{s}$ before clearing the ESU1 and ESU2 bits to exit erase mode. After exiting erase mode, the watchdog timer setting is also cleared. The operating mode is then switched to erase-verify mode by setting the EV1 and EV2 bits in FLMCR1 and FLMCR2. Before reading in erase-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of (tsev) $\mu \mathrm{s}$ or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least (tsevr) $\mu \mathrm{s}$ after the dummy write before performing this read operation. If the read data has been erased (all 1), a dummy write is performed to the next address, and erase-verify is performed. If the read data is unerased, set erase mode again, and repeat the erase/erase-verify sequence as before. The maximum value for repetition of the erase/erase-verify sequence is indicated by the maximum erase count ( N ). When verification is completed, exit erase-verify mode, and wait for at least (tcev) $\mu$ s. If erasure has been completed on all the erase blocks, clear bits SWE1 and SWE2 in FLMCR1 and FLMCR2, and leave a wait time of at least (tcswe) $\mu \mathrm{s}$.

If erasing multiple blocks, set a single bit in EBR1/EBR2 for the next block to be erased, and repeat the erase/erase-verify sequence as before.

Rev. 2.0, 03/01, page 592 of 822


Notes: 1. Prewriting (setting erase block data to all 0s) is not necessary.
2. Verify data is read in 16 -bit (W) units.
3. Make only a single-bit specification in the erase block registers (EBR1 and EBR2). Two or more bits must not be set simultaneously.
4. The wait times and the value of N are shown in section 21.2.5, Flash Memory Characteristics.

Figure 18.13 Erase/Erase-Verify Flowchart (Single-Block Erase)

### 18.8 Protection

There are three kinds of flash memory program/erase protection: hardware protection, software protection, and error protection.

### 18.8.1 Hardware Protection

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted. Hardware protection is reset by settings in flash memory control register 1 (FLMCR1), flash memory control register 2 (FLMCR2), erase block register 1 (EBR1), and erase block register 2 (EBR2). In the error-protected state, the FLMCR1, FLMCR2, EBR1, and EBR2 settings are retained; the P1 and P2 bits can be set, but a transition is not made to program mode or erase mode. (See table 18.10.)

Table 18.10 Hardware Protection

| Item | Description | Functions |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  | Program | Erase | Verify* ${ }^{1}$ |
| FWE pin protection | - When a low level is input to the FWE pin, FLMCR1, FLMCR2, (except bit FLER) EBR1, and EBR2 are initialized, and the program/ erase-protected state is entered. | No* ${ }^{2}$ | No* ${ }^{3}$ | - |
| Reset/ standby protection | - In a power-on reset (including a WDT power-on reset) and in standby mode, FLMCR1, FLMCR2, EBR1, and EBR2 are initialized, and the program/erase-protected state is entered. <br> - In a reset via the $\overline{\mathrm{RES}}$ pin, the reset state is not entered unless the RES pin is held low until oscillation stabilizes after powering on. In the case of a reset during operation, hold the $\overline{\operatorname{RES}}$ pin low for the $\overline{R E S}$ pulse width specified in the AC Characteristics section. | No | No* ${ }^{3}$ | - |
| Error protection | - When a microcomputer operation error (error generation (FLER=1)) was detected while flash memory was being programmed/erased, error protection is enabled. At this time, the FLMCR1, FLMCR2, EBR1, and EBR2 settings are held, but programming/erasing is aborted at the time the error was generated. Error protection is released only by a reset via the RES pin or a WDT reset, or in the hardware standby mode. | No | No* ${ }^{3}$ | Yes** |

Notes: 1. Two modes: program-verify and erase-verify.
2. Excluding a RAM area overlapping flash memory.
3. All blocks are unerasable and block-by-block specification is not possible.
4. It is possible to perform a program-verify operation on the 128 bytes being programmed, or an erase-verify operation on the block being erased.
5. For details see section 18.11, Flash Memory Programming and Erasing Precautions.
6. See section 4.2.2, Reset Sequence, and section 18.11, Flash Memory Programming and Erasing Precautions. The H8/3052F requires at least 20 system clocks for a reset during operation.

### 18.8.2 Software Protection

Software protection can be implemented by setting the SWE1 bit in FLMCR1, the SWE2 bit in FLMCR2, erase block register 1 (EBR1), erase block register 2 (EBR2), and the RAMS bit in the RAM control register (RAMCR). When software protection is in effect, setting the P1 or E1 bit in flash memory control register 1 (FLMCR1), or the P2 or E2 bit in flash memory control register 2 (FLMCR2) does not cause a transition to program mode or erase mode. (See table 18.11.)

Table 18.11 Software Protection

|  |  | Functions |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Item | Description | Program | Erase | Verify* ${ }^{1}$ |
| Block specification protection | - Erase protection can be set for individual blocks by settings in erase block register 1 (EBR1)*2 and erase block register 2 (EBR2)*². However, programming protection is disabled. <br> - Setting EBR1 and EBR2 to H'00 places all blocks in the erase-protected state. | - | No | Yes |
| Emulation protection | - Setting the RAMS bit to 1 in the RAM control register (RAMCR) places all blocks in the program/erase-protected state. | No* ${ }^{3}$ | No* ${ }^{4}$ | Yes |

Notes: 1. Two modes: program-verify and erase-verify.
2. When not erasing, clear all EBR1, EBR2 bits to 0 .
3. A RAM area overlapping flash memory can be written to.
4. All blocks are unerasable and block-by-block specification is not possible.

### 18.8.3 Error Protection

In error protection, an error is detected when $\mathrm{H} 8 / 3052 \mathrm{~F}$ runaway occurs during flash memory programming/erasing*1, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to overprogramming or overerasing.

If the $\mathrm{H} 8 / 3052 \mathrm{~F}$ malfunctions during flash memory programming/erasing, the FLER bit is set to 1 in FLMCR2 and the error protection state is entered. The FLMCR1, FLMCR2, EBR1, and EBR2 settings* ${ }^{* 3}$ are retained, but program mode or erase mode is aborted at the point at which the error occurred. Program mode or erase mode cannot be re-entered by re-setting the P1, P2, E1, or E2 bit. However, PV1, PV2, EV1 and EV2 bit setting is enabled, and a transition can be made to verify mode. ${ }^{2}$

FLER bit setting conditions are as follows:

1. When the flash memory of the relevant address area is read during programming/erasing (including vector read and instruction fetch)
2. Immediately after exception handling (excluding a reset) during programming/erasing
3. When a SLEEP instruction (including software standby) is executed during programming/erasing
4. When the CPU releases the bus to the DMAC during programming/erasing

Error protection is released only by a power-on reset and in hardware standby mode.
Notes: 1. State in which the P1 bit or E1 bit in FLMCR1, or the P2 bit or E2 bit in FLMCR2, is set to 1 . Note that NMI input is disabled in this state.
2. It is possible to perform a program-verify operation on the 128 bytes being programmed, or an erase-verify on the block being erased.
3. FLMCR1, FLMCR2, EBR1, and EBR2 can be written to. However, the registers are initialized if a transition is made to software standby mode while in the error-protected state.

Figure 18.14 shows the flash memory state transition diagram.


Legend
RD: Memory read possible
VF: Verify-read possible
PR: Programming possible
ER: Erasing possible
$\overline{\mathrm{RD}}$ : Memory read not possible
$\overline{\mathrm{VF}}$ : Verify-read not possible
$\overline{\mathrm{PR}}$ : Programming not possible
$\overline{\mathrm{ER}}$ : Erasing not possible
INIT: Register initialization state

Figure 18.14 Flash Memory State Transitions (Modes 5, 6, and 7 (on-chip ROM enabled), high level applied to FWE pin)

### 18.8.4 NMI Input Disable Conditions

While flash memory is being programed/erased and the boot program is executing in the boot mode (however, period up to branching to on-chip RAM area) ${ }^{*}$, NMI input is disabled because the programming/erasing operations have priority.

This is done to avoid the following operation states:

1. Generation of an NMI input during programming/erasing violates the program/erase algorithms and normal operation can not longer be assured.
2. Vector-read cannot be carried out normally*2 ${ }^{2}$ during NMI exception handling during programming/erasing and the microcomputer runs away as a result.
3. If an NMI input is generated during boot program execution, the normal boot mode sequence cannot be executed.

Therefore, this LSI has conditions that exceptionally disable NMI inputs only in the on-board programming mode. However, this does not assure normal programming/erasing and microcomputer operation.

Thus, in the FWE application state, all requests, including NMI, inside and outside the microcomputer, exception handling, and bus release must be restricted. NMI input is also disabled ${ }^{* 3}$ in the error-protected state and when the P1 bit or E1 bit in FLMCR1, or the P2 bit or E2 bit in FLMCR2, is retained during flash memory emulation by RAM.

Notes: 1. Indicates the period up to branching to the on-chip RAM boot program area (H'FFDF10). (This branch occurs immediately after user program transfer was completed.)
Therefore, after branching to RAM area, NMI input is enabled in states other than the program/erase state. Thus, interrupt requests inside and outside the microcomputer must be disabled until initial writing by user program (writing of vector table and NMI processing program, etc.) is completed.
2. In this case, vector read is not performed normally for the following two reasons:
a. The correct value cannot be read even by reading the flash memory during programming/erasing. (Value is undefined.)
b. If a value has not yet been written to the NMI vector table, NMI exception handling will not be performed correctly.
3. When the emulation function is used, NMI input is prohibited when the P1 bit or E1 bit in FLMCR1, or the P2 bit or E2 bit in FLMCR2, is set to 1, in the same way as with normal programming and erasing. The P1 and E1 bits and the P2 and E2 bits are cleared by a reset (including a watchdog timer reset), in standby mode, when a high level is not being input to the FWE pin, or when the SWE1 bit in FLMCR1 is 0 , or the SWE2 bit in FLMCR2 is 0, while a high level is being input to the FWE pin.

### 18.9 Flash Memory Emulation in RAM

Making a setting in the RAM control register (RAMCR) enables part of RAM to be overlapped onto the flash memory area so that data to be written to flash memory can be emulated in RAM in real time. After the RAMCR setting has been made, accesses can be made from the flash memory area or the RAM area overlapping flash memory. Emulation can be performed in user mode and user program mode. Figure 18.15 shows an example of emulation of real-time flash memory programming.


Figure 18.15 Flowchart for Flash Memory Emulation in RAM


Figure 18.16 Example of RAM Overlap Operation

## Example in which Flash Memory Block Area EB0 is Overlapped

1. Set bits RAMS, RAM2 to RAM0 in RAMCR to $1,0,0,0$, to overlap part of RAM onto the area (EB0) for which real-time programming is required.
2. Real-time programming is performed using the overlapping RAM.
3. After the program data has been confirmed, clear the RAMS bit to release RAM overlap.
4. Write the data written in the overlapping RAM into the flash memory space (EB0).

Notes: 1. When the RAMS bit is set to 1, program/erase protection is enabled for all blocks regardless of the value of RAM2 to RAM0 (emulation protection). In this state, setting the P1 or E1 bit in flash memory control register 1 (FLMCR1), or the P2 or E2 bit in flash memory control register 2 (FLMCR2), will not cause a transition to program mode or erase mode. When actually programming or erasing a flash memory area, the RAMS bit should be cleared to 0 .
2. A RAM area cannot be erased by execution of software in accordance with the erase algorithm while flash memory emulation in RAM is being used.
3. Block area EB0 includes the vector table. When performing RAM emulation, the vector table is needed by the overlap RAM.
4. Flash write enable (FWE) application and releasing

As in on-board programming mode, care is required when applying and releasing FWE to prevent erroneous programming or erasing. To prevent erroneous programming and erasing due to program runaway during FWE application, in particular, the watchdog timer should be set when the P1 or E1 bit in FLMCR1, or the P2 or E2 bit in FLMCR2, is set to 1 , even while the emulation function is being used. For details, see section 18.11, Flash Memory Programming and Erasing Precautions.
5. When the emulation function is used, NMI input is prohibited when the P1 bit or E1 bit in FLMCR1, or the P2 bit or E2 bit in FLMCR2, is set to 1, in the same way as with normal programming and erasing. The P1 and E1 bits and the P2 and E2 bits are cleared by a reset (including a watchdog timer reset), in standby mode, when a high level is not being input to the FWE pin, or when the SWE1 bit in FLMCR1 is 0 , or the SWE2 bit in FLMCR2 is 0, while a high level is being input to the FWE pin.

### 18.10 Flash Memory PROM Mode

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ has a PROM mode as well as the on-board programming modes for programming and erasing flash memory. In PROM mode, the on-chip ROM can be freely programmed using a general-purpose PROM writer that supports the Hitachi microcomputer device type with 512kbyte on-chip flash memory.

### 18.10.1 Socket Adapters and Memory Map

In PROM mode using a PROM writer, memory reading (verification) and writing and flash memory initialization (total erasure) can be performed. For these operations, a special socket adapter is mounted in the PROM writer. The socket adapter product codes are given in table 18.12. In the H8/3052F PROM mode, only the socket adapters shown in this table should be used.

Table 18.12 H8/3052F Socket Adapter Product Codes

| Product Code | Package | Socket Adapter Product Code | Manufacturer |
| :---: | :---: | :---: | :---: |
| HD64F3052F | $\begin{aligned} & \text { 100-pin QFP } \\ & (\text { FP-100B }) \end{aligned}$ | ME3064ESHF1H | Minato Electronics |
| HD64F3052BF |  |  |  |
| HD64F3052BVF |  |  |  |
| HD64F3052TE | $\begin{aligned} & \text { 100-pin TQFP } \\ & \text { (TFP-100B) } \end{aligned}$ | ME3064ESNF1H |  |
| HD64F3052BTE |  |  |  |
| HD64F3052BVTE |  |  |  |
| HD64F3052F | $\begin{aligned} & \text { 100-pin QFP } \\ & \text { (FP-100B) } \end{aligned}$ | HF306BQ100D4001 | Data IO Japan |
| HD64F3052BF |  |  |  |
| HD64F3052BVF |  |  |  |
| HD64F3052TE | $\begin{aligned} & \text { 100-pin TQFP } \\ & \text { (TFP-100B) } \end{aligned}$ | HF306BT100D4001 |  |
| HD64F3052BTE |  |  |  |
| HD64F3052BVTE |  |  |  |

Figure 18.17 shows the memory map in PROM mode.

| $\begin{aligned} & \text { MCU mode } \\ & \text { H'000000 } \end{aligned}$ | H8/3052F | PROM mode H'00000 |
| :---: | :---: | :---: |
|  |  |  |
|  | On-chip ROM |  |
| H'07FFFF |  | H'7FFFF |

Figure 18.17 Memory Map in PROM Mode

### 18.10.2 Notes on Use of PROM Mode

1. A write to a 128 -byte programming unit in PROM mode should be performed once only.

Erasing must be carried out before reprogramming an address that has already been programmed.
2. When using a PROM writer to reprogram a device on which on-board programming/erasing has been performed, it is recommended that erasing be carried out before executing programming.
3. The memory is initially in the erased state when the device is shipped by Hitachi. For samples for which the erasure history is unknown, it is recommended that erasing be executed to check and correct the initialization (erase) level.
4. The H8/3052F does not support a product identification mode as used with general-purpose EPROMs, and therefore the device name cannot be set automatically in the PROM writer.
5. Refer to the instruction manual provided with the socket adapter, or other relevant documentation, for information on PROM writers and associated program versions that are compatible with the PROM mode of the $\mathrm{H} 8 / 3052 \mathrm{~F}$.

### 18.11 Notes on Flash Memory Programming/Erasing

The following describes notes when using the on-board programming mode, RAM emulation function, and PROM mode.

1. Program/erase with the specified voltage and timing.

Applied voltages in excess of the rating can permanently damage the device.
Use a PROM writer that supports the Hitachi 512 kbytes flash memory on-board microcomputer device type.
If the wrong device type is set, a high level may be input to the FWE pin, resulting in permanent damage to the device.
2. Notes on powering on/powering off (See figures 18.18 to 18.20)

Input a high level to the FWE pin after verifying Vcc. Before turning off Vcc, set the FWE pin to a low level.
When powering on and powering off the Vcc power supply, fix the FWE pin low and set the flash memory to the hardware protection mode.
Be sure that the powering on and powering off timing is satisfied even when the power is turned off and back on in the event of a power interruption, etc. If this timing is not satisfied, microcomputer runaway, etc., may cause overprogramming or overerasing and the memory cells may not operate normally.
3. Notes on FWE pin High/Low switching (See figures 18.18 to 18.20) Input FWE in the state microcomputer operation is verified. If the microcomputer does not satisfy the operation confirmation state, fix the FWE pin low to set the protection mode. To prevent erroneous programming/erasing of flash memory, note the following in FWE pin High/Low switching:
a. Apply an input to the FWE pin after the Vcc voltage has stabilized within the rated voltage. If an input is applied to the FWE pin when the microcomputer Vcc voltage does not satisfy the rated voltage, flash memory may be erroneously programmed or erased because the microcomputer is in the unconfirmed state.
b. Apply an input to the FWE pin when the oscillation has stabilized (after the oscillation stabilization time).
When turning on the Vcc power, apply an input to the FWE pin after holding the $\overline{\mathrm{RES}}$ pin at a low level during the oscillation stabilization time ( $\mathrm{t}_{\text {oscl }}=20 \mathrm{~ms}$ ). Do not apply an input to the FWE pin when oscillation is stopped or unstable.
c. In the boot mode, perform FWE pin High/Low switching during reset.

In transition to the boot mode, input $\mathrm{FWE}=$ High level and set $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ while the $\overline{\mathrm{RES}}$ input is low. At this time, the FWE and $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ inputs must satisfy the mode programming setup time ( $\mathrm{t}_{\mathrm{MDS}}$ ) relative to the reset clear timing. The mode programming setup time is necessary for $\overline{\mathrm{RES}}$ reset timing even in transition from the boot mode to another mode.
In reset during operation, the $\overline{\mathrm{RES}}$ pin must be held at a low level for at least 20 system clocks.
d. In the user program mode, $\mathrm{FWE}=\mathrm{High} / \mathrm{Low}$ switching is possible regardless of the $\overline{\mathrm{RES}}$ input.
FWE input switching is also possible during program execution on flash memory.
e. Apply an input to FWE when the program is not running away.

When applying an input to the FWE pin, the program execution state must be supervised using a watchdog timer, etc.
f. Release FWE pin input only when the SWE1, ESU1, PSU1, EV1, PV1, E1, and P1 bits in FLMCR1, and the SWE2, ESU2, PSU2, EV2, PV2, E2, and P2 bits in FLMCR2, are cleared.

Do not erroneously set any of bits SWE1, ESU1, PSU1, EV1, PV1, E1, P1, SWE2, ESU2, PSU2, EV2, PV2, E2, or P2 when applying or releasing FWE.
4. Do not input a constant high level to the FWE pin.

To prevent erroneous programming/erasing in the event of program runaway, etc., input a high level to the FWE pin only when programming/erasing flash memory (including flash memory emulation by RAM). Avoid system configurations that constantly input a high level to the FWE pin. Handle program runaway, etc. by starting the watchdog timer so that flash memory is not overprogrammed/overerased even while a high level is input to the FWE pin.
5. Program/erase the flash memory in accordance with the recommended algorithms.

The recommended algorithms can program/erase the flash memory without applying voltage stress to the device or sacrificing the reliability of the program data.
When setting the PSU1 and ESU1 bits in FLMCR1, or PSU2 and ESU2 bits in FLMCR2 set the watchdog timer for program runaway, etc.

## Accesses to flash memory by means of an MOV instruction, etc., are prohibited while bit $\mathbf{P 1 / P 2}$ or bit $\mathbf{E} 1 / \mathrm{E} 2$ is set.

6. Do not set/clear the SWE bit while a program is executing on flash memory.

Before performing flash memory program execution or data read, clear the SWE bit.
If the SWE bit is set, the flash data can be reprogrammed, but flash memory cannot be accessed for purposes other than verify (verify during programming/erase).
Similarly perform flash memory program execution and data read after clearing the SWE bit even when using the RAM emulation function with a high level input to the FWE pin. However, RAM area that overlaps flash memory space can be read/programmed whether the SWE bit is set or cleared.
7. Do not use an interrupt during flash memory programming or erasing.

Since programming/erase operations (including emulation by RAM) have priority when a high level is input to the FWE pin, disable all interrupt requests, including NMI.
8. Do not perform additional programming. Reprogram flash memory after erasing.

With on-board programming, program to 128 -byte programming unit blocks one time only. Program to 128 -byte programming unit blocks one time only even in the writer mode. Erase all the programming unit blocks before reprogramming.
9. Before programming, check that the chip is correctly mounted in the PROM programmer. Overcurrent damage to the device can result if the index marks on the PROM programmer socket, socket adapter, and chip are not correctly aligned.
10. Do not touch the socket adapter or chip during programming. Touching either of these can cause contact faults and write errors.

$\square$ Flash memory access disabled period
( $x$ : Wait time after SWE1 (2) setting) ${ }^{* 2}$
$\square$ Flash memory reprogrammable period
(Flash memory program execution and data read, other than verify, are disabled.)
Notes: 1. Always fix the level by pulling down or pulling up the mode pins $\left(M D_{2}\right.$ to $\left.M D_{0}\right)$ until powering off, except for mode switching.
2. See 21.2.5 Flash Memory Characteristics.

Figure 18.18 Powering On/Off Timing (Boot Mode)
Rev. 2.0, 03/01, page 606 of 822
WE1(2) bit $\qquad$
SWE1 (2) clear

Flash memory access disabled period (x: Wait time after SWE1(2) setting)*2
Flash memory reprogrammable period (Flash memory program execution and data read, other than verify, are disabled.)

Notes: 1. Always fix the level by pulling down or pulling up the mode pins $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$ up to powering off, except for mode switching.
2. See 21.2.5 Flash Memory Characteristics.

Figure 18.19 Powering On/Off Timing (User Program Mode)


Flash memory access disabled time
(x: Wait time after SWE1 (2) setting)*3
Flash memory reprogammable period
(Flash memory program execution and data read, other than verify, are disabled.)

Notes: 1. In transition to the boot mode and transition from the boot mode to another mode, mode switching via $\overline{\operatorname{RES}}$ input is necessary.
During this switching period (period during which a low level is input to the $\overline{\mathrm{RES}}$ pin), the state of the address dual port and bus control output signals ( $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{WR}})$ changes.
Therefore, do not use these pins as output signals during this switching period.
2. When making a transition from the boot mode to another mode, the mode programming setup time $t_{\text {MDS }}$ relative to the $\overline{R E S}$ clear timing is necessary.
3. See 21.2.5 Flash Memory Characteristics.

Figure 18.20 Mode Transition Timing
(Example: Boot mode $\rightarrow$ User mode $\leftrightarrow$ User program mode)

## Section 19 Clock Pulse Generator

### 19.1 Overview

The H8/3052F has a built-in clock pulse generator (CPG) that generates the system clock ( $\phi$ ) and other internal clock signals ( $\phi / 2$ to $\phi / 4096$ ). After duty adjustment, a frequency divider divides the clock frequency to generate the system clock ( $\phi$ ). The system clock is output at the $\phi$ pin*1 and furnished as a master clock to prescalers that supply clock signals to the on-chip supporting modules. Frequency division ratios of $1 / 1,1 / 2,1 / 4$, and $1 / 8$ can be selected ${ }^{* 2}$ for the frequency divider by settings in a division control register (DIVCR). Power consumption in the chip is reduced in almost direct proportion to the frequency division ratio.

Notes: 1. Usage of the $\phi$ pin differs depending on the chip operating mode and the PSTOP bit setting in the module standby control register (MSTCR). For details, see section 20.7, System Clock Output Disabling Function.
2. The division ratio of the frequency divider can be changed dynamically during operation. The clock output at the $\phi$ pin also changes when the division ratio is changed. The frequency output at the $\phi$ pin is shown below.

$$
\phi=\operatorname{EXTAL} \times \mathrm{n}
$$

where, EXTAL: Frequency of crystal resonator or external clock signal $\mathrm{n}: \quad$ Frequency division ratio ( $\mathrm{n}=1 / 1,1 / 2,1 / 4$, or $1 / 8$ )

### 19.1.1 Block Diagram

Figure 19.1 shows a block diagram of the clock pulse generator.


Figure 19.1 Block Diagram of Clock Pulse Generator

### 19.2 Oscillator Circuit

Clock pulses can be supplied by connecting a crystal resonator, or by input of an external clock signal.

### 19.2.1 Connecting a Crystal Resonator

Circuit Configuration: A crystal resonator can be connected as in the example in figure 19.2. The damping resistance Rd should be selected according to table 19.1(1). An AT-cut parallelresonance crystal should be used.


Figure 19.2 Connection of Crystal Resonator (Example)
If a crystal resonator with a frequency higher than 20 MHz is connected, the external load capacitance values in table 20.1(2)should not exceed $10[\mathrm{pF}]$ Also,in order to improve the accuracy of the oscillation frequency a thorough study of oscillation matching evaluation etc. should be carried out when deciding the circuit constants.

Table 19.1(1) Damping Resistance Value


Crystal Resonator: Figure 19.3 shows an equivalent circuit of the crystal resonator. The crystal resonator should have the characteristics listed in table 19.2.


Figure 19.3 Crystal Resonator Equivalent Circuit
Table 19.2 Crystal Resonator Parameters

| Frequency (MHz) | $\mathbf{2}$ | $\mathbf{4}$ | $\mathbf{8}$ | $\mathbf{1 0}$ | $\mathbf{1 2}$ | $\mathbf{1 6}$ | $\mathbf{1 8}$ | $\mathbf{2 0}$ | $\mathbf{2 5}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Rs max $(\Omega)$ | 500 | 120 | 80 | 70 | 60 | 50 | 40 | 40 | 40 |
| Co (pF) | 7 pF |  |  |  |  |  |  |  |  |
|  | $\max$ |  |  |  |  |  |  |  |  |

Use a crystal resonator with a frequency equal to the system clock frequency $(\phi)$.
Notes on Board Design: When a crystal resonator is connected, the following points should be noted:

Other signal lines should be routed away from the oscillator circuit to prevent induction from interfering with correct oscillation. See figure 19.4.

When the board is designed, the crystal resonator and its load capacitors should be placed as close as possible to the XTAL and EXTAL pins.


Figure 19.4 Example of Incorrect Board Design
Rev. 2.0, 03/01, page 612 of 822

### 19.2.2 External Clock Input

Circuit Configuration: An external clock signal can be input as shown in the examples in figure 19.5. If the XTAL pin is left open, the stray capacitance should not exceed 10 pF . If the stray capacitance at the XTAL pin exceeds 10 pF in configuration a, use configuration b instead and hold the clock high in standby mode.


Figure 19.5 External Clock Input (Examples)

External Clock: The external clock frequency should be equal to the system clock frequency ( $\phi$ ) when not divided by the on-chip frequency divider. Table 19.3 shows the clock timing, and figure 19.6 shows the external clock input timing. Figure 19.7 shows the external clock output stabilization delay timing.

When the appropriate external clock is input via the EXTAL pin, its waveform is corrected by the on-chip oscillator and duty adjustment circuit. The resulting stable clock is output to external devices after the external clock settling time ( $\mathrm{t}_{\text {DEXT }}$ ) has passed after the clock input. The system must remain reset with the reset signal low during $\mathrm{t}_{\mathrm{DEXT}}$, while the clock output is unstable.

Table 19.3 Clock Timing

| Item | Symbol | $\begin{aligned} & \mathrm{V}_{\mathrm{cc}}=5.0 \mathrm{~V} \pm 10 \% \\ & \mathrm{~V}_{\mathrm{cc}}=3.0 \mathrm{~V} \text { to } 3.6 \mathrm{~V} \end{aligned}$ |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max |  |  |
| External clock input low pulse width | $\mathrm{t}_{\text {ExL }}$ | 15 | - | ns | Figure 19.6 |
| External clock input high pulse width | $\mathrm{t}_{\text {EXH }}$ | 15 | - | ns |  |
| External clock rise time | $\mathrm{t}_{\text {EXI }}$ | - | 5 | ns |  |
| External clock fall time | $\mathrm{t}_{\text {EXf }}$ | - | 5 | ns |  |
| Clock low pulse width | $\mathrm{t}_{\mathrm{cL}}$ | 0.4 | 0.6 | $t_{\text {cyc }}$ | $\phi \geq 5 \mathrm{MHz}$ Figure 21.4 |
|  |  | 80 | - | ns | $\phi<5 \mathrm{MHz}$ |
| Clock high pulse width | $\mathrm{t}_{\mathrm{CH}}$ | 0.4 | 0.6 | $\mathrm{t}_{\text {cyc }}$ | $\phi \geq 5 \mathrm{MHz}$ |
|  |  | 80 | - | ns | $\phi<5 \mathrm{MHz}$ |
| External clock output settling delay time | $\mathrm{t}_{\mathrm{DEXT}}$ * | 500 | - | $\mu \mathrm{s}$ | Figure 19.7 |

Note: * $t_{\text {DEXT }}$ includes $10 t_{\text {cyc }}$ of $\overline{R E S}$ pulse width ( $\left.t_{\text {RESw }}\right)$.


Figure 19.6 External Clock Input Timing
Rev. 2.0, 03/01, page 614 of 822


Note: * $t_{\text {DEXT }}$ includes $10 \mathrm{t}_{\text {cyc }}$ of $\overline{\mathrm{RES}}$ pulse width ( $\mathrm{t}_{\text {RESw }}$ ).

Figure 19.7 External Clock Output Settling Delay Timing

### 19.3 Duty Adjustment Circuit

When the oscillator frequency is 5 MHz or higher, the duty adjustment circuit adjusts the duty cycle of the clock signal from the oscillator to generate the signal that becomes the system clock.

### 19.4 Prescalers

The prescalers divide the system clock ( $\phi$ ) to generate internal clocks ( $\phi / 2$ to $\phi / 4096$ ).

### 19.5 Frequency Divider

The frequency divider divides the duty-adjusted clock signal to generate the system clock $(\phi)$. The frequency division ratio can be changed dynamically by modifying the value in DIVCR, as described below. Power consumption in the chip is reduced in almost direct proportion to the frequency division ratio. The system clock generated by the frequency divider can be output at the $\phi$ pin.

### 19.5.1 Register Configuration

Table 19.4 summarizes the frequency division register.
Table 19.4 Frequency Division Register

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FF5D | Division control register | DIVCR | R/W | H'FC |

Note: * The lower 16 bits of the address are shown.

### 19.5.2 Division Control Register (DIVCR)

DIVCR is an 8-bit readable/writable register that selects the division ratio of the frequency divider.


DIVCR is initialized to H'FC by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 2—Reserved: Read-only bits, always read as 1.
Bits 1 and 0—Divide (DIV1 and DIV0): These bits select the frequency division ratio, as follows.

| Bit 1: DIV1 | Bit 0: DIV0 | Frequency Division Ratio |  |
| :--- | :--- | :--- | :--- |
| 0 | 0 | $1 / 1$ | (Initial value) |
|  | 1 | $1 / 2$ |  |
| 1 | 0 | $1 / 4$ |  |
| 1 | $1 / 8$ |  |  |

### 19.5.3 Usage Notes

The DIVCR setting changes the $\phi$ frequency, so note the following points.

- Select a frequency division ratio that stays within the assured operation range specified for the clock cycle time $t_{\text {cyc }}$ in the AC electrical characteristics. Note that $\phi_{\text {min }}=2 \mathrm{MHz}$. Avoid settings that give system clock frequencies less than 2 MHz .
- All on-chip module operations are based on $\phi$. Note that the timing of timer operations, serial communication, and other time-dependent processing differs before and after any change in the division ratio. The waiting time for exit from software standby mode also changes when the division ratio is changed. For details, see section 20.4.3, Selection of Waiting Time for Exit from Software Standby Mode.


## Section 20 Power-Down State

### 20.1 Overview

The $\mathrm{H} 8 / 3052 \mathrm{~F}$ has a power-down state that greatly reduces power consumption by halting the CPU, and a module standby function that reduces power consumption by selectively halting onchip modules.

The power-down state includes the following three modes:

- Sleep mode
- Software standby mode
- Hardware standby mode

The module standby function can halt on-chip supporting modules independently of the powerdown state. The modules that can be halted are the ITU, SCI0, SCI1, DMAC, refresh controller, and $\mathrm{A} / \mathrm{D}$ converter.

Table 20.1 indicates the methods of entering and exiting the power-down modes and module standby mode, and gives the status of the CPU and on-chip supporting modules in each mode.

Table 20.1 Power-Down State and Module Standby Function

|  |  | State |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mode | Entering Conditions | Clock | CPU | CPU <br> Registers | DMAC | Refresh Controller | ITU | SCIO | SCl1 | A/D | Other Modules | RAM | $\phi$ Clock Output | I/O Ports | Exiting Conditions |
| Sleep mode | SLEEP instruction executed while SSBY = 0 in SYSCR | Active | Halted | Held | Active | Active | Active | Active | Active | Active | Active | Held | ¢ output | Held | - Interrupt <br> - RES <br> - STBY |
| Software standby mode | SLEEP <br> instruction <br> executed <br> while <br> SSBY = 1 <br> in SYSCR | Halted | Halted | Held | Halted and reset | Halted and held ${ }^{* 1}$ | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Held | High output | Held | - NMI <br> - RRQ $_{0}$ to $\mathrm{IRQ}_{2}$ <br> - RES <br> - STBY |
| Hardware standby mode | Low input at STBY pin | Halted | Halted | Undetermined | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Halted and reset | Held*3 | High impedance | High impedance | - STBY <br> - RES |
| Module standby | Corresponding bit set to 1 in MSTCR | Active | Active | - | Halted*2 and reset | Halted ${ }^{* 2}$ and held ${ }^{* 1}$ | Halted ${ }^{* 2}$ and reset | Halted ${ }^{* 2}$ and reset | Halted ${ }^{* 2}$ and reset | Halted ${ }^{* 2}$ and reset | Active | - | High impedance ${ }^{* 2}$ |  | - STBY <br> - RES <br> - Clear MSTCR bit to $0^{* 4}$ |

[^3]
### 20.2 Register Configuration

The H8/3052F has a system control register (SYSCR) that controls the power-down state, and a module standby control register (MSTCR) that controls the module standby function. Table 20.2 summarizes these registers.

Table 20.2 Control Register

| Address* | Name | Abbreviation | R/W | Initial Value |
| :--- | :--- | :--- | :--- | :--- |
| H'FFF2 | System control register | SYSCR | R/W | H'0B |
| H'FF5E | Module standby control register | MSTCR | R/W | H'40 |

Note: * Lower 16 bits of the address.

### 20.2.1 System Control Register (SYSCR)



SYSCR is an 8-bit readable/writable register. Bit 7 (SSBY) and bits 6 to 4 (STS2 to STS0) control the power-down state. For information on the other SYSCR bits, see section 3.3, System Control Register (SYSCR).

Bit 7—Software Standby (SSBY): Enables transition to software standby mode. When software standby mode is exited by an external interrupt, this bit remains set to 1 after the return to normal operation. To clear this bit, write 0 .

Bit 7: SSBY Description

| 0 | SLEEP instruction causes transition to sleep mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | SLEEP instruction causes transition to software standby mode |  |

Bits 6 to 4—Standby Timer Select (STS2 to STS0): These bits select the length of time the CPU and on-chip supporting modules wait for the clock to settle when software standby mode is exited by an external interrupt. If the clock is generated by a crystal resonator, set these bits according to the clock frequency so that the waiting time will be at least 7 ms . See table 20.3. If an external clock is used, Set these bits according to the operating frequency so that the waiting time will be at least 100 $\mu \mathrm{s}$.

| Bit 6: STS2 | Bit 5: STS1 | Bit 4: STS0 | Description |  |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Waiting time $=8,192$ states | (Initial value) |
|  |  | 1 | Waiting time $=16,384$ states |  |
|  | 1 | 0 | Waiting time $=32,768$ states |  |
| 1 | 0 | 0 | Waiting time $=65,536$ states |  |
|  |  | 1 | Waiting time $=131,072$ states |  |

### 20.2.2 Module Standby Control Register (MSTCR)

MSTCR is an 8-bit readable/writable register that controls output of the system clock $(\phi)$. It also controls the module standby function, which places individual on-chip supporting modules in the standby state. Module standby can be designated for the ITU, SCI0, SCI1, DMAC, refresh controller, and A/D converter modules.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PSTOP | - | MSTOP5 | MSTOP4 | MSTOP3 | MSTOP2 | MSTOP1 | MSTOPO |
| Initial value | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | - | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  | Res | ed bit |  | Module standby 5 to 0 <br> These bits select modules to be placed in standby |  |  |  |
|  | ø cloc Enable output | top <br> r dis the sy | s <br> clock |  |  |  |  |  |

MSTCR is initialized to H'40 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bit 7— $\phi$ Clock Stop (PSTOP): Enables or disables output of the system clock $(\phi)$.
Bit 7: PSTOP Description

| 0 | System clock output is enabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | System clock output is disabled |  |

Bit 6-Reserved: Read-only bit, always read as 1.
Bit 5—Module Standby 5 (MSTOP5): Selects whether to place the ITU in standby.
Bit 5: MSTOP5 Description

| 0 | ITU operates normally | (Initial value) |
| :--- | :--- | :---: |
| 1 | ITU is in standby state |  |

Bit 4—Module Standby 4 (MSTOP4): Selects whether to place SCI0 in standby.

## Bit 4: MSTOP4 Description

| 0 | SCIO operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | SCIO is in standby state |  |

Bit 3—Module Standby 3 (MSTOP3): Selects whether to place SCI1 in standby.

## Bit 3: MSTOP3 Description

| 0 | SCl1 operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | SCl1 is in standby state |  |

Bit 2—Module Standby 2 (MSTOP2): Selects whether to place the DMAC in standby.
Bit 2: MSTOP2 Description

| 0 | DMAC operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | DMAC is in standby state |  |

Bit 1—Module Standby 1 (MSTOP1): Selects whether to place the refresh controller in standby.
Bit 1: MSTOP1 Description

| 0 | Refresh controller operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | Refresh controller is in standby state |  |

Bit 0—Module Standby 0 (MSTOP0): Selects whether to place the A/D converter in standby. Bit 0: MSTOPO Description

| 0 | A/D converter operates normally | (Initial value) |
| :--- | :--- | :--- |
| 1 | A/D converter is in standby state |  |

### 20.3 Sleep Mode

### 20.3.1 Transition to Sleep Mode

When the SSBY bit is cleared to 0 in SYSCR, execution of the SLEEP instruction causes a transition from the program execution state to sleep mode. Immediately after executing the SLEEP instruction the CPU halts, but the contents of its internal registers are retained. The DMA controller (DMAC), refresh controller, and on-chip supporting modules do not halt in sleep mode. Modules which have been placed in standby by the module standby function, however, remain halted.

### 20.3.2 Exit from Sleep Mode

Sleep mode is exited by an interrupt, or by input at the $\overline{\operatorname{RES}}$ or $\overline{\mathrm{STBY}}$ pin.
Exit by Interrupt: An interrupt terminates sleep mode and causes a transition to the interrupt exception handling state. Sleep mode is not exited by an interrupt source in an on-chip supporting module if the interrupt is disabled in the on-chip supporting module. Sleep mode is not exited by an interrupt other than NMI if the interrupt is masked by the I and UI bits in CCR and IPR.

Exit by $\overline{\text { RES }}$ Input: Low input at the $\overline{\text { RES }}$ pin exits from sleep mode to the reset state.
Exit by $\overline{\text { STBY }}$ Input: Low input at the $\overline{\text { STBY }}$ pin exits from sleep mode to hardware standby mode.

### 20.4.1 Transition to Software Standby Mode

To enter software standby mode, execute the SLEEP instruction while the SSBY bit is set to 1 in SYSCR.

In software standby mode, current dissipation is reduced to an extremely low level because the CPU, clock, and on-chip supporting modules all halt. The DMAC and on-chip supporting modules are reset. As long as the specified voltage is supplied, however, CPU register contents and on-chip RAM data are retained. The settings of the I/O ports and refresh controller* are also held.

Note: * RTCNT and bits 7 and 6 of RTMCSR are initialized. Other bits and registers hold their previous states.

### 20.4.2 Exit from Software Standby Mode

Software standby mode can be exited by input of an external interrupt at the NMI, $\overline{\operatorname{IRQ}}_{0}, \overline{\operatorname{IRQ}}_{1}$, or $\overline{\mathrm{IRQ}}_{2}$ pin, or by input at the $\overline{\mathrm{RES}}$ or $\overline{\text { STBY }}$ pin.

Exit by Interrupt: When an NMI, $\mathrm{IRQ}_{0}, \mathrm{IRQ}_{1}$, or $\mathrm{IRQ}_{2}$ interrupt request signal is received, the clock oscillator begins operating. After the oscillator settling time selected by bits STS2 to STS0 in SYSCR, stable clock signals are supplied to the entire chip, software standby mode ends, and interrupt exception handling begins. Software standby mode is not exited if the interrupt enable bits of interrupts $\mathrm{IRQ}_{0}, I R Q_{1}$, and $\mathrm{IRQ}_{2}$ are cleared to 0 , or if these interrupts are masked in the CPU.

Exit by $\overline{\text { RES }}$ Input: When the $\overline{\mathrm{RES}}$ input goes low, the clock oscillator starts and clock pulses are supplied immediately to the entire chip. The $\overline{\mathrm{RES}}$ signal must be held low long enough for the clock oscillator to stabilize. When $\overline{\mathrm{RES}}$ goes high, the CPU starts reset exception handling.

Exit by $\overline{\text { STBY }}$ Input: Low input at the $\overline{\text { STBY }}$ pin causes a transition to hardware standby mode.

### 20.4.3 Selection of Waiting Time for Exit from Software Standby Mode

Bits STS2 to STS0 in SYSCR and bits DIV1 and DIV0 in DIVCR should be set as follows.
Crystal Resonator: Set STS2 to STS0, DIV1, and DIV0 so that the waiting time (for the clock to stabilize) is at least 7 ms . Table 20.3 indicates the waiting times that are selected by STS2 to STS0, DIV1, and DIV0 settings at various system clock frequencies.

External Clock: Set STS2 to STS0, DIV0, and DIV1 so that the waiting time is at least $100 \mu \mathrm{~s}$.
Table 20.3 Clock Frequency and Waiting Time for Clock to Settle

| DIV1 | DIVO | STS2 | STS1 | STS0 | Waiting Time | 18 MHz | 16 MHz | 12 MHz | 10 MHz | 8 MHz | 6 MHz | 4 MHz | 2 MHz | 1 MHz | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | $\begin{aligned} & \hline 8192 \\ & \text { states } \end{aligned}$ | 0.46 | 0.51 | 0.65 | 0.8 | 1.0 | 1.3 | 2.0 | 4.1 | 8.2 | ms |
|  |  | 0 | 0 | 1 | $\begin{aligned} & 16384 \\ & \text { states } \end{aligned}$ | 0.91 | 1.0 | 1.3 | 1.6 | 2.0 | 2.7 | 4.1 | 8.2 | 16.4 |  |
|  |  | 0 | 1 | 0 | $\begin{aligned} & 32768 \\ & \text { states } \end{aligned}$ | 1.8 | 2.0 | 2.7 | 3.3 | 4.1 | 5.5 | 8.2 | 16.4 | 32.8 |  |
|  |  | 0 | 1 | 1 | $\begin{aligned} & 65536 \\ & \text { states } \end{aligned}$ | 3.6 | 4.1 | 5.5 | 6.6 | 8.2 | 10.9 | 16.4 | 32.8 | 65.5 |  |
|  |  | 1 | 0 | 0 | $\begin{aligned} & 131072 \\ & \text { states } \end{aligned}$ | 7.3 | 8.2 | 10.9 | 13.1 | 16.4 | 21.8 | 32.8 | 65.5 | 131.1 |  |
|  |  | 1 | 0 | 1 | $\begin{aligned} & \hline 1024 \\ & \text { states } \end{aligned}$ | 0.057 | 0.064 | 0.085 | 0.10 | 0.13 | 0.17 | 0.26 | 0.51 | 1.0 |  |
|  |  | 1 | 1 | - | Illegal setting |  |  |  |  |  |  |  |  |  |  |
| 0 | 1 | 0 | 0 | 0 | $\begin{aligned} & 8192 \\ & \text { states } \end{aligned}$ | 0.91 | 1.02 | 1.4 | 1.6 | 2.0 | 2.7 | 4.0 | 8.2 | 16.4 | ms |
|  |  | 0 | 0 | 1 | $\begin{aligned} & 16384 \\ & \text { states } \end{aligned}$ | 1.8 | 2.0 | 2.7 | 3.3 | 4.1 | 5.5 | 8.2 | 16.4 | 32.8 |  |
|  |  | 0 | 1 | 0 | $\begin{aligned} & 32768 \\ & \text { states } \end{aligned}$ | 3.6 | 4.1 | 5.5 | 6.6 | 8.2 | 10.9 | 16.4 | 32.8 | 65.5 |  |
|  |  | 0 | 1 | 1 | $\begin{aligned} & 65536 \\ & \text { states } \end{aligned}$ | 7.3 | 8.2 | 10.9 | 13.1 | 16.4 | 21.8 | 32.8 | 65.5 | 131.1 |  |
|  |  | 1 | 0 | 0 | $\begin{aligned} & 131072 \\ & \text { states } \end{aligned}$ | 14.6 | 16.4 | 21.8 | 26.2 | 32.8 | 43.7 | 65.5 | 131.1 | 262.1 |  |
|  |  | 1 | 0 | 1 | $\begin{aligned} & 1024 \\ & \text { states } \end{aligned}$ | 0.11 | 0.13 | 0.17 | 0.20 | 0.26 | 0.34 | 0.51 | 1.0 | 2.0 |  |
|  |  | 1 | 1 | - | Illegal setting |  |  |  |  |  |  |  |  |  |  |


| DIV1 | DIVO | STS2 | STS1 | STS0 | Waiting Time | 18 MHz | 16 MHz | 12 MHz | 10 MHz | 8 MHz | 6 MHz | 4 MHz | 2 MHz | 1 MHz | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 0 | 0 | $\begin{aligned} & \hline 8192 \\ & \text { states } \end{aligned}$ | 1.8 | 2.0 | 2.7 | 3.3 | 4.1 | 5.5 | 8.2 | 16.4 | 32.8 | ms |
|  |  | 0 | 0 | 1 | $\begin{aligned} & 16384 \\ & \text { states } \end{aligned}$ | 3.6 | 4.1 | 5.5 | 6.6 | 8.2 | 10.9 | 16.4 | 32.8 | 65.5 |  |
|  |  | 0 | 1 | 0 | $\begin{aligned} & 32768 \\ & \text { states } \end{aligned}$ | 7.3 | 8.2 | 10.9 | 13.1 | 16.4 | 21.8 | 32.8 | 65.5 | 131.1 |  |
|  |  | 0 | 1 | 1 | $\begin{aligned} & 65536 \\ & \text { states } \end{aligned}$ | 14.6 | 16.4 | 21.8 | 26.2 | 32.8 | 43.7 | 65.5 | 131.1 | 262.1 |  |
|  |  | 1 | 0 | 0 | $131072$ states | 29.1 | 32.8 | 43.7 | 52.4 | 65.5 | 87.4 | 131.1 | 262.1 | 524.3 |  |
|  |  | 1 | 0 | 1 | $\begin{aligned} & 1024 \\ & \text { states } \end{aligned}$ | 0.23 | 0.26 | 0.34 | 0.41 | 0.51 | 0.68 | 1.02 | 2.0 | 4.1 |  |
|  |  | 1 | 1 | - | Illegal setting |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 0 | 0 | 0 | $\begin{aligned} & 8192 \\ & \text { states } \end{aligned}$ | 3.6 | 4.1 | 5.5 | 6.6 | 8.2 | 10.9 | 16.4 | 32.8 | 65.5 | ms |
|  |  | 0 | 0 | 1 | $\begin{aligned} & 16384 \\ & \text { states } \end{aligned}$ | 7.3 | 8.2 | 10.9 | 13.1 | 16.4 | 21.8 | 32.8 | 65.5 | 131.1 |  |
|  |  | 0 | 1 | 0 | $\begin{aligned} & 32768 \\ & \text { states } \end{aligned}$ | 14.6 | 16.4 | 21.8 | 26.2 | 32.8 | 43.7 | 65.5 | 131.1 | 262.1 |  |
|  |  | 0 | 1 | 1 | $\begin{aligned} & 65536 \\ & \text { states } \end{aligned}$ | 29.1 | 32.8 | 43.7 | 52.4 | 65.5 | 87.4 | 131.1 | 262.1 | 524.3 |  |
|  |  | 1 | 0 | 0 | $\begin{aligned} & 131072 \\ & \text { states } \end{aligned}$ | 58.3 | 65.5 | 87.4 | 104.9 | 131.1 | 174.8 | 262.1 | 524.3 | 1048.6 |  |
|  |  | 1 | 0 | 1 | $\begin{aligned} & \hline 1024 \\ & \text { states } \end{aligned}$ | 0.46 | 0.51 | 0.68 | 0.82 | 1.0 | 1.4 | 2.0 | 4.1 | 8.2 |  |
|  |  | 1 | 1 | - | Illegal setting |  |  |  |  |  |  |  |  |  |  |

Bold face is recommended setting

### 20.4.4 Sample Application of Software Standby Mode

Figure 20.1 shows an example in which software standby mode is entered at the fall of NMI and exited at the rise of NMI.

With the NMI edge select bit (NMIEG) cleared to 0 in SYSCR (selecting the falling edge), an NMI interrupt occurs. Next the NMIEG bit is set to 1 (selecting the rising edge) and the SSBY bit is set to 1 ; then the SLEEP instruction is executed to enter software standby mode.

Software standby mode is exited at the next rising edge of the NMI signal.


Figure 20.1 NMI Timing for Software Standby Mode (Example)

### 20.4.5 Note

The I/O ports retain their existing states in software standby mode. If a port is in the high output state, its output current is not reduced.

### 20.5 Hardware Standby Mode

### 20.5.1 Transition to Hardware Standby Mode

Regardless of its current state, the chip enters hardware standby mode whenever the $\overline{\text { STBY }}$ pin goes low. Hardware standby mode reduces power consumption drastically by halting all functions of the CPU, DMAC, refresh controller, and on-chip supporting modules. All modules are reset except the on-chip RAM. As long as the specified voltage is supplied, on-chip RAM data is retained. I/O ports are placed in the high-impedance state.

Clear the RAME bit to 0 in SYSCR before $\overline{\text { STBY }}$ goes low to retain on-chip RAM data.
The inputs at the mode pins $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$ should not be changed during hardware standby mode.

### 20.5.2 Exit from Hardware Standby Mode

Hardware standby mode is exited by inputs at the $\overline{\text { STBY }}$ and $\overline{\text { RES }}$ pins. While $\overline{\text { RES }}$ is low, when $\overline{\text { STBY }}$ goes high, the clock oscillator starts running. $\overline{\text { RES }}$ should be held low long enough for the clock oscillator to settle. When $\overline{\mathrm{RES}}$ goes high, reset exception handling begins, followed by a transition to the program execution state.

### 20.5.3 Timing for Hardware Standby Mode

Figure 20.2 shows the timing relationships for hardware standby mode. To enter hardware standby mode, first drive $\overline{\operatorname{RES}}$ low, then drive $\overline{\mathrm{STBY}}$ low. To exit hardware standby mode, first drive $\overline{\text { STBY }}$ high, wait for the clock to settle, then bring $\overline{\mathrm{RES}}$ from low to high.


Figure 20.2 Hardware Standby Mode Timing

### 20.6 Module Standby Function

### 20.6.1 Module Standby Timing

The module standby function can halt several of the on-chip supporting modules (the ITU, SCIO, SCI1, DMAC, refresh controller, and A/D converter) independently of the power-down state. This standby function is controlled by bits MSTOP5 to MSTOP0 in MSTCR. When one of these bits is set to 1 , the corresponding on-chip supporting module is placed in standby and halts at the beginning of the next bus cycle after the MSTCR write cycle.

### 20.6.2 Read/Write in Module Standby

When an on-chip supporting module is in module standby, read/write access to its registers is disabled. Read access always results in H'FF data. Write access is ignored.

### 20.6.3 Usage Notes

When using the module standby function, note the following points.
DMAC and Refresh Controller: When setting bit MSTOP2 or MSTOP1 to 1 to place the DMAC or refresh controller in module standby, make sure that the DMAC or refresh controller is not currently requesting the bus right. If bit MSTOP2 or MSTOP1 is set to 1 when a bus request is present, operation of the bus arbiter becomes ambiguous and a malfunction may occur.

Internal Peripheral Module Interrupt: When MSTCR is set to " 1 ", prevent module interrupt in advance. When an on-chip supporting module is placed in standby by the module standby function, its registers are initialized.

Pin States: Pins used by an on-chip supporting module lose their module functions when the module is placed in module standby. What happens after that depends on the particular pin. For details, see section 9, I/O Ports. Pins that change from the input to the output state require special care. For example, if SCI1 is placed in module standby, the receive data pin loses its receive data function and becomes a generic I/O pin. If its data direction bit is set to 1 , the pin becomes a data output pin, and its output may collide with external serial data. Data collisions should be prevented by clearing the data direction bit to 0 or taking other appropriate action.

Register Resetting: When an on-chip supporting module is halted by the module standby function, all its registers are initialized. To restart the module, after its MSTOP bit is cleared to 0 , its registers must be set up again. It is not possible to write to the registers while the MSTOP bit is set to 1 .

MSTCR Access from DMAC Disabled: To prevent malfunctions, MSTCR can only be accessed from the CPU. It can be read by the DMAC, but it cannot be written by the DMAC.

### 20.7 System Clock Output Disabling Function

Output of the system clock ( $\phi$ ) can be controlled by the PSTOP bit in MSTCR. When the PSTOP bit is set to 1 , output of the system clock halts and the $\phi$ pin is placed in the high-impedance state. Figure 20.3 shows the timing of the stopping and starting of system clock output. When the PSTOP bit is cleared to 0 , output of the system clock is enabled. Table 20.4 indicates the state of the $\phi$ pin in various operating states.


Figure 20.3 Starting and Stopping of System Clock Output
Table 20.4 $\phi$ Pin State in Various Operating States

| Operating State | PSTOP $=\mathbf{0}$ | PSTOP $=\mathbf{1}$ |
| :--- | :--- | :--- |
| Hardware standby | High impedance | High impedance |
| Software standby | Always high | High impedance |
| Sleep mode | System clock output | High impedance |
| Normal operation | System clock output | High impedance |

## Section 21 Electrical Characteristics (Preliminary)

### 21.1 Absolute Maximum Ratings

Table 21.1 lists the absolute maximum ratings.

Table 21.1 Absolute Maximum Ratings

| Item | Symbol | Value | Unit |
| :---: | :---: | :---: | :---: |
| Power supply voltage | $\mathrm{V}_{\text {cc }}$ | 5 V operation : -0.3 to +7.0 | V |
|  |  | 3 V operation : -0.3 to +4.3 |  |
| Programming voltage HD64F3052 (FWE) | $\mathrm{V}_{\text {in }}$ | -0.3 to $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |
| Input voltage (except for port 7) | $\mathrm{V}_{\text {in }}$ | -0.3 to $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |
| Input voltage (port 7) | $V_{\text {in }}$ | -0.3 to $\mathrm{AV}_{\mathrm{cc}}+0.3$ | V |
| Reference voltage | $\mathrm{V}_{\text {REF }}$ | -0.3 to $\mathrm{AV}_{\mathrm{cc}}+0.3$ | V |
| Analog power supply voltage | $\mathrm{AV}_{\mathrm{cc}}$ | 5 V operation : -0.3 to +7.0 | V |
|  |  | 3 V operation : -0.3 to +4.3 |  |
| Analog input voltage | $\mathrm{V}_{\text {AN }}$ | -0.3 to $\mathrm{AV}_{\mathrm{cc}}+0.3$ | V |
| Operating temperature | $\mathrm{T}_{\text {opr }}$ | -20 to +75* | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |

Notes: During 5V operation, connect an external capacitor to the $\mathrm{V}_{\mathrm{cL}}$ pin. During 3V operation, the $\mathrm{V}_{\mathrm{cL}}$ pin functions as the $\mathrm{V}_{\mathrm{cc}}$ pin so the power supply voltage should be applied.Connect an external capacitor between this pin and ground.

* For flash memory program/erase operations, the operating temperature range is $\mathrm{T}_{\mathrm{a}}=0$ to $+75^{\circ} \mathrm{C}$.

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

### 21.2 Electrical Characteristics

### 21.2.1 DC Characteristics

Table 21.2 lists the DC characteristics. Table 21.3 lists the permissible output currents.
Table 21.2(1) DC Characteristics
Conditions: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{sS}}=\mathrm{AV}_{\mathrm{sS}}=0 \mathrm{~V}^{* 1}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item |  | Symbol | Min | Typ | Max | Unit Test Conditions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Schmitt trigger input voltages | Port A, $\mathrm{P} 8{ }_{0}$ to $\mathrm{P8}_{2}$, $\mathrm{PB}_{0}$ to $\mathrm{PB}_{3}$ | $\mathrm{V}_{T}{ }^{-}$ | 1.0 | - | - | V |  |
|  |  | $\mathrm{V}_{T}+$ | - | - | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | V |  |
|  |  | $\mathrm{V}_{T}^{+}-\mathrm{V}_{T}{ }^{-}$ | 0.4 | - | - | V |  |
| Input high voltage | $\overline{\mathrm{RES}}, \overline{\mathrm{STBY}}$, FWE, NMI, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ | $\mathrm{V}_{\text {IH }}$ | $\mathrm{V}_{\mathrm{cc}}-0.7$ | - | $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |  |
|  | EXTAL |  | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | - | $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |  |
|  | Port 7 |  | 2.0 | - | $\mathrm{AV}_{\mathrm{cc}}+0.3$ | V |  |
|  | $\begin{aligned} & \text { Ports 1, 2, 3, 4, } \\ & 5,6,9, \mathrm{~PB}_{3}, \mathrm{~PB}_{4}, \\ & \mathrm{~PB}_{4} \text { to } \mathrm{PB}_{7} \end{aligned}$ |  | 2.0 | - | $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |  |
| Input low voltage | $\overline{\text { RES }}, \overline{\text { STBY }}$, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$, FWE | $\mathrm{V}_{\mathrm{IL}}$ | -0.3 | - | 0.5 | V |  |
|  | NMI, EXTAL, ports 1, 2, 3, 4, $5,6,7,9, \mathrm{P}_{3}$, $\mathrm{P8}_{4}, \mathrm{~PB}_{4}$ to $\mathrm{PB}_{7}$ |  | -0.3 | - | 0.8 | V |  |
| Output high voltage | All output pins | $\mathrm{V}_{\text {OH }}$ | $\mathrm{V}_{\mathrm{cc}}-0.5$ | - | - | V | $\mathrm{I}_{\text {OH }}=-200 \mu \mathrm{~A}$ |
|  |  |  | 3.5 | - | - | V | $\mathrm{IOH}_{\mathrm{O}}=-1 \mathrm{~mA}$ |
| Output low voltage | All output pins | $\mathrm{V}_{\mathrm{oL}}$ | - | - | 0.4 | V | $\mathrm{I}_{\mathrm{oL}}=1.6 \mathrm{~mA}$ |
|  | Ports 1, 2, 5, and $B$ |  | - | - | 1.0 | V | $\mathrm{I}_{\mathrm{OL}}=10 \mathrm{~mA}$ |


| Item |  | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input leakage current | STBY, NMI, RES, FWE, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ | $\|l\|_{\text {in }} \mid$ | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & V_{i n}=0.5 \text { to } \\ & V_{c c}-0.5 \mathrm{~V} \end{aligned}$ |
|  | Port 7 |  | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{in}}=0.5 \mathrm{to} \\ & \mathrm{AV}_{\mathrm{cc}}-0.5 \mathrm{~V} \end{aligned}$ |
| Three-state leakage current (off state) | Ports 1, 2, 3, $4,5,6,8$ to B | $\mid \mathrm{Itss}_{\text {sil }}$ | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{in}}=0.5 \mathrm{to} \\ & \mathrm{~V}_{\mathrm{cc}}-0.5 \mathrm{~V} \end{aligned}$ |
| Input pull-up current | Ports 2, 4, and 5 | $-l_{p}$ | 50 | - | 300 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{in}}=0 \mathrm{~V}$ |
| Input capacitance | FWE | $\mathrm{C}_{\text {in }}$ | - | - | 60 | pF | $\begin{gathered} V_{\text {IN }}=0 \mathrm{~V} \\ \mathrm{f}=1 \mathrm{MHz} \\ \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \end{gathered}$ |
|  | NMI |  | - | - | 50 | pF |  |
|  | All input pins except NMI |  | - | - | 15 | pF |  |
| Current dissipation*2 | Normal operation | $\mathrm{I}_{\mathrm{cc}}$ | - | 25 | 48 | mA | $\mathrm{f}=18 \mathrm{MHz}$ |
|  |  |  | - | 35 | 60 | mA | $f=25 \mathrm{MHz}$ |
|  | Sleep mode |  | - | 23 | 38 | mA | $\mathrm{f}=18 \mathrm{MHz}$ |
|  |  |  | - | 33 | 50 | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
|  | Module standby mode*4 |  | - | 18 | 25 | mA | $\mathrm{f}=18 \mathrm{MHz}$ |
|  |  |  | - | 25 | 40 | mA | $f=25 \mathrm{MHz}$ |
|  | Standby mode*3 |  | - | 1.0 | 10 | $\mu \mathrm{A}$ | $\mathrm{T}_{\mathrm{a}} \leq 50^{\circ} \mathrm{C}$ |
|  |  |  | - | - | 80 | $\mu \mathrm{A}$ | $50^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{a}}$ |
|  | Flash programming / erasing |  | - | 35 | 58 | mA | $f=18 \mathrm{MHz}$ |
|  |  |  | - | 45 | 70 | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
| Analog power supply current | During A/D conversion | $\mathrm{Al}_{\text {cc }}$ | - | 0.5 | 1.5 | mA |  |
|  | During A/D and D/A conversion |  | - | 0.5 | 1.5 | mA |  |
|  | Idle |  | - | 0.01 | 5.0 | $\mu \mathrm{A}$ | DASTE $=0$ |


| Item | Symbol | Min | Typ | Max | Unit | Test Conditions |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Reference <br> current | During A/D <br> conversion | $\mathrm{Al}_{\mathrm{cc}}$ | - | 0.4 | 0.8 | mA | $\mathrm{~V}_{\text {REF }}=5.0 \mathrm{~V}$ |
|  | During A/D and <br> $\mathrm{D} / \mathrm{A}$ conversion |  | - | 1.5 | 3.0 | mA |  |
|  |  |  | - | 0.01 | 5.0 | $\mu \mathrm{~A}$ | DASTE $=0$ |
| RAM standby voltage | $\mathrm{V}_{\text {RAM }}$ | 2.0 | - | - | V |  |  |

Notes: 1. If the $A / D$ and $D / A$ converters are not used, do not leave the $A V_{C C}, A V_{S S}$, and $V_{\text {REF }}$ pins open. Connect $A V_{c C}$ and $\mathrm{V}_{\text {REF }}$ to $\mathrm{V}_{\mathrm{cc}}$, and connect $\mathrm{AV}_{\text {ss }}$ to $\mathrm{V}_{\mathrm{SS}}$.
2. Current dissipation values are for $\mathrm{V}_{\text {HImin }}=\mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V}$ and $\mathrm{V}_{\text {Itmax }}=0.5 \mathrm{~V}$ with all output pins unloaded and the on-chip pull-up transistors in the off state.
$\mathrm{I}_{\mathrm{cc}}$ max. (under normal operations) $=3.0(\mathrm{~mA})+0.45(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}$
$\mathrm{I}_{\mathrm{cc}}$ max.(when using the sleeve) $=3.0(\mathrm{~mA})+0.35(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}$
$\mathrm{I}_{\mathrm{cc}}$ max. (when the sleeve + module are standing by)

$$
=3.0(\mathrm{~mA})+0.26(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}
$$

Also,the typ.values for current dissipation are reference values.
3. The values are for $\mathrm{V}_{\text {RAM }} \leq \mathrm{V}_{\mathrm{CC}}<4.5 \mathrm{~V}, \mathrm{~V}_{\text {Itmin }}=\mathrm{V}_{\mathrm{cc}} \times 0.9$, and $\mathrm{V}_{\text {ILmax }}=0.3 \mathrm{~V}$.
4. Module standby current values apply in sleep mode with all modules halted.

Table 21.2(2) DC Characteristics
Conditions: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}^{* 1}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item |  | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Schmitt trigger input voltages | Port A, $P 8{ }_{0}$ to $P 8_{2}$, $\mathrm{PB}_{0}$ to $\mathrm{PB}_{3}$ | $\mathrm{V}_{T}^{-}$ | $\mathrm{V}_{\mathrm{cc}} \times 0.2$ | - | - | V |  |
|  |  | $\mathrm{V}_{T}^{+}$ | - | - | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | V |  |
|  |  | $\mathrm{V}_{T}^{+}-\mathrm{V}_{T}{ }^{-}$ | $\mathrm{V}_{\mathrm{cc}} \times 0.5$ | - | - | V |  |
| Input high voltage | $\overline{\mathrm{RES}}, \overline{\mathrm{STBY}}$, <br> FWE, NMI, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ | $\mathrm{V}_{\text {IH }}$ | $\mathrm{V}_{\mathrm{cc}} \times 0.9$ | - | $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |  |
|  | EXTAL |  | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | - | $V_{c c}+0.3$ | V |  |
|  | Port 7 |  | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | - | $\mathrm{AV}_{\mathrm{cc}}+0.3$ | V |  |
|  | Ports 1, 2, 3, 4, 5, $6,9, \mathrm{P8}_{3}, \mathrm{P8}_{4}, \mathrm{~PB}_{4}$ to $\mathrm{PB}_{7}$ |  | $\mathrm{V}_{\mathrm{cc}} \times 0.7$ | - | $\mathrm{V}_{\mathrm{cc}}+0.3$ | V |  |
| Input low voltage | $\overline{R E S}, \overline{S T B Y}$, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$, FWE | $\mathrm{V}_{\mathrm{LL}}$ | -0.3 | - | $\mathrm{V}_{\mathrm{cc}} \times 0.1$ | V |  |
|  | NMI, EXTAL, ports 1, 2, 3, 4, 5, $6,7,9, \mathrm{P8}_{3}, \mathrm{P} 8_{4}$, $\mathrm{PB}_{4}$ to $\mathrm{PB}_{7}$ |  | -0.3 | - | $\mathrm{V}_{\mathrm{cc}} \times 0.2$ | V |  |
| Output high voltage | All output pins | $\mathrm{V}_{\text {OH }}$ | $V_{c c}-0.5$ | - | - | V | $\mathrm{I}_{\text {OH }}=-200 \mu \mathrm{~A}$ |
|  |  |  | $\mathrm{V}_{\mathrm{cc}}-1.0$ | - | - | V | $\mathrm{IOH}^{\text {O }}=-1 \mathrm{~mA}$ |
| Output low voltage | All output pins | $\mathrm{V}_{\mathrm{oL}}$ | - | - | 0.4 | V | $\mathrm{I}_{\mathrm{OL}}=1.6 \mathrm{~mA}$ |
|  | Ports 1, 2, 5, and B |  | - | - | 1.0 | V | $\mathrm{I}_{\mathrm{oL}}=5 \mathrm{~mA}$ |
| Input leakage current | $\overline{S T B Y}$, NMI, RES, FWE, $\mathrm{MD}_{2}$ to $\mathrm{MD}_{0}$ | \| ${ }_{\text {in }} \mid$ | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\text {in }}=0.5 \text { to } \\ & \mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V} \end{aligned}$ |
|  | Port 7 |  | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\text {in }}=0.5 \text { to } \\ & A V_{c \mathrm{c}}-0.5 \mathrm{~V} \end{aligned}$ |
| Three-state leakage current (off state) | Ports 1, 2, 3, 4, 5, 6, 8 to B | $\left\|\mathrm{I}_{\text {TSI }}\right\|$ | - | - | 1.0 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\text {in }}=0.5 \mathrm{to} \\ & \mathrm{~V}_{\mathrm{cc}}-0.5 \mathrm{~V} \end{aligned}$ |
| Input pullup current | Ports 2, 4, and 5 | $-l_{p}$ | 10 | - | 300 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{in}}=0 \mathrm{~V}$ |


| Item |  | Symbol |  | Typ | Max | Unit Test Conditions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input capacitance | FWE | $\mathrm{C}_{\text {in }}$ | - | - | 60 | pF | $\mathrm{V}_{\text {IN }}=0 \mathrm{~V}$ |
|  | NMI |  | - | - | 50 | pF | $\mathrm{f}=1 \mathrm{MHz}$ |
|  | All input pins except NMI |  | - | - | 15 | pF | $\mathrm{T}_{\mathrm{a}}=25^{\circ} \mathrm{C}$ |
| Current dissipation*2 | Normal operation | $I_{\text {cc }}$ | - | 33 | 45 | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
|  | Sleep mode |  | - | 31 | 35 | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
|  | Module standby mode*4 |  | - | 23 | 27 | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
|  | Standby mode*3 |  | - | 1.0 | 10 | $\mu \mathrm{A}$ | $\mathrm{T}_{\mathrm{a}} \leq 50^{\circ} \mathrm{C}$ |
|  |  |  | - | - | 80 | $\mu \mathrm{A}$ | $50^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{a}}$ |
|  | Flash programming / erasing |  | - | TBD | TBD | mA | $\mathrm{f}=25 \mathrm{MHz}$ |
| Analog power supply current | During A/D conversion | $\mathrm{Al}_{\mathrm{cc}}$ | - | 0.6 | 1.5 | mA |  |
|  | During $A / D$ and $D / A$ conversion |  | - | 0.6 | 1.5 | mA |  |
|  | Idle |  | - | 0.01 | 5.0 | $\mu \mathrm{A}$ | DASTE $=0$ |
| Reference current | During A/D conversion | $\mathrm{Al}_{\text {cc }}$ | - | 0.45 | 0.8 | mA | $\mathrm{V}_{\text {REF }}=5.0 \mathrm{~V}$ |
|  | During A/D and D/A conversion |  | - | 2.0 | 3.0 | mA |  |
|  | Idle |  | - | 0.01 | 5.0 | $\mu \mathrm{A}$ | DASTE $=0$ |
| RAM standby voltage |  | $V_{\text {RAM }}$ | 2.0 | - | - | V |  |

Notes: 1. If the $A / D$ and $D / A$ converters are not used, do not leave the $A V_{C C}, A V_{\text {Ss }}$, and $V_{\text {REF }}$ pins open. Connect $\mathrm{AV}_{\mathrm{CC}}$ and $\mathrm{V}_{\text {REF }}$ to $\mathrm{V}_{\mathrm{cc}}$, and connect $\mathrm{AV}_{\mathrm{SS}}$ to $\mathrm{V}_{\mathrm{SS}}$.
2. Current dissipation values are for $\mathrm{V}_{\text {HImin }}=\mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V}$ and $\mathrm{V}_{\text {Itmax }}=0.5 \mathrm{~V}$ with all output pins unloaded and the on-chip pull-up transistors in the off state.
$\mathrm{I}_{\mathrm{cc}}$ max. (under normal operations) $=3.0(\mathrm{~mA})+0.45(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}$
$\mathrm{I}_{\mathrm{cc}}$ max. (when using the sleeve) $=3.0(\mathrm{~mA})+0.35(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}$
$\mathrm{I}_{\mathrm{cc}}$ max. (when the sleeve + module are standing by)

$$
=3.0(\mathrm{~mA})+0.26(\mathrm{~mA} /(\mathrm{MHz} \times \mathrm{V})) \times \mathrm{V}_{\mathrm{cc}} \times \mathrm{f}
$$

Also,the typ.values for current dissipation are reference values.
3. The values are for $\mathrm{V}_{\text {RaM }} \leq \mathrm{V}_{\mathrm{cc}}<4.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{H} \text { min }}=\mathrm{V}_{\mathrm{cc}} \times 0.9$, and $\mathrm{V}_{\mathrm{IL} \max }=0.3 \mathrm{~V}$.
4. Module standby current values apply in sleep mode with all modules halted.

Table 21.3 Permissible Output Currents
Conditions: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,
$\mathrm{V}_{\text {ss }}=\mathrm{AV}_{\text {ss }}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item |  | Symbol | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Permissible output low current (per pin) | Ports 1, 2, 5, and B | $\mathrm{I}_{\mathrm{ol}}$ | - | - | 10 | mA |
|  | Other output pins |  | - | - | 2.0 | mA |
| Permissible output low current (total) | Total of 28 pins in ports 1, 2, 5, and B | $\Sigma_{\text {o }}$ | - | - | 80 | mA |
|  | Total of all output pins, including the above |  | - | - | 120 | mA |
| Permissible output high current (per pin) | All output pins | $\mathrm{I}_{\mathrm{OH}}$ | - | - | 2.0 | mA |
| Permissible output high current (total) | Total of all output pins | $\Sigma \mathrm{I}_{\text {OH }}$ | - | - | 40 | mA |

Notes: 1. To protect chip reliability, do not exceed the output current values in table 21.3.
2. When driving a darlington pair or LED, always insert a current-limiting resistor in the output line, as shown in figures 21.1 and 21.2.


Figure 21.1 Darlington Pair Drive Circuit (Example)


Figure 21.2 LED Drive Circuit (Example)

### 21.2.2 AC Characteristics

Bus timing parameters are listed in table 21.4. Refresh controller bus timing parameters are listed in table 21.5. Control signal timing parameters are listed in table 21.6. Timing parameters of the on-chip supporting modules are listed in table 21.7.

## Table 21.4 Bus Timing

Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$ - Preliminary -

Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $18 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

Condition C: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$



Note: * At 18 MHz , the times below depend as indicated on the clock cycle time.

$$
\begin{aligned}
\mathrm{t}_{\mathrm{ACC} 1}=1.5 \times \mathrm{t}_{\mathrm{cyc}}-34(\mathrm{~ns}) & \mathrm{t}_{\mathrm{wsw} 1}=1.0 \times \mathrm{t}_{\mathrm{cyc}}-24(\mathrm{~ns}) \\
\mathrm{t}_{\mathrm{ACC} 2}=2.5 \times \mathrm{t}_{\mathrm{cyc}}-34(\mathrm{~ns}) & \mathrm{t}_{\mathrm{wsw} 2}=1.5 \times \mathrm{t}_{\mathrm{cyc}}-22(\mathrm{~ns}) \\
\mathrm{t}_{\mathrm{ACC} 3}=1.0 \times \mathrm{t}_{\mathrm{cyc}}-36(\mathrm{~ns}) & \mathrm{t}_{\mathrm{PCH}}=1.0 \times \mathrm{t}_{\mathrm{cyc}}-21(\mathrm{~ns}) \\
\mathrm{t}_{\mathrm{ACC} 4}=2.0 \times \mathrm{t}_{\mathrm{cyc}}-31(\mathrm{~ns}) &
\end{aligned}
$$

Table 21.5 Refresh Controller Bus Timing
Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\text {REF }}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,
$\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\text {SS }}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

- Preliminary -

Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $18 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

Condition C: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\text {ss }}=\mathrm{AV}_{\text {ss }}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item | Symbol | Condition A |  | Condition B |  | Condition C |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |  |
| $\overline{\text { RAS }}$ delay time 1 | $\mathrm{t}_{\text {RAD } 1}$ | - | 25 | - | 30 | - | 18 | ns | Figure 21.7 |
| $\overline{\text { RAS }}$ delay time 2 | $\mathrm{t}_{\text {RAD2 }}$ | - | 25 | - | 30 | - | 18 |  |  |


| $\overline{\text { RAS delay time } 3}$ | $\mathrm{t}_{\text {RAO3 }}$ | - | 25 | - | - | - | 18 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Row address hold <br> time* | $\mathrm{t}_{\text {RAH }}$ | $0.5 \mathrm{t}_{\mathrm{cyc}}-5$ | - | 15 | - | $0.5 \mathrm{t}_{\mathrm{cyc}}-5$ | - |
| $\overline{\text { RAS precharge time* }}$ | $\mathrm{t}_{\mathrm{RP}}$ | $1.0 \mathrm{t}_{\mathrm{cyc}}-15$ | - | 45 | - | $1.0 \mathrm{t}_{\mathrm{cyc}}-15$ | - |
| $\overline{\mathrm{CAS}}$ to RAS precharge | $\mathrm{t}_{\mathrm{CRP}}$ | $1.0 \mathrm{t}_{\mathrm{cyc}}-15$ | - | 45 | - | $1.0 \mathrm{t}_{\mathrm{cyc}}-15$ | - |
| time* |  |  |  |  |  |  |  |

Note: * At 18 MHz , the times below depend as indicated on the clock cycle time.

$$
\begin{aligned}
& \mathrm{t}_{\mathrm{RAH}}=0.5 \times \mathrm{t}_{\mathrm{ccc}}-17(\mathrm{~ns}) \\
& \mathrm{t}_{\mathrm{fAC}}=2.0 \times \mathrm{t}_{\mathrm{ccc}}-40(\mathrm{~ns}) \\
& \mathrm{t}_{\mathrm{RP}}=\mathrm{t}_{\mathrm{cfP}}=1.0 \times \mathrm{t}_{\mathrm{ccc}}-18(\mathrm{~ns})
\end{aligned}
$$

$$
\mathrm{t}_{\mathrm{cAC}}=1.0 \times \mathrm{t}_{\mathrm{cyc}}-33(\mathrm{~ns})
$$

$$
\mathrm{t}_{\mathrm{CSR}}=0.5 \times \mathrm{t}_{\mathrm{cyc}}-17(\mathrm{~ns})
$$

Table 21.6 Control Signal Timing
Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\text {ss }}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$ - Preliminary -

Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $18 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

Condition C: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{sS}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item | Symbol | $\underset{A}{\text { Condition }}$ |  | $\underset{B}{\text { Condition }}$ |  | $\underset{C}{\text { Condition }}$ |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max | Min | Max |  |  |
| $\overline{\mathrm{RES}}$ setup time | $\mathrm{t}_{\text {RESS }}$ | 200 | - | 200 | - | 200 | - | ns | Figure 21.15 |
| $\overline{\mathrm{RES}}$ pulse width | $\mathrm{t}_{\text {RESW }}$ | 20 | - | 20 | - | 20 | - | $\mathrm{t}_{\mathrm{cyc}}$ |  |
| Mode programming setup time | $\mathrm{t}_{\text {MOS }}$ | 200 | - | 200 | - | 200 | - | ns |  |
| NMI setup time ( $\mathrm{NMI}, \overline{\mathrm{RQ}}_{5}$ to $\overline{\mathrm{RQ}}_{0}$ ) | $\mathrm{t}_{\text {MMI }}$ | 150 | - | 150 | - | 150 | - | ns | Figure 21.16 |
| NMI hold time ( $\mathrm{NMI}, \overline{\mathrm{IRQ}}_{5}$ to $\overline{\mathrm{RQ}}_{0}$ ) | $\mathrm{t}_{\text {мМН }}$ | 10 | - | 10 | - | 10 | - |  |  |
| Interrupt pulse width ( $\mathrm{NMI}, \mathrm{IRQ}_{5}$ to $\overline{\mathrm{RQ}}_{2}$ when exiting software standby mode) | $t_{\text {mum }}$ | 200 | - | 200 | - | 200 | - |  |  |
| Clock oscillator settling time at reset (crystal) | $\mathrm{t}_{\text {osc } 1}$ | 20 | - | 20 | - | 20 | - | ms | Figure 21.18 |
| Clock oscillator settling time in software standby (crystal) | $\mathrm{t}_{\text {osc2 }}$ | 7 | - | 7 | - | 7 | - | ns | Figure 20.1 |

## Table 21.7 Timing of On-Chip Supporting Modules

Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,
$\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

- Preliminary -

Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $18 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

Condition C: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\text {ss }}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item |  | Symbol | Condition A |  | Condition B |  | $\begin{gathered} \text { Condition } \end{gathered}$ |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Max | Min | Max | Min | Max |  |  |
| DMAC | $\overline{\text { DREQ }}$ setup time | $\mathrm{t}_{\text {DRas }}$ | 20 | - | 30 | - | 20 | - | ns | Figure 21.16 |
|  | $\overline{\text { DREQ }}$ hold time | $\mathrm{t}_{\text {DROH }}$ | 10 | - | 10 | - | 10 | - |  |  |
|  | TEND delay time 1 | $\mathrm{t}_{\text {TED } 1}$ | - | 50 | - | 50 | - | 50 |  | Figure 21.24, Figure 21.25 |
|  | $\overline{\text { TEND }}$ delay time 2 | $\mathrm{t}_{\text {TED2 }}$ | - | 50 | - | 50 | - | 50 |  |  |
| ITU | Timer output delay time | $\mathrm{t}_{\text {TocD }}$ | - | 100 | - | 100 | - | 50 | ns | Figure 21.20 |
|  | Timer input setup time | $\mathrm{t}_{\text {TICS }}$ | 40 | - | 50 | - | 40 | - |  |  |
|  | Timer clock input setup time | $\mathrm{t}_{\text {TCKS }}$ | 40 | - | 50 | - | 40 | - |  | Figure 21.21 |
|  | Timer clockpulse width | $\mathrm{t}_{\text {TCKWH }}$ | 1.5 | - | 1.5 | - | 1.5 | - | $\mathrm{t}_{\mathrm{cyc}}$ |  |
|  |  | $\mathrm{t}_{\text {TCKWL }}$ | 2.5 | - | 2.5 | - | 2.5 | - |  |  |
| SCl | Input clock Asynchronous | $\mathrm{t}_{\text {scrc }}$ | 4 | - | 4 | - | 4 | - | $\mathrm{t}_{\mathrm{cyc}}$ | Figure 21.22 |
|  | cycle Synchronous | $\mathrm{t}_{\text {scrc }}$ | 6 | - | 6 | - | 6 | - |  |  |
|  | Input clock rise time | $\mathrm{t}_{\text {sckr }}$ | - | 1.5 | - | 1.5 | - | 1.5 |  |  |
|  | Input clock fall time | $\mathrm{t}_{\text {sckf }}$ | - | 1.5 | - | 1.5 | - | 1.5 |  |  |
|  | Input clock pulse width | $\mathrm{t}_{\text {sckw }}$ | 0.4 | 0.6 | 0.4 | 0.6 | 0.4 | 0.6 | $\mathrm{t}_{\text {scy }}$ |  |
|  | Transmit data delay time | $\mathrm{t}_{\text {TX0 }}$ | - | 100 | - | 100 | - | 100 | ns | Figure 21.23 |
|  | Receive data setup time (synchronous) | $\mathrm{t}_{\text {Rxs }}$ | 100 | - | 100 | - | 100 | - |  |  |
|  | Receive data hold time (synchronous) | $\mathrm{t}_{\text {RXH }}$ | 100 | - | 100 | - | 100 | - |  |  |
|  |  | $\mathrm{t}_{\text {RXH }}$ | 0 | - | 0 | - | 0 | - |  |  |
| Ports and TPC | Output data delay time | $t_{\text {pwo }}$ | - | 100 | - | 100 | - | 50 | ns | Figure 21.19 |
|  | Input data setup time | $\mathrm{t}_{\text {PRS }}$ | 50 | - | 50 | - | 50 | - |  |  |
|  | Input data hold time | $\mathrm{t}_{\text {PRH }}$ | 50 | - | 50 | - | 50 | - |  |  |



Figure 21.3 Output Load Circuit

### 21.2.3 A/D Conversion Characteristics

Table 21.8 lists the $\mathrm{A} / \mathrm{D}$ conversion characteristics.
Table 21.8 A/D Converter Characteristics
Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,
$\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$
Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $18 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

Condition $\mathrm{C}: \mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{Ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz}$ to $25 \mathrm{MHz}, \mathrm{T}_{\mathrm{a}}=-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$

| Item | Condition A |  |  | Condition B |  |  | Condition C |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Typ | Max | Min | Typ | Max | Min | Typ | Max |  |
| Resolution | 10 | 10 | 10 | 10 | 10 | 10 | 10 | 10 | 10 | bits |
| Conversion time | - | - | $134 t_{\text {cyc }}$ | - | - | 7.5 | - | - | $134 t_{\text {cyc }}$ | $\mu \mathrm{s}$ |
| Analog input capacitance | - | - | 20 | - | - | 20 | - | - | 20 | pF |
| Permissible signal-source impedance | - | - | $10^{* 3}$ | - | - | $10^{* 1}$ | - | - | $10^{* 1}$ | $\mathrm{k} \Omega$ |
|  | - | - | $5^{* 4}$ | - | - | $5^{* 2}$ | - | - | $5^{* 2}$ |  |
| Nonlinearity error | - | - | $\pm 7.5$ | - | - | $\pm 3.5$ | - | - | $\pm 3.5$ | LSB |
| Offset error | - | - | $\pm 7.5$ | - | - | $\pm 3.5$ | - | - | $\pm 3.5$ | LSB |
| Full-scale error | - | - | $\pm 7.5$ | - | - | $\pm 3.5$ | - | - | $\pm 3.5$ | LSB |
| Quantization error | - | - | $\pm 0.5$ | - | - | $\pm 0.5$ | - | - | $\pm 0.5$ | LSB |
| Absolute accuracy | - | - | $\pm 8.0$ | - | - | $\pm 4.0$ | - | - | $\pm 4.0$ | LSB |

Notes: 1 The value is for $\phi \leq 12 \mathrm{MHz}$.
2 The value is for $\phi>12 \mathrm{MHz}$.
3 The value is for $\phi \leq 12 \mathrm{MHz}$ and $\mathrm{AV}_{c \mathrm{c}}=4.0 \mathrm{~V}$ to 5.5 V
4 The value is for $\phi>12 \mathrm{MHz}$ and $A V_{c c}=3.3 \mathrm{~V}$ to 4.0 V

### 21.2.4 D/A Conversion Characteristics

Table 21.9 lists the D/A conversion characteristics.

## Table 21.9 D/A Converter Characteristics

Condition A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{REF}}=3.0 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,

$$
\mathrm{V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz} \text { to } 25 \mathrm{MHz}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C} \text { to }+75^{\circ} \mathrm{C}
$$

Condition B: $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,

$$
\mathrm{V}_{\mathrm{sS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \phi=1 \mathrm{MHz} \text { to } 18 \mathrm{MHz}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C} \text { to }+75^{\circ} \mathrm{C}
$$

Condition $\mathrm{C}: \mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{REF}}=4.5 \mathrm{~V}$ to $\mathrm{AV}_{\mathrm{CC}}$,

$$
\mathrm{V}_{\mathrm{ss}}=\mathrm{AV}_{\mathrm{ss}}=0 \mathrm{~V}, \phi=2 \mathrm{MHz} \text { to } 25 \mathrm{MHz}, \mathrm{~T}_{\mathrm{a}}=-20^{\circ} \mathrm{C} \text { to }+75^{\circ} \mathrm{C}
$$

|  | Condition A |  |  | Condition B |  |  | Condition C |  |  | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Item | Min | Typ | Max | Min | Typ | Max | Min | Typ | Max |  |  |
| Resolution | 8 | 8 | 8 | 8 | 8 | 8 | 8 | 8 | 8 | Bits |  |
| Conversion time | - | - | 10 | - | - | 10 | - | - | 10 | $\mu \mathrm{s}$ | 20-pF capacitive load |
| Absolute accuracy | - | $\pm 2.0$ | $\pm 3.0$ | - | $\pm 1.0$ | $\pm 1.5$ | - | $\pm 1.5$ | $\pm 2.0$ | LSB | 2-M M resistive load |
|  | - | - | $\pm 2.0$ | - | - | $\pm 1.0$ | - | - | $\pm 1.5$ | LSB | 4-M $\Omega$ resistive load |

### 21.2.5 Flash Memory Characteristics

## Table 21.10 Flash Memory Characteristics

Conditions A: $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=0^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$ (program/erase operating temperature range)

Conditions B: $\mathrm{V}_{\mathrm{CC}}=4.5 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=4.5 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathrm{AV}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=0^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$ (program/erase operating temperature range)

| Item |  | Symbol | Condition A,B |  |  | Unit | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max |  |  |
| Programming time ${ }^{* 1, *^{2}, *^{4}}$ |  |  | $\mathrm{t}_{\mathrm{p}}$ | - | 10 | 200 | $\mathrm{ms} / 128$ bytes |  |
| Erase time ${ }^{* 1 *^{1} *^{3} *^{5}}$ |  | $\mathrm{t}_{\mathrm{E}}$ | - | 100 | 1200 | $\mathrm{ms} / \mathrm{block}$ |  |
| Reprogramming count |  | $\mathrm{N}_{\text {wEC }}$ | - | - | 100 | Times |  |
| Programming | Wait time after SWE bit setting*1 | tsswe | 1 | 1 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after PSU bit setting*1 | tspsu | 50 | 50 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after P bit setting* ${ }^{\text {+1, }{ }^{4}}$ | tsp30 | 28 | 30 | 32 | $\mu \mathrm{s}$ | Programming time wait |
|  |  | tsp200 | 198 | 200 | 202 | $\mu \mathrm{s}$ | Programming time wait |
|  |  | tsp10 | 8 | 10 | 12 | $\mu \mathrm{s}$ | Additional programming time wait |
|  | Wait time after P bit clear*1 | tcp | 5 | 5 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after PSU bit clear*1 | tcpsu | 5 | 5 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after PV bit setting*1 | tspv | 4 | 4 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after H'FF dummy write*1 | tspur | 2 | 2 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after PV bit clear* ${ }^{1}$ | tcpv | 2 | 2 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after SWE bit clear*1 | tcswe | 100 | 100 | - | $\mu \mathrm{s}$ |  |
|  | Maximum programming count***** | N | - | - | 1000 | Times |  |
| Erase | Wait time after SWE bit setting*1 | tsswe | 1 | 1 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after ESU bit setting*1 | tsesu | 100 | 100 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after E bit setting* ${ }^{* 1, *^{5}}$ | tse | 10 | 10 | 100 | ms | Erase time wait |
|  | Wait time after E bit clear** | tce | 10 | 10 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after ESU bit clear*1 | tcesu | 10 | 10 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after EV bit setting*1 | tsev | 20 | 20 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after H'FF dummy write*1 | tsevr | 2 | 2 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after EV bit clear*1 | tcev | 4 | 4 | - | $\mu \mathrm{s}$ |  |
|  | Wait time after SWE bit clear*1 | tcswe | 100 | 100 | - | $\mu \mathrm{s}$ |  |
|  | Maximum erase count**** ${ }^{\text {/ }}$ | N | 12 | - | 120 | Times |  |

Notes: 1. Set the times according to the program/erase algorithms.
2. Programming time per 128 bytes. (Shows the total time the P 1 bit or P 2 bit in the flash memory control register (FLMCR1 or FLMCR2) is set. It does not include the programming verification time.)
3. Block erase time. (Shows the total time the E1 bit in FLMCR1 or E2 bit in FLMCR2 is set. It does not include the erase verification time.)
4. To specify the maximum programming time value $\left(t_{p}(\max )\right)$ in the 128-byte programming algorithm, set the max. value (1000) for the maximum programming count (N).

The wait time after $P$ bit setting should be changed as follows according to the value of the programming counter (n).
Programming counter $(n)=1$ to 6 :
Programming counter (n) $=7$ to 1000:
tsp30 $=30 \mu \mathrm{~s}$

Programming counter ( n ) [in additional programming] = 1 to 6:
tsp200 $=200 \mu \mathrm{~s}$
tsp10 $=10 \mu \mathrm{~s}$
5. For the maximum erase time ( $\mathrm{t}_{\mathrm{E}}(\max )$ ), the following relationship applies between the wait time after $E$ bit setting (tse) and the maximum erase count ( N ): $t_{E}(\max )=$ Wait time after $E$ bit setting (tse) $x$ maximum erase count ( $N$ )
To set the maximum erase time, the values of tse and $N$ should be set so as to satisfy the above formula.
Examples: When tse $=100$ [ms], $\mathrm{N}=12$
When tse $=10$ [ms], $N=120$

### 21.3 Operational Timing

This section shows timing diagrams.

### 21.3.1 Bus Timing

Bus timing is shown as follows:

- Basic bus cycle: two-state access

Figure 21.4 shows the timing of the external two-state access cycle.

- Basic bus cycle: three-state access

Figure 21.5 shows the timing of the external three-state access cycle.

- Basic bus cycle: three-state access with one wait state

Figure 21.6 shows the timing of the external three-state access cycle with one wait state inserted.


Figure 21.4 Basic Bus Cycle: Two-State Access


Figure 21.5 Basic Bus Cycle: Three-State Access


Figure 21.6 Basic Bus Cycle: Three-State Access with One Wait State

### 21.3.2 Refresh Controller Bus Timing

Refresh controller bus timing is shown as follows:

- DRAM bus timing

Figures 21.7 to 21.12 show the DRAM bus timing in each operating mode.

- PSRAM bus timing

Figures 21.13 and 21.14 show the pseudo-static RAM bus timing in each operating mode.


Figure 21.7 DRAM Bus Timing (Read/Write): Three-State Access

- $2 \overline{\mathrm{WE}}$ Mode -


Figure 21.8 DRAM Bus Timing (Refresh Cycle): Three-State Access - 2 $\overline{\mathrm{WE}}$ Mode -


Figure 21.9 DRAM Bus Timing (Self-Refresh Mode) - $2 \overline{\mathrm{WE}}$ Mode -


Figure 21.10 DRAM Bus Timing (Read/Write): Three-State Access - 2 $\overline{\mathrm{CAS}}$ Mode -


Figure 21.11 DRAM Bus Timing (Refresh Cycle): Three-State Access - $2 \overline{\text { CAS }}$ Mode -


Figure 21.12 DRAM Bus Timing (Self-Refresh Mode)

$$
-2 \overline{\mathrm{CAS}} \text { Mode — }
$$



Figure 21.13 PSRAM Bus Timing (Read/Write): Three-State Access


Figure 21.14 PSRAM Bus Timing (Refresh Cycle): Three-State Access

### 21.3.3 Control Signal Timing

Control signal timing is shown as follows:

- Reset input timing

Figure 21.15 shows the reset input timing.

- Interrupt input timing

Figure 21.16 shows the input timing for NMI and $\overline{\mathrm{IRQ}}_{5}$ to $\overline{\mathrm{IRQ}}_{0}$.

- Bus-release mode timing

Figure 21.17 shows the bus-release mode timing.


Figure 21.15 Reset Input Timing

$\overline{\operatorname{IRQ}}_{\mathrm{E}}$ : Edge-sensitive $\overline{\mathrm{RQ}}_{\mathrm{i}}$
$\overline{\mathrm{IRQ}}_{\mathrm{L}}$ : Level-sensitive $\overline{\mathrm{IRQ}}_{\mathrm{i}}(\mathrm{i}=0$ to 5$)$


$$
(\mathrm{j}=0 \text { to } 2)
$$

Figure 21.16 Interrupt Input Timing


Figure 21.17 Bus-Release Mode Timing

### 21.3.4 Clock Timing

Clock timing is shown as follows:

- Oscillator settling timing

Figure 21.18 shows the oscillator settling timing.


Figure 21.18 Oscillator Settling Timing

### 21.3.5 TPC and I/O Port Timing

Figure 21.19 shows the TPC and I/O port timing.


Figure 21.19 TPC and I/O Port Input/Output Timing

### 21.3.6 ITU Timing

ITU timing is shown as follows:

- ITU input/output timing

Figure 21.20 shows the ITU input/output timing.

- ITU external clock input timing

Figure 21.21 shows the ITU external clock input timing.


Figure 21.20 ITU Input/Output Timing


Figure 21.21 ITU External Clock Input Timing

### 21.3.7 SCI Input/Output Timing

SCI timing is shown as follows:

- SCI input clock timing

Figure 21.22 shows the SCK input clock timing.

- SCI input/output timing (synchronous mode)

Figure 21.23 shows the SCI input/output timing in synchronous mode.


Figure 21.22 SCK Input Clock Timing


Figure 21.23 SCI Input/Output Timing in Synchronous Mode

### 21.3.8 DMAC Timing

DMAC timing is shown as follows.

- DMAC $\overline{\text { TEND }}$ output timing for 2 state access

Figure 21.24 shows the DMAC $\overline{\text { TEND }}$ output timing for 2 state access.

- DMAC $\overline{\text { TEND }}$ output timing for 3 state access

Figure 21.25 shows the DMAC $\overline{\text { TEND }}$ output timing for 3 state access.

- DMAC $\overline{\mathrm{DREQ}}$ input timing

Figure 21.26 shows DMAC $\overline{\mathrm{DREQ}}$ input timing.


Figure 21.24 DMAC $\overline{\text { TEND }}$ Output Timing for 2 State Access


Figure 21.25 DMAC $\overline{\text { TEND }}$ Output Timing for 3 State Access


Figure 21.26 DMAC $\overline{\text { DREQ }}$ Input Timing

## Appendix A Instruction Set

## A. 1 Instruction List

Operand Notation

| Symbol | Description |
| :---: | :---: |
| Rd | General destination register |
| Rs | General source register |
| Rn | General register |
| ERd | General destination register (address register or 32-bit register) |
| ERs | General source register (address register or 32-bit register) |
| ERn | General register (32-bit register) |
| (EAd) | Destination operand |
| (EAs) | Source operand |
| PC | Program counter |
| SP | Stack pointer |
| CCR | Condition code register |
| N | N (negative) flag in CCR |
| Z | Z (zero) flag in CCR |
| V | V (overflow) flag in CCR |
| C | C (carry) flag in CCR |
| disp | Displacement |
| $\rightarrow$ | Transfer from the operand on the left to the operand on the right, or transition from the state on the left to the state on the right |
| + | Addition of the operands on both sides |
| - | Subtraction of the operand on the right from the operand on the left |
| $\times$ | Multiplication of the operands on both sides |
| $\div$ | Division of the operand on the left by the operand on the right |
| $\wedge$ | Logical AND of the operands on both sides |
| $\checkmark$ | Logical OR of the operands on both sides |
| $\oplus$ | Exclusive logical OR of the operands on both sides |
| $\neg$ | NOT (logical complement) |
| ( ), < > | Contents of operand |

Note: General registers include 8-bit registers (ROH to R7H and ROL to R7L) and 16-bit registers (R0 to R7 and E0 to E7).

Condition Code Notation

| Symbol | Description |
| :--- | :--- |
| $\downarrow$ | Changed according to execution result |
| $*$ | Undetermined (no guaranteed value) |
| 0 | Cleared to 0 |
| 1 | Set to 1 |
| - | Not affected by execution of the instruction |
| $\Delta$ | Varies depending on conditions, described in notes |

Table A. 1 Instruction Set

## 1. Data transfer instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  | $\underset{\sim}{〔}$ | (8) | (B) | (8) | (8) | (8) | (8) | 1 | 1 | H | N | Z | V | C | 2 | ¢ |
| MOV.B \#xx:8, Rd | B | \#xx:8 $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| MOV.B Rs, Rd | B | Rs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| MOV.B @ERs, Rd | B | @ERs $\rightarrow$ Rd8 |  |  | 2 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.B @(d:16, ERs), Rd | B | @(d:16, ERs) $\rightarrow$ Rd8 |  |  |  | 4 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B @ (d:24, ERs), Rd | B | @(d:24, ERs) $\rightarrow$ Rd8 |  |  |  | 8 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.B @ERs+, Rd | B | $\begin{aligned} & @ E R s \rightarrow \text { Rd8 } \\ & \text { ERs32+1 } \rightarrow \text { ERs32 } \end{aligned}$ |  |  |  |  | 2 |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B @aa:8, Rd | B | @aa:8 $\rightarrow$ Rd8 |  |  |  |  |  | 2 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.B @aa:16, Rd | B | @aa:16 $\rightarrow$ Rd8 |  |  |  |  |  | 4 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B @aa:24, Rd | B | @aa:24 $\rightarrow$ Rd8 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.B Rs, @ERd | B | Rs8 $\rightarrow$ @ERd |  |  | 2 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.B Rs, @(d:16, ERd) | B | Rs8 $\rightarrow$ @(d:16, ERd) |  |  |  | 4 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B Rs, @(d:24, ERd) | B | Rs8 $\rightarrow$ @(d:24, ERd) |  |  |  | 8 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.B Rs, @-ERd | B | $\begin{aligned} & \text { ERd32-1 } \rightarrow \text { ERd32 } \\ & \text { Rs8 } \rightarrow \text { @ERd } \end{aligned}$ |  |  |  |  | 2 |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B Rs, @aa:8 | B | Rs8 $\rightarrow$ @aa:8 |  |  |  |  |  | 2 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.B Rs, @aa:16 | B | Rs8 $\rightarrow$ @aa:16 |  |  |  |  |  | 4 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.B Rs, @aa:24 | B | Rs8 $\rightarrow$ @aa:24 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.W \#xx:16, Rd | W | \#xx:16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.W Rs, Rd | W | Rs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| MOV.W @ERs, Rd | W | @ERs $\rightarrow$ Rd16 |  |  | 2 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.W @(d:16, ERs), Rd | W | @(d:16, ERs) $\rightarrow$ Rd16 |  |  |  | 4 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W @(d:24, ERs), Rd | W | @(d:24, ERs) $\rightarrow$ Rd16 |  |  |  | 8 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.W@ERs+, Rd | W | @ERs $\rightarrow$ Rd16 ERs32+2 $\rightarrow$ @ERd32 |  |  |  |  | 2 |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W @aa:16, Rd | W | @aa:16 $\rightarrow$ Rd16 |  |  |  |  |  | 4 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W @aa:24, Rd | W | @aa:24 $\rightarrow$ Rd16 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.W Rs, @ERd | W | Rs16 $\rightarrow$ @ERd |  |  | 2 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.W Rs, @(d:16, ERd) | W | Rs16 $\rightarrow$ @(d:16, ERd) |  |  |  | 4 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W Rs, @(d:24, ERd) | W | Rs16 $\rightarrow$ @(d:24, ERd) |  |  |  | 8 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |


| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | * | ¢ |  |  |  | © |  | $\begin{gathered} \text { ๙0 } \\ \text { ©区 } \\ \text { (8) } \end{gathered}$ | 1 |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  | 1 | H | N | Z | v | c | 2 | 8 |
| MOV.W Rs, @-ERd | W | $\begin{aligned} & \text { ERd32-2 } \rightarrow \text { ERd32 } \\ & \text { Rs16 } \rightarrow \text { @ERd } \end{aligned}$ |  |  |  |  | 2 |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W Rs, @aa:16 | W | Rs16 $\rightarrow$ @aa:16 |  |  |  |  |  | 4 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.W Rs, @aa:24 | W | Rs16 $\rightarrow$ @aa:24 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.L \#xx:32, Rd | L | \#xx:32 $\rightarrow$ Rd32 | 6 |  |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 6 |
| MOV.L ERs, ERd | L | ERs32 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| MOV.L @ERs, ERd | L | @ERs $\rightarrow$ ERd32 |  |  | 4 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.L @(d:16, ERs), ERd | L | @(d:16, ERs) $\rightarrow$ ERd32 |  |  |  | 6 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L @(d:24, ERs), ERd | L | @(d:24, ERs) $\rightarrow$ ERd32 |  |  |  | 10 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.L @ERs+, ERd | L | $\begin{aligned} & @ E R s \rightarrow \text { ERd32 } \\ & \text { ERs32+4 } \rightarrow \text { ERs32 } \end{aligned}$ |  |  |  |  | 4 |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L @aa:16, ERd | L | @aa:16 $\rightarrow$ ERd32 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L @aa:24, ERd | L | @aa:24 $\rightarrow$ ERd32 |  |  |  |  |  | 8 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| MOV.L ERs, @ERd | L | ERs32 $\rightarrow$ @ERd |  |  | 4 |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 8 |
| MOV.L ERs, @(d:16, ERd) | L | ERs32 $\rightarrow$ @(d:16, ERd) |  |  |  | 6 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L ERs, @(d:24, ERd) | L | ERs32 $\rightarrow$ @(d:24, ERd) |  |  |  | 10 |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 4 |
| MOV.L ERs, @-ERd | L | $\begin{aligned} & \text { ERd32-4 } \rightarrow \text { ERd32 } \\ & \text { ERs32 } \rightarrow \text { @ERd } \end{aligned}$ |  |  |  |  | 4 |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L ERs, @aa:16 | L | ERs32 $\rightarrow$ @aa:16 |  |  |  |  |  | 6 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOV.L ERs, @aa:24 | L | ERs32 $\rightarrow$ @aa:24 |  |  |  |  |  | 8 |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| POP.W Rn | W | $\begin{aligned} & @ S P \rightarrow \text { Rn16 } \\ & \mathrm{SP}+2 \rightarrow \mathrm{SP} \end{aligned}$ |  |  |  |  |  |  |  |  | 2 | - | - | $\uparrow$ | $\downarrow$ | 0 | - |  | 6 |
| POP.L ERn | L | $\begin{aligned} & @ \mathrm{SP} \rightarrow \mathrm{ERn} 32 \\ & \mathrm{SP}+4 \rightarrow \mathrm{SP} \end{aligned}$ |  |  |  |  |  |  |  |  | 4 | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| PUSH.W Rn | W | $\begin{aligned} & \mathrm{SP}-2 \rightarrow \mathrm{SP} \\ & \mathrm{Rn} 16 \rightarrow \text { @SP } \end{aligned}$ |  |  |  |  |  |  |  |  | 2 | - | - | $\uparrow$ | $\downarrow$ | 0 | - |  | 6 |
| PUSH.L ERn | L | $\begin{aligned} & \mathrm{SP}-4 \rightarrow \mathrm{SP} \\ & \mathrm{ERn} 32 \rightarrow @ \text { SP } \end{aligned}$ |  |  |  |  |  |  |  |  | 4 | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 0 |
| MOVFPE @aa:16, Rd | B | Cannot be used in the $\mathrm{H} 8 / 3052 \mathrm{~F}$ |  |  |  |  |  | 4 |  |  |  |  | $\begin{aligned} & \text { annot } \\ & \text { e H8/ } \end{aligned}$ |  |  |  |  |  |  |
| MOVTPE Rs, @aa:16 | B | Cannot be used in the $\mathrm{H} 8 / 3052 \mathrm{~F}$ |  |  |  |  |  | 4 |  |  |  |  | $\begin{aligned} & \text { annot } \\ & \text { e } \mathrm{H} 8 / 4 \end{aligned}$ | /305 |  |  |  |  |  |

## 2. Arithmetic instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{aligned} & \times \\ & \times \\ & \hline \end{aligned}$ | $\underset{\sim}{\boldsymbol{\sim}}$ |  |  | +uyヨ(®)/uyヨ-® | $\begin{aligned} & \mathbb{\pi} \\ & \text { ®in } \end{aligned}$ |  |  | 1 |  |  |  |  |  |  | $\begin{aligned} & \bar{\sigma} \\ & \text { E0 } \\ & \text { Z } \end{aligned}$ | O <br> d <br> C <br> N <br> Z <br> 1 |
|  |  |  |  |  |  |  |  |  |  |  |  | 1 | H | N | Z | V | C |  |  |
| ADD.B \#xx:8, Rd | B | Rd8+\#xx:8 $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| ADD.B Rs, Rd | B | Rd8+Rs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| ADD.W \#xx:16, Rd | W | Rd16+\#xx:16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | - (1) | (1) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 4 |
| ADD.W Rs, Rd | W | Rd16+Rs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | (1) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| ADD.L \#xx:32, ERd | L | $\begin{aligned} & \text { ERd32+\#xx:32 } \rightarrow \\ & \text { ERd32 } \end{aligned}$ | 6 |  |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 6 |
| ADD.L ERs, ERd | L | $\begin{aligned} & \text { ERd32+ERs32 } \rightarrow \\ & \text { ERd32 } \end{aligned}$ |  | 2 |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| ADDX.B \#xx:8, Rd | B | Rd8+\#xx:8 +C $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | - | 1 | $\downarrow$ | (3) | $\downarrow$ | $\downarrow$ |  | 2 |
| ADDX.B Rs, Rd | B | Rd8+Rs8 +C $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | (3) | $\downarrow$ | $\downarrow$ |  | 2 |
| ADDS.L \#1, ERd | L | ERd32+1 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| ADDS.L \#2, ERd | L | ERd32+2 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| ADDS.L \#4, ERd | L | ERd32+4 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| INC.B Rd | B | Rd8+1 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\uparrow$ | $\downarrow$ | - |  | 2 |
| INC.W \#1, Rd | W | Rd16+1 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| INC.W \#2, Rd | W | Rd16+2 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\uparrow$ | $\downarrow$ | - |  | 2 |
| INC.L \#1, ERd | L | ERd32+1 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| INC.L \#2, ERd | L | ERd32+2 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| DAA Rd | B | Rd8 decimal adjust $\rightarrow \text { Rd8 }$ |  | 2 |  |  |  |  |  |  |  | - | * | $\downarrow$ | $\downarrow$ | * | - |  | 2 |
| SUB.B Rs, Rd | B | Rd8-Rs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SUB.W \#xx:16, Rd | W | Rd16-\#xx:16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | - | (1) | $\downarrow$ | $\downarrow$ | 1 | $\downarrow$ |  | 4 |
| SUB.W Rs, Rd | W | Rd16-Rs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | (1) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SUB.L \#xx:32, ERd | L | ERd32-\#xx:32 $\rightarrow$ ERd32 | 6 |  |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 6 |
| SUB.L ERs, ERd | L | ERd32-ERs32 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SUBX.B \#xx:8, Rd | B | Rd8-\#xx:8-C $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | (3) | $\downarrow$ | $\downarrow$ |  | 2 |
| SUBX.B Rs, Rd | B | Rd8-Rs8-C $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | I | $\downarrow$ | (3) | $\downarrow$ | $\downarrow$ |  | 2 |
| SUBS.L \#1, ERd | L | ERd32-1 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| SUBS.L \#2, ERd | L | ERd32-2 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| SUBS.L \#4, ERd | L | ERd32-4 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| DEC.B Rd | B | Rd8-1 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| DEC.W \#1, Rd | W | Rd16-1 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| DEC.W \#2, Rd | W | Rd16-2 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |


| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{aligned} & \text { X } \\ & \text { 相 } \end{aligned}$ | ¢ |  |  |  | (ᄌ) | $\begin{aligned} & \text { O} \\ & \text { 2 } \\ & \text { OU } \end{aligned}$ |  | 1 |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  | 1 | H | N | Z | V | C | 2 | 4 |
| DEC.L \#1, ERd | L | ERd32-1 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| DEC.L \#2, ERd | L | ERd32-2 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | - |  | 2 |
| DAS.Rd | B | Rd8 decimal adjust $\rightarrow \text { Rd8 }$ |  | 2 |  |  |  |  |  |  |  | - | * | $\downarrow$ | $\downarrow$ | * | - |  | 2 |
| MULXU. B Rs, Rd | B | Rd8 $\times$ Rs8 $\rightarrow$ Rd16 (unsigned multiplication) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 4 |
| MULXU. W Rs, ERd | W | Rd16 $\times$ Rs16 $\rightarrow$ ERd32 (unsigned multiplication) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| MULXS. B Rs, Rd | B | Rd8 $\times$ Rs8 $\rightarrow$ Rd16 (signed multiplication) |  | 4 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | - | - |  | 6 |
| MULXS. W Rs, ERd | W | Rd16 $\times$ Rs16 $\rightarrow$ ERd32 (signed multiplication) |  | 4 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | - | - |  | 4 |
| DIVXU. B Rs, Rd | B | Rd16 $\div$ Rs8 $\rightarrow$ Rd16 <br> (RdH: remainder, <br> RdL: quotient) (unsigned division) |  | 2 |  |  |  |  |  |  |  | - | - | (6) | (7) | - | - |  | 14 |
| DIVXU. W Rs, ERd | W | ERd32 $\div$ Rs $16 \rightarrow$ ERd32 <br> (Ed: remainder, <br> Rd: quotient) (unsigned division) |  | 2 |  |  |  |  |  |  |  | - | - | (6) | (7) | - | - |  | 22 |
| DIVXS. B Rs, Rd | B | Rd16 $\div$ Rs8 $\rightarrow$ Rd16 <br> (RdH: remainder, <br> RdL: quotient) <br> (signed division) |  | 4 |  |  |  |  |  |  |  | - | - | (8) | (7) | - | - |  | 16 |
| DIVXS. W Rs, ERd | W | ERd32 $\div$ Rs16 $\rightarrow$ ERd32 <br> (Ed: remainder, <br> Rd: quotient) (signed division) |  | 4 |  |  |  |  |  |  |  | - | - | (8) | (7) | - | - |  | 24 |
| CMP.B \#xx:8, Rd | B | Rd8-\#xx:8 | 2 |  |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| CMP.B Rs, Rd | B | Rd8-Rs8 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\uparrow$ | $\downarrow$ | $\uparrow$ | $\downarrow$ |  | 2 |
| CMP.W \#xx:16, Rd | W | Rd16-\#xx:16 | 4 |  |  |  |  |  |  |  |  | - | (1) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 4 |
| CMP.W Rs, Rd | W | Rd16-Rs16 |  | 2 |  |  |  |  |  |  |  | - | (1) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| CMP.L \#xx:32, ERd | L | ERd32-\#xx:32 | 6 |  |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\uparrow$ | $\downarrow$ | $\downarrow$ |  | 4 |
| CMP.L ERs, ERd | L | ERd32-ERs32 |  | 2 |  |  |  |  |  |  |  | - | (2) | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |


| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | ¢ |  |  |  |  |  |  | 1 |  |  |  |  |  |  |  | $\begin{aligned} & \text { ס } \\ & \text { O } \\ & \text { U } \\ & \text { N } \\ & \text { O } \end{aligned}$ |
|  |  |  |  |  |  |  |  |  |  |  |  | 1 | H | N | Z | v | C |  |  |
| NEG.B Rd | B | 0-Rd8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | I | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| NEG.W Rd | W | 0-Rd16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| NEG.L ERd | L | 0-ERd32 $\rightarrow$ ERd32 |  | 2 |  |  |  |  |  |  |  | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| EXTU.W Rd | W | $0 \rightarrow \text { (<bits } 15 \text { to 8> }$ of Rd16) |  | 2 |  |  |  |  |  |  |  | - | - | 0 | $\downarrow$ | 0 | - |  | 2 |
| EXTU.L ERd | L | $0 \rightarrow$ (<bits 31 to 16> of ERd32) |  | 2 |  |  |  |  |  |  |  | - | - | 0 | $\downarrow$ | 0 | - |  | 2 |
| EXTS.W Rd | W | (<bit 7> of Rd16) $\rightarrow$ (<bits 15 to 8> of Rd16) |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |
| EXTS.L ERd | L | (<bit 15> of ERd32) $\rightarrow$ (<bits 31 to 16> of ERd32) |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | - |  | 2 |

## 3．Logic instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length（bytes） |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No．of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | $\underset{\sim}{\boldsymbol{x}}$ | $\begin{aligned} & \stackrel{c}{\underset{\sim}{\sim}} \\ & \stackrel{\rightharpoonup}{山} \\ & \hline \end{aligned}$ |  |  | $\begin{aligned} & \mathbb{\pi} \\ & \text { (8) } \end{aligned}$ |  |  | 1 |  |  |  |  |  |  |  |  |
|  |  |  | $\begin{aligned} & \text { 㐅} \\ & \text { 相 } \end{aligned}$ |  |  |  |  |  |  |  |  | 1 | H | N | Z | V | C |  |  |
| AND．B \＃xx：8，Rd | B | Rd8＾\＃xx：8 $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| AND．B Rs，Rd | B | Rd8＾Rs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| AND．W \＃xx：16，Rd | W | Rd16＾\＃xx：16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 4 |
| AND．W Rs，Rd | W | Rd16＾Rs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| AND．L \＃xx：32，ERd | L | ERd32＾\＃xx：32 $\rightarrow$ ERd32 | 6 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\uparrow$ | 0 | － |  | 6 |
| AND．L ERs，ERd | L | ERd32＾ERs32 $\rightarrow$ ERd32 |  | 4 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 4 |
| OR．B \＃xx：8，Rd | B | Rd8v\＃xx：8 $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| OR．B Rs，Rd | B | Rd8vRs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| OR．W \＃xx：16，Rd | W | Rd16v\＃xx：16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 4 |
| OR．W Rs，Rd | W | Rd16vRs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| OR．L \＃xx：32，ERd | L | ERd32v\＃xx：32 $\rightarrow$ ERd32 | 6 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 6 |
| OR．L ERs，ERd | L | ERd32vERs32 $\rightarrow$ ERd32 |  | 4 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\uparrow$ | 0 | － |  | 4 |
| XOR．B \＃xx：8，Rd | B | Rd8 $\oplus$ \＃xx：8 $\rightarrow$ Rd8 | 2 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| XOR．B Rs，Rd | B | Rd8 $\oplus$ Rs8 $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| XOR．W \＃xx：16，Rd | W | Rd16 $\oplus$ \＃xx：16 $\rightarrow$ Rd16 | 4 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\uparrow$ | 0 | － |  | 4 |
| XOR．W Rs，Rd | W | Rd16 $\oplus$ Rs16 $\rightarrow$ Rd16 |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| XOR．L \＃xx：32，ERd | L | ERd32 $\oplus$ \＃xx：32 $\rightarrow$ ERd32 | 6 |  |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 6 |
| XOR．L ERs，ERd | L | ERd32 $\oplus$ ERs32 $\rightarrow$ ERd32 |  | 4 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 4 |
| NOT．B Rd | B | $\neg \mathrm{Rd} 8 \rightarrow \mathrm{Rd} 8$ |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| NOT．W Rd | W | $\neg \mathrm{Rd} 16 \rightarrow \mathrm{Rd} 16$ |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |
| NOT．L ERd | L | $\neg \mathrm{Rd} 32 \rightarrow \mathrm{Rd} 32$ |  | 2 |  |  |  |  |  |  |  | － | － | $\downarrow$ | $\downarrow$ | 0 | － |  | 2 |

4. Shift instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{*}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\underset{~}{\underset{X}{x}}$ | c |  |  |  |  |  | $\begin{aligned} & 0 \\ & 0 \\ & \text { B } \\ & \text { B6 } \end{aligned}$ | ® |  |  |  |  |  |  |  | 융 |
|  |  |  |  |  |  |  |  |  |  | 1 |  | H | N | z | v | c |  |  |
| SHAL.B Rd | B |  |  | 2 |  |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SHAL.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SHAL.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| SHAR.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHAR.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHAR.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHLL.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | 1 | 0 | $\downarrow$ |  | 2 |
| SHLL.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHLL.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHLR.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHLR.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| SHLR.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTXL.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTXL.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\mathfrak{\downarrow}$ | 0 | $\downarrow$ |  | 2 |
| ROTXL.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTXR.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTXR.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTXR.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\mathfrak{\downarrow}$ | 0 | $\downarrow$ |  | 2 |
| ROTL.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTL.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTL.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTR.B Rd | B |  |  | 2 |  |  |  |  |  |  |  | - | - | $\downarrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTR.W Rd | W |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\downarrow$ | 0 | $\downarrow$ |  | 2 |
| ROTR.L ERd | L |  |  | 2 |  |  |  |  |  |  |  | - | - | $\uparrow$ | $\uparrow$ | 0 | 1 |  | 2 |

## 5. Bit manipulation instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  | * | $\underset{\sim}{1}$ |  | (8) | (B) | (8) | (8) | (8) | 1 | 1 | H | N | z | v | C | z | $\stackrel{8}{4}$ |
| BSET \#xx:3, Rd | B | $(\# x x: 3$ of Rd8) $\leftarrow 1$ |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BSET \#xx:3, @ERd | B | $(\# x x: 3$ of @ERd) $\leftarrow 1$ |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BSET \#xx:3, @aa:8 | B | $(\# x x: 3$ of @aa:8) $\leftarrow 1$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BSET Rn, Rd | B | $($ Rn8 of Rd8) $\leftarrow 1$ |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BSET Rn, @ERd | B | $($ Rn8 of @ERd $) \leftarrow 1$ |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BSET Rn, @aa:8 | B | $($ Rn8 of @aa:8) $\leftarrow 1$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BCLR \#xx:3, Rd | B | $(\# x x: 3$ of Rd8) $\leftarrow 0$ |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BCLR \#xx:3, @ERd | B | $(\# x x: 3$ of @ERd $) \leftarrow 0$ |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BCLR \#xx:3, @aa:8 | B | $(\# x x: 3$ of @aa:8) $\leftarrow 0$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BCLR Rn, Rd | B | $($ Rn8 of Rd8) $\leftarrow 0$ |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BCLR Rn, @ERd | B | $($ Rn8 of @ERd $) \leftarrow 0$ |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BCLR Rn, @aa:8 | B | $($ Rn8 of @aa:8) $\leftarrow 0$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BNOT \#xx:3, Rd | B | (\#xx:3 of Rd8) $\leftarrow$ $\neg$ (\#xx:3 of Rd8) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BNOT \#xx:3, @ERd | B | (\#xx:3 of @ERd) $\leftarrow$ $\neg(\# x x: 3$ of @ERd) |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BNOT \#xx:3, @aa:8 | B | $\begin{aligned} & (\# x x: 3 \text { of @aa:8) } \leftarrow \\ & \neg(\# x x: 3 \text { of @aa:8) } \end{aligned}$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BNOT Rn, Rd | B | (Rn8 of Rd8) $\leftarrow$ <br> $\neg(R n 8$ of Rd8) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BNOT Rn, @ERd | B | $($ Rn8 of @ERd) $\leftarrow$ <br> $\neg($ Rn8 of @ERd) |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BNOT Rn, @aa:8 | B | (Rn8 of @aa:8) $\leftarrow$ <br> ᄀ (Rn8 of @aa:8) |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BTST \#xx:3, Rd | B | $\neg(\# x x: 3$ of Rd8) $\rightarrow$ Z |  | 2 |  |  |  |  |  |  |  | - | - | - | $\downarrow$ | - | - |  | 2 |
| BTST \#xx:3, @ERd | B | $\neg(\# x x: 3$ of @ERd) $\rightarrow$ Z |  |  | 4 |  |  |  |  |  |  | - | - | - | $\downarrow$ | - | - |  | 6 |
| BTST \#xx:3, @aa:8 | B | $\neg(\# x x: 3$ of @aa:8) $\rightarrow$ Z |  |  |  |  |  | 4 |  |  |  | - | - | - | $\downarrow$ | - | - |  | 6 |
| BTST Rn, Rd | B | $\neg($ Rn8 of @Rd8) $\rightarrow$ Z |  | 2 |  |  |  |  |  |  |  | - | - | - | $\downarrow$ | - | - |  | 2 |
| BTST Rn, @ERd | B | $\neg($ Rn8 of @ERd $) \rightarrow$ Z |  |  | 4 |  |  |  |  |  |  | - | - | - | $\downarrow$ | - | - |  | 6 |
| BTST Rn, @aa:8 | B | $\neg($ Rn8 of @aa:8) $\rightarrow$ Z |  |  |  |  |  | 4 |  |  |  | - | - | - | $\downarrow$ | - | - |  | 6 |
| BLD \#xx:3, Rd | B | (\#xx:3 of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |


| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  | $\begin{aligned} & \stackrel{\text { さ }}{\underset{\sim}{\boldsymbol{\sim}}} \\ & \hline \end{aligned}$ |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  | $\begin{aligned} & \text { 㐅} \\ & \text { x } \end{aligned}$ | ¢ | (8) | (8) | (8) | (8) | (8) | (8) | 1 | 1 | H | N | Z | V | C | 2 | 8 |
| BLD \#xx:3, @ERd | B | (\#xx:3 of @ERd) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BLD \#xx:3, @aa:8 | B | (\#xx:3 of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BILD \#xx:3, Rd | B | $\neg(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BILD \#xx:3, @ERd | B | $\neg(\# x x: 3$ of @ERd) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BILD \#xx:3, @aa:8 | B | $\neg(\# x x: 3$ of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BST \#xx:3, Rd | B | $C \rightarrow$ (\#xx:3 of Rd8) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BST \#xx:3, @ERd | B | C $\rightarrow$ (\#xx:3 of @ERd24) |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BST \#xx:3, @aa:8 | B | C $\rightarrow$ (\#xx:3 of @aa:8) |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BIST \#xx:3, Rd | B | $\neg \mathrm{C} \rightarrow$ (\#xx:3 of Rd8) |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| BIST \#xx:3, @ERd | B | $\neg \mathrm{C} \rightarrow$ (\#xx:3 of @ERd24) |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| BIST \#xx:3, @aa:8 | B | $\neg \mathrm{C} \rightarrow$ (\#xx:3 of @aa:8) |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 8 |
| BAND \#xx:3, Rd | B | $\mathrm{C} \wedge(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BAND \#xx:3, @ERd | B | C^(\#xx:3 of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BAND \#xx:3, @aa:8 | B | C^(\#xx:3 of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIAND \#xx:3, Rd | B | C^ ᄀ (\#xx:3 of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BIAND \#xx:3, @ERd | B | C^ ᄀ (\#xx:3 of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIAND \#xx:3, @aa:8 | B | C^ ᄀ (\#xx:3 of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BOR \#xx:3, Rd | B | $C \vee(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BOR \#xx:3, @ERd | B | $C \vee(\# x x: 3$ of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BOR \#xx:3, @aa:8 | B | C $(\#$ ( $x$ x:3 of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIOR \#xx:3, Rd | B | $C \vee \neg(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BIOR \#xx:3, @ERd | B | $C \vee \neg(\# x x: 3$ of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIOR \#xx:3, @aa:8 | B | C $\vee \neg(\# x x: 3$ of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BXOR \#xx:3, Rd | B | $C \oplus(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BXOR \#xx:3, @ERd | B | $C \oplus(\# x x: 3$ of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BXOR \#xx:3, @aa:8 | B | C¢(\#xx:3 of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIXOR \#xx:3, Rd | B | $C \oplus \neg(\# x x: 3$ of Rd8) $\rightarrow$ C |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 2 |
| BIXOR \#xx:3, @ERd | B | $C \oplus \neg(\# x x: 3$ of @ERd24) $\rightarrow$ C |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |
| BIXOR \#xx:3, @aa:8 | B | $\mathrm{C} \oplus \neg(\# \mathrm{xx}: 3$ of @aa:8) $\rightarrow$ C |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | $\downarrow$ |  | 6 |

6. Branching instructions

| Mnemonic |  | Operation¢ $\begin{gathered}\text { Branch } \\ \text { Condition }\end{gathered}$ |  | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  | $\underset{\text { x }}{\text { п }}$ | $\begin{array}{\|l} \overline{\mathbf{m}} \\ \stackrel{\rightharpoonup}{\mathbf{B}} \end{array}$ | $\begin{aligned} & \text { Bे } \\ & \hline(8) \end{aligned}$ | $\begin{array}{\|l} \mathbf{u} \\ \mathbf{1} \\ \text { ® } \end{array}$ | $\begin{aligned} & \text { 觅 } \\ & \text { Be } \end{aligned}$ | $\frac{\overrightarrow{8}}{8}$ | $\begin{aligned} & \text { © } \\ & \text { © } \\ & \text { Bi } \end{aligned}$ | 1 | 1 | H | N | Z | V | C | E | $\stackrel{\text { 20 }}{4}$ |
| BRA d:8 (BT d:8) | - | If condition is true then $P C \leftarrow P C+d$ else next; | Always |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BRA d:16 (BT d:16) | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BRN d:8 (BF d:8) | - |  | Never |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BRN d:16 (BF d:16) | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BHI d:8 | - |  | $C \vee Z=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BHI d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BLS d:8 | - |  | $C \vee Z=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BLS d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BCC d:8 (BHS d:8) | - |  | $C=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BCC d:16 (BHS d:16) | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BCS d:8 (BLO d:8) | - |  | $C=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BCS d:16 (BLO d:16) | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BNE d:8 | - |  | $Z=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BNE d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BEQ d:8 | - |  | $Z=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BEQ d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BVC d:8 | - |  | $\mathrm{V}=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BVC d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BVS d:8 | - |  | $V=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BVS d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BPL d:8 | - |  | $\mathrm{N}=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BPL d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BMI d:8 | - |  | $\mathrm{N}=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BMI d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BGE d:8 | - |  | $\mathrm{N} \oplus \mathrm{V}=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BGE d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BLT d:8 | - |  | $\mathrm{N} \oplus \mathrm{V}=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BLT d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BGT d:8 | - |  | $\mathrm{Z} \vee(\mathrm{N} \oplus \mathrm{V})=0$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BGT d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |
| BLE d:8 | - |  | $\mathrm{Z} \vee(\mathrm{N} \oplus \mathrm{V})=1$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - |  | 4 |
| BLE d:16 | - |  |  |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - |  | 6 |


| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{array}{\|l} \text { 㐅 } \\ \text { ج } \end{array}$ | 듣 |  |  |  | $\begin{aligned} & \text { K } \\ & \text { ®í } \end{aligned}$ | $\begin{aligned} & 0 \\ & 0 \\ & \text { के } \\ & \text { Be } \end{aligned}$ |  | 1 |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  | 1 | H | N | Z | v | c |  |  |
| JMP @ERn | - | $\mathrm{PC} \leftarrow \mathrm{ERn}$ |  |  | 2 |  |  |  |  |  |  | - | - | - | - | - | - |  | 4 |
| JMP @aa:24 | - | $\mathrm{PC} \leftarrow \mathrm{aa}: 24$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - |  | 6 |
| JMP @@aa:8 | - | $\mathrm{PC} \leftarrow$ @aa:8 |  |  |  |  |  |  |  | 2 |  | - | - | - | - | - | - | 8 | 10 |
| BSR d:8 | - | $\begin{aligned} & \mathrm{PC} \rightarrow @-\mathrm{SP} \\ & \mathrm{PC} \leftarrow \mathrm{PC}+\mathrm{d}: 8 \end{aligned}$ |  |  |  |  |  |  | 2 |  |  | - | - | - | - | - | - | 6 | 8 |
| BSR d:16 | - | $\begin{aligned} & \mathrm{PC} \rightarrow @-\mathrm{SP} \\ & \mathrm{PC} \leftarrow \mathrm{PC}+\mathrm{d}: 16 \end{aligned}$ |  |  |  |  |  |  | 4 |  |  | - | - | - | - | - | - | 8 | 10 |
| JSR @ERn | - | $\begin{aligned} & \mathrm{PC} \rightarrow @-S P \\ & \mathrm{PC} \leftarrow @ E R n \end{aligned}$ |  |  | 2 |  |  |  |  |  |  | - | - | - | - | - | - | 6 | 8 |
| JSR @aa:24 | - | $\begin{aligned} & \mathrm{PC} \rightarrow \text { @-SP } \\ & \mathrm{PC} \leftarrow \text { @aa:24 } \end{aligned}$ |  |  |  |  |  | 4 |  |  |  | - | - | - | - | - | - | 8 | 10 |
| JSR @@aa:8 | - | $\begin{aligned} & \mathrm{PC} \rightarrow \text { @-SP } \\ & \mathrm{PC} \leftarrow \text { @aa: } 8 \end{aligned}$ |  |  |  |  |  |  |  | 2 |  | - | - | - | - | - | - | 8 | 12 |
| RTS | - | $\mathrm{PC} \leftarrow$ @SP+ |  |  |  |  |  |  |  |  | 2 | - | - | - | - | - | - | 8 | 10 |

## 7. System control instructions

| Mnemonic |  | Operation | Addressing Mode and Instruction Length (bytes) |  |  |  |  |  |  |  |  | Condition Code |  |  |  |  |  | No. of States ${ }^{* 1}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  | $\begin{aligned} & \stackrel{+}{\underset{\sim}{\boldsymbol{\sim}}} \end{aligned}$ |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  |  | * | ¢ | (8) | (8) | (8) | (8) | (3) | (8) | 1 | 1 | H | N | z | v | C | 2 | ¢ |
| TRAPA \#x:2 | - | $\begin{aligned} & \mathrm{PC} \rightarrow @-\mathrm{SP} \\ & \mathrm{CCR} \rightarrow \text { @-SP } \\ & \text { <vector> } \rightarrow \mathrm{PC} \end{aligned}$ |  |  |  |  |  |  |  |  | 2 | 1 | - | - | - | - | - | 14 | 16 |
| RTE | - | $\begin{aligned} & \mathrm{CCR} \leftarrow @ \mathrm{SP}+ \\ & \mathrm{PC} \leftarrow @ \mathrm{SP}+ \end{aligned}$ |  |  |  |  |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 10 |
| SLEEP | - | Transition to powerdown state |  |  |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| LDC \#xx:8, CCR | B | \#xx:8 $\rightarrow$ CCR | 2 |  |  |  |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| LDC Rs, CCR | B | Rs8 $\rightarrow$ CCR |  | 2 |  |  |  |  |  |  |  | $\downarrow$ | 1 | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| LDC @ERs, CCR | W | @ERs $\rightarrow$ CCR |  |  | 4 |  |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 6 |
| LDC @(d:16, ERs), CCR | W | @(d:16, ERs) $\rightarrow$ CCR |  |  |  | 6 |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 8 |
| LDC @(d:24, ERs), CCR | W | @(d:24, ERs) $\rightarrow$ CCR |  |  |  | 10 |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 12 |
| LDC @ERs+, CCR | W | $\begin{aligned} & @ E R s \rightarrow \text { CCR } \\ & \text { ERs32+2 } \rightarrow \text { ERs32 } \end{aligned}$ |  |  |  |  | 4 |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 8 |
| LDC @aa:16, CCR | W | @aa:16 $\rightarrow$ CCR |  |  |  |  |  | 6 |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 8 |
| LDC @aa:24, CCR | W | @aa:24 $\rightarrow$ CCR |  |  |  |  |  | 8 |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 10 |
| STC CCR, Rd | B | CCR $\rightarrow$ Rd8 |  | 2 |  |  |  |  |  |  |  | - | - | - | - | - | - |  | 2 |
| STC CCR, @ERd | W | CCR $\rightarrow$ @ERd |  |  | 4 |  |  |  |  |  |  | - | - | - | - | - | - |  | 6 |
| STC CCR, @(d:16, ERd) | W | CCR $\rightarrow$ @(d:16, ERd) |  |  |  | 6 |  |  |  |  |  | - | - | - | - | - | - |  | 8 |
| STC CCR, @(d:24, ERd) | W | CCR $\rightarrow$ @(d:24, ERd) |  |  |  | 10 |  |  |  |  |  | - | - | - | - | - | - |  | 12 |
| STC CCR, @-ERd | W | $\begin{aligned} & \text { ERd32-2 } \rightarrow \text { ERd32 } \\ & \text { CCR } \rightarrow \text { @ERd } \end{aligned}$ |  |  |  |  | 4 |  |  |  |  | - | - | - | - | - | - |  | 8 |
| STC CCR, @aa:16 | W | CCR $\rightarrow$ @aa:16 |  |  |  |  |  | 6 |  |  |  | - | - | - | - | - | - |  | 8 |
| STC CCR, @aa:24 | W | CCR $\rightarrow$ @aa:24 |  |  |  |  |  | 8 |  |  |  | - | - | - | - | - | - |  | 10 |
| ANDC \#xx:8, CCR | B | CCR^\#xx:8 $\rightarrow$ CCR | 2 |  |  |  |  |  |  |  |  | $\uparrow$ | 1 | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| ORC \#xx:8, CCR | B | CCRv\#xx:8 $\rightarrow$ CCR | 2 |  |  |  |  |  |  |  |  | $\downarrow$ | 1 | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| XORC \#xx:8, CCR | B | CCR $\oplus$ \#xx: $8 \rightarrow$ CCR | 2 |  |  |  |  |  |  |  |  | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ | $\downarrow$ |  | 2 |
| NOP | - | $\mathrm{PC} \leftarrow \mathrm{PC}+2$ |  |  |  |  |  |  |  |  | 2 | - | - | - | - | - | - |  | 2 |

8. Block transfer instructions

|  |  |  |  | Add | ddres ructio | ssing <br> ion L | ng M Leng | Mode gth | $\begin{aligned} & \text { de and } \\ & \text { I (byte } \end{aligned}$ | nd ytes) |  |  |  |  |  |  |  | No. Stat | $\text { ofs }^{* 1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | $\stackrel{\sim}{N}$ | Operation |  |  |  |  | $\begin{aligned} & \stackrel{ \pm}{\underset{\sim}{x}} \\ & \underset{W}{*} \end{aligned}$ |  |  |  |  |  | Cond | ditio | Co |  |  |  |  |
|  |  |  | X | $\underset{\sim}{\square}$ |  |  |  |  |  |  | 1 | 1 | H | N | z | v | c | 2 |  |
| EEPMOV. B |  | if $R 4 L \neq 0$ then  <br> repeat $@ R 5$ <br>  $\rightarrow @ R 6$ <br> $R 5+1$ $\rightarrow R 5$ <br>  $R 6+1 \rightarrow R 6$ <br>  $R 4 L-1 \rightarrow R 4 L$ <br> until $R 4 L=0$ <br> else next  |  |  |  |  |  |  |  |  | 4 | - | - | - | - | - |  | $\left.\begin{array}{\|c\|} \hline 8+ \\ 4 n^{* 2} \end{array} \right\rvert\,$ |  |
| EEPMOV. W |  |  |  |  |  |  |  |  |  |  | 4 | - | - | - | - | - |  | $\left.\begin{array}{\|c\|} \hline 8+ \\ 4 n^{* 2} \end{array} \right\rvert\,$ |  |

Notes: 1. The number of states is the number of states required for execution when the instruction and its operands are located in on-chip memory. For other cases see section A.3, Number of States Required for Execution.
2. $n$ is the value set in register $R 4 L$ or $R 4$.
(1) Set to 1 when a carry or borrow occurs at bit 11 ; otherwise cleared to 0 .
(2) Set to 1 when a carry or borrow occurs at bit 27 ; otherwise cleared to 0 .
(3) Retains its previous value when the result is zero; otherwise cleared to 0 .
(4) Set to 1 when the adjustment produces a carry; otherwise retains its previous value.
(5) The number of states required for execution of an instruction that transfers data in synchronization with the E clock is variable.
(6) Set to 1 when the divisor is negative; otherwise cleared to 0 .
(7) Set to 1 when the divisor is zero; otherwise cleared to 0 .
(8) Set to 1 when the quotient is negative; otherwise cleared to 0.

## A. 2 Operation Code Map

Table A. 2 Operation Code Map

| Instruction code: |  |  | 1st byte | 2nd byte |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | AH | H AL | BH | BL |  |  |  |  |  |  |  |  |  |  |  |
|  | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | A | B | C | D | E | F |
| 0 | NOP | Table A-2 <br> (2) | STC | LDC | ORC | XORC | ANDC | LDC | ADD |  | Table A-2 <br> (2) | Table A-2 <br> (2) | MOV |  | ADDX | Table A-2 <br> (2) |
| 1 | Table A-2 <br> (2) | Table A-2 <br> (2) | Table A-2 <br> (2) | Table A-2 <br> (2) | OR.B | XOR.B | AND.B | Table A-2 <br> (2) | SUB |  | Table A-2 <br> (2) | Table A-2 <br> (2) | CMP |  | SUBX | Table A-2 <br> (2) |
| 2 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 3 | MOV.B |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 4 | BRA | BRN | BHI | BLS | BCC | BCS | BNE | BEQ | BVC | BVS | BPL | BMI | BGE | BLT | BGT | BLE |
| 5 | MULXU | DIVXU | MULXU | DIVXU | RTS | BSR | RTE | TRAPA | Table A-2 <br> (2) |  | JMP |  | BSR |  | JSR |  |
| 6 | BSET | BNOT | BCLR | BTST | OR | XOR | AND |  | MOV |  |  |  |  |  |  |  |
| 7 |  |  |  |  | BOR | BXOR | BAND BIAND | BILD | MOV | Table A-2 <br> (2) | Table A-2 <br> (2) | EEPMOV |  |  | A-2 |  |
| 8 | ADD |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 9 | ADDX |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| A | CMP |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| B | SUBX |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| C | OR |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| D | XOR |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| E | AND |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| F | MOV |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Table A． 2 Operation Code Map

Instruction code： | 1 1st byte | 2nd byte |
| :--- | :--- |
|  |  |

| « | 等 |  | $\underset{\sim}{0}$ |  |  |  |  |  | 気 |  | O |  | 岗 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ш |  |  |  |  |  |  |  |  |  |  |  |  | ¢ |  |  |
| － | （ |  | O |  |  |  |  |  | $\stackrel{\varrho}{\underset{\sim}{x}}$ |  | 吕 |  | $\stackrel{\text { ¢ }}{\text { ¢ }}$ |  |  |
| 0 |  |  |  |  |  |  |  |  |  |  |  |  | 㞤 |  |  |
| $\propto$ |  |  |  |  | $\frac{\overrightarrow{\mathrm{T}}}{\stackrel{\rightharpoonup}{\top}}$ | $\begin{aligned} & \frac{\mathfrak{x}}{\frac{1}{5}} \\ & \hline \end{aligned}$ | $\underset{\sim}{\text { B }}$ | $\stackrel{\text { r }}{\substack{\text { O}}}$ |  |  |  |  | $\sum_{\infty}$ |  |  |
| $<$ |  |  |  |  |  |  |  |  |  |  |  |  | ${ }_{\text {¢ }}^{0}$ |  |  |
| の |  |  |  |  |  |  |  |  |  |  |  |  | $\stackrel{\text { a }}{\text { m }}$ |  |  |
| $\infty$ | $\begin{gathered} \text { 岂 } \\ \underset{\omega}{u} \end{gathered}$ |  |  |  |  |  |  |  |  |  |  |  | 号 |  |  |
| $\wedge$ |  |  | ¢ |  |  |  |  |  | 긎 |  | 吕 |  | ¢ |  |  |
| $\omega$ |  |  |  |  |  |  |  |  |  |  |  |  | 岗 | 号 | 号 |
| $\sim$ |  |  | ¢ |  |  |  |  |  | 른 |  | 邑 |  | \％ | $\stackrel{\text { ¢ }}{\substack{\text { ¢ }}}$ | $\stackrel{\sim}{\circ}$ |
| ＊ | $\begin{aligned} & 0 \\ & 0 \\ & 0 \\ & 0 \\ & \hline \end{aligned}$ |  |  |  |  |  |  |  |  |  |  |  | O | ® | ® |
| $\infty$ |  |  |  |  | $\begin{aligned} & \text { コ1 } \\ & \text { コ } \end{aligned}$ | $\stackrel{\text { ๙ }}{\stackrel{y}{5}}$ | $\stackrel{\rightharpoonup}{x}$ $\underset{\sim}{x}$ |  | $\frac{5}{2}$ |  |  |  | $\stackrel{9}{\square}$ | ¢ | $\stackrel{\text { ¢ }}{ }$ |
| $\sim$ |  |  |  |  |  |  |  |  |  |  |  |  | 피 | $\sum_{0}^{0}$ | $\sum_{0}^{0}$ |
| － |  |  |  |  |  |  |  |  |  |  |  |  | $\underset{\substack{\text { ¢ }}}{\text { z }}$ | 会 | ） |
| － | 仓̀ | $\stackrel{0}{\underline{1}}$ | $\begin{aligned} & \text { N } \\ & \text { 号 } \end{aligned}$ | $\frac{\square}{4}$ |  |  |  |  |  | O | $\begin{aligned} & \infty \\ & \stackrel{\infty}{\omega} \end{aligned}$ | $\stackrel{N}{\Delta}$ | ¢ | $\stackrel{\text { ® }}{ }$ | 를 |
| I <br> $\stackrel{\text { a }}{\text { a }}$ | ¢ | ¢ | ® | 낭 | 안 | F | $\cong$ | $\stackrel{\text { r }}{\sim}$ | ＾ | 【 | $\stackrel{\text { ¢ }}{\sim}$ | $\stackrel{\text { แ }}{\sim}$ | $\stackrel{\infty}{\sim}$ | 용 | $\stackrel{\text { ® }}{ }$ |

Table A. 2 Operation Code Map

Notes: 1. $r$ is the register designation field.

## A. 3 Number of States Required for Execution

The tables in this section can be used to calculate the number of states required for instruction execution by the $\mathrm{H} 8 / 300 \mathrm{H}$ CPU. Table A. 4 indicates the number of instruction fetch, data read/write, and other cycles occurring in each instruction. Table A. 3 indicates the number of states required per cycle according to the bus size. The number of states required for execution of an instruction can be calculated from these two tables as follows:

Number of states $=I \times S_{I}+J \times S_{J}+K \times S_{K}+L \times S_{L}+M \times S_{M}+N \times S_{N}$

## Examples of Calculation of Number of States Required for Execution

Examples: Advanced mode, stack located in external address space, on-chip supporting modules accessed with 8 -bit bus width, external devices accessed in three states with one wait state and 16-bit bus width.

BSET \#0, @FFFFC7:8
From table A.4, $\mathrm{I}=\mathrm{L}=2$ and $\mathrm{J}=\mathrm{K}=\mathrm{M}=\mathrm{N}=0$
From table A. $3, S_{\mathrm{I}}=4$ and $\mathrm{S}_{\mathrm{L}}=3$
Number of states $=2 \times 4+2 \times 3=14$
JSR @ @30

From table A.4, $\mathrm{I}=\mathrm{J}=\mathrm{K}=2$ and $\mathrm{L}=\mathrm{M}=\mathrm{N}=0$
From table A.3, $S_{I}=S_{\mathrm{J}}=\mathrm{S}_{\mathrm{K}}=4$
Number of states $=2 \times 4+2 \times 4+2 \times 4=24$

Table A. 3 Number of States per Cycle
Access Conditions

| Cycle |  | On-Chip Memory | On-Chip Supporting Module |  | External Device |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | 8-Bit Bus |  | 16-Bit Bus |  |
|  |  |  | 8-Bit <br> Bus | 16-Bit <br> Bus | 2-State Access | 3-State <br> Access | 2-State Access | 3-State <br> Access |
| Instruction fetch | S | 2 | 6 | 3 | 4 | $6+2 m$ | 2 | $3+m$ |
| Branch address read | $\mathrm{S}_{J}$ |  |  |  |  |  |  |  |
| Stack operation | $\mathrm{S}_{\text {K }}$ |  |  |  |  |  |  |  |
| Byte data access | $\mathrm{S}_{\llcorner }$ |  | 3 |  | 2 | $3+m$ |  |  |
| Word data access | $\mathrm{S}_{\mathrm{M}}$ |  | 6 |  | 4 | $6+2 m$ |  |  |
| Internal operation | $\mathrm{S}_{\mathrm{N}}$ | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Legend
m : Number of wait states inserted into external device access

Table A. 4 Number of Cycles per Instruction

| Instruction | Mnemonic | Instruction <br> Fetch <br> I | Branch <br> Addr. Read J | Stack <br> Operation K | Byte Data <br> Access <br> L | Word Data <br> Access <br> M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADD | ADD.B \#xx:8, Rd | 1 |  |  |  |  |  |
|  | ADD.B Rs, Rd | 1 |  |  |  |  |  |
|  | ADD.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | ADD.W Rs, Rd | 1 |  |  |  |  |  |
|  | ADD.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | ADD.L ERs, ERd | 1 |  |  |  |  |  |
| ADDS | ADDS \#1/2/4, ERd | 1 |  |  |  |  |  |
| ADDX | ADDX \#xx:8, Rd | 1 |  |  |  |  |  |
|  | ADDX Rs, Rd | 1 |  |  |  |  |  |
| AND | AND.B \#xx:8, Rd | 1 |  |  |  |  |  |
|  | AND.B Rs, Rd | 1 |  |  |  |  |  |
|  | AND.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | AND.W Rs, Rd | 1 |  |  |  |  |  |
|  | AND.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | AND.L ERs, ERd | 2 |  |  |  |  |  |
| ANDC | ANDC \#xx:8, CCR | 1 |  |  |  |  |  |
| BAND | BAND \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BAND \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BAND \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
| Bcc | BRA d:8 (BT d:8) | 2 |  |  |  |  |  |
|  | BRN d:8 (BF d:8) | 2 |  |  |  |  |  |
|  | BHI d:8 | 2 |  |  |  |  |  |
|  | BLS d:8 | 2 |  |  |  |  |  |
|  | BCC d:8 (BHS d:8) | 2 |  |  |  |  |  |
|  | BCS d:8 (BLO d:8) | 2 |  |  |  |  |  |
|  | BNE d:8 | 2 |  |  |  |  |  |
|  | BEQ d:8 | 2 |  |  |  |  |  |
|  | BVC d:8 | 2 |  |  |  |  |  |
|  | BVS d:8 | 2 |  |  |  |  |  |
|  | BPL d:8 | 2 |  |  |  |  |  |
|  | BMI d:8 | 2 |  |  |  |  |  |
|  | BGE d:8 | 2 |  |  |  |  |  |
|  | BLT d:8 | 2 |  |  |  |  |  |
|  | BGT d:8 | 2 |  |  |  |  |  |
|  | BLE d:8 | 2 |  |  |  |  |  |

$\left.\begin{array}{llllll} & & \begin{array}{l}\text { Instruction } \\ \text { Fetch } \\ \text { Instruction }\end{array} & \begin{array}{l}\text { Branch } \\ \text { Addr. Read } \\ \text { J }\end{array} & \begin{array}{l}\text { Stack } \\ \text { Operation } \\ \text { K }\end{array} & \begin{array}{l}\text { Byte Data } \\ \text { Access } \\ \text { L }\end{array}\end{array} \begin{array}{l}\text { Word Data } \\ \text { Access } \\ \text { M }\end{array}\right)$

| Instruction | Mnemonic | Instruction Fetch I | Branch Addr. Read J | Stack <br> Operation <br> K | Byte Data Access L | Word Data <br> Access <br> M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BIXOR | BIXOR \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BIXOR \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BIXOR \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
| BLD | BLD \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BLD \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BLD \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
| BNOT | BNOT \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BNOT \#xx:3, @ERd | 2 |  |  | 2 |  |  |
|  | BNOT \#xx:3, @aa:8 | 2 |  |  | 2 |  |  |
|  | BNOT Rn, Rd | 1 |  |  |  |  |  |
|  | BNOT Rn, @ERd | 2 |  |  | 2 |  |  |
|  | BNOT Rn, @aa:8 | 2 |  |  | 2 |  |  |
| BOR | BOR \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BOR \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BOR \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
| BSET | BSET \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BSET \#xx:3, @ERd | 2 |  |  | 2 |  |  |
|  | BSET \#xx:3, @aa:8 | 2 |  |  | 2 |  |  |
|  | BSET Rn, Rd | 1 |  |  |  |  |  |
|  | BSET Rn, @ERd | 2 |  |  | 2 |  |  |
|  | BSET Rn, @aa:8 | 2 |  |  | 2 |  |  |
| BSR | BSR d:8 Normal* ${ }^{* 1}$ | 2 |  | 1 |  |  |  |
|  | Advanced | 2 |  | 2 |  |  |  |
|  | BSR d:16 Normal*1 | 2 |  | 1 |  |  | 2 |
|  | Advanced | 2 |  | 2 |  |  | 2 |
| BST | BST \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BST \#xx:3, @ERd | 2 |  |  | 2 |  |  |
|  | BST \#xx:3, @aa:8 | 2 |  |  | 2 |  |  |
| BTST | BTST \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BTST \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BTST \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
|  | BTST Rn, Rd | 1 |  |  |  |  |  |
|  | BTST Rn, @ERd | 2 |  |  | 1 |  |  |
|  | BTST Rn, @aa:8 | 2 |  |  | 1 |  |  |


| Instruction Mnemonic |  | Instruction Fetch I | Branch Addr. Read J | Stack <br> Operation K | Byte Data <br> Access <br> L | Word Data Access M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BXOR | BXOR \#xx:3, Rd | 1 |  |  |  |  |  |
|  | BXOR \#xx:3, @ERd | 2 |  |  | 1 |  |  |
|  | BXOR \#xx:3, @aa:8 | 2 |  |  | 1 |  |  |
| CMP | CMP.B \#xx:8, Rd | 1 |  |  |  |  |  |
|  | CMP.B Rs, Rd | 1 |  |  |  |  |  |
|  | CMP.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | CMP.W Rs, Rd | 1 |  |  |  |  |  |
|  | CMP.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | CMP.L ERs, ERd | 1 |  |  |  |  |  |
| DAA | DAA Rd | 1 |  |  |  |  |  |
| DAS | DAS Rd | 1 |  |  |  |  |  |
| DEC | DEC.B Rd | 1 |  |  |  |  |  |
|  | DEC.W \#1/2, Rd | 1 |  |  |  |  |  |
|  | DEC.L \#1/2, ERd | 1 |  |  |  |  |  |
| DIVXS | DIVXS.B Rs, Rd | 2 |  |  |  |  | 12 |
|  | DIVXS.W Rs, ERd | 2 |  |  |  |  | 20 |
| DIVXU | DIVXU.B Rs, Rd | 1 |  |  |  |  | 12 |
|  | DIVXU.W Rs, ERd | 1 |  |  |  |  | 20 |
| EEPMOV | EEPMOV.B | 2 |  |  | $2 \mathrm{n}+2^{* 2}$ |  |  |
|  | EEPMOV.W | 2 |  |  | $2 \mathrm{n}+2^{* 2}$ |  |  |
| EXTS | EXTS.W Rd | 1 |  |  |  |  |  |
|  | EXTS.L ERd | 1 |  |  |  |  |  |
| EXTU | EXTU.W Rd | 1 |  |  |  |  |  |
|  | EXTU.L ERd | 1 |  |  |  |  |  |
| INC | INC.B Rd | 1 |  |  |  |  |  |
|  | INC.W \#1/2, Rd | 1 |  |  |  |  |  |
|  | INC.L \#1/2, ERd | 1 |  |  |  |  |  |
| JMP | JMP @ERn | 2 |  |  |  |  |  |
|  | JMP @aa:24 | 2 |  |  |  |  | 2 |
|  | JMP @@aa:8 Norma**1 | 2 | 1 |  |  |  | 2 |
|  | Advanced | 2 | 2 |  |  |  | 2 |


| Instruction Mnemonic |  |  | Instruction <br> Fetch <br> I | Branch Addr. Read J | Stack <br> Operation <br> K | Byte Data Access L | Word Data <br> Access <br> M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| JSR | JSR @ERn | Normal*1 | 2 |  | 1 |  |  |  |
|  |  | Advanced | 2 |  | 2 |  |  |  |
|  | JSR @aa:24 | Normal*1 | 2 |  | 1 |  |  | 2 |
|  |  | Advanced | 2 |  | 2 |  |  | 2 |
|  | JSR @@aa:8 | Normal** | 2 | 1 | 1 |  |  |  |
|  |  | Advanced | 2 | 2 | 2 |  |  |  |
| LDC | LDC \#xx:8, CCR |  | 1 |  |  |  |  |  |
|  | LDC Rs, CCR |  | 1 |  |  |  |  |  |
|  | LDC @ERs, CCR |  | 2 |  |  |  | 1 |  |
|  | LDC @(d:16, ERs), CCR |  | 3 |  |  |  | 1 |  |
|  | LDC @(d:24, ERs), CCR |  | 5 |  |  |  | 1 |  |
|  | LDC @ERs+, CCR |  | 2 |  |  |  | 1 | 2 |
|  | LDC @aa:16, CCR |  | 3 |  |  |  | 1 |  |
|  | LDC @aa:24, CCR |  | 4 |  |  |  | 1 |  |
| MOV | MOV.B \#xx:8, Rd |  | 1 |  |  |  |  |  |
|  | MOV.B Rs, Rd |  | 1 |  |  |  |  |  |
|  | MOV.B@ERs, Rd |  | 1 |  |  | 1 |  |  |
|  | MOV.B @(d:16, ERs), Rd |  | 2 |  |  | 1 |  |  |
|  | MOV.B @(d:24, ERs), Rd |  | 4 |  |  | 1 |  |  |
|  | MOV.B @ERs+, Rd |  | 1 |  |  | 1 |  | 2 |
|  | MOV.B @aa:8, Rd |  | 1 |  |  | 1 |  |  |
|  | MOV.B @aa:16, Rd |  | 2 |  |  | 1 |  |  |
|  | MOV.B @aa:24, Rd |  | 3 |  |  | 1 |  |  |
|  | MOV.B Rs, @ERd |  | 1 |  |  | 1 |  |  |
|  | MOV.B Rs, @(d:16, ERd) |  | 2 |  |  | 1 |  |  |
|  | MOV.B Rs, @(d:24, ERd) |  | 4 |  |  | 1 |  |  |
|  | MOV.B Rs, @-ERd |  | 1 |  |  | 1 |  | 2 |
|  | MOV.B Rs, @aa:8 |  | 1 |  |  | 1 |  |  |
|  | MOV.B Rs, @aa:16 |  | 2 |  |  | 1 |  |  |
|  | MOV.B Rs, @aa:24 |  | 3 |  |  | 1 |  |  |
|  | MOV.W \#xx:16, Rd |  | 2 |  |  |  |  |  |
|  | MOV.W Rs, Rd |  | 1 |  |  |  |  |  |
|  | MOV.W @ERs, Rd |  | 1 |  |  |  | 1 |  |
|  | MOV.W @(d:16, ERs), Rd |  | 2 |  |  |  | 1 |  |


| Instruction | Mnemonic | Instruction Fetch I | Branch <br> Addr. Read <br> J | Stack <br> Operation K | Byte Data Access L | Word Data <br> Access <br> M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MOV | MOV.W @(d:24, ERs), Rd | 4 |  |  |  | 1 |  |
|  | MOV.W @ERs+, Rd | 1 |  |  |  | 1 | 2 |
|  | MOV.W @aa:16, Rd | 2 |  |  |  | 1 |  |
|  | MOV.W @aa:24, Rd | 3 |  |  |  | 1 |  |
|  | MOV.W Rs, @ERd | 1 |  |  |  | 1 |  |
|  | MOV.WRs, @(d:16, ERd) | 2 |  |  |  | 1 |  |
|  | MOV.WRs, @(d:24, ERd) | 4 |  |  |  | 1 |  |
|  | MOV.W Rs, @-ERd | 1 |  |  |  | 1 | 2 |
|  | MOV.W Rs, @aa:16 | 2 |  |  |  | 1 |  |
|  | MOV.W Rs, @aa:24 | 3 |  |  |  | 1 |  |
|  | MOV.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | MOV.L ERs, ERd | 1 |  |  |  |  |  |
|  | MOV.L @ERs, ERd | 2 |  |  |  | 2 |  |
|  | MOV.L @(d:16, ERs), ERd | 3 |  |  |  | 2 |  |
|  | MOV.L @(d:24, ERs), ERd | 5 |  |  |  | 2 |  |
|  | MOV.L@ERs+, ERd | 2 |  |  |  | 2 | 2 |
|  | MOV.L @aa:16, ERd | 3 |  |  |  | 2 |  |
|  | MOV.L @aa:24, ERd | 4 |  |  |  | 2 |  |
|  | MOV.LERs, @ERd | 2 |  |  |  | 2 |  |
|  | MOV.LERs, @(d:16, ERd) | 3 |  |  |  | 2 |  |
|  | MOV.LERs, @(d:24, ERd) | 5 |  |  |  | 2 |  |
|  | MOV.L ERs, @-ERd | 2 |  |  |  | 2 | 2 |
|  | MOV.L ERs, @aa:16 | 3 |  |  |  | 2 |  |
|  | MOV.L ERs, @aa:24 | 4 |  |  |  | 2 |  |
| MOVFPE | MOVFPE @aa:16, Rd*1 | 2 |  |  | 1 |  |  |
| MOVTPE | MOVTPE Rs, @aa:16*1 | 2 |  |  | 1 |  |  |
| MULXS | MULXS.B Rs, Rd | 2 |  |  |  |  | 12 |
|  | MULXS.W Rs, ERd | 2 |  |  |  |  | 20 |
| MULXU | MULXU.B Rs, Rd | 1 |  |  |  |  | 12 |
|  | MULXU.W Rs, ERd | 1 |  |  |  |  | 20 |
| NEG | NEG.B Rd | 1 |  |  |  |  |  |
|  | NEG.W Rd | 1 |  |  |  |  |  |
|  | NEG.L ERd | 1 |  |  |  |  |  |
| NOP | NOP | 1 |  |  |  |  |  |


| Instruction | Mnemonic | Instruction Fetch I | Branch Addr. Read J | Stack <br> Operation <br> K | Byte Data Access L | Word Data Access M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NOT | NOT.B Rd | 1 |  |  |  |  |  |
|  | NOT.W Rd | 1 |  |  |  |  |  |
|  | NOT.L ERd | 1 |  |  |  |  |  |
| OR | OR.B \#xx:8, Rd | 1 |  |  |  |  |  |
|  | OR.B Rs, Rd | 1 |  |  |  |  |  |
|  | OR.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | OR.W Rs, Rd | 1 |  |  |  |  |  |
|  | OR.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | OR.L ERs, ERd | 2 |  |  |  |  |  |
| ORC | ORC \#xx:8, CCR | 1 |  |  |  |  |  |
| POP | POP.W Rn | 1 |  |  |  | 1 | 2 |
|  | POP.L ERn | 2 |  |  |  | 2 | 2 |
| PUSH | PUSH.W Rn | 1 |  |  |  | 1 | 2 |
|  | PUSH.L ERn | 2 |  |  |  | 2 | 2 |
| ROTL | ROTL.B Rd | 1 |  |  |  |  |  |
|  | ROTL.W Rd | 1 |  |  |  |  |  |
|  | ROTL.L ERd | 1 |  |  |  |  |  |
| ROTR | ROTR.B Rd | 1 |  |  |  |  |  |
|  | ROTR.W Rd | 1 |  |  |  |  |  |
|  | ROTR.L ERd | 1 |  |  |  |  |  |
| ROTXL | ROTXL.B Rd | 1 |  |  |  |  |  |
|  | ROTXL.W Rd | 1 |  |  |  |  |  |
|  | ROTXL.L ERd | 1 |  |  |  |  |  |
| ROTXR | ROTXR.B Rd | 1 |  |  |  |  |  |
|  | ROTXR.W Rd | 1 |  |  |  |  |  |
|  | ROTXR.L ERd | 1 |  |  |  |  |  |
| RTE | RTE | 2 |  | 2 |  |  | 2 |
| RTS | RTS Normal* | 2 |  | 1 |  |  | 2 |
|  | Advanced | 2 |  | 2 |  |  | 2 |
| SHAL | SHAL.B Rd | 1 |  |  |  |  |  |
|  | SHAL.W Rd | 1 |  |  |  |  |  |
|  | SHAL.L ERd | 1 |  |  |  |  |  |
| SHAR | SHAR.B Rd | 1 |  |  |  |  |  |
|  | SHAR.W Rd | 1 |  |  |  |  |  |
|  | SHAR.L ERd | 1 |  |  |  |  |  |


| Instruction | Mnemonic | Instruction <br> Fetch <br> I | Branch Addr. Read J | Stack <br> Operation K | Byte Data Access L | Word Data <br> Access <br> M | Internal Operation N |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SHLL | SHLL.B Rd | 1 |  |  |  |  |  |
|  | SHLL.W Rd | 1 |  |  |  |  |  |
|  | SHLL.L ERd | 1 |  |  |  |  |  |
| SHLR | SHLR.B Rd | 1 |  |  |  |  |  |
|  | SHLR.W Rd | 1 |  |  |  |  |  |
|  | SHLR.L ERd | 1 |  |  |  |  |  |
| SLEEP | SLEEP | 1 |  |  |  |  |  |
| STC | STC CCR, Rd | 1 |  |  |  |  |  |
|  | STC CCR, @ERd | 2 |  |  |  | 1 |  |
|  | STC CCR, @(d:16, ERd) | 3 |  |  |  | 1 |  |
|  | STC CCR, @(d:24, ERd) | 5 |  |  |  | 1 |  |
|  | STC CCR, @-ERd | 2 |  |  |  | 1 | 2 |
|  | STC CCR, @aa:16 | 3 |  |  |  | 1 |  |
|  | STC CCR, @aa:24 | 4 |  |  |  | 1 |  |
| SUB | SUB.B Rs, Rd | 1 |  |  |  |  |  |
|  | SUB.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | SUB.W Rs, Rd | 1 |  |  |  |  |  |
|  | SUB.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | SUB.L ERs, ERd | 1 |  |  |  |  |  |
| SUBS | SUBS \#1/2/4, ERd | 1 |  |  |  |  |  |
| SUBX | SUBX \#xx:8, Rd | 1 |  |  |  |  |  |
|  | SUBX Rs, Rd | 1 |  |  |  |  |  |
| TRAPA | TRAPA \#x:2 Normal*1 | 2 | 1 | 2 |  |  | 4 |
|  | Advanced | 2 | 2 | 2 |  |  | 4 |
| XOR | XOR.B \#xx:8, Rd | 1 |  |  |  |  |  |
|  | XOR.B Rs, Rd | 1 |  |  |  |  |  |
|  | XOR.W \#xx:16, Rd | 2 |  |  |  |  |  |
|  | XOR.W Rs, Rd | 1 |  |  |  |  |  |
|  | XOR.L \#xx:32, ERd | 3 |  |  |  |  |  |
|  | XOR.L ERs, ERd | 2 |  |  |  |  |  |
| XORC | XORC \#xx:8, CCR | 1 |  |  |  |  |  |

Notes: 1. Not available in the H83052F.
2. $n$ is the value set in register R4L or R4. The source and destination are accessed $n+1$ times each.

## Appendix B Internal I/O Register

## B. 1 Addresses

| Address (low) | Register Name | Data <br> Bus <br> Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'1C | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'1D |  |  |  |  |  |  |  |  |  |  |  |
| H'1E |  |  |  |  |  |  |  |  |  |  |  |
| H'1F |  |  |  |  |  |  |  |  |  |  |  |
| H'20 | MAR0AR | 8 |  |  |  |  |  |  |  |  | DMAC channel 0A |
| H'21 | MAROAE | 8 |  |  |  |  |  |  |  |  |  |
| H'22 | MAROAH | 8 |  |  |  |  |  |  |  |  |  |
| H'23 | MAR0AL | 8 |  |  |  |  |  |  |  |  |  |
| H'24 | ETCROAH | 8 |  |  |  |  |  |  |  |  |  |
| H'25 | ETCROAL | 8 |  |  |  |  |  |  |  |  |  |
| H'26 | IOAROA | 8 |  |  |  |  |  |  |  |  |  |
| H'27 | DTCR0A | 8 | DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 | Short address mode |
|  |  |  | DTE | DTSZ | SAID | SAIDE | DTIE | DTS2A | DTS1A | DTS0A | Full address mode |
| H'28 | MAR0BR | 8 |  |  |  |  |  |  |  |  | DMAC channel OB |
| H'29 | MAROBE | 8 |  |  |  |  |  |  |  |  |  |
| H'2A | MAROBH | 8 |  |  |  |  |  |  |  |  |  |
| $\mathrm{H}^{\prime} 2 \mathrm{~B}$ | MAROBL | 8 |  |  |  |  |  |  |  |  |  |
| $\mathrm{H}^{\prime} 2 \mathrm{C}$ | ETCROBH | 8 |  |  |  |  |  |  |  |  |  |
| H'2D | ETCROBL | 8 |  |  |  |  |  |  |  |  |  |
| H'2E | IOAR0B | 8 |  |  |  |  |  |  |  |  |  |
| H'2F | DTCR0B | 8 | DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 | Short address mode |
|  |  |  | DTME | - | DAID | DAIDE | TMS | DTS2B | DTS1B | DTS0B | Full address mode |


| Address (low) | Register Name | Data <br> Bus <br> Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'30 | MAR1AR | 8 |  |  |  |  |  |  |  |  |  |
| H'31 | MAR1AE | 8 |  |  |  |  |  |  |  |  | nel |
| H'32 | MAR1AH | 8 |  |  |  |  |  |  |  |  |  |
| H'33 | MAR1AL | 8 |  |  |  |  |  |  |  |  |  |
| H'34 | ETCR1AH | 8 |  |  |  |  |  |  |  |  |  |
| H'35 | ETCR1AL | 8 |  |  |  |  |  |  |  |  |  |
| H'36 | IOAR1A | 8 |  |  |  |  |  |  |  |  |  |
| H'37 | DTCR1A | 8 | DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 | Short address mode |
|  |  |  | DTE | DTSZ | SAID | SAIDE | DTIE | DTS2A | DTS1A | DTS0A | Full address mode |
| H'38 | MAR1BR | 8 |  |  |  |  |  |  |  |  | DMAC |
| H'39 | MAR1BE | 8 |  |  |  |  |  |  |  |  | channel 1B |
| H'3A | MAR1BH | 8 |  |  |  |  |  |  |  |  |  |
| H'3B | MAR1BL | 8 |  |  |  |  |  |  |  |  |  |
| H'3C | ETCR1BH | 8 |  |  |  |  |  |  |  |  |  |
| H'3D | ETCR1BL | 8 |  |  |  |  |  |  |  |  |  |
| H'3E | IOAR1B | 8 |  |  |  |  |  |  |  |  |  |
| H'3F | DTCR1B | 8 | DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 | Short address mode |
|  |  |  | DTME | - | DAID | DAIDE | TMS | DTS2B | DTS1B | DTS0B | Full address mode |
| H'40 | FLMCR1 | 8 | FWE | SWE1 | ESU1 | PSU1 | EV1 | PV1 | E1 | P1 | Flash |
| H'41 | FLMCR2 | 8 | FLER | SWE2 | ESU2 | PSU2 | EV2 | PV2 | E2 | P2 | memory |
| H'42 | EBR1 | 8 | EB7 | EB6 | EB5 | EB4 | EB3 | EB2 | EB1 | EB0 |  |
| H'43 | EBR2 | 8 | EB15 | EB14 | EB13 | EB12 | EB11 | EB10 | EB9 | EB8 |  |
| H'44 | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'45 |  |  |  |  |  |  |  |  |  |  |  |
| H'46 |  |  |  |  |  |  |  |  |  |  |  |
| H'47 | RAMCR | 8 | - | - | - | - | RAMS | RAM2 | RAM1 | RAM0 |  |
| H'48 | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'49 |  |  |  |  |  |  |  |  |  |  |  |
| H'4A |  |  |  |  |  |  |  |  |  |  |  |
| H'4B |  |  |  |  |  |  |  |  |  |  |  |


| Address (low) | Register Name | Data Bus Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'4C | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'4D |  |  |  |  |  |  |  |  |  |  |  |
| H'4E |  |  |  |  |  |  |  |  |  |  |  |
| H'4F |  |  |  |  |  |  |  |  |  |  |  |
| H'50 |  |  |  |  |  |  |  |  |  |  |  |
| H'51 |  |  |  |  |  |  |  |  |  |  |  |
| H'52 |  |  |  |  |  |  |  |  |  |  |  |
| H'53 |  |  |  |  |  |  |  |  |  |  |  |
| H'54 |  |  |  |  |  |  |  |  |  |  |  |
| H'55 |  |  |  |  |  |  |  |  |  |  |  |
| H'56 |  |  |  |  |  |  |  |  |  |  |  |
| H'57 |  |  |  |  |  |  |  |  |  |  |  |
| H'58 |  |  |  |  |  |  |  |  |  |  |  |
| H'59 |  |  |  |  |  |  |  |  |  |  |  |
| H'5A |  |  |  |  |  |  |  |  |  |  |  |
| H'5B |  |  |  |  |  |  |  |  |  |  |  |
| H'5C | DASTCR | 8 | - | - | - | - | - | - | - | DASTE | D/A converter |
| H'5D | DIVCR | 8 | - | - | - | - | - | - | DIV1 | DIV0 | System |
| H'5E | MSTCR | 8 | PSTOP | - | MSTOP5 | MSTOP4 | MSTOP3 | MSTOP2 | MSTOP1 | MSTOP0 | control |
| H'5F | CSCR | 8 | CS7E | CS6E | CS5E | CS4E | - | - | - | - | Bus controller |
| H'60 | TSTR | 8 | - | - | - | STR4 | STR3 | STR2 | STR1 | STR0 | ITU |
| H'61 | TSNC | 8 | - | - | - | SYNC4 | SYNC3 | SYNC2 | SYNC1 | SYNC0 | all channels) |
| H'62 | TMDR | 8 | - | MDF | FDIR | PWM4 | PWM3 | PWM2 | PWM1 | PWM0 |  |
| H'63 | TFCR | 8 | - | - | CMD1 | CMD0 | BFB4 | BFA4 | BFB3 | BFA3 |  |
| H'64 | TCR0 | 8 | - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 |  |
| H'65 | TIORO | 8 | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOAO | channel 0 |
| H'66 | TIER0 | 8 | - | - | - | - | - | OVIE | IMIEB | IMIEA |  |
| H'67 | TSR0 | 8 | - | - | - | - | - | OVF | IMFB | IMFA |  |
| H'68 | TCNTOH | 16 |  |  |  |  |  |  |  |  |  |
| H'69 | TCNTOL |  |  |  |  |  |  |  |  |  |  |
| H'6A | GRAOH | 16 |  |  |  |  |  |  |  |  |  |
| H'6B | GRAOL |  |  |  |  |  |  |  |  |  |  |
| H'6C | GRBOH | 16 |  |  |  |  |  |  |  |  |  |
| H'6D | GRB0L |  |  |  |  |  |  |  |  |  |  |
| H'6E | TCR1 | 8 | - | CCLR1 | CCLR0 | CKEG1 | CKEGO | TPSC2 | TPSC1 | TPSC0 | ITU |
| H'6F | TIOR1 | 8 | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 | hannel 1 |


| Address (low) | Register Name | Data <br> Bus <br> Width | Bit Names |  |  |  |  |  |  |  | - Module <br> Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'70 | TIER1 | 8 | - | - | - | - | - | OVIE | IMIEB | IMIEA | ITU channel 1 |
| H'71 | TSR1 | 8 | - | - | - | - | - | OVF | IMFB | IMFA |  |
| H'72 | TCNT1H | 16 |  |  |  |  |  |  |  |  |  |
| H'73 | TCNT1L |  |  |  |  |  |  |  |  |  |  |
| H'74 | GRA1H | 16 |  |  |  |  |  |  |  |  |  |
| H'75 | GRA1L |  |  |  |  |  |  |  |  |  |  |
| H'76 | GRB1H | 16 |  |  |  |  |  |  |  |  |  |
| H'77 | GRB1L |  |  |  |  |  |  |  |  |  |  |
| H'78 | TCR2 | 8 | - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 | ITU channel 2 |
| H'79 | TIOR2 | 8 | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOAO |  |
| H'7A | TIER2 | 8 | - | - | - | - | - | OVIE | IMIEB | IMIEA |  |
| H'7B | TSR2 | 8 | - | - | - | - | - | OVF | IMFB | IMFA |  |
| H'7C | TCNT2H | 16 |  |  |  |  |  |  |  |  |  |
| H'7D | TCNT2L |  |  |  |  |  |  |  |  |  |  |
| H'7E | GRA2H | 16 |  |  |  |  |  |  |  |  |  |
| H'7F | GRA2L |  |  |  |  |  |  |  |  |  |  |
| H'80 | GRB2H | 16 |  |  |  |  |  |  |  |  |  |
| H'81 | GRB2L |  |  |  |  |  |  |  |  |  |  |
| H'82 | TCR3 | 8 | - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 | ITU channel 3 |
| H'83 | TIOR3 | 8 | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOAO |  |
| H'84 | TIER3 | 8 | - | - | - | - | - | OVIE | IMIEB | IMIEA |  |
| H'85 | TSR3 | 8 | - | - | - | - | - | OVF | IMFB | IMFA |  |
| H'86 | TCNT3H | 16 |  |  |  |  |  |  |  |  |  |
| H'87 | TCNT3L |  |  |  |  |  |  |  |  |  |  |
| H'88 | GRA3H | 16 |  |  |  |  |  |  |  |  |  |
| H'89 | GRA3L |  |  |  |  |  |  |  |  |  |  |
| H'8A | GRB3H | 16 |  |  |  |  |  |  |  |  |  |
| H'8B | GRB3L |  |  |  |  |  |  |  |  |  |  |
| $\mathrm{H}^{\prime} 8 \mathrm{C}$ | BRA3H | 16 |  |  |  |  |  |  |  |  |  |
| H'8D | BRA3L |  |  |  |  |  |  |  |  |  |  |
| H'8E | BRB3H | 16 |  |  |  |  |  |  |  |  |  |
| H'8F | BRB3L |  |  |  |  |  |  |  |  |  |  |
| H'90 | TOER | 8 | - | - | EXB4 | EXA4 | EB3 | EB4 | EA4 | EA3 | TU (all |
| H'91 | TOCR | 8 | - | - | - | XTGD | - | - | OLS4 | OLS3 | hannels) |
| H'92 | TCR4 | 8 | - | CCLR1 | CCLR0 | CKEG1 | CKEGO | TPSC2 | TPSC1 | TPSC0 | ITU channel 4 |
| H'93 | TIOR4 | 8 | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 |  |

Rev. 2.0, 03/01, page 698 of 822

| Address (low) | Register Name | Data <br> Bus <br> Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'94 | TIER4 | 8 | - | - | - | - | - | OVIE | IMIEB | IMIEA | ITU channel 4 |
| H'95 | TSR4 | 8 | - | - | - | - | - | OVF | IMFB | IMFA |  |
| H'96 | TCNT4H | 16 |  |  |  |  |  |  |  |  |  |
| H'97 | TCNT4L |  |  |  |  |  |  |  |  |  |  |
| H'98 | GRA4H | 16 |  |  |  |  |  |  |  |  |  |
| H'99 | GRA4L |  |  |  |  |  |  |  |  |  |  |
| H'9A | GRB4H | 16 |  |  |  |  |  |  |  |  |  |
| H'9B | GRB4L |  |  |  |  |  |  |  |  |  |  |
| H'9C | BRA4H | 16 |  |  |  |  |  |  |  |  |  |
| H'9D | BRA4L |  |  |  |  |  |  |  |  |  |  |
| H'9E | BRB4H | 16 |  |  |  |  |  |  |  |  |  |
| H'9F | BRB4L |  |  |  |  |  |  |  |  |  |  |
| H'AO | TPMR | 8 | - | - | - | - | G3NOV | G2NOV | G1NOV | GONOV | TPC |
| H'A1 | TPCR | 8 | G3CMS1 | G3CMS0 | G2CMS1 | G2CMS0 | G1CMS1 | G1CMS0 | G0CMS1 | GOCMSO |  |
| H'A2 | NDERB | 8 | NDER15 | NDER14 | NDER13 | NDER12 | NDER11 | NDER10 | NDER9 | NDER8 |  |
| H'A3 | NDERA | 8 | NDER7 | NDER6 | NDER5 | NDER4 | NDER3 | NDER2 | NDER1 | NDER0 |  |
| H'A4 | NDRB** | 8 | NDR15 | NDR14 | NDR13 | NDR12 | NDR11 | NDR10 | NDR9 | NDR8 |  |
|  |  | 8 | NDR15 | NDR14 | NDR13 | NDR12 | - | - | - | - |  |
| H'A5 | NDRA** | 8 | NDR7 | NDR6 | NDR5 | NDR4 | NDR3 | NDR2 | NDR1 | NDR0 |  |
|  |  | 8 | NDR7 | NDR6 | NDR5 | NDR4 | - | - | - | - |  |
| H'A6 | NDRB* ${ }^{*}$ | 8 | - | - | - | - | - | - | - | - |  |
|  |  | 8 | - | - | - | - | NDR11 | NDR10 | NDR9 | NDR8 |  |
| H'A7 | NDRA $^{*}{ }^{*}$ | 8 | - | - | - | - | - | - | - | - |  |
|  |  | 8 | - | - | - | - | NDR3 | NDR2 | NDR1 | NDR0 |  |
| H'A8 | TCSR ${ }^{* 2}$ | 8 | OVF | WT/IT | TME | - | - | CKS2 | CKS1 | CKSO | WDT |
| H'A9 | TCNT* ${ }^{\text {2 }}$ | 8 |  |  |  |  |  |  |  |  |  |
| H'AA | - |  | - | - | - | - | - | - | - | - |  |
| H'AB | RSTCSR ${ }^{*}$ | 8 | WRST | - | - | - | - | - | - | - |  |
| H'AC | RFSHCR | 8 | SRFMD | PSRAME | DRAME | CAS/WE | M9/M8 | RFSHE | - | RCYCE | Refresh |
| H'AD | RTMCSR | 8 | CMF | CMIE | CKS2 | CKS1 | CKSO | - | - | - | controller |
| H'AE | RTCNT | 8 |  |  |  |  |  |  |  |  |  |
| H'AF | RTCOR | 8 |  |  |  |  |  |  |  |  |  |


| Address (low) | Register Name | Data <br> Bus <br> Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'B0 | SMR | 8 | C//̄/GM | CHR | PE | O/E | STOP | MP | CKS1 | CKS0 | SCI channel 0 |
| H'B1 | BRR | 8 |  |  |  |  |  |  |  |  |  |
| H'B2 | SCR | 8 | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKE0 |  |
| H'B3 | TDR | 8 |  |  |  |  |  |  |  |  |  |
| H'B4 | SSR | 8 | TDRE | RDRF | ORER | $\begin{aligned} & \text { FER/ } \\ & \text { ERS } \end{aligned}$ | PER | TEND | MPB | MPBT |  |
| H'B5 | RDR | 8 |  |  |  |  |  |  |  |  |  |
| H'B6 | SCMR | 8 | - | - | - | - | SDIR | SINV | - | SMIF |  |
| H'B7 | Reserved | a (acce | ess prohib | ited) |  |  |  |  |  |  |  |
| H'B8 | SMR | 8 | $\mathrm{C} / \overline{\mathrm{A}}$ | CHR | PE | O/E | STOP | MP | CKS1 | CKS0 | SCI channel 1 |
| H'B9 | BRR | 8 |  |  |  |  |  |  |  |  |  |
| H'BA | SCR | 8 | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKE0 |  |
| H'BB | TDR | 8 |  |  |  |  |  |  |  |  |  |
| H'BC | SSR | 8 | TDRE | RDRF | ORER | FER | PER | TEND | MPB | MPBT |  |
| H'BD | RDR | 8 |  |  |  |  |  |  |  |  |  |
| H'BE | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'BF |  |  |  |  |  |  |  |  |  |  |  |
| H'C0 | P1DDR | 8 | P17 DDR | $\mathrm{P} 1_{6}$ DDR | $\mathrm{P} 1_{5} \mathrm{DDR}$ | P 14 DDR | $\mathrm{P} 1_{3} \mathrm{DDR}$ | $\mathrm{P} 1_{2} \mathrm{DDR}$ | P1,DDR | P10DDR | Port 1 |
| H'C1 | P2DDR | 8 | P2 ${ }_{7}$ DDR | P2 ${ }_{6}$ DDR | $\mathrm{P} 2_{5} \mathrm{DDR}$ | P 24 DDR | $\mathrm{P}_{3} \mathrm{DDR}$ | P22DDR | P2,DDR | P20DDR | Port 2 |
| H'C2 | P1DR | 8 | $\mathrm{P} 1_{7}$ | $\mathrm{P} 1{ }_{6}$ | $\mathrm{P} 1_{5}$ | P 14 | P 13 | $\mathrm{P} 1_{2}$ | $\mathrm{P} 1_{1}$ | P 10 | Port 1 |
| H'C3 | P2DR | 8 | $\mathrm{P} 2_{7}$ | $\mathrm{P} 2{ }_{6}$ | $\mathrm{P} 2_{5}$ | P 24 | $\mathrm{P}_{3}$ | $\mathrm{P} 2_{2}$ | $\mathrm{P} 2_{1}$ | P 20 | Port 2 |
| H'C4 | P3DDR | 8 | P37DDR | P36DDR | $\mathrm{P}_{5} \mathrm{DDR}$ | $\mathrm{P}_{4} \mathrm{DDR}$ | $\mathrm{P}_{3} \mathrm{DDR}$ | P32DDR | P3,DDR | P30DDR | Port 3 |
| H'C5 | P4DDR | 8 | P47DDR | $\mathrm{P}_{6}$ DDR | $\mathrm{P}_{5} \mathrm{DDR}$ | P 44 DDR | $\mathrm{P}_{3} \mathrm{DDR}$ | $\mathrm{P}_{2} \mathrm{DDR}$ | P4, DDR | P40DDR | Port 4 |
| H'C6 | P3DR | 8 | P 37 | P 36 | $\mathrm{P} 3_{5}$ | P 34 | $\mathrm{P}_{3}$ | P 32 | P3 ${ }_{1}$ | P 30 | Port 3 |
| H'C7 | P4DR | 8 | P 47 | P46 | P 45 | P44 | $\mathrm{P}_{3}$ | P 42 | P4, | P40 | Port 4 |
| H'C8 | P5DDR | 8 | - | - | - | - | $\mathrm{P5}_{3} \mathrm{DDR}$ | P5 ${ }_{2}$ DDR | P5,DDR | P50DDR | Port 5 |
| H'C9 | P6DDR | 8 | - | P6, DDR | $\mathrm{Pb}_{5}$ DDR | P64DDR | $\mathrm{P}_{3} \mathrm{DDR}$ | P6, DDR | P6,DDR | P60DDR | Port 6 |
| H'CA | P5DR | 8 | - | - | - | - | $\mathrm{P5}_{3}$ | $\mathrm{P5}_{2}$ | P5 ${ }_{1}$ | P 50 | Port 5 |
| H'CB | P6DR | 8 | - | P6 $6_{6}$ | P6 5 | P64 | $\mathrm{P6}_{3}$ | P6 ${ }_{2}$ | P6 ${ }_{1}$ | P60 | Port 6 |
| H'CC | - |  | - | - | - | - | - | - | - | - |  |
| H'CD | P8DDR | 8 | - | - | - | P84DDR | $\mathrm{P}_{3} \mathrm{DDR}$ | P82DDR | P8,DDR | P80DDR | Port 8 |
| H'CE | P7DR | 8 | $\mathrm{P} 7_{7}$ | P7 ${ }_{6}$ | P 75 | P 74 | $\mathrm{P}_{3}$ | $\mathrm{P} 7{ }_{2}$ | P7 ${ }_{1}$ | P 70 | Port 7 |
| H'CF | P8DR | 8 | - | - | - | $\mathrm{P}_{4}$ | $\mathrm{P}_{3}$ | $\mathrm{P8} 8_{2}$ | P8, | P 80 | Port 8 |
| H'D0 | P9DDR | 8 | - | - | $\mathrm{P9}_{5} \mathrm{DDR}$ | $\mathrm{Pg}_{4} \mathrm{DDR}$ | $\mathrm{Pg}_{3} \mathrm{DDR}$ | $\mathrm{Pg}_{2} \mathrm{DDR}$ | P9, DDR | P9, DDR | Port 9 |
| H'D1 | PADDR | 8 | PA_DDR | $\mathrm{PA}_{6}$ DDR | $\mathrm{PA}_{5} \mathrm{DDR}$ | $\mathrm{PA}_{4} \mathrm{DDR}$ | $\mathrm{PA}_{3} \mathrm{DDR}$ | $\mathrm{PA}_{2} \mathrm{DDR}$ | PA,DDR | PA ${ }_{0}$ DDR | Port A |
| H'D2 | P9DR | 8 | - | - | $\mathrm{P9}_{5}$ | $\mathrm{P9}_{4}$ | $\mathrm{P9}_{3}$ | $\mathrm{P9} 2$ | $\mathrm{P} 9_{1}$ | $\mathrm{P9}{ }_{0}$ | Port 9 |
| H'D3 | PADR | 8 | PA ${ }_{7}$ | PA ${ }_{6}$ | $\mathrm{PA}_{5}$ | $\mathrm{PA}_{4}$ | $\mathrm{PA}_{3}$ | $\mathrm{PA}_{2}$ | PA ${ }_{1}$ | PA ${ }_{0}$ | Port A |


| Address (low) | Register Name | Data Bus <br> Width | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'D4 | PBDDR | 8 | PB77DR | $\mathrm{PB}_{6}$ DDR | $\mathrm{PB}_{5}$ DDR | $\mathrm{PB}_{4}$ DDR | $\mathrm{PB}_{3} \mathrm{DDR}$ | PB2DDR | PB,DDR | PB ${ }_{0}$ DRR | Port B |
| H'D5 | - | - | - | - | - | - | - | - | - | - | - |
| H'D6 | PBDR | 8 | $\mathrm{PB}_{7}$ | $\mathrm{PB}_{6}$ | $\mathrm{PB}_{5}$ | $\mathrm{PB}_{4}$ | $\mathrm{PB}_{3}$ | $\mathrm{PB}_{2}$ | PB ${ }_{1}$ | PB 0 | Port B |
| H'D7 | - | - | - | - | - | - | - | - | - | - | - |
| H'D8 | P2PCR | 8 | P2 ${ }_{7} \mathrm{PCR}$ | P2 ${ }_{6} \mathrm{PCR}$ | P2 ${ }_{5} \mathrm{PCR}$ | P2 ${ }_{4} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{3} \mathrm{PCR}$ | P2 $2_{2} \mathrm{PCR}$ | P2,PCR | P2 $2_{0} \mathrm{PCR}$ | Port 2 |
| H'D9 | - |  | - | - | - | - | - | - | - | - |  |
| H'DA | P4PCR | 8 | P47PCR | P46PCR | P4 ${ }_{5} \mathrm{PCR}$ | P44 ${ }_{4}$ PCR | P43 PCR | P4 $4_{2} \mathrm{PCR}$ | P4, PCR | P40 PCR | Port 4 |
| H'DB | P5PCR | 8 | - | - | - | - | $\mathrm{P5}_{3} \mathrm{PCR}$ | P5 $5_{2} \mathrm{PCR}$ | P5,PCR | P50PCR | Port 5 |
| H'DC | DADR0 | 8 |  |  |  |  |  |  |  |  | D/A converter |
| H'DD | DADR1 | 8 |  |  |  |  |  |  |  |  |  |
| H'DE | DACR | 8 | DAOE1 | DAOE0 | DAE | - | - | - | - | - |  |
| H'DF | Reserved | (ac | ess prohi |  |  |  |  |  |  |  |  |
| H'E0 | ADDRAH | 8 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | A/D converter |
| H'E1 | ADDRAL | 8 | AD1 | AD0 | - | - | - | - | - | - |  |
| H'E2 | ADDRBH | 8 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 |  |
| H'E3 | ADDRBL | 8 | AD1 | AD0 | - | - | - | - | - | - |  |
| H'E4 | ADDRCH | 8 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 |  |
| H'E5 | ADDRCL | 8 | AD1 | AD0 | - | - | - | - | - | - |  |
| H'E6 | ADDRDH | 8 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 |  |
| H'E7 | ADDRDL | 8 | AD1 | AD0 | - | - | - | - | - | - |  |
| H'E8 | ADCSR | 8 | ADF | ADIE | ADST | SCAN | CKS | CH 2 | CH 1 | CHO |  |
| H'E9 | ADCR | 8 | TRGE | - | - | - | - | - | - | - |  |
| H'EA | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'EB |  |  |  |  |  |  |  |  |  |  |  |
| H'EC | ABWCR | 8 | ABW7 | ABW6 | ABW5 | ABW4 | ABW3 | ABW2 | ABW1 | ABWO | Bus controller |
| H'ED | ASTCR | 8 | AST7 | AST6 | AST5 | AST4 | AST3 | AST2 | AST1 | ASTO |  |
| H'EE | WCR | 8 | - | - | - | - | WMS1 | WMSO | WC1 | WC0 |  |
| H'EF | WCER | 8 | WCE7 | WCE6 | WCE5 | WCE4 | WCE3 | WCE2 | WCE1 | WCE0 |  |
| H'F0 | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'F1 | MDCR | 8 | - | - | - | - | - | MDS2 | MDS1 | MDS0 | System |
| H'F2 | SYSCR | 8 | SSBY | STS2 | STS1 | STS0 | UE | NMIEG | - | RAME | control |
| H'F3 | BRCR | 8 | A23E | A22E | A21E | - | - | - | - | BRLE | Bus controller |
| H'F4 | ISCR | 8 | - | - | IRQ5SC | IRQ4SC | IRQ3SC | IRQ2SC | IRQ1SC | IRQ0SC | Interrupt |
| H'F5 | IER | 8 | - | - | IRQ5E | IRQ4E | IRQ3E | IRQ2E | IRQ1E | IRQ0E | controller |
| H'F6 | ISR | 8 | - | - | IRQ5F | IRQ4F | IRQ3F | IRQ2F | IRQ1F | IRQ0F |  |
| H'F7 | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'F8 | IPRA | 8 | IPRA7 | IPRA6 | IPRA5 | IPRA4 | IPRA3 | IPRA2 | IPRA1 | IPRA0 |  |
| H'F9 | IPRB | 8 | IPRB7 | IPRB6 | IPRB5 | - | IPRB3 | IPRB2 | IPRB1 | - |  |


|  |  | Data | Bit Names |  |  |  |  |  |  |  | Module Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| (low) | Register Name |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 |  |
| H'FA | Reserved area (access prohibited) |  |  |  |  |  |  |  |  |  |  |
| H'FB |  |  |  |  |  |  |  |  |  |  |  |
| H'FC |  |  |  |  |  |  |  |  |  |  |  |
| H'FD |  |  |  |  |  |  |  |  |  |  |  |
| H'FE |  |  |  |  |  |  |  |  |  |  |  |
| H'FF |  |  |  |  |  |  |  |  |  |  |  |

Notes: 1. The address depends on the output trigger setting.
2. For write access to TCSR TCNT, and RSTCR see section 12.2.4, Notes on Register Access.
Legend
DMAC: DMA controller
ITU: 16-bit integrated timer unit
TPC: Programmable timing pattern controller
WDT: Watchdog timer
SCI: Serial communication interface
B. 2 Function


Bit

Bit

Initial value


Read/Write



- Short address mode
— I/O mode and idle mode
Bit

Initial value


Read/Write


Transfer counter

- Repeat mode

Bit
Initial value
Read/Write


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | ETCROAL |  |  |  |  |  |  |  |
|  |  |  |  |  | unt |  |  |  |

ETCR0A H/L—Execute Transfer Count Register 0A H/L (cont)

- Full address mode
- Normal mode

Bit


Initial value
Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

- Block transfer mode


- Short address mode

| Bit | 7 |  | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DTE |  | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 |
| Initial value | 0 |  | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W |  | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Data tr | ransfer | r selec |  |  |  |  |  |  |
|  | Bit 2 | Bit 1 | Bit 0 |  |  |  |  |  |  |
|  | DTS2 | DTS1 | DTS0 | Data Tran | Activ | Source |  |  |  |
|  | 0 | 0 | 0 | Compare | atch/inp | apture A | errupt fror | ITU cha |  |
|  |  |  | 1 | Compare | atch/inp | apture A | errupt fror | ITU cha | el 1 |
|  |  | 1 | 0 | Compare | atch/inp | apture A | errupt fror | ITU cha | el 2 |
|  |  |  | 1 | Compare | atch/inp | apture A | errupt from | ITU cha | el 3 |
|  | 1 | 0 | 0 | SCIO tran | it-data- | ty interr |  |  |  |
|  |  |  | 1 | SCIO rece | -data-fu | terrupt |  |  |  |
|  |  | 1 | 0 | Transfer | ull addr | mode |  |  |  |
|  |  |  | 1 | Transfer | ull addr | mode |  |  |  |

## Data transfer interrupt enable

| 0 | Interrupt requested by DTE bit is disabled |
| :---: | :--- |
| 1 | Interrupt requested by DTE bit is enabled |

Repeat enable

| RPE | DTIE | Description |
| :---: | :---: | :--- |
| 0 | 0 | I/O mode |
|  | 1 |  |
| 1 | 0 | Repeat mode |
|  | 1 | Idle mode |

Data transfer increment/decrement

| 0 | Incremented: If DTSZ $=0$, MAR is incremented by 1 after each transfer |
| ---: | ---: |
| If DTSZ $=1$, MAR is incremented by 2 after each transfer |  | \left\lvert\, | 1 |
| ---: | | Decremented: If DTSZ $=0$, MAR is decremented by 1 after each transfer |
| ---: |
| If DTSZ $=1$, MAR is decremented by 2 after each transfer |\right.

## Data transfer size

| 0 | Byte-size transfer |
| :--- | :--- |
| 1 | Word-size transfer |

## Data transfer enable

| 0 | Data transfer is disabled |
| :--- | :--- |
| 1 | Data transfer is enabled |

- Full address mode


Data transfer size

| 0 | Byte-size transfer |
| :--- | :--- |
| 1 | Word-size transfer |

Data transfer enable

| 0 | Data transfer is disabled |
| :--- | :--- |
| 1 | Data transfer is enabled |

Bit

Bit

Initial value
Read/Write


- Short address mode
— I/O mode and idle mode
Bit

Initial value


Read/Write


Transfer counter
— Repeat mode
Bit
Initial value
Read/Write

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Undetermined |  |  |  |  |  |  |  |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| ETCROBH |  |  |  |  |  |  |  |
| Transfer counter |  |  |  |  |  |  |  |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | ETCROBL |  |  |  |  |  |  |  |
|  |  |  |  |  | ount |  |  |  |

ETCR0B H/L—Execute Transfer Count Register 0B H/L (cont)

- Full address mode
- Normal mode

Bit

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Initial value

## Read/Write

- Block transfer mode

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Read/Write


Block transfer counter
IOAR0B—I/O Address Register 0B $\quad H^{\prime} 2 E \quad$ DMAC0

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value |  |  |  | Und | ined |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Short address mode: source or destination address
Full address mode: not used

## DTCR0B—Data Transfer Control Register 0B

$\mathbf{H}^{\prime} \mathbf{2 F}$
DMAC0

- Short address mode


Data transfer interrupt enable

| 0 | Interrupt requested by DTE bit is disabled |
| :---: | :--- |

1 Interrupt requested by DTE bit is enabled
An interrupt request is issued to the CPU when the DTE bit $=0$
Repeat enable

| RPE | DTIE | Description |
| :---: | :---: | :--- |
| 0 | 0 | I/O mode |
|  | 1 |  |
| 1 | 0 | Repeat mode |
|  | 1 | Idle mode |

Data transfer increment/decrement
0 Incremented: If DTSZ $=0$, MAR is incremented by 1 after each transfer If DTSZ $=1, M A R$ is incremented by 2 after each transfer
1 Decremented: If DTSZ $=0$, MAR is decremented by 1 after each transfer If DTSZ $=1, M A R$ is decremented by 2 after each transfer

## Data transfer size

| 0 | Byte-size transfer |
| :--- | :--- |
| 1 | Word-size transfer |

## Data transfer enable

| 0 | Data transfer is disabled |
| :--- | :--- |
| 1 | Data transfer is enabled |

## DTCR0B—Data Transfer Control Register 0B (cont)

$\mathbf{H}^{\prime} \mathbf{2 F}$
DMAC0

- Full address mode

| Bit | 7 |  | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | DTME | - | DAID | DAIDE | TMS | DTS2B | DTS1B | DTS0B |
| Initial valu | 0 |  | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Writ | - R/W |  | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Data transfer select 2B to 0B |  |  |  |  |  |  |  |  |
|  | Bit 2 | Bit 1 | Bit 0 | Data Transfer Activation Source |  |  |  |  |  |
|  | DTS2B | DTS1B | DTS0B | Normal M |  | Block Transfer Mode |  |  |  |
|  | 0 | 0 | 0 | Auto-requ (burst mod |  | Compare match/input capture A from ITU channel 0 |  |  |  |
|  |  |  | 1 | Not availa |  | Compare match/input capture A from ITU channel 1 |  |  |  |
|  |  | 1 | 0 | Auto-requ (cycle-ste | mode) | Compare match/input capture A from ITU channel 2 |  |  |  |
|  |  |  | 1 | Not availa |  | Compare match/input capture A from ITU channel 3 |  |  |  |
|  | 1 | 0 | 0 | Not availa |  | Not available |  |  |  |
|  |  |  | 1 | Not availa |  | Not available |  |  |  |
|  |  | 1 | 0 | Falling edg | of $\overline{\text { DREQ }}$ | Falling edge of $\overline{\text { DREQ }}$ |  |  |  |
|  |  |  | 1 | Low level input at $\overline{\text { DREQ }}$ |  | Not available |  |  |  |

Transfer mode select

| 0 | Destination is the block area in block transfer mode |
| :--- | :--- |
| 1 | Source is the block area in block transfer mode |

Destination address increment/decrement (bit 5)
Destination address increment/decrement enable (bit 4)
$\begin{array}{|c|c|l|}\hline \text { Bit } 5 & \text { Bit } 4 & \\ \hline \text { DAID } & \text { DAIDE } & \text { Increment/Decrement Enable } \\ \hline 0 & 0 & \text { MARB is held fixed } \\$\cline { 2 - 3 } \& 1 \& Incremented: If DTSZ$\left.=0, \text { MARB is incremented by } 1 \text { after each transfer } \\ \text { If DTSZ }=1, \text { MARB is incremented by } 2 \text { after each transfer }\end{array}\right]$

Data transfer master enable

| 0 | Data transfer is disabled |
| :--- | :--- |
| 1 | Data transfer is enabled |

## Bit

Initial value


Bit

Initial value


Read/Write


Note: Bit functions are the same as for DMAC0.

Bit

Initial value
Read/Write

Bit
Initial value
Read/Write

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Undetermined
R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W


Undetermined
R/W R/W R/W R/W R/W
ETCR1AH


Note: Bit functions are the same as for DMAC0.

IOAR1A-I/O Address Register 1A
H'36
DMAC1

Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |
| Undetermined |  |  |  |  |  |  |  |
| Initial value |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W

Note: Bit functions are the same as for DMAC0.

- Short address mode

Bit

Initial value
Read/Write

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Full address mode

Bit

|  | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for DMAC0.


Note: Bit functions are the same as for DMAC0.

Bit

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Initial value
Undetermined
Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for DMACO.
IOAR1B—I/O Address Register 1B H'3E DMAC1

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | Undetermined |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for DMACO.

- Short address mode

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DTE | DTSZ | DTID | RPE | DTIE | DTS2 | DTS1 | DTS0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Full address mode

Bit

Initial value
Read/Write

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DTME | - | DAID | DAIDE | TMS | DTS2B | DTS1B | DTS0B |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for DMAC0.

Bit Initial value* Read/Write


Flash write enable bit

| 0 | When a low level is input to the FWE pin (hardware protection state) |
| :--- | :--- |
| 1 | When a high level is input to the FWE pin |

Note: * The initial value is $\mathrm{H}^{\prime} 00$ in modes 5, 6, and 7 (on-chip flash memory enabled). In modes 1, 2, 3 , and 4 (on-chip flash memory disabled), this register cannot be modified and is always read as H'FF.


## Flash memory error

| 0 | Flash memory is operating normally <br> Flash memory program/erase protection (error protection) is disabled | (Initial value) |
| :--- | :--- | :--- |$|$| An error occurred during flash memory programming/erasing |
| :--- | :--- | :--- |
| Flash memory program/erase protection (error protection) is enabled |

Note: * The initial value is $\mathrm{H}^{\prime} 00$ in modes 5,6 , and 7 (on-chip flash memory enabled). In modes 1, 2 , 3 , and 4 (on-chip flash memory disabled), this register cannot be modified and is always read as H'FF.

Bit

|  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | EB7 | EB6 | EB5 | EB4 | EB3 | EB2 | EB1 | EB1 |
| Initial value* | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* |
|  |  |  | rase bl | speci | ation b |  |  |  |
|  |  |  | 0 Eras | rotectio | state |  |  |  |
|  |  |  | 1 Eras | e state |  |  |  |  |

Note: * The initial value is $\mathrm{H}^{\prime} 00$ in modes 5,6 and 7 (on-chip ROM enabled). In modes $1,2,3$, and 4 (on-chip ROM disabled), this register cannot be modified and is always read as H'FF.

EBR2—Erase Block Register 2 H'43 Flash memory

| Bit | 7 | 6 | 5 | 4 | 3 | 2 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | EB15 | EB14 | EB13 | EB12 | EB11 | EB10 | EB |
| Initial value* | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* | R/W |
|  |  | Erase block specification bits (2) |  |  |  |  |  |
|  |  |  | D Erase protection state |  |  |  |  |
|  |  |  | 1 Erasable state |  |  |  |  |

Note: * The initial value is $\mathrm{H}^{\prime} 00$ in modes 5,6 and 7 (on-chip ROM enabled). In modes $1,2,3$, and 4 (on-chip ROM disabled), this register cannot be modified and is always read as H'FF.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | RAMS | RAM2 | RAM1 | RAMO |
| Initial value* | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | R | R | R | R | R/W | R/W | R/W | R/W |

RAM select, RAM 2 to RAM 0

| Bit 3 | Bit 2 | Bit 1 | Bit 0 | RAM Area |
| :---: | :---: | :---: | :---: | :---: |
| RAMS | RAM 2 | RAM 1 | RAM 0 |  |
| 0 | 1/0 | 1/0 | 1/0 | H'FFE000 to H'FFEFFF |
| 1 | 0 | 0 | 0 | H'000000 to H'000FFF |
|  |  |  | 1 | H'001000 to H'001FFF |
|  |  | 1 | 0 | H'002000 to H'002FFF |
|  |  |  | 1 | H'003000 to H'003FFF |
|  | 1 | 0 | 0 | H'004000 to H'004FFF |
|  |  |  | 1 | H'005000 to H'005FFF |
|  |  | 1 | 0 | H'006000 to H'006FFF |
|  |  |  | 1 | H'007000 to H'007FFF |

Bit

Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | - | DASTE |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |

Read/Write

D/A standby enable

| 0 | D/A output is disabled in software standby mode |
| :--- | :--- |
| 1 | D/A output is enabled in software standby mode |

DIVCR—Division Control Register H'5D System control

| Bit | 7 | 6 | 5 | 7 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | DIV1 | DIVO |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 0 |
| Read/Write | - | - | - | - | - | - | R/W | R/W |

Divide 1 and 0

| Bit 1 | Bit 0 | Frequency |
| :---: | :---: | :--- |
| DIV1 | DIV0 | Division Ratio |
| 0 | 0 | $1 / 1$ (Initial value) |
|  | 1 | $1 / 2$ |
| 1 | 0 | $1 / 4$ |
|  | 1 | $1 / 8$ |

Bit
Initial value




Counter start 4

| 0 | TCNT4 is halted |
| :--- | :--- |
| 1 | TCNT4 is counting |



Bit
Initial value


Phase counting mode flag

| 0 | Channel 2 operates normally |
| :--- | :--- |
| 1 | Channel 2 operates in phase counting mode |



## Buffer mode B3

| 0 | GRB3 operates normally |
| :--- | :--- |
| 1 | GRB3 is buffered by BRB3 |

## Buffer mode A4

0 GRA4 operates normally
1 GRA4 is buffered by BRA4
Buffer mode B4

| 0 | GRB4 operates normally |
| :--- | :--- |
| 1 | GRB4 is buffered by BRB4 |

Combination mode 1 and 0

| Bit 5 | Bit 4 |  |
| :---: | :---: | :--- |
| CMD1 | CMD0 |  |
| 0 | 0 | Operating Mode of Channels 3 and 4 |
|  | 1 |  |
| 1 | 0 | Channels 3 and 4 operate normally |
|  | 1 | Channels 3 and 4 operate together in complementary PWM mode 4 operate together in reset-synchronized PWM mode |

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 |

Read/Write


| Bit 2 | Bit 1 | Bit 0 |  |
| :---: | :---: | :---: | :--- |
| TPSC2 | TPSC1 | TPSC0 | TCNT Clock Source |
| 0 | 0 | 0 | Internal clock: $\varnothing$ |
|  |  | 1 | Internal clock: $\varnothing / 2$ |
|  | 1 | 0 | Internal clock: $\varnothing / 4$ |
|  |  | 1 | Internal clock: $\varnothing / 8$ |
| 1 | 0 | 0 | External clock A: TCLKA input |
|  |  | 1 | External clock B: TCLKB input |
|  | 1 | 0 | External clock C: TCLKC input |
|  |  | 1 | External clock D: TCLKD input |

## Clock edge 1 and 0

| Bit 4 | Bit 3 |  |
| :---: | :---: | :--- |
| CKEG1 | CKEG0 | Counted Edges of External Clock |
| 0 | 0 | Rising edges counted |
|  | 1 | Falling edges counted |
| 1 | - | Both edges counted |

Counter clear 1 and 0

| Bit 6 | Bit 5 |  |
| :---: | :---: | :--- |
| CCLR1 | CCLR0 | TCNT Clear Source |
| 0 | 0 | TCNT is not cleared |
|  | 1 | TCNT is cleared by GRA compare match or input capture |
| 1 | 0 | TCNT is cleared by GRB compare match or input capture |
|  | 1 | Synchronous clear: TCNT is cleared in synchronization <br> with other synchronized timers |


| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOAO |
| Initial value | 1 |  | 0 | 00 |  | 1 | 0 | 0 | 0 |
| Read/Write |  | - | R/W | R/W | R/W | - | R/W | R/W | R/W |
|  | I/O control A2 to A0 |  |  |  |  |  |  |  |  |
|  | Bit 2 | Bit 1 | Bit 0 |  |  |  |  |  |  |
|  | IOA2 | IOA1 | IOAO | GRA Function |  |  |  |  |  |
|  | 0 | 0 | 0 | GRA is an output compare register |  | No output at compare match |  |  |  |
|  |  |  | 1 |  |  | 0 output at GRA compare match |  |  |  |
|  |  | 1 | 0 |  |  | 1 output at GRA compare match |  |  |  |
|  |  |  | 1 |  |  | Output toggles at GRA compare match |  |  |  |
|  | 1 | 0 | 0 | GRA is an input capture register |  | GRA captures rising edge of input |  |  |  |
|  |  |  | 1 |  |  | GRA captures falling edge of input |  |  |  |
|  |  | 1 | 0 |  |  | GRA captures both edges of input |  |  |  |
|  |  |  | 1 |  |  |  |  |  |  |

1/O control B2 to B0

| Bit 6 | Bit 5 | Bit 4 | GRB Function |  |
| :---: | :---: | :---: | :---: | :---: |
| IOB2 | IOB1 | IOBO |  |  |
| 0 | 0 | 0 | GRB is an output compare register | No output at compare match |
|  |  | 1 |  | 0 output at GRB compare match |
|  | 1 | 0 |  | 1 output at GRB compare match |
|  |  | 1 |  | Output toggles at GRB compare match |
| 1 | 0 | 0 | GRB is an input capture register | GRB captures rising edge of input |
|  |  | 1 |  | GRB captures falling edge of input |
|  | 1 | 0 |  | GRB captures both edges of input |
|  |  | 1 |  |  |



Overflow interrupt enable

| 0 | OVI interrupt requested by OVF flag is disabled |
| :--- | :--- |
| 1 |  |

1 OVI interrupt requested by OVF flag is enabled


Note: * Only 0 can be written, to clear the flag.

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Read/Write

> Up-counter

GRA0 H/L—General Register A0 H/L
H'6A, H'6B
ITU0

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | 0 |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Output compare or input capture register

GRB0 H/L—General Register B0 H/L H'6C, H'6D

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write
R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W
Output compare or input capture register
TCR1—Timer Control Register $1 \quad$ H'6E $\quad$ ITU1

| Bit | 7 | 6 | 5 | 4 | $3 \quad 2$ |  | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | CCLR1 | CCLRO | CKEG1 | CKEGO | TPSC2 | TPSC1 | TPSC0 |
| Initial value | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for ITUO.

Bit

|  | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Initial value | 1 | 0 | 0 | 0 | 1 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.
TIER1—Timer Interrupt Enable Register $1 \quad$ H'70 ITU1

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVIE | IMIEB | IMIEA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

## TSR1—Timer Status Register 1

H'71
ITU1

Bit

Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | OVF | IMFB | IMFA |
| 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| - | - | - | - | - | R/(W)* | R/(W)* | R/(W)* |

Notes: Bit functions are the same as for ITU0.

* Only 0 can be written, to clear the flag.


## TCNT1 H/L—Timer Counter 1 H/L

H'72, H'73
ITU1

Bit

Initial value
Read/Write

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU0.

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU0.

GRB1 H/L—General Register B1 H/L
H'76, H'77
ITU1

Bit

Initial value


Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU0.

TCR2—Timer Control Register 2
H'78
ITU2

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | CCLR1 | CCLRO | CKEG1 | CKEGO | TPSC2 | TPSC1 | TPSC0 |
| Initial value | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Notes: 1. Bit functions are the same as for ITU0.
2. When channel 2 is used in phase counting mode, the counter clock source selection by bits TPSC2 to TPSC0 is ignored.

Bit

|  | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Initial value | 1 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 |  |  |  |  |  |  |  |  |
| Read/Write | - | R/W | R/W | R/W | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

TIER2-Timer Interrupt Enable Register 2

$$
\mathrm{H}^{\prime} 7 \mathrm{~A}
$$

ITU2

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVIE | IMIEB | IMIEA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITUO.

TSR2—Timer Status Register 2
H'7B
ITU2

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVF | IMFB | IMFA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | R/(W)* | $\mathrm{R} / \mathrm{W})^{*}$ | R/(W)* |
| Note:* Only 0 can be written, to clear the flag. |  |  |  |  |  |  |  |  |
|  |  |  | rflow | flag |  | functio | is the sa | e as ITUO |
|  |  | 0 |  | [Clearing condition] <br> Read OVF when OVF $=1$, then write 0 in OVF. |  |  |  |  |
|  |  |  |  | [Setting condition] The TCNT value overflows (from H'FFFF to $\mathrm{H}^{\prime} 0000$ ) or underflows (from H'OOOO to H'FFFF) |  |  |  |  |

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Phase counting mode: up/down counter Other modes: up-counter

GRA2 H/L—General Register A2 H/L
H'7E, H'7F
ITU2

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITUO.

GRB2 H/L—General Register B2 H/L
H'80, H'81
ITU2
Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITUO.

Bit

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| - | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.
TIOR3—Timer I/O Control Register 3 H'83 ITU3

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 |
| Initial value | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

| TIER3—Timer Interrupt Enable Register 3 | H'84 | ITU3 |
| :--- | :--- | :--- |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVIE | IMIEB | IMIEA |
|  | - | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Initial value | 1 | 1 | - | - | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

| Bit | 7 | 6 |  | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - |  |  | - | - | OVF | IMFB | IMFA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | $\underline{R /(W) *}$ | $\mathrm{R} /(\mathrm{W})^{*} \mathrm{R} /(\mathrm{W})^{*}$ |  |
|  |  | Overflow flag |  |  |  |  | Bit functions are the same as for ITUO |  |
|  |  | 0 | [Clearing condition] |  |  |  |  |  |
|  |  |  | TCNT overflowed from H'FFFF to $\mathrm{H}^{\prime} 0000$ or underflowed from H'0000 to H'FFFF |  |  |  |  |  |

Note: * Only 0 can be written, to clear the flag.


Bit

Initial value
Read/Write

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W Output compare or input capture register (can be buffered)

BRA3 H/L—Buffer Register A3 H/L H'8C, H'8D ITU3

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write
R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Used in combination with GRA when buffer operation is selected
BRB3 H/L—Buffer Register B3 H/L H'8E, H'8F ITU3

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write $\quad$ R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W
Used in combination with GRB when buffer operation is selected


Bit
Initial value


Note:* When an external trigger occurs, bits 5 to 0 in TOER are cleared to 0 , disabling ITU output.

Bit

|  | - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Initial value | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 |  |  |  |  |  |  |  |  |
| Read/Write | - | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.
TIOR4—Timer I/O Control Register $4 \quad$ H'93 ITU4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | IOB2 | IOB1 | IOB0 | - | IOA2 | IOA1 | IOA0 |
| Initial value | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

## TIER4-Timer Interrupt Enable Register 4

H'94
ITU4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVIE | IMIEB | IMIEA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | R/W | R/W | R/W |

Note: Bit functions are the same as for ITU0.

## TSR4—Timer Status Register 4

H'95
ITU4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | OVF | IMFB | IMFA |
| Initial value | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | - | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ |

Notes: Bit functions are the same as for ITU0.

* Only 0 can be written, to clear the flag.

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU3.

GRA4 H/L—General Register A4 H/L H'98, H'99 ITU4

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU3.

GRB4 H/L—General Register B4 H/L
H'9A, H'9B
ITU4

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write
R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU3.

BRA4 H/L—Buffer Register A4 H/L
H'9C, H'9D
ITU4

Bit

Initial value
Read/Write

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU3.

Bit

Initial value

| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Read/Write R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W

Note: Bit functions are the same as for ITU3.

| TPMR—TPC Output Mode Register | H'A0 | TPC |
| :--- | :--- | :--- |



## Group 1 non-overlap

0 Normal TPC output in group 1
Output values change at compare match A in the selected ITU channel
1 Non-overlapping TPC output in group 1, controlled by compare match $A$ and $B$ in the selected ITU channel

## Group 2 non-overlap

0 Normal TPC output in group 2 Output values change at compare match A in the selected ITU channel
1 Non-overlapping TPC output in group 2, controlled by compare match $A$ and $B$ in the selected ITU channel

## Group 3 non-overlap

0 Normal TPC output in group 3
Output values change at compare match A in the selected ITU channel
1 Non-overlapping TPC output in group 3, controlled by compare match $A$ and $B$ in the selected ITU channel

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | G3CMS1 G3CMS0 |  | G2CMS1 | G2CMS0 | G1CMS1 | G1CMS0 | G0CMS1 | G0CMS0 |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| coup 0 compare match select 1 and 0 |  |  |  |  |  |  |  |  |
| Bit 1 | Bit 0 | ITU Channel Selected as Output Trigger |  |  |  |  |  |  |
| G0CMS1 | G0CMS0 I |  |  |  |  |  |  |  |
| 0 | 0 T | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare match in ITU channel 0 |  |  |  |  |  |  |
|  | 1 T | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare match in ITU channel 1 |  |  |  |  |  |  |
| 1 | 0 T | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare match in ITU channel 2 |  |  |  |  |  |  |
|  | 1 T | TPC output group $0\left(\mathrm{TP}_{3}\right.$ to $\left.\mathrm{TP}_{0}\right)$ is triggered by compare match in ITU channel 3 |  |  |  |  |  |  |

Group 1 compare match select 1 and 0

| Bit 3 | Bit 2 |  |
| :---: | :---: | :--- |
| G1CMS1 | G1CMS0 | ITU Channel Selected as Output Trigger |
| 0 | 0 | TPC output group $1\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare match in ITU channel 0 |
|  | 1 | TPC output group $1\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare match in ITU channel 1 |
| 1 | 0 | TPC output group $1\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare match in ITU channel 2 |
|  | 1 | TPC output group $1\left(\mathrm{TP}_{7}\right.$ to $\left.\mathrm{TP}_{4}\right)$ is triggered by compare match in ITU channel 3 |

Group 2 compare match select 1 and 0

| Bit 5 | Bit 4 |  |
| :---: | :---: | :--- |
| G2CMS1 | G2CMS0 | ITU Channel Selected as Output Trigger |
| 0 | 0 | TPC output group 2 $\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare match in ITU channel 0 |
|  | 1 | TPC output group 2 $\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare match in ITU channel 1 |
| 1 | 0 | TPC output group 2 $\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare match in ITU channel 2 |
|  | 1 | TPC output group 2 $\left(\mathrm{TP}_{11}\right.$ to $\left.\mathrm{TP}_{8}\right)$ is triggered by compare match in ITU channel 3 |

## Group 3 compare match select 1 and 0

| Bit 7 | Bit 6 |  |
| :---: | :---: | :--- |
| G3CMS1 | G3CMS0 | ITU Channel Selected as Output Trigger |
| 0 | 0 | TPC output group 3 $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare match in ITU channel 0 |
|  | 1 | TPC output group 3 $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare match in ITU channel 1 |
| 1 | 0 | TPC output group 3 $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare match in ITU channel 2 |
|  | 1 | TPC output group 3 $\left(\mathrm{TP}_{15}\right.$ to $\left.\mathrm{TP}_{12}\right)$ is triggered by compare match in ITU channel 3 |


NDERA—Next Data Enable Register A H'A3 TPC

| Bit | 7 6 |  | 5 | 43 |  | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDER7 | NDER6 | NDER5 | NDER4 | NDER3 | NDER2 | NDER1 | NDER0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Next data enable 7 to 0 |  |  |  |  |  |  |  |
|  |  | 7 to 0 | Description |  |  |  |  |  |
|  | NDER7 | to NDER |  |  |  |  |  |  |
|  |  | 0 | TPC outputs $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$ are disabled (NDR7 to NDR0 are not transferred to $\mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$ ) |  |  |  |  |  |
|  |  | 1 | TPC outputs $\mathrm{TP}_{7}$ to $\mathrm{TP}_{0}$ are enabled (NDR7 to NDR0 are transferred to $\mathrm{PA}_{7}$ to $\mathrm{PA}_{0}$ ) |  |  |  |  |  |

- Same trigger for TPC output groups 2 and 3
- Address H'FFA4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR15 | NDR14 | NDR13 | NDR12 | NDR11 | NDR10 | NDR9 | NDR8 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Store the next output data for TPC output group 3 |  |  |  | Store the next output data for TPC output group 2 |  |  |  |

- Address H'FFA6

Bit

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | - | - |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | - | - | - | - | - | - | - | - |

- Different triggers for TPC output groups 2 and 3
— Address H'FFA4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR15 | NDR14 | NDR13 | NDR12 | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | - | - | - | - |

Store the next output data for
TPC output group 3

- Address H'FFA6

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | NDR11 | NDR10 | NDR9 | NDR8 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
|  |  |  |  |  | Store the next output data for TPC output group 2 |  |  |  |

- Same trigger for TPC output groups 0 and 1
- Address H'FFA5

Bit

|  | NDR7 | NDR6 | NDR5 | NDR4 | NDR3 | NDR2 | NDR1 | NDR0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | Store the next output data for TPC output group 1 |  |  |  | Store the next output data for TPC output group 0 |  |  |  |

- Address H'FFA7

Bit

| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Read/Write | - | - | - | - | - | - | - | - |

- Different triggers for TPC output groups 0 and 1
- Address H'FFA5

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NDR7 | NDR6 | NDR5 | NDR4 | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | - | - | - | - |
|  | Store the next output data for TPC output group 1 |  |  |  |  |  |  |  |

— Address H'FFA7

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | NDR3 | NDR2 | NDR1 | NDR0 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |

Store the next output data for TPC output group 0


## Overflow flag

0 [Clearing condition] Read OVF when OVF $=1$, then write 0 in OVF
1 [Setting condition] TCNT changes from H'FF to $\mathrm{H}^{\prime} \mathrm{OO}$

Note: * Only 0 can be written, to clear the flag.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |


| RSTCSR—Reset Control/Status Register | H'AB (read), <br> H'AA (write) | WDT |
| :--- | :--- | :--- |


| Bit | 7 |  | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | RST | - | - | - | - | - | - | - |
| Initial value | 0 |  | 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | $\mathrm{R} /(\mathrm{W})^{* 1}$ |  | -*2 | - | - | - | - | - | - |
|  |  | chdo | mer re |  |  |  |  |  |  |
|  | 0 | [Clea <br> - Res <br> - When | cond <br> signal WRST | $\begin{aligned} & \text { n] } \\ & \text { ut at } \\ & \text { 1", wit } \end{aligned}$ | pin "aft |  |  |  |  |
|  | 1 | [Settin <br> TCN | conditio verflow | nera | ese |  |  |  |  |

Notes: 1. Only 0 can be written in bit 7 , to clear the flag.
2. Bit 6 must not be set to 1 ; in a write, 0 must always be written in this bit.

Bit
Initial value


Address multiplex mode select

| 0 | 8-bit column mode |
| :--- | :--- |
| 1 | 9 -bit column mode |

## Strobe mode select

| 0 | $2 \overline{\mathrm{WE}}$ mode |
| :--- | :--- |
| 1 | $2 \overline{\mathrm{CAS}}$ mode |

PSRAM enable, DRAM enable

| Bit 6 | Bit 5 |  |
| :---: | :---: | :--- |
| PSRAME | DRAME | RAM Interface |
| 0 | 0 | Can be used as an interval timer <br> (DRAM and PSRAM cannot be <br> directly connected) |
| 1 | 1 | DRAM can be directly connected |
|  | 0 | PSRAM can be directly connected |
|  | 1 | Illegal setting |

Self-refresh mode

| 0 | DRAM or PSRAM self-refresh is disabled in software standby mode |
| :--- | :--- |
| 1 | DRAM or PSRAM self-refresh is enabled in software standby mode |

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CMF | CMIE | CKS2 C | CKS1 | CKSO | - | - |
| 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| R/(W)* | R/W | R/W | R/W | R/W | - | - |
|  |  | Clock | select | to 0 |  |  |
|  |  | Bit 5 | Bit 4 | Bit 3 |  |  |
|  |  | CKS2 | 2 CKS1 | CKSO | Counter | ck Source |
|  |  | 0 | 0 | 0 | Clock inp | s disabled |
|  |  |  |  | 1 | ¢/2 |  |
|  |  |  | 1 | 0 | ø/8 |  |
|  |  |  |  | 1 | ø/32 |  |
|  |  | 1 | 0 | 0 | ø/128 |  |
|  |  |  |  | 1 | ø/512 |  |
|  |  |  | 1 | 0 | ब/2048 |  |
|  |  |  |  | 1 | ø/4096 |  |

Compare match interrupt enable

| 0 | The CMI interrupt requested by CMF is disabled |
| :---: | :--- |
| 1 | The CMI interrupt requested by CMF is enabled |

Compare match flag

| 0 | [Clearing condition $]$ <br> Read CMF when CMF $=1$, then write 0 in CMF |
| :--- | :--- |
| 1 | $[$ Setting condition $]$ <br> RTCNT $=$ RTCOR |

Note: * Only 0 can be written, to clear the flag.

Bit

Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Count value |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |

RTCOR—Refresh Time Constant Register H'AF Refresh controller





## Transmit-end interrupt enable

| 0 | Transmit-end interrupt requests (TEI) are disabled |
| :---: | :---: |
| 1 | Transmit-end interrupt requests (TEI) are enabled |

## Multiprocessor interrupt enable

| 0 | Multiprocessor interrupts are disabled (normal receive operation) |
| :--- | :--- |
|  |  |

1 Multiprocessor interrupts are enabled

## Transmit enable

| 0 | Transmitting is disabled |
| :--- | :--- |
| 1 | Transmitting is enabled |

## Receive enable

| 0 | Receiving is disabled |
| :--- | :--- |
| 1 | Receiving is enabled |

## Receive interrupt enable

| 0 | Receive-data-full (RXI) and receive-error (ERI) interrupt requests are disabled |
| :--- | :--- |
| 1 | Receive-data-full (RXI) and receive-error (ERI) interrupt requests are enabled |

Transmit interrupt enable

| 0 | Transmit-data-empty interrupt request (TXI) is disabled |
| :---: | :---: |
| 1 | Transmit-data-empty interrupt request (TXI) is enabled |




Note: * Only 0 can be written, to clear the flag.

|  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Bit |  |  |  |  |  |  |  |
|  | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Initial value | 0 | $R$ | $R$ | $R$ | $R$ | $R$ | $R$ | $R$ |


| SCMR—Smart Card Mode Register | H'B6 | SCI0 |
| :--- | :---: | :---: |

Bit

Initial value


Smart card interface mode select

| 0 | Smart card interface function is disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | Smart card interface function is enabled |  |

Smart card data invert
0 Unmodified TDR contents are transmitted (Initial value) Received data is stored unmodified in RDR
1 Inverted TDR contents are transmitted
Received data are inverted before storage in RDR

## Smart card data transfer direction

| 0 | TDR contents are transmitted LSB-first <br> Received data is stored LSB-first in RDR | (Initial value) |
| :--- | :--- | :--- |
| 1 | TDR contents are transmitted MSB-first <br> Received data is stored MSB-first in RDR |  |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | C/ $\bar{A}$ | CHR | PE | O/E | STOP | MP | CKS1 | CKSO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for SCIO.
BRR—Bit Rate Register H'B9 SCI1

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for SCIO.

| SCR—Serial Control Register |  |  |  |  | H'BA |  |  | SCI1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
|  | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKEO |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for SCIO.

Bit

|  |  |  |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Initial value | 1 |  |  |  |  |  |  |  |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: Bit functions are the same as for SCIO.

| SSR-Serial Status Register |  |  |  |  | H'BC |  |  | SCI1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 0 |  |
|  | TDRE | RDRF | ORER | FER | PER | TEND | MPB | MPBT |
| Initial value | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| Read/Write | R/(W)* | R/(W)* | R/(W)* | $\mathrm{R} / \mathrm{W})^{*}$ | R/(W)* | R | R | R/W |

Notes: Bit functions are the same as for SCIO.

* Only 0 can be written, to clear the flag.


## RDR—Receive Data Register

H'BD
SCI1

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R | R | R | R | R | R | R |

Note: Bit functions are the same as for SCIO.


| P2DDR—Port 2 Data Direction Register |  |  |  |  |  | H'C1 |  |  | Port 2 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  | $\mathrm{P} 2{ }_{7}$ DDR | $\mathrm{P} 26_{6}$ DDR | $\mathrm{P} 2_{5}$ DDR | P 24 DDR | $\mathrm{P} 2_{3}$ DDR | $\mathrm{P} 22_{2}$ DRR | $\mathrm{P} 2{ }_{1}$ DDR | $\mathrm{P} 2{ }_{0} \mathrm{DDR}$ |
| Modes <br> 1 to 4 | Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
|  | Read/Write | - | - | - | - | - | - | - | - |
| Modes 5 to 7 | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
|  | Read/Write | W | W | W | W | W | W | W | W |
|  |  |  |  | Port 2 input/output select |  |  |  |  |  |
|  |  |  |  | 0 | Generic input pin |  |  |  |  |
|  |  |  |  | 1 | Generic output pin |  |  |  |  |
| P1DR—Port 1 Data Register |  |  |  |  | H'C2 |  |  |  | Port 1 |
| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  | P 17 | P16 | P 15 | P 14 | P 13 | P 12 | $\mathrm{P} 1{ }_{1}$ | P 10 |
| Initia | value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read | /Write P/ | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  | Data for port 1 pins |  |  |  |  |  |  |  |

Bit

|  | $\mathrm{P} 2_{7}$ | $\mathrm{P} 2_{6}$ | $\mathrm{P} 2_{5}$ | $\mathrm{P} 2_{4}$ | $\mathrm{P} 2_{3}$ | $\mathrm{P} 2_{2}$ | $\mathrm{P} 2_{1}$ | $\mathrm{P} 2_{0}$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |  |

$$
\text { Data for port } 2 \text { pins }
$$

P3DDR—Port 3 Data Direction Register H'C4 Port 3

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P3 ${ }_{7}$ DDR | P3 ${ }_{6}$ DDR | P3 ${ }_{5}$ DDR | P3 ${ }_{4}$ DDR | P3 ${ }_{3}$ DDR | $\mathrm{P} 3{ }_{2} \mathrm{DDR}$ | P3 ${ }_{1}$ DDR | $\mathrm{P} 3_{0}$ DDR |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | W | W | W | W | W | W | W | W |

Port 3 input/output select

| 0 | Generic input pin |
| :--- | :--- |
| 1 | Generic output pin |

P4DDR—Port 4 Data Direction Register
H'C5
Port 4


Bit

|  | $\mathrm{P} 3_{7}$ | $\mathrm{P}_{6}$ | $\mathrm{P} 3_{5}$ | $\mathrm{P}_{4}$ | $\mathrm{P} 3_{3}$ | $\mathrm{P} 3_{2}$ | $\mathrm{P} 3_{1}$ | $\mathrm{P} 3_{0}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Data for port 3 pins
P4DR—Port 4 Data Register
$\mathbf{H}^{\prime} \mathrm{C} 7$
Port 4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{P}_{7}$ | P46 | P45 | P 44 | P4 3 | P 42 | P4 1 | P40 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

P5DDR—Port 5 Data Direction Register H'C8 Port 5

| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | - | - | - | - | P53 DDR | $\mathrm{P} 5{ }_{2} \mathrm{DDR}$ | P5 ${ }_{1}$ DDR | $\mathrm{P} 5{ }_{0}$ DDR |
| Modes 1 to 4 | Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
|  | Read/Write | - | - | - | - | - | - | - | - |
| Modes 5 to 7 | Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
|  | Read/Write | - | - | - | - | W | W | W | W |

Port 5 input/output select

| 0 | Generic input pin |
| :--- | :--- |
| 1 | Generic output pin |

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | $\mathrm{P} 6_{6} \mathrm{DDR}$ | $\mathrm{P} 6_{5} \mathrm{DDR}$ | $\mathrm{P} 6_{4} \mathrm{DDR}$ | $\mathrm{P} 6_{3} \mathrm{DDR}$ | $\mathrm{P} 6_{2} \mathrm{DDR}$ | $\mathrm{P} 6{ }_{1} \mathrm{DDR}$ | $\mathrm{P} 6_{0} \mathrm{DDR}$ |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| - | W | W | W | W | W | W | W |

Port 6 input/output select

| 0 | Generic input pin |
| :---: | :--- |
| 1 | Generic output pin |


| P5DR—Port 5 Data Register |  |  |  |  | H'CA |  |  | Port 5$0$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
|  | - | - | - | - | P53 | P52 | P5 ${ }_{1}$ | P50 |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
|  |  |  |  |  |  | ta fo | 5 pin |  |

P6DR—Port 6 Data Register H'CB Port 6

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | P66 | P65 | P64 | $\mathrm{P6}_{3}$ | P6 2 | P6 1 | P60 |
| Initial value | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | R/W | R/W | R/W | R/W | R/W | R/W | R/W |



| P7DR—Port 7 Data Register |  |  |  |  | H'CE |  |  | Port 7 <br> 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
|  | P 77 | P 76 | P 75 | P 74 | $\mathrm{P}_{3}$ | P 72 | P7 ${ }_{1}$ | P 70 |
| Initial value | —* | —* | —* | —* | -* | —* | —* | —* |
| Read/Write | R | R | R | R | R | R | R | R |
|  | Read the pin levels for port 7 |  |  |  |  |  |  |  |

Note: * Determined by pins $\mathrm{P} 7_{7}$ to $\mathrm{P} 7_{0}$.
P8DR—Port 8 Data Register $\quad$ H'CF $\quad$ Port 8

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | P84 | P83 | P82 | P8 ${ }_{1}$ | P80 |
| Initial value | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | R/W | R/W | R/W | R/W | R/W |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | P9 ${ }_{5}$ DDR | $\mathrm{P9}{ }_{4} \mathrm{DDR}$ | P9 ${ }_{3}$ DDR | P9 ${ }_{2}$ DDR | P9 ${ }_{1}$ DDR | P9 ${ }_{0}$ DDR |
| Initial value | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | W | W | W | W | W | W |
|  |  |  |  | Port 9 input/output select |  |  |  |  |
|  |  |  |  | 0 Gen | eric input pin |  |  |  |
|  |  |  |  | 1 Gen | eric output | pin |  |  |


| PADDR-Port A Data Direction Register |  |  |  |  |  | H'D1 |  |  | Port A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  | $\mathrm{PA}_{7} \mathrm{DDR}$ | $\mathrm{PA}_{6} \mathrm{DDR}$ | PA ${ }_{5}$ DDR | $\mathrm{PA}_{4} \mathrm{DDR}$ | $\mathrm{PA}_{3} \mathrm{DDR}$ | PA ${ }_{2}$ DDR | $\mathrm{PA}_{1}$ DDR | PA ${ }_{0}$ DDR |
| Modes | Initial value | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 3,4,6 | Read/Write | - | W | W | W | W | W | W | W |
| Modes | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| $\begin{aligned} & 1,2, \\ & 5,7 \end{aligned}$ | Read/Write | W | W | W | W | W | W | W | W |
| Port A input/output select |  |  |  |  |  |  |  |  |  |
|  |  |  |  | 0 | Generic in | input pin |  |  |  |
|  |  |  |  | 1 | Generic | output pin |  |  |  |


| P9DR-Port 9 Data Register |  |  |  |  | H'D2 |  |  | Port 9 <br> 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
|  | - | - | $\mathrm{P9}_{5}$ | $\mathrm{P9}_{4}$ | $\mathrm{P9}_{3}$ | $\mathrm{P9}_{2}$ | P9 ${ }_{1}$ | P9 0 |
| Initial value | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  |  | Data for | rt 9 pin |  |  |

Bit


| PBDDR—Port B Data Direction Register | H'D4 | Port B |
| :--- | :--- | :--- |



| PBDR—Port B Data Register |  |  |  |  | H'D6 |  |  | Port B$0$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
|  | $\mathrm{PB}_{7}$ | $\mathrm{PB}_{6}$ | $\mathrm{PB}_{5}$ | $\mathrm{PB}_{4}$ | $\mathrm{PB}_{3}$ | $\mathrm{PB}_{2}$ | $\mathrm{PB}_{1}$ | $\mathrm{PB}_{0}$ |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Bit

|  | $\mathrm{P} 2{ }_{7} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{6} \mathrm{PCR}$ | $\mathrm{P} 2_{5} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{4} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{3} \mathrm{PCR}$ | $\mathrm{P} 2_{2} \mathrm{PCR}$ | $\mathrm{P} 2_{1} \mathrm{PCR}$ | $\mathrm{P} 2{ }_{0} \mathrm{PCR}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

## Port 2 input pull-up MOS control 7 to 0

| 0 | Input pull-up transistor is off |
| :--- | :--- |
| 1 | Input pull-up transistor is on |

Note: Valid when the corresponding P2DDR bit is cleared to 0 (designating generic input).
P4PCR—Port 4 Input Pull-Up MOS Control Register H'DA Port 4

| Bit | 7 | 6 | 5 | 4 | 3 | 2 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P 47 PCR | $\mathrm{P} 46{ }_{6} \mathrm{PCR}$ | P 45 PCR | $\mathrm{P} 44{ }_{4} \mathrm{PCR}$ | $\mathrm{P} 43{ }_{3} \mathrm{PCR}$ | P 42 PCR | P 41 P |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/V |
|  |  |  | Port 4 input pull-up MOS control 7 to 0 |  |  |  |  |
|  |  |  | 0 Input | ut pull-up | transistor | is off |  |
|  |  |  | 1 Input | ut pull-up | transistor | is on |  |

Note: Valid when the corresponding P4DDR bit is cleared to 0 (designating generic input).

Bit

|  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | P5 ${ }_{3} \mathrm{PCR}$ | P 52 PCR | P5 ${ }_{1} \mathrm{PCR}$ | P50PCR |
| Initial value | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| Read/Write | - | - | - | - | R/W | R/W | R/W | R/W |
|  |  |  |  |  | Port 5 inpu | t pull-up | IOS cont | ol 3 to 0 |
|  |  |  |  |  | 0 Input pu | ull-up trans | sistor is off |  |
|  |  |  |  |  | 1 Input pu | ull-up trans | sistor is on |  |

Note: Valid when the corresponding P5DDR bit is cleared to 0 (designating generic input).

| DADR0-D/A Data Register 0 |  |  |  | H'DC |  |  |  | D/A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| D/A conversion data |  |  |  |  |  |  |  |  |
| DADR1-D/A Data Register 1 |  |  | 5 | H'DD |  |  |  | D/A |
| Bit | 7 | 6 |  | 4 | 3 | 2 | 1 | 0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  | A con | ion da |  |  |  |



## D/A output enable 0

| 0 | $\mathrm{DA}_{0}$ analog output is disabled |
| :--- | :--- |

1 Channel-0 D/A conversion and $\mathrm{DA}_{0}$ analog output are enabled

## D/A output enable 1

| 0 | $\mathrm{DA}_{1}$ analog output is disabled |
| :--- | :--- |
| 1 | Channel-1 D/A conversion and $\mathrm{DA}_{1}$ analog output are enabled |

ADDRA H/L—A/D Data Register A H/L H'E0, H'E1 A/D

| Bit | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 | - | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

$\begin{array}{llllllllllllllllllllll}\text { Read/Write } & R & R & R & R & R & R & R & R & R & R & R & R & R & R & R & R\end{array}$
$\underbrace{R}_{\text {ADDRAH }} \underbrace{R}_{\text {ADDRAL }}$

## A/D conversion data

10-bit data giving an
A/D conversion result

| Bit | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 | - | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R |
|  | ADDRBH |  |  |  |  |  |  |  |  |  | ADDRBL |  |  |  |  |  |
| A/D conversion data 10-bit data giving an A/D conversion result |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

ADDRC H/L—A/D Data Register C H/L H'E4, H'E5 A/D

| Bit | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 | - | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R |
|  | ADDRCH |  |  |  |  |  |  |  |  |  | ADDRCL |  |  |  |  |  |
| A/D conversion data 10-bit data giving an A/D conversion result |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

ADDRD H/L—A/D Data Register D H/L H'E6, H'E7 A/D

| Bit | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | ADO | - | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R |
|  | ADDRDH |  |  |  |  |  |  |  |  |  | ADDRDL |  |  |  |  |  |
| A/D conversion data 10-bit data giving an A/D conversion result |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |



## A/D interrupt enable

| 0 | A/D end interrupt request is disabled |
| :--- | :--- |
| 1 | A/D end interrupt request is enabled |

## A/D end flag

| 0 | [Clearing condition] <br> Read ADF while ADF $=1$, then write 0 in ADF |
| :--- | :--- |
| 1 | [Se |

1 [Setting conditions]
Single mode: A/D conversion ends
Scan mode: A/D conversion ends in all selected channels
Note: * Only 0 can be written, to clear flag.


Note: * Bit 0 must not be set to 1 ; in a write, 0 must always be written in this bit.

| ABWCR-Bus Width Control Register |  |  |  |  |  | H'EC |  | Bus controller |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit |  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  |  | ABW7 | ABW6 | ABW5 | ABW4 | ABW3 | ABW2 | ABW1 | ABW0 |
| Initial value | Mode 1,3,5,6 | 61 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
|  | Mode 2,4,7 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write |  | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Area 7 to 0 bus width control

| Bits 7 to 0 |  |
| :---: | :--- |
| ABW7 to ABW0 | Bus Width of Access Area |
| 0 | Areas 7 to 0 are 16 -bit access areas |
| 1 | Areas 7 to 0 are 8 -bit access areas |


| ASTCR—Access State Control Register |  |  |  |  | H'ED |  | Bus controller |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | AST7 | AST6 | AST5 | AST4 | AST3 | AST2 | AST1 | AST0 |
| Initial value | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  | ea 7 to | access | tate con |  |  |  |  |  |
|  | Bits |  |  |  |  |  |  |  |
|  | AST7 | ASTO | Number | States | Access | ycle |  |  |
|  |  |  | Areas 7 | 0 are t | -state | ess are |  |  |
|  |  |  | Areas 7 | 0 are t | e-state | ccess ar |  |  |



Bit

Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | MDS2 | MDS1 | MDS0 |
| 1 | 1 | 0 | 0 | 0 | $—^{*}$ | $—^{*}$ | $—^{*}$ |
| - | - | - | - | - | R | R | R |


| Bit 2 | Bit 1 | Bit 0 |  |
| :---: | :---: | :---: | :--- |
| $\mathrm{MD}_{2}$ | $\mathrm{MD}_{1}$ | $\mathrm{MD}_{0}$ | Operating mode |
| 0 | 0 | 0 | - |
|  |  | 1 | Mode 1 |
|  | 1 | 0 | Mode 2 |
|  |  | 1 | Mode 3 |
|  |  | 0 | Mode 4 |
|  |  | 1 | Mode 5 |
|  | 1 | 0 | Mode 6 |
|  |  | 1 | Mode 7 |

Note: * Determined by the state of the mode pins $\left(\mathrm{MD}_{2}\right.$ to $\left.\mathrm{MD}_{0}\right)$.

Bit
Initial value

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SSBY | STS2 | STS1 | STS0 | UE | NMIEG | - | RAME |
| 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |
| R/W | R/W | R/W | R/W | R/W | R/W | - | R/W |
|  |  |  |  |  | RAM en |  |  |
|  |  |  |  |  | 0 On-ch | RA | disabled |
|  |  |  |  |  | 1 On- | RA | enabled |
|  |  |  | dge se |  |  |  |  |
|  |  | 0 | interru | is req | ted at th | lling | e of NMI |
|  |  | 1 | interru | is req | ted at th | ing | e of NMI |
|  |  | User | enable |  |  |  |  |
|  |  | 0 | bit 6 (U) | is use | an inte | ma |  |
|  |  | 1 C | bit 6 (U) | is use | a user |  |  |

Standby timer select 2 to 0

| Bit 6 | Bit 5 | Bit 4 |  |
| :---: | :---: | :---: | :--- |
| STS2 | STS1 | STS0 | Standby Timer |
| 0 | 0 | 0 | Waiting time $=8,192$ states |
|  |  | 1 | Waiting time $=16,384$ states |
|  | 1 | 0 | Waiting time $=32,768$ states |
|  |  | 1 | Waiting time $=65,536$ states |
| 1 | 0 | 0 | Waiting time $=131,072$ states |
|  |  | 1 | Waiting time $=1,024$ states |
|  | 1 | - | Illegal setting |

Software standby

| 0 | SLEEP instruction causes transition to sleep mode |
| :--- | :--- |
| 1 | SLEEP instruction causes transition to software standby mode |



## Address 23 to 21 enable

| 0 | Address output |
| :--- | :--- |
| 1 | Other input/output |

## ISCR—IRQ Sense Control Register

H'F4 Interrupt controller

| Bit | 7 |  | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | IRQ5SC | IRQ4SC | IRQ3SC | IRQ2SC | IRQ1SC | IRQ0SC |  |
|  | Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |  |

## $\mathrm{IRQ}_{5}$ to $\mathrm{IRQ}_{\mathbf{0}}$ sense control

| 0 | Interrupts are requested when $\overline{\mathrm{RQ}_{5}}$ to $\overline{\mathrm{IQ}_{0}}$ inputs are low |
| :--- | :--- |
| 1 | Interrupts are requested by falling-edge input at $\overline{\mathrm{RQ}}{ }_{5}$ to $\overline{\mathrm{IQ}_{0}}$ |


| IER—IRQ Enable Register | H'F5 | Interrupt controller |
| :--- | :--- | :--- |


| Bit | 7 | 6 | 5 | 4 | 3 | 2 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | IRQ5E | IRQ4E | IRQ3E | IRQ2E | IRQ1E |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/(W) | R/(W) | R/(W) | R/(W) | R/(W) | R/(W) | R/(W) |
|  |  |  | $\mathrm{IRQ} \mathrm{S}_{5}$ to $\mathrm{IRQ} \mathrm{Q}_{0}$ enable |  |  |  |  |
|  |  |  |  | $I R Q_{5}$ to $I R Q_{0}$ interrupts are disabled |  |  |  |
|  |  |  |  | $\mathrm{IRQ}_{5}$ to $\mathrm{IRQ} \mathrm{O}_{0}$ interrupts are enabled |  |  |  |

Bit
Initial value


Note: * Only 0 can be written, to clear the flag.

IPRA—Interrupt Priority Register A
H'F8 Interrupt controller

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | IPRA7 | IPRA6 | IPRA5 | IPRA4 | IPRA3 | IPRA2 | IPRA1 | IPRA0 |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Read/Write | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  | Priority level A7 to A0 |  |  |  |  |  |
|  |  |  |  | Priority | vel 0 (low | priority) |  |  |
|  |  |  |  | Priority | vel 1 (high | priority) |  |  |

- Interrupt sources controlled by each bit

Bit 7: Bit 6: Bit 5: Bit 4: Bit 3: Bit 2: Bit 1: Bit 0:

|  | IPRA7 | IPRA6 | IPRA5 | IPRA4 | IPRA3 | IPRA2 | IPRA1 | IPRA0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Interrupt | $I R Q_{0}$ | $I R Q_{1}$ | $I \mathrm{IRQ}_{2}$, | $\mathrm{IRQ}_{4}$, | WDT, | ITU | ITU | ITU |
| source |  |  | $I R Q_{3}$ | $\mathrm{IRQ}_{5}$ | Refresh | channel | channel | channel |
|  |  |  |  |  |  | Controller | 0 | 1 |

IPRB—Interrupt Priority Register B
H'F9 Interrupt controller


- Interrupt sources controlled by each bit

|  | Bit 7: | Bit 6: | Bit 5: | Bit 4: | Bit 3: <br> IPRB7 | Bit 2: | Bit 1: | Bit 0: |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | IPRB6 | IPRB5 | - | IPRB3 | IPRB2 | IPRB1 | - |  |
| Interrupt | ITU | ITU | DMAC | - |  | SCI | SCI | A/D |
| source | channel |  |  |  |  |  |  |  |
|  | 3 | channel |  |  | channel | channel | converter |  |
|  | 4 |  |  | 0 | 1 |  |  |  |

## Appendix C I/O Port Block Diagrams

## C. 1 Port 1 Block Diagram



Figure C. 1 Port 1 Block Diagram

## C. 2 Port 2 Block Diagram



Figure C. 2 Port 2 Block Diagram

## C. 3 Port 3 Block Diagram



Figure C. 3 Port 3 Block Diagram

## C. 4 Port 4 Block Diagram



Figure C. 4 Port 4 Block Diagram

## C. 5 Port 5 Block Diagram



Figure C. 5 Port 5 Block Diagram

## C. 6 Port 6 Block Diagrams



Figure C. 6 (a) Port 6 Block Diagram (Pin P6 ${ }_{0}$ )


WP6D: Write to P6DDR
WP6: Write to port 6
RP6: Read port 6

Figure C. 6 (b) Port 6 Block Diagram (Pin P6 $\mathbf{I}_{1}$ )


Figure C. 6 (c) Port 6 Block Diagram (Pin P6 $\mathbf{F}_{2}$ )


WP6D: Write to P6DDR
WP6: Write to port 6
RP6: Read port 6
$\mathrm{n}=6$ to 3

Figure C. 6 (d) Port 6 Block Diagram (Pins P6 $_{6}$ to $\mathbf{P 6}_{3}$ )

## C. 7 Port 7 Block Diagrams



Figure C. 7 (a) Port 7 Block Diagram (Pins P7 to P7 ${ }_{5}$ )


RP7: Read port 7
$\mathrm{n}=6$ or 7

Figure C. 7 (b) Port 7 Block Diagram (Pins P7 ${ }_{6}$, P7 ${ }_{7}$ )


Figure C. 8 (a) Port 8 Block Diagram (Pin P8 ${ }_{0}$ )


Figure C. 8 (b) Port 8 Block Diagram (Pins $\mathrm{P8}_{1}$ to $\mathrm{P8}_{\mathbf{3}}$ )


Figure C. 8 (c) Port 8 Block Diagram (Pin P84)

## C. 9 Port 9 Block Diagrams



WP9D: Write to P9DDR
WP9: Write to port 9
RP9: Read port 9

Figure C. 9 (a) Port 9 Block Diagram (Pin P9 ${ }_{0}$ )


WP9D: Write to P9DDR
WP9: Write to port 9
RP9: Read port 9

Figure C. 9 (b) Port 9 Block Diagram (Pin P9 ${ }_{1}$ )


Figure C. 9 (c) Port 9 Block Diagram (Pins P9 ${ }_{2}, \mathrm{P9}_{3}$ )


Figure C. 9 (d) Port 9 Block Diagram (Pins $\mathbf{P 9}_{4}, \mathrm{P9}_{5}$ )

## C. 10 Port A Block Diagrams



Figure C. 10 (a) Port A Block Diagram (Pins $\mathbf{P A}_{\mathbf{0}}, \mathbf{P A}_{1}$ )


Figure C. 10 (b) Port A Block Diagram ( Pins $^{\mathbf{P A}} \mathbf{2}_{2}, \mathbf{P A}_{3}$ )


Figure C. 10 (c) Port A Block Diagram (Pins PA $_{4}$ to PA $_{6}$ )


Figure C. 10 (d) Port A Block Diagram (Pin PA $\mathbf{P}_{7}$ )

## C. 11 Port B Block Diagrams



Figure C. 11 (a) Port B Block Diagram (Pins $\mathbf{P B}_{0}$ to $\mathbf{P B}_{3}$ )


WPBD: Write to PBDDR
WPB: Write to port B
RPB: Read port B
$\mathrm{n}=4$ or 5

Figure C. 11 (b) Port B Block Diagram (Pins PB $_{4}$, PB $_{5}$ )


Figure C. 11 (c) Port B Block Diagram (Pin PB ${ }_{6}$ )


Figure C. 11 (d) Port B Block Diagram (Pin PB ${ }_{7}$ )

## Appendix D Pin States

## D. 1 Port States in Each Mode

Table D. 1 Port States

| Pin Name | Mode |  | Reset | Hardware Standby Mode | Software Standby Mode | Bus- <br> Released Mode | Program <br> Execution, <br> Sleep Mode |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\phi$ | - |  | Clock output | T | H | Clock output | Clock output |
| P1, to P1 | 1 to 4 |  | L | T | T | T | $\mathrm{A}_{7}$ to $\mathrm{A}_{0}$ |
|  | 5, 6 |  | T | T | keep | T | Input port (DDR = 0) |
|  |  |  |  |  | T | T | $\begin{aligned} & A_{7} \text { to } A_{0} \\ & (D D R=1) \end{aligned}$ |
|  | 7 |  | T | T | keep | - | I/O port |
| P2 ${ }_{7}$ to $\mathrm{P} 2_{0}$ | 1 to 4 |  | L | T | T | T | $\mathrm{A}_{15}$ to $\mathrm{A}_{8}$ |
|  | 5, 6 |  | T | T | keep | T | Input port $(\mathrm{DDR}=0)$ |
|  |  |  |  |  | T | T | $\begin{aligned} & A_{15} \text { to } A_{8} \\ & (D D R=1) \end{aligned}$ |
|  | 7 |  | T | T | keep | - | I/O port |
| $\mathrm{P}_{7}$ to $\mathrm{P} 3_{0}$ | 1 to 6 |  | T | T | T | T | $\mathrm{D}_{15}$ to $\mathrm{D}_{8}$ |
|  | 7 |  | T | T | keep | - | I/O port |
| P4 ${ }_{7}$ to P40 | 1 to 6 | 8-bit bus | T | T | keep | keep | I/O port |
|  |  | 16-bit bus | T | T | T | T | $\mathrm{D}_{7}$ to $\mathrm{D}_{0}$ |
|  | 7 |  | T | T | keep | - | I/O port |
| $\mathrm{P}_{3}$ to P5 | 1 to 4 |  | L | T | T | T | $\mathrm{A}_{19}$ to $\mathrm{A}_{16}$ |
|  | 5, 6 |  | T | T | keep | T | Input port <br> (DDR = 0) |
|  |  |  |  |  | T | T | $\begin{aligned} & A_{19} \text { to } A_{16} \\ & (D D R=1) \end{aligned}$ |
|  | 7 |  | T | T | keep | - | I/O port |


| Pin Name | Mode | Reset | Hardware <br> Standby <br> Mode | Software <br> Standby <br> Mode | Bus- <br> Released <br> Mode | Program <br> Execution, <br> Sleep Mode |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{P} 6_{0}$ | 1 to 6 | T | T | keep | keep | I/O port |


|  | 7 | T | T | keep | - | I/O port |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| P6, | 1 to 6 | T | T | $\begin{aligned} & \text { keep } \\ & (\text { BRLE }=0) \end{aligned}$ | T | I/O port BREQ |
|  |  |  |  | $\begin{aligned} & \mathrm{T} \\ & (\mathrm{BRLE}=1) \end{aligned}$ |  |  |


|  | 7 | T | T | keep | - | I/O port |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| P6 ${ }_{2}$ | 1 to 6 | T | T | keep $(\mathrm{BRLE}=0)$ | L | I/O port (BRLE = 0 ) |
|  |  |  |  | $\begin{aligned} & \mathrm{H} \\ & (\mathrm{BRLE}=1) \end{aligned}$ |  | or BACK <br> (BRLE = 1) |
|  | 7 | T | T | keep | - | I/O port |
| $\mathrm{Pb}_{6}$ to $\mathrm{P6}_{3}$ | 1 to 6 | $\mathrm{H}^{* 2}$ | T | T | T | $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}$, HWR, LWR |


|  | 7 | T | T | keep | - | I/O port |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{P7}_{7}$ to P70 | 1 to 7 | T | T | T | T*' | Input port |
| P80 | 1 to 6 | T | T | $\begin{aligned} & \text { keep } \\ & (\text { RFSHE }=0) \end{aligned}$ | $\begin{aligned} & \text { keep } \\ & (\text { RFSHE }=0) \end{aligned}$ | $\begin{aligned} & \text { I/O port } \\ & (\text { RFSHE }=0) \end{aligned}$ |
|  |  |  |  | $\begin{aligned} & \overline{\mathrm{RFSH}} \\ & (\mathrm{RFSHE}=1) \end{aligned}$ | $\begin{aligned} & \mathrm{H} \\ & (\text { RFSHE }=1) \end{aligned}$ | $\begin{aligned} & \text { or } \overline{\text { RFSH }} \\ & \text { (RFSHE }=1) \end{aligned}$ |


|  | 7 | T | T | keep | - | I/O port |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{P8}_{3}$ to $\mathrm{P8}_{1}$ | 1 to 6 | T | T | T | keep | Input port |
|  |  |  | $(\mathrm{DDR}=0)$ | $(\mathrm{DDR}=0)$ | $(\mathrm{DDR}=0)$ or |  |
|  |  |  | H | H | $\mathrm{CS}_{3}$ to $\overline{\mathrm{CS}}_{1}$ |  |
|  |  |  | $(\mathrm{DDR}=1)$ | $(\mathrm{DDR}=1)$ | $(\mathrm{DDR}=1)$ |  |


|  | 7 | T | T | keep | - | 1/O port |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\mathrm{P} 8_{4}}$ | 1 to 6 | L | T | $\begin{aligned} & \mathrm{T} \\ & (\mathrm{DDR}=0) \end{aligned}$ | keep $(\mathrm{DDR}=0)$ | Input port <br> (DDR = 0) |
|  |  |  |  | $\begin{aligned} & \mathrm{L} \\ & (\mathrm{DDR}=1) \end{aligned}$ | $\begin{aligned} & \mathrm{H} \\ & (\mathrm{DDR}=1) \end{aligned}$ | or $\mathrm{CS}_{0}$ $(\mathrm{DDR}=1)$ |
|  | 7 | T | T | keep | - | I/O port |
| $\mathrm{P9}_{6}$ to $\mathrm{P9}$ 。 | 1 to 7 | T | T | keep | keep*1 | I/O port |


| Pin Name | Mode | Reset | Hardware <br> Standby <br> Mode | Software Standby Mode | Bus- <br> Released <br> Mode | Program Execution, Sleep Mode |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{PA}_{3}$ to $\mathrm{PA}_{0}$ | 1 to 7 | T | T | keep | keep*1 | I/O port |
| $\mathrm{PA}_{6}$ to $\mathrm{PA}_{4}$ | 3,4,6 | T*3 | T | H (CS output) | H (CS output) | $\overline{\mathrm{CS}}_{6}$ to $\overline{\mathrm{CS}}_{4}$ (CS output) |
|  |  |  |  | T (address output) | T (address output) | A23 to A21 (address |
|  |  |  |  | keep (otherwise) | keep (otherwise) | output) <br> I/O port (otherwise) |
|  | 1,2,5,7 | T*3 | T | keep | keep*1 | I/O port |
| $\mathrm{PA}_{7}$ | 3, 4, 6 | L*3 | T | T | T | $\mathrm{A}_{20}$ |
|  | 1,2,5,7 | T*3 | T | keep | keep*1 | I/O port |
| $\begin{aligned} & \mathrm{PB}_{7}, \\ & \mathrm{~PB}_{5} \text { to } \mathrm{PB}_{0} \\ & \hline \end{aligned}$ | 1 to 7 | T | T | keep | keep*1 | I/O port |
| $\mathrm{PB}_{6}$ | 3, 4, 6 | T | T | H (CS output) | H (CS output) | $\overline{\mathrm{CS}}_{7}$ (CS output) |
|  |  |  |  | keep (otherwise) | keep (otherwise) | I/O port (otherwise) |
|  | 1,2, 5, 7 | T | T | keep | keep*1 | I/O port |

## Legend

H: High
L: Low
T: High-impedance state
keep: Input pins are in the high-impedance state; output pins maintain their previous state.
DDR: Data direction register bit
Notes: 1. The bus cannot be released in mode 7.
2. During direct power supply, oscillation damping time is "H" or "T".
3. During direct power supply, oscillation damping time differs between " H ", " $L$ " and "T".

## D. 2 Pin States at Reset

Reset in $\mathbf{T}_{1}$ State: Figure D. 1 is a timing diagram for the case in which $\overline{\text { RES }}$ goes low during the $\mathrm{T}_{1}$ state of an external memory access cycle. As soon as $\overline{\mathrm{RES}}$ goes low, all ports are initialized to the input state. $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{HWR}}$, and $\overline{\mathrm{LWR}}$ go high, and the data bus goes to the high-impedance state. The address bus is initialized to the low output level 0.5 state after the low level of $\overline{\operatorname{RES}}$ is sampled. Sampling of $\overline{\text { RES }}$ takes place at the fall of the system clock $(\phi)$.


Figure D. 1 Reset during Memory Access (Reset during T $\mathbf{T}_{1}$ State)
Rev. 2.0, 03/01, page 812 of 822

Reset in $\mathbf{T}_{2}$ State: Figure D. 2 is a timing diagram for the case in which $\overline{\text { RES }}$ goes low during the $\mathrm{T}_{2}$ state of an external memory access cycle. As soon as $\overline{\mathrm{RES}}$ goes low, all ports are initialized to the input state. $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{HWR}}$, and $\overline{\mathrm{LWR}}$ go high, and the data bus goes to the high-impedance state. The address bus is initialized to the low output level 0.5 state after the low level of $\overline{\operatorname{RES}}$ is sampled. The same timing applies when a reset occurs during a wait state ( $\mathrm{T}_{\mathrm{w}}$ ).


Figure D. 2 Reset during Memory Access (Reset during $\mathrm{T}_{2}$ State)

Reset in $\mathbf{T}_{3}$ State: Figure D. 3 is a timing diagram for the case in which RES goes low during the $\mathrm{T}_{3}$ state of an external three-state space access cycle. As soon as $\overline{\mathrm{RES}}$ goes low, all ports are initialized to the input state. $\overline{\mathrm{AS}}, \overline{\mathrm{RD}}, \overline{\mathrm{HWR}}$, and $\overline{\mathrm{LWR}}$ go high, and the data bus goes to the highimpedance state. The address bus outputs are held during the $T_{3}$ state. The same timing applies when a reset occurs in the $\mathrm{T}_{2}$ state of an access cycle to a two-state-access area.


Figure D. 3 Reset during Memory Access (Reset during T3 State)

# Appendix E Timing of Transition to and Recovery from Hardware Standby Mode 

## E. 1 Timing of Transition to Hardware Standby Mode

(1) To retain RAM contents with the RAME bit set to 1 in SYSCR, drive the $\overline{\operatorname{RES}}$ signal low 10 system clock cycles before the $\overline{\text { STBY }}$ signal goes low, as shown below. The minimum delay from the fall of the $\overline{\text { STBY }}$ signal to the rise of the $\overline{\mathrm{RES}}$ signal is 0 ns .


Figure E. 1 Timing of Recovery from Hardware Standby Mode (1)
(2) To retain RAM contents with the RAME bit cleared to 0 in SYSCR, or when RAM contents do not need to be retained, $\overline{\mathrm{RES}}$ does not have to be driven low as in (1).

## E. 2 Timing of Recovery from Hardware Standby Mode

Drive the $\overline{\mathrm{RES}}$ signal low approximately 100 ns before $\overline{\text { STBY }}$ goes high.


Figure E. 1 Timing of Recovery from Hardware Standby Mode (2)

## Appendix F Product Code Lineup

Table F. 1 H8/3052F Product Code Lineup

| Product Type |  | Product Code | Mark Code | Package <br> (Hitachi Package Code) |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline \text { H8/3052 } \\ & \text { F-ZTAT } \end{aligned}$ | 5 V version | HD64F3052TE | HD64F3052TE | 100-pin TQFP (TFP-100B) |
|  |  | HD64F3052F | HD64F3052F | 100-pin QFP (FP-100B) |
| H8/3052 F-ZTAT <br> $B$ mask version | 5 V version | HD64F3052BTE | HD64F3052BTE | 100-pin TQFP (TFP-100B) |
|  |  | HD64F3052BF | HD64F3052BF | 100-pin QFP (FP-100B) |
|  | 3 V version | HD64F3052BVTE | HD64F3052BVTE | 100-pin TQFP (TFP-100B) |
|  |  | HD64F3052BVF | HD64F3052BVF | 100-pin QFP (FP-100B) |

## Appendix G Package Dimensions

Figure G. 1 shows the FP-100B package dimensions of the H8/3052 F-ZTAT. Figure G. 2 shows the TFP-100B package dimensions.


Figure G. 1 Package Dimensions (FP-100B)

*Dimension including the plating thickness
Base material dimension

| Hitachi Code | TFP-100B |
| :--- | :--- |
| JEDEC | - |
| EIAJ | Conforms |
| Weight (reference value) | 0.5 g |

Figure G. 2 Package Dimensions (TFP-100B)

## Appendix H Differences from H8/3048F-ZTAT

Table H. 1 Differences between H8/3052F-ZTAT and H8/3048F-ZTAT

| Item | H8/3048F-ZTAT |  |  |  | H8/3052F-ZTAT |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Pin specifications | Pin $1 \rightarrow \mathrm{~V}_{\mathrm{cc}}$ |  |  |  | 5V Operation <br> Pin $1 \rightarrow \mathrm{~V}_{\mathrm{c}}$ <br> Connected to $\mathrm{V}_{\text {ss }}$, with external connection of $0.1 \mu \mathrm{~F}$ capacitor |  |  |  |
|  |  |  |  |  | 3V Operation <br> Pin $1 \rightarrow V_{\text {cc }}$ <br> Connected to system power supply |  |  |  |
|  | Pin $10 \rightarrow \mathrm{~V}_{\text {pp }} / \overline{\text { RESO }}$ |  |  |  | Pin $10 \rightarrow$ FWE |  |  |  |
| ROM/RAM | 128 kbytes dual-power-supply flash memory <br> 4 kbytes RAM |  |  |  | 512 kbytes single-power-supply flash memory <br> 8 kbytes RAM |  |  |  |
| Program/erase voltage | 12 V application |  |  |  | $\mathrm{V}_{\mathrm{cc}}$ single power supply |  |  |  |
| $\mathrm{V}_{\mathrm{pp}}$ pin function | Multiplexed as RESO pin |  |  |  | FWE function only (RESO function eliminated) |  |  |  |
| Boot mode setting method | $\overline{\mathrm{RESO}}=$ | MD2 | MD1 | MDO | FWE = 1 | MD2 | MD1 | MDO |
|  | Mode 5 | 12 V | 0 | 1 | Mode 5 | 0 | 0 | 1 |
|  | Mode 6 | 12 V | 1 | 0 | Mode 6 | 0 | 1 | 0 |
|  | Mode 7 | 12 V | 1 | 1 | Mode 7 | 0 | 1 | 1 |
|  | Reset release |  |  |  | Set to: <br> mode 1 in case of mode 5 mode 2 in case of mode 6 mode 3 in case of mode 7 Reset release |  |  |  |
| User program mode setting method | $\overline{\text { RESO }}=$ | MD2 | MD1 | MDO | FWE = 1 | MD2 | MD1 | MDO |
|  | Mode 5 | 1 | 0 | 1 | Mode 5 | 1 | 0 | 1 |
|  | Mode 6 | 1 | 1 | 0 | Mode 6 | 1 | 1 | 0 |
|  | Mode 7 | 1 | 1 | 1 | Mode 7 | 1 | 1 | 1 |
|  | Reset release |  |  |  | Reset release |  |  |  |
| Programming processing | Block corresponding to programming addresses is set in EBR1/EBR2 before programming |  |  |  | No setting |  |  |  |


| Item | H8/3048F-ZTAT |  |  |  |  |  |  |  | H8/3052F-ZTAT |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FLMCR | FLMCR (H'FF40) |  |  |  |  |  |  |  | FLMCR1 (H'FF40) |  |  |  |  |  |  |  |
|  | $\mathrm{V}_{\text {PP }}$ | $\mathrm{V}_{\mathrm{PP}} \mathrm{E}$ | - | - | EV | PV | E | P | FWE | SWE1 | ESU1 | PSU1 | EV1 | PV1 | E1 | P1 |
|  |  |  |  |  |  |  |  |  | FLMCR2 (H'FF41) |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  | FLER | SWE2 | ESU2 | PSU2 | EV2 | PV2 | E2 | P2 |



Item


| Refresh controller | In modes 1 to 6, DRAM or PSRAM can be directly connected to area 3. | In modes 1, 2, 3, 4, and 6, DRAM or PSRAM can be directly connected to area 3. <br> Cannot be used in mode 5 (because flash area overlaps area 3). |
| :---: | :---: | :---: |
| DMAC registers MAROAR, MAROBR, MAR1AR, MAR1BR | MAR0AR (H'FF20), MAR0BR (H'FF28), MAR1AR (H'FF30), MAR1BR (H'FF38) <br> All bits are reserved; they always return 1 if read, and cannot be modified. | MAR0AR (H'FF20), MAR0BR (H'FF28), MAR1AR (H'FF30), MAR1BR (H'FF38) <br> All bits are reserved; they return an undefined value if read, and cannot be modified. |
| A/D register ADCR | ADCR (H'FFE9) <br> Initial value: $\mathrm{H}^{\prime} 7 \mathrm{~F}$ <br> Bit 7 only is readable/writable. <br> Other bits are reserved; they always return 1 if read, and cannot be modified. | ADCR (H'FFE9) <br> Initial value: H'7E <br> Bit 7 only is readable/writable. Bit 0 is reserved, and must not be set to 1 . Other bits are reserved; they always return 1 if read, and cannot be modified. |
| WDT register RSTCSR | RSTCSR (H'FFAB) <br> Initial value: '3F <br> Bits 7 and 6 only are readable/writable. Other bits are reserved; they always return 1 if read, and cannot be modified. | RSTCSR (H'FFAB) <br> Initial value: '3F <br> Bit 7 only is readable/writable. <br> Bit 6 is reserved, and must not be set to 1 . <br> Other bits are reserved; they always return 1 if read, and cannot be modified. |

Note: The H8/3052F-ZTAT program/erase procedures are different from those of the H8/3048F-ZTAT.

Rev. 2.0, 03/01, page 821 of 822

## H8/3052 F-ZTAT ${ }^{\text {TM }}$ Hardware Manual

Publication Date: 1st Edition, January 2000
2nd Edition, March 2001
Published by: Electronic Devices Sales \& Marketing Group Semiconductor \& Integrated Circuits Hitachi, Ltd.
Edited by: Technical Documentation Group Hitachi Kodaira Semiconductor Co., Ltd.
Copyright © Hitachi, Ltd., 2000. All rights reserved. Printed in Japan.


[^0]:    Legend: OSetting available (valid). - Setting does not affect this mode.

[^1]:    Note: * Do not execute a SLEEP instruction while the SWE1 bit is set to 1 .

[^2]:    Note: * Do not access flash memory while the E2 bit is set to 1 .

[^3]:    Notes: 1. RTCNT and bits 7 and 6 of RTMCSR are initialized. Other bits and registers hold their previous states.
    N. State in which the cor (MSTCR). up the module registers again.

    SSBY: Software standby bit
    MSTCR: Module standby control register

